

SPECIAL ISSUE-Part 1 Product Showcase No 27 Highlighting key trends in power sources, software, integrated circuits, and hardware and interconnects

Expanded literature section

## ELECTRONIC TECHNOLOGY FOR ENGINEERS AND ENGINEERING MANAGERS



## When World Class Precision and Performance Are Required



## Highest precision available with $10 \mu V$ input offset voltage from Raytheon.

Raytheon's RC4077 Series high-precision op amp family offers the highest performance in the industry. Looking for the lowest input offset voltage-Raytheon has it. The lowest power dissipationRaytheon has it. The 4077 Series can upgrade your system to new heights of precision and performance. You can depend on Raytheon's reliability and advanced design techniques to enhance your system.
$\square$ Ultimate prec̣ision: $\pm 10 \mu \mathrm{~V}$ maximum guaranteed input offset voltage, delivered in a variety of package types including low-cost commercial plastic DIPs, sets the RC4077 series apart from other precision op amps. No monolithic op amp-except noisy chopperstabilized types-has better $V_{\text {OS }}$ perfor-
mance. Additionally Raytheon offers an 8 -lead SOIC specified at $\pm 25 \mu \mathrm{~V}$.
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PSRR: 110 dB minimum
$\square$ Companion product: Raytheon's LT1001 high-precision, high-performance op amp follows RC4077's lead with a very low $15 \mu \mathrm{~V}$ offset voltage. The LT1001 offers 2 nA offset current and gain of .45 million minimum.

No wait: The RC4077, LT1001 and other members of Raytheon's broad line of op amps are available now from your
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There are two sides to this story...

## Side One:

Highly integrated 16-bit industrial computer Ziatech's NEC V50-based single board computer, the ZT 8816, packages the features of several STD boards into a unique, dual-sided surface-mount design. The ZT 8816 tackles demanding industrial applications with a 16 -bit data bus, an 832 K on-board memory capacity, a real-time battery-backed clock, AC/DC power-fail protection, DMA controller, an interrupt controller, two serial channels, and three counter-timers.

## For the rest of the story...

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Call today for the ZT 8816 Technical Data Sheet and the 24page STD DOS Technical Brochure. With more information on what the ZT 8816 can do for your industrial application, you may start seeing the Ziatech side of the story.
(805) 541-0488

## Side Two:

IBM AT-compatible industrial computer
The ZT 8816 is more than just the most advanced STD Bus computer hardware on the market today. It is designed to operate PC DOS or ROM-based user programs such as the VRTX multitasking kernel. Development tools are available to provide a large range of target system software architectures. STD DOS V50 on the ZT 8816 delivers IBM AT performance and compatibility with optional networking, EGA video, disk and bubble memory subsystems, multiprocessing, and a device driver library. Ziatech's exclusive Virtual System Console supports easy development through a host PC by transparent resource sharing.

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#  

## "For a bunch of companies that don't always agree on everything, we sure were unanimous on VTC."

The VME Consortium needed an economical, yet highly functional VME bus interface chip, to minimize design time . . . and to help raise the VME standard to higher levels.
"We looked at the leading suppliers," said Joe Ramunni, consortium chairman (and president of Mizar), "and VTC came out on top. Their CMOS standard-cell ASIC approach gave us the high drive capability we needed, optimized for bus interfacing. And, it proved much more cost-effective, with higher performance, than gate array technology.'

The VME Consortium is made up of such firms as Plessey Microsystems, Omnibyte Corporation, Mizar Inc., Ironics Inc., Heurikon Corporation, Matrix Corporation, and Clearpoint Inc., among others. What did they look for in a supplier?
"We needed a credible business partner," said Ramunni, "with a proven track record, who could provide a turnkey package . . . both design and fab. A supplier that could produce in quantity, and provide technical support to the market at large.
"We also needed a firm with an international marketing structure, because we expect this chip to be the de facto standard worldwide.
"But, we needed people we could work with, too. VTC had the right 'comfort factor'.'
Jack Regula, consortium technical director (and VP-R\&D, Ironics) added: "Our requirements for high speed, high gate-count, low power consumption, and VME bus drive capability were all met well with VTC's 1-micron CMOS standard cell library. And we were extremely impressed with VTC's facilities, its people, and its customer list."

In the future, the VME bus chip (VIC) will become a standard cell within VTC's CMOS library, to allow customers to further customize the chip.

Shouldn't you be getting to know VTC, too? You'll be in good company when you do. Call or write us today, and we'll send you our short-form product catalog, which describes our product offerings in linear signal processing, high-speed CMOS logic, mass storage ICs, bipolar ASIC, and CMOS ASIC.

VTC Incorporated, 2401 East 86th Street, Bloomington, MN 55420. (In Minnesota, 612/851-5200.) Telex 857113.

Joseph Ramunni, chairman (left), and Jack Regul Joseph Ramunni, chairman (echnology Consortium



## Side by side comparison of our data logger and theirs.

Wavetek has brought an exciting new dimension to data logging-small. Our Series 50 Data Logger is a fraction the size and considerably less expensive than the one on the right. And our data logger is light enough to be easily carried by mere mortals. When you compare the rest of the features, the competition drops right out of sight.

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- dBw, dBm
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- Pulse Width, Time Interval
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- Diode Junction Voltage

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We could go on for pages, but rather than weigh you down with specifications, we'd rather show you how Series 50 will make your job easier. Please call, or write for our brochure. Wavetek San Diego, Inc., P.O. Box 85265, San Diego, CA 92138. Tel. (619) 279-2200; TWX 910-335-2007.

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| Si9110/11 | High-voltage (10-120 V) switch- <br> mode controller with high-speed, <br> source-sirik output drive |

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On the cover: Part I of EIDN's Product Showcase No 27 puts products in their places. Staff-mritten articles lead off cach product section and cover various topics: the advantages of using current-feedback op amps (pg 84); bigh-power supplies and the question of efficiency (pg 162); desimu comsiderations in the use of multilaver backplanes (pa 208); and the necd for-cross-development softurate tools (pa 2.52). (Plootopraplyy by Dana Sifall; art dinection (oy Katblecn Rutbl)

## DESIGN FEATURES <br> Integrated Circuits

## Current-feedback op amps 84 ease high-speed circuit design

Because they're free from gain-bandwidth limitations, currentfeedback op amps excel in high-frequency and fast-settling circuitry. The latest models also achieve a fair amount of de accuracy without sacrificing speed-an important consideration when you're designing high-frequency circuits such as flash A/D converters.-Peter Harold, European Editor

## Power Sources

## High-power switching supplies 162 stress efficiency

Manufacturers of high-power supplies tend to emphasize efficiency over other performance considerations. Most of these supplies operate at frequencies well below 100 kHz and use half- or full-bridge circuits. Yet some manufacturers are shifting to higher frequencies and to the use of power MOSFETs instead of bipolar transistors.-Dave Pryct, Associate Editor

## Hardware and Interconnect Devices

## Multilayer backplanes require careful design specs

Dense, multilayer backplanes are necessary for connecting today's heavily populated daughter boards. But to get the highest performance out of your system, you have to keep in mind design considerations that reduce noise and prevent transmission degradation.- $/$ I) Mosley, Regional Editor.

## Software

## Integrated tool sets simplify 252 soft ware cross-development

The prevalence of various types of workstations has not only automated some software-development tasks-it's also created a need for sophisticated cross-development tools. Using these packages, you can now write software more easily and efficiently for embedded systems or for structurally dissimilar $\mu \mathrm{Ps}$.-Cloris Tov?Y, Associate Editor

Continued on page 7

[^0]
## Fluke breaks the old mold.



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Product coverage in this issuc bigins with a mevicm of ICs and scmiconductons (prg 100). (encrafe continutes with meriows of power sommes (pic 174), bardwan and intcronuact devices (per 219), and softwame (pll 200 ).

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[^1]
## We've Invented the Future of Instrumentation Software . . . Twice.

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## EDITORIAL

Another computer myth-that of the paperless office-finally comes to an end.

## PROFESSIONAL ISSUES

How two enginecrs built-and nearly lost-their business.-Deborah
Asbrand, Associate Editor.

## LOOKING AHEAD

Electronic still imaging to gross $\$ 540 \mathrm{M}$ in 1992 . . Demand for T3 test devices is expected to boom.

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## A product-oriented design aid

To save you time in your efforts to keep current, EIDN's editors have survered the new-product offerings from thousands of companies, screening and selecting only the most significant of those offerings introduced in the last six months. We present our findings-the best of the best-in a format devised to make your product selection as casy as possible. You can keep this Product Showease as a reference until the next one that covers these four key product areas appears in December.

## This year,you'll <br> hear a lot of claims <br> that "systems" <br> design automation has arrived.

# At Mentor Graphics, we know better. And so do our customers. 

## Skeptical about "systems" electronic design automation?

You should be. Because in many cases, it's a triumph of form over content. Look behind the facade of so-called "systems" design automation tools, and you'll find little substance, if any.

Buy into this kind of systems design methodology, and you're participating in a very costly experiment. With highly uncertain results.

## They preach. We practice.

There's only one practical yardstick for evaluating a systems design solution. And that's how many successful products it has produced. Apply this measure and the field narrows dramatically. Essentially, down to a single vendor.

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## To be continued.

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# Hitachi's HMCS400 Series of CMOS Microcontrollers <br> The intelligent answer for small system control problems 

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## Resourceful

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## How to crack 386 protected mode.



Unlock selectors and descriptor tables. Break open task state segments (TSS) and call gates. Microtek's In-Circuit-Emulator (MICE) cracks 80386 protected mode with real-time, non-intrusive emulation to 20 MHz .

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Crack 80386 protected mode, and open the door to exciting design possibilities-with the best 80386 emulation system now available. From MicroCASE.


## NEWS BREAKS

EDITED BY JOANNE CLAY

## ISDN TRANSCEIVER ICs EXCEED ANSI SPECS

The MCl45474 and MCl45475 CMOS single-chip ISDN transceiver ICs have passed CCITT and ANSI conformance testing, and are available in sample quantities from Motorola Inc (Austin, TX, (512) 928-7944). These transceivers offer multiframing Sand Q-channel operation and maintenance-signaling channels. An adaptive receiver circuit automatically selects the optimum sampling phase and detection threshold of incoming signals, thereby permitting both devices to double the point-to-point range required by ISDN's I. 430 specification. The 28 -pin MCl45475 is suitable for use in an NTl Star bus configuration. The 22-pin MCl45474 costs $\$ 17$ (1000), and the MC145475 sells for $\$ 18$ (1000).

Both chips come with selectable NT and TE modes for both line-card and terminal applications. An Interchip Digital Link (IDL) interface lets these devices exchange Band D-channel information among a variety of ISDN components and systems. Other features include loop-back support, activation and deactivation functions, extendedrange operation, and a line driver that can tolerate a l:l transformer ratio without the need for complex protection circuitry.-J D Mosley

## 12-BIT A/D CONVERTER MOVES ONTO ASIC

Sierra Semiconductor (San Jose, CA, (408) 263-9300) now offers a standard cell for an A/D converter (ADC) that can resolve 12 bits. The ADC12B is a 12 -bit-plus-sign, dualslope integrating ADC standard cell for the company's CMOS-based analog/digital ASICs. The ADC12B is similar to the standard 7109 ADC; it can resolve a minimum of $50 \mu \mathrm{~V}$ and has a typical conversion rate of 10 conversions/sec. The device requires 1.5 mA in active mode and $1 \mu \mathrm{~A}$ in power-down mode. The ADCl 2 B is the newest member of a cell library that contains 50 analog cells, 300 digital cells, and 20 EEPROM cells.
-David Shear

## THREE VENDORS OFFER TRANSLATORS FOR GENRAD 179X PROGRAMS

The news that ATE vendor GenRad Inc (Concord, MA, (617) 369-4400) will no longer be producing the 179X series of board-test systems has prompted announcements by two of the firm's competitors as well as by GenRad itself. Both Teradyne Inc (Boston, MA, (617) 482-2700) and Schlumberger Technologies (San Jose, CA, (408) 998-0123) now offer facilities that permit users of the GenRad systems to migrate to other testers without compromising their investment in board-test programs and fixtures. That ability is particularly important to defense-electronics manufacturers and armedforces repair depots, because they deal with assemblies whose life cycles are longer than those of most commercial products.

Teradyne's 179X/L200 Program Translator, which starts at \$25,000, creates programs that run on the company's L200 test systems. These programs support go/no-go testing and guided-probe fault diagnosis; an option converts the programs to the format of Teradyne's Lasar Version 6 simulator. Schlumberger's CAPS/ITG, which adapts programs to run on the firm's 700 Series testers, includes similar capabilites. In addition, Schlumberger offers an adapter that lets you use 179X test fixtures on the 700 Series. Including the hardware adapter, CAPS/ITG for the 179 X costs under $\$ 50,000$. Meanwhile, GenRad continues to support the l79X Series in several ways. The company supplies spare parts and maintenance service for the testers. It has also refurbished used systems and furnished them to customers who wanted to expand capacity without migrating to different equipment, and it may continue to offer that service depending on the availability of used machines. Finally, the company offers software
(priced from $\$ 5000$ ) and fixture adapters to let you run 179X Series programs on the newer 2750 Series.-Dan Strassberg

## MONOLITHIC BUFFER AMPLIFIER ACHIEVES '730-MHz BANDWIDTH

The CLCllo unity-gain buffer amplifier from Comlinear Corp (Fort Collins, CO, (303) 226-0500) features a $-3-\mathrm{dB}$ bandwidth of 730 MHz with a guaranteed gain flatness of better than 1 dB from dc to 200 MHz . It thus allows you to buffer very-high-frequency signals in circuits such as phase-locked loops and flash A/D converters. At dc, the device exhibits a maximum output offset voltage of 13 mV . Packaged in an 8-pin plastic DIP, the commercial version of the device (the CLCllOAJP) costs $\$ 15.50$ (100).
-Steven H Leibson

## SHORT-DISTANCE OPTICAL LINK CARRIES THREE 40-MHz ANALOG SIGNALS

A 3-channel, fiber-optic transmission system from Molex Inc (Lisle, IL, (312) 969-4550) accommodates analog signals with frequencies of 40 Hz to 40 MHz over as much as 5 m of cable. Inputs require full-scale drive currents of 30 mA p -p, which produce an output signal of 108 mV p-p riding on a 3.3 V dc offset. A 3-channel system, including transmitters, receivers, and a terminated 2 m cable assembly, costs $\$ 98$.
-Steven H Leibson

## 1- $\mu$ m CHIP FAMILY IS FASTEST ECL TECHNOLOGY AVAILABLE

A new family of ICs that implement standard logic functions represents the fastest ECL technology available, according to Sony Corp of America's Component Products Div (Cypress, CA). Introduced at the Electro show in Boston (May 10 through May 12, 1988), the ECL chips feature $1-\mu \mathrm{m}$ geometry, $60-\mathrm{psec}$ gate delays, and $4-\mathrm{GHz}$ operation, which rivals the speed of GaAs devices. The 24 initial offerings cost from $\$ 40$ to $\$ 70$ (1000). They come in metal flat packs; more economical versions in plastic leaded chip carriers will be available soon. All the functions are derived from a 200 -gate chip that will become available in August as a product for custom digital applications. For linear-ASIC applications, the company will introduce an ECL-technology analog array in October 1988. The array includes $3-\mathrm{GHz}$ npn transistors and $350-\mathrm{MHz}$ vertical pnp transistors.-Tarlton Fleming

## SCSI-CONTROLLER CHIP INTERFACES DIRECTLY TO PC BUS

The 53C400 SCSI controller from NCR Microelectronics Div (Colorado Springs, CO, (800) 334-5454) comprises bus-interface circuitry, two 128-byte RAM buffers, addressdecode and interrupt logic, and a 53C80 core cell with high-current SCSI-bus drivers. The controller interfaces directly to IBM's PC, PC/AT, and PS/2 buses. The $\$ 13.51$ (1000) chip performs SCSI-bus transfers at greater than 2 M bytes/sec and accomplishes PC-bus burst transfers at 1 M byte/sec.-Steven H Leibson

## CLAMPING AMPLIFIER GUARDS A/D CONVERTERS FROM OVERLOAD

You can protect circuits from off-scale voltage excursions with the $\$ 17.10$ (100) CLC501 clamping op amp. The high-speed device features a $-3-\mathrm{dB}$ bandwidth of 80 MHz (at a gain of 32), but unlike conventional op amps, it includes two extra clamping-voltage inputs, $\mathrm{V}_{\mathrm{H}}$ and $\mathrm{V}_{\mathrm{L}}$, that you can use to confine its output between two voltage levels. One possible way to apply the part would be to use it as a residue amplifier in a 2-stage, subranging A/D-conversion system. The op amp's clamping action protects the second-stage A/D converter from overload damage. In addition, the CLC501 speeds up the subranging A/D-conversion system's recovery rate from overloads, because the op amp recovers more quickly than an A/D converter (the op amp typically recovers in 1 nsec ).-Steven $H$ Leibson

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## NEWS BREAKS: international

## DELTA CODEC MEETS MILITARY COMMUNICATIONS REQUIREMENTS

Meeting the requirements of the Eurocom D1-IA8 specification, the FX619 singlechip CVSD (continuously variable slope delta) codec is suitable for use in a variety of military communications equipment. The device is manufactured by Consumer Microcircuits Ltd (Witham, UK, TLX 99382; in the US: Mx-Com Inc, Winston-Salem, NC, (919) 744-5050). It operates in full-duplex mode, and it includes input and output audio filters. An on-chip oscillator clocks the device's on-chip switched-capacitor filters and its CVSD modulator and demodulator. The oscillator accepts a crystal or an external clock input.

You can program the device's sampling rate to $16 \mathrm{k}, 32 \mathrm{k}$, or 64 k bps. The chip provides clock outputs so that you can synchronize external circuitry to the codec. Other programmable features include an encoder-enable/disable facility, encoder/decoder idle modes, 3 - or 4 -bit companding algorithms, and a low-power standby mode. The FX619 is a CMOS device that operates from a single 5 V supply and is available in a 22-pin DIP or a 28-lead surface-mount package. The company expects to sell the chip for less than $£ 20$ (1000).-Peter Harold

## X. 25 COMMUNICATIONS CARD IMPLEMENTS OSI LAYERS 1 TO 3

Simplifying the software that you have to write during systems integration, the $\$ 3675$ CC-125 VME Bus X. 25 communications card conforms to the X. 213 specification for communication to layer-4 functions of the OSI model. The card, which is available from Compcontrol (Eindhoven, The Netherlands, TLX 51603; in the US: Los Gatos, CA (408) 356-3817), implements layers 1 to 3 of the OSI model.

Layer 1 operates in accordance with the X. 21 bis (V.24/V.28) standard, and layer 2 conforms to the LAPB (link access procedures balanced) protocol standard. The board's layer-3 software handles the X. 25 packet-level protocol and imposes flow control to prevent the communications link from overloading. Messages are passed between the CC-125 and the VME Bus system's layer-4 functions via dual-port RAM on the CC-125. The messages, coded according to X. 409 recommendations, are transferred to the dualport RAM in the form of network-service data units that can be as long as 4 k bytes. The 68000-based board is suitable for transmitting packet-switched data over public or private data networks at bit rates as high as 64 k bps.-Peter Harold

## MODULE PUTS IBM PC/AT-COMPATIBLE COMPUTER IN VME BUS SYSTEMS

The $\$ 2500$ PX4000 VME Bus module from Philips' Industrial and Electroacoustic Systems Div (Eindhoven, The Netherlands, TLX 35000; in the US: (201) 529-3800) allows you to implement an IBM PC/AT-compatible computer in VME Bus systems. The 2-board sandwich runs an $80286 \mu \mathrm{P}$ and an optional 80287 math coprocessor at 8 MHz . It includes all the ports you'd expect to find on a standard PC/AT (for example, keyboard, monitor, printer, and floppy-disk interfaces), and it runs the MS-DOS operating system. However, the module's video output provides CGA-compatible color graphics as a standard feature. In addition to accessing the board's 512 k bytes of onboard dynamic RAM, processor and DMA channels can transparently access additional memory via the VME Bus. If you use the $\mu$ P's protected-addressing mode, the $\mu \mathrm{P}$ can access as much as l6M bytes of memory. The onboard RAM is also accessible to other VME Bus masters. A hard-disk-controller board that operates in conjunction with the PX4000 is also available.-Peter Harold

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|  | +03 | $\left\|\begin{array}{c}\text { DC-1000 } \\ 0.6\end{array}\right\|$ | 1000-1500 0.8 | $\begin{gathered} \mathrm{DC}-1000 \mathrm{MHz} \\ 1.3 \end{gathered}$ | $\left\lvert\, \begin{gathered} 1000-1500 \mathrm{MHz} \\ 1.5 \end{gathered}\right.$ |

*DC-1000 MHz (all 75 ohm or 30 dB models) $\mathrm{DC}-500 \mathrm{MHz}$ (all 40 dB models)
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| :--- | :---: | :---: | :---: |
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| L 293 | 576 | 80 MHz | $\pm 1.5 \mathrm{~ns}$ |
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[^2]

## Delta or Aorta? Which is Which?

No problem here because both of these images were processed on Raytheon's new TDU-850 Thermal Display Unit.
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Phone: (401) 847-8000.

## Raytheon

## A. Satellite view of river delta. <br> B. Arterial angiogram.

Note: These began as continuous tone images which were processed in black and grey by a TDU-850. The TDU-850 images, however, had to be converted to conventional halftones in order to be shown in this magazine. Thus the high quality of the original TDU-850 images have been obscured. For true results ask to see a demonstration.

## SIGNALS \& NOISE

## IEEE is alive and well

IEEE membership will pass the 300,000 mark this year, so the organization is far from heading toward extinction. With 267 sections, over 900 chapters, and over 550 student branches all over the world, and a steady annual growth in spite of editorials such as EDN's ("The IEEE faces extinction," EDN, February $4,1988, \mathrm{pg} 53$ ), it is not hard to see that the IEEE is indeed meeting the needs and desires of many engineers. It is the noisy whiners on the sidelines who make news sometimes-and apparently catch the attention of EDN's editor. The literally thousands of volunteers who man committees, serve as officers in student branches, chapters, subsections, sections, societies, and the national and international boards are dedicated people who work for no pay... just the
desire to make the IEEE better and stronger-and more responsive.

It really hurts to see an editorial like EDN's, because it is obviously giving recognition to the few very vocal dissidents (who give no volunteer effort to the IEEE and who have never served in any office), while ignoring the thousands of volunteers who are moving the IEEE forward through their quiet, dedicated efforts. I hope Jon Titus will take a little time to look at the other side of the coin. The IEEE is doing a lot for the profession! We could do even more if the complainers would join with the many other volunteer workers and help pull the load in improving our Institute.
Robert S Duggan, Jr
Candidate for IEEE Executive
Vice President (1989)
IEEE
Atlanta, GA

## Criticism of IEEE was undeserved

In his editorial of February 4 (pg 53), Jon Titus has indulged a desire to create controversy at the expense of any reasonable approximation of good journalism and attention to facts. "Yellow journalism" was the term used to describe the type of unfounded allegations that abound in the editorial. He certainly has a right to a biased opinion on the vitality of the IEEE and on the merit or lack of it of actions taken by the IEEE's board of directors. But there is a minimum standard of integrity and factual information that should be observed.

Let me be very specific and to the point:

1. Approval voting. Jon says that "the board of directors modified the election procedure by changing the organization's bylaws, not by asking


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## SIGNALS \& NOISE

members to amend the organization's constitution," implying some sort of violation of proper procedure in our action. The facts are that the voting method for member election of officers and directors of the IEEE has not previously been specified in either the constitution or the bylaws; the board of directors felt it would be desirable to specify in the governing documents of the Institute the procedure to be followed, particularly with the introduction of approval voting; and the principle stated in the IEEE's constitution is that " . . . methods of nomination and election shall be specified in the Bylaws."
The board acted properly in placing this information in the bylaws. It would not, in fact, have been in keeping with the guidelines in our constitution to have proposed a constitutional change. Of the three governing documents for the IEEE, the constitution is a very general,

4-page document, while the bylaws provide more-detailed guidance and are about 50 pages long. Our Policies and Procedures Manual is still longer and provides further procedural information.
In his criticism of the board's action, Jon Titus has also conveniently chosen to ignore the fact that the new approval-voting procedure offers the best opportunity for a nonestablishment candidate to win, if he or she truly represents the interests of a preponderance of the voting members. It should be obvious that this method gives voting members greater flexibility to express their views, and helps ensure that the winning candidate most closely reflects the will of the majority. If Jon's point is that he is opposed to majority governance, perhaps he should come right out and say it, rather than leave EDN's readers with a false impression.
2. Professional concerns. Jon

Titus says that "engineers will observe no action on tax or pension reforms, no action on age discrimination, and no action on other professional concerns. In short, no action at all." Perhaps if he took the time to look into these areas, he might find out what's going on, and the extent of the IEEE's role. Pen-sion-reform legislation has been an active target of the IEEE for over a decade. The IRA bill owed much of its momentum to IEEE work. The reduction in tax breaks in the course of the overwhelming pressure for tax reform in 1986 was fought by the IEEE, and we are actively supporting the passage of the newest pension-reform bill dealing with portability and nonintegration. Can Jon point to anyone or any organization that is even close to this record?
Let me go on about other professional concerns. Who stimulated the government guidelines against wage busting and other unfair prac-


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3. Publications. The editorial dis-


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We developed the science of listening.
}
misses journals and publications as not being at "the heart of a vital professional organization." I'm aghast! Jon seems unable to grasp what constitutes a professional in our field. First and foremost is technical competence. We maintain that competence in a variety of ways, not the least of which is via access to technical journals and publications. The only aspect of our professional life that he seems to recognizethat is, the part dealing with the working environment-is covered primarily by the IEEE's US Activities Board (USAB). Its activities, just a few of which I cited earlier, are too numerous and wide-ranging to cover adequately in a letter.

I believe the IEEE has demonstrated and continues to demonstrate its value to the profession in a great number of areas. We're not solely technical and not solely "professional" either. We serve both these aspects and many others besides. We do not believe that we do everything to perfection, and we encourage communication from those who have constructive suggestions on how we can improve. Unfortunately, editorials such as the one I have cited are not constructive and only mislead the casual reader. The engineering community deserves better coverage, and I hope EDN will accept the challenge to provide it.

Perhaps you recall the old adage that it's better to light a single candle than to curse the darkness. The IEEE is lighting the candles.
Russell C Drew
1988 President
IEEE
Sterling, VA

\section*{WRITE IN}

Send your letters to the Signals and Noise Editor, 275 Washington St, Newton MA 02158. We welcome all comments, pro or con. All letters must be signed, but we will withhold your name upon request. We reserve the right to edit letters for space and clarity.

\title{
In the ever-changing microcomputer business, it takes a different kind of thinking \\ to stay one step ahead.
}

\section*{Four years ago, Chips and entered a traditional with a very untraditional}

\section*{Since then, some people have recognized its merits. Including 80\% of the AT-compatible market.}

Microcomputers have evolved considerably since we first started out. But market needs remain much the same: higher performance at lower cost.

The semiconductor industry has always approached these needs from a traditional, discretebased point of view. As systems have become more powerful and sophisticated, this perspective has proven increasingly inadequate.

More and more, systems-level problems call for integrated systems-level solutions. To develop them, we've pioneered a dramatic departure from traditional thinking: a systems approach to semiconductors.

\section*{The systems approach.}

Our design process begins with a broader perspective-a thorough examination of the market trends and advanced technologies that affect the future of microcomputing.

We focus on industry-standard hardware architectures and software environments. To ensure \(100 \%\) compatibility. And to avoid the kind of shortsighted silicon implementations that might require expensive OEM redesigns.

We look closely at complex subsystem relationships, using sophisticated mainframe design techniques to optimize performance. Finally we pull it all together, drawing on our inhouse microcomputer design expertise.

This "tops-down" design process results in highly integrated CHIPSets.' Providing higher system performance, increased functionality and \(100 \%\) compatibility.

\section*{CHIPSets"' make better systems.}

CHIPSets solve system-level hardware and software design problems in silicon, so you don't have to. Which greatly reduces OEM system design time. And lowers development costs.

That means you can get to market faster. With systems that offer higher performance and longer product life cycles. And are less expensive to upgrade. Our systems approach has been so successful that in just a few short years, it's changed the face of the industry.

\section*{Technologies industry approach.}
"Since itsinception, Chips and Technologies has pioneered an unusual concept in the semiconductor industry, involving the sale not just of silicon circuits, but of entire CHIPSets and design services, which incorporate a substantial amount of personal computer systems expertise."

Edward White
John Hancock Financial Services


\section*{Achip is bound by and shor}
"CHIPS offers manufacturers of IBM-compatible personal computers an essentially complete solution, even down to the layout of the printed circuit board that uses CHIPS components."

Drew Peck
Donaldson, Lufkin \& Jenrette

\section*{that merely"clones" all the flaws, imperfections comings of the original.}

\section*{CHIPSets" transcend these limitations.}

In 1985, Chips and Technologies introduced its very first products-the EGA \({ }^{m \mathrm{~m}}\) CHIPSet and the 286/AT CHIPSet for AT-compatible microcomputers.

Both products outperformed standard industry implementations. And dramatically reduced OEM manufacturing costs and system development time as well.

\section*{The second generation: architectural innovation.}

As microprocessors got much faster, system performance was being constrained by the comparatively slow-moving AT bus. Using our systems expertise, we arrived at an ideal solution: a mainframe-like asynchronous architecture that decouples the bus from the processor and memory. Enabling each to run comfortably at maximum speed, with no loss in compatibility.

With this innovation came the first CHIPSets for 386 and high-speed 286 processorsour New Enhanced \(A T\) products. Or NEAT, \({ }^{\text {™ }}\) for short.

These high-performance CHIPSets incorporate advanced microcomputer features, like 4 -way page interleaved memory and shadow RAM. They offer OEMs the ultimate in cost-effective solutions: the ability to accommodate high-speed

286 and \(386^{\mathrm{m} \times}\) processors, with a seamless upgrade path from 12 to 25 MHz and beyond.

That means customers can build 16 MHz AT-compatibles based on lower cost 286 microprocessors. And still run MS-DOS \({ }^{\text {T" }}\) software faster than most 16 MHz 386 -based systems.

NEAT CHIPSets also offer an extra measure of flexibility: automatic mode switching between MS-DOS and OS/2." A real advantage for OEMs and end-users alike.

Unprecedented performance.
NEAT CHIPSets improve the speed and performance of OEM products through architectural innovations that allow faster data communications between the tightly coupled subsystems. And by extending systems logic interface techniques throughout the entire microcomputer.

By implementing mainframe-level techniques in our CHIPSets, we continue to work with OEMs to push AT performance and economy to new levels. For both desktop and laptop systems.

In doing this, we've developed the expertise to do precisely the same for the PS/ \(2^{\text {m" }}\) market.

\section*{With a micro-mainframe proprietary architecture fundamentally changec}

\section*{Systems-level thinking and mainframe expertise are no longer optional. Theyre requirements.}

The features that distinguish PS/2 -mainframe-like architecture, a higher degree of functional integration, and close coupling of hardware and software - are a long way from where traditional semiconductor companies presently find themselves.

\section*{Meeting the new standard.}

While compatibility is only the starting point, it's now much more difficult to achieve.

Chips and Technologies has approached the PS/2 market with four years and five million CHIPSets of directly applicable experience. Systems expertise that has enabled us to achieve gate-level compatibility with PS/2 and the Micro Channel \({ }^{T \mathrm{TM}}\) architecture.

Our PS/2-compatible offerings, \(\mathrm{CHIPS} / 250^{\text {Tw }}\) and CHIPS/280" CHIPSets , incorporate all the innovations we've developed to date.

Beyond compatibility.
The CHIPS/250 and CHIPS/280 CHIPSets operate up to twice as fast as the industry standard.

Both are available for high-performance, costeffective, desktop configurations, which provide Model 80 functionality in a Model 50 chassis.

We even offer a special CHIPS/280 version which supports both cache and page interleaved memory systems. Giving OEMs the option to use low-cost, slower DRAMs with no reduction in system performance. What's more, the matched memory technique used in the IBM PS/2 Model 80 is incorporated into our CHIPS/250 CHIPSetdoubling the rate of data transfer over the Micro Channel bus for Model 50-compatible designs.

Our CHIPS/450 \({ }^{\text {min }}\) family provides VGA \({ }^{m}\) compatibility down to the gate level as well. Including implementation of hidden modes-the only real assurance of compatibility. And close coupling with our systems logic CHIPSets enables our VGA chips to run twice as fast as the industry standard. Without the need for special drivers.

Along with other enhancements in our VGA chips, we've also implemented a bus architecture that's configurable for either AT or PS/2 environments. So one chip can serve two applications, with no compromise in performance.

\section*{System design support at every level.}

We furnish extensive technical documenta tion and applications support with all our CHIPSets

And to help OEMs make the most of our products, we offer the most comprehensive Design Services Organization in the business. A partnership that can cover the ground from technical consultation to the design of an entire turnkey microcomputer system.

It's the kind of support that can make the difference in achieving successful entry in the advanced AT-class and PS/2 arenas. Markets we now serve with a full range of CHIPSet solutions.

\section*{approach and the PS/2 has the rules.}
"When the clone makers do line up, they will be talking performance...so that is where Chips and Technologies is concentrating. The goal is not mly 100\% compatibility, but also significantly faster data rates, much greater mass-storage capacities, superior graphics and a line-up of originalfeatures."

Stan Runyon
Electronics

CHIP P82C223

\section*{CHIPS P82C226}



\section*{Where we go from here.}

By providing OEMs with the most timely and cost-effective microcomputer solutions available, our CHIPSets have fundamentally changed the way systems are designed-creating vast new markets in the process.

And this is only the beginning. We continue to enhance standard architectures and develop other performance innovations, while maintaining compatibility with hardware and software standards.

That's the difference that untraditional thinking makes. In CHIPSets. And in the success of your products.

To experience the CHIPSet differencein more powerful, cost-effective systems-call 800-323-4477. Or write Chips and Technologies, Inc., 3050 Zanker Road, Department MC-1, San Jose, California 95134.


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CASE '88 (2nd International Workshop on Computer-Aided Software Engineering), Cambridge, MA. Pamela Meyer, Index Technology Corp, 1 Main St, Cambridge, MA 02142. (617) 494-8200, ext 1988. July 12 to 15.

Principles of RF and Microwave Circuit Design (short course), College Park, MD. Besser Associates, 3975 E Bayshore Rd, Palo Alto, CA 94303. (415) 969-3400. July 25 to 27.

Siggraph, Atlanta, GA. Barbara Voss, Robert P Kenworthy Inc, 866 United Nations Plaza, Suite 424, New York, NY 10017. (212) 752 0911. August 1 to 5.

Midcon, Dallas, TX. Electronic Conventions Management, 8110 Airport Blvd, Los Angeles, CA 90045. (800) 421-6816; in CA, (213) 772-2965. August 30 to September 1 .

Surface Mount '88, Marlborough, MA. MG Expositions Group, 1050 Commonwealth Ave, Boston, MA 02215. (800) 223-7126; in MA, (617) 232-3976. August 30 to September 1 .

Modern Electronic Packaging (seminar), Santa Clara, CA. Technology Seminars, Box 487, Lutherville, MD 21093. (301) 2694102. September 7 to 9 .

International Test Conference, Washington, DC. Doris Thomas, ITC, Box 264, Mount Freedom, NJ 07970. (201) 267-7120. September 12 to 14.

Worst-Case Circuit Analysis (seminar), Boston, MA. Design and Evaluation, 1000 White Horse Rd, Suite 304, Voorhees, NJ 08043. (609) 7700800. September 12 to 14.

C Programming Workshop (short course), Seattle, WA. SSC, Box 55549, Seattle, WA 98155. (206) 5273385. September 12 to 15.

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\section*{CIRCLE NO 37}

\section*{DID YOU KNOW?} Half of all EDN's
articles are staff-written.

12th International Fiber Optic Communications and Local Area Networks Exposition, Atlanta, GA. Information Gatekeepers, 214 Harvard Ave, Boston, MA 02134. (800) \(323-1088\); in MA, (617) 2323111. September 12 to 16.

Connector and Interconnection Technology Symposium, Dallas, TX. Electronic Connector Study Group, 104 Wilmot Rd, Suite 201, Deerfield, IL 60015. (312) 940-8800. October 3 to 5.

Autotestcon, Minneapolis, MN. Steve Palmer, Unisys, 3333 Pilot Knob Rd, Eagan, MN 55121. (612) 456-2349. October 4 to 6.

Buscon/88 East, New York, NY. Conference Management Corp, 200 Connecticut Ave, Norwalk, CT 06856. (203) 852-0500. October 4 to 6 .

Electronic Imaging Conference East, Boston, MA. MG Expositions Group, 1050 Commonwealth Ave, Boston, MA 02215. (800) 223-7126; in MA, (617) 232-3976. October 4 to 6 .

Power Electronics East, New York, NY. Conference Management Corp, 200 Connecticut Ave, Norwalk, CT 06856. (203) 852-0500. October 4 to 6 .

Frontiers '88: The 2nd Symposium on the Frontiers of Massively Parallel Computers, Fairfax, VA. Frontiers Symposium, Box 334, Greenbelt, MD 20770. October 10 to 12 .

International Electronic Manufacturing Technology (IEMT) Symposium, Lake Buena Vista, FL. Bill Moody, 2529 Eaton Rd, Wilmington, DE 19810. (302) 4784143. October 10 to 12.

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Siemens presents the world's first programmable Codec filter device in CMOS technology with digital signal processing. This breakthrough is the cornerstone for a new generation of digital switching systems: Signal Processing Codec Filter (SICOFI) PEB 2060 SICOFI can handle different international postal specifications. It is programmable for all systems in all countries. And it is easily adaptable - in its amplifier setting, in its input impedance, in its \(2 / 4\)-wire conversion, and in its equalization of frequency response.
That means greater flexibility and convenience for producers of switching systems. Manufacturing is simpler and more straightforward. One board is enough for all specs.

And, just look at these performance features of the SICOFI PEB2060:
- Capability for three-party conferencing.
- 22-pin, low-dissipation package - so 16 instead of 8 subscribers per module can be serviced. And that means dramatic space savings.
- Simple board structure when used in conjunction with Peripheral Board Controller (PBC) PEB2050 for reduced outlay in circuit design.
- Flexible signaling interface - for ease in connecting monolithic SLIC components.
- Programmable digital filters control - impedance matching - frequency response
- gain
- hybrid balancing


SICOFI does all this without any extra external components. It is one of the most innovative circuits of the big telecom IC family from Siemens - the complete IC family for the telecommunications of the future and for simplicity of design in digital communication systems that leads the way.

Europe: Siemens AG, Infoservice 12/Z072,
Postfach 2348, D-8510 Fürth
USA: Siemens Semiconductor Group,
2191 Laurelwood Road, Santa Clara,
CA 95054, mentioning "SICOFI PEB 2060«.
The Key to digital communication systems

Siemens is ready to lead the way for you - with experience, know-how, and pace-setting technology. And we'll back you with comprehensive support for the products you buy. You'll get fully detailed technical documentation, training courses, prototype devices, and test boards, plus dependable second-sources.



\section*{If you built your own logic analyzer... ...it would be}

Why make just an analyzer when you could build a complete, flexible logic analysis system? A modular system with a 400 megasample/s digitizing oscilloscope, an interactive \(50 \mathrm{Mbit} / \mathrm{s}\) pattern generator, up to 400 state/timing channels available, and a variety of features that make it versatile and easy to use. Plus a configurable price that you could easily justify to the boss.

Fortunately, you don't have to build it yourself to get all this. Just choose the modular HP 16500A Logic Analysis System. HP has already done it for you.

\section*{HP Logic Analyzers: the extra measure.}

Like all HP logic analysis tools, the HP 16500A delivers an extra measure of value. There are four different measurement modules to choose from. To maximize your investment, configure the HP 16500A, using its five mainframe slots, with the module combination that best matches your budget and application. Just imagine the possibilities for mixing and matching:
- \(80-\) channel, 25 MHz state \(/ 100 \mathrm{MHz}\) transitional timing module (400 channels maximum)
- 2-channel, 400 megasample/s digitizing scope module (8 channels maximum)
- \(50 \mathrm{Mbit} / \mathrm{s}\) pattern generator module (204 channels maximum)
- 16 -channel, 1 GHz timing analyzer module ( 80 channels maximum)
If your applications are in hardware, powerful inter-module cross-domain triggering allows high speed waveforms from both the scope and analyzer modules to be time correlated, giving you unmatched analysis capability. And you



\section*{more than a logic analyzer.}
ment results for simulation by using HP's bi-directional CAE link.

If you're a software developer the HP 16500A offers complete support for popular 8-, 16- and 32-bit microprocessors. You can also develop your own support package for use with your custom processors or ASICs.

But we didn't stop there. We designed a full set of features into the HP 16500A that make it easy to use. A touch screen or mouse controls functions through pop-up menus. Use your choice of a front panel knob or screen keyboard to change values. Plus, there are two built-in disc drives for data set-up and storage, HP Auto-Scale, EDN July 7, 1988
and color hardcopy output at the push of a button. And the proverbial much more. Call for your free video! 1-800-752-0900, Ext. H215.

See for yourself how the modular HP 16500A Logic Analysis System is the logic analysis tool you'd design yourself. Like every logic analyzer in the HP family, it delivers the extra measure of value you're looking for!

Performance: 25 ns .
Low power: \(90 \mathrm{~mA}(55 \mathrm{~mA}\) max " \(L\) ").
Variable product terms Variable product terms.
User programmable Macrocell.
Up User programmable Macroceli.
Up to 22 input terms, 10 outputs.

\title{
Why our high performance 22V10 PLD is the market leader:
}

First, all the great architectural features of the standard 22V10.

Including Macrocell I/O, so one part can be programmed to replace up to 10 different PLD devices, plus many additional logic configurations. Inverted or non-inverted, registered or combinatorial operating modes. As many registers as you want -1 to 10 . Synchronous Preset and Asynchronous Reset features. Configurable with up to 22 inputs and 10 outputs. A Variable Product Term architecture lets you easily tailor the 22 V 10 for high performance in a wide variety of applications, without burdening the product term structure.
And you get the benefit of easy programming, using industry standard languages like \(\mathrm{CUPL}^{\oplus}\) or \(\mathrm{ABEL}^{\mathrm{TM}}\), standard programmers, or our handy QuickPro \({ }^{\text {TM }}\) programming accessory for PC or PC-compatible.

In other words, you get superb flexibility and ease of use. The 22 VI 0 gives you the convenience of PLD design for logic functions in the 500 to 800 gate array complexity. But without the design complexity of gate arrays.

\section*{Now add Cypress CMOS leadership and you have the market-leading \(22 V 10\).}

Blazing performance, with speeds to 25 ns combinatorial \(/ 33.3 \mathrm{MHz}\) registered.

Quarter-Power for cool performance.
Optional windowed versions for the convenience of reprogrammability.

Skinny DIP or surface mount packaging. Greater than 2000 V ESD tolerance on every pin, and the ability to tolerate \(\pm 10 \%\) power supply fluctuations.

No wonder this is such a best seller.
Get the databook that has the information you need on this great part, and you'll have the information you need on ALL our high performance parts.

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\section*{EDITORIAL}

\section*{The paperless-office myth}


Back at the birth of business computers, various industry prophets predicted paperless offices. In those days, when timesharing was a computing technique and not a vacation style, people envisioned sharing great quantities of information that was stored in a large central computer. Thus, they thought, offices would no longer need vast quantities of paper or the facilities to store reports and memos.
Such talk spread panic among paper and office-equipment manufacturers. After all, they couldn't survive if offices no longer needed paper and file cabinets. It turns out that the words "paperless office" have become an oxymoron, much like "plastic silverware" and "solid tubing." In fact, developments in the last few years have increased rather than decreased the amount of printed material that offices create.
Hardly a day goes by, for example, that I don't receive a computer-products catalog in the mail. Over a week or two, those catalogs add considerable bulk to the household paper trash. Not only that, but many flyers and advertisements entice me to buy special paper, labels, envelopes, and forms-more paper. And the advent of the personal computer also spawned many, many computer publications, magazine supplements, and newspaper sectionseven more paper.
These days, people talk about diskless workstations, but it's hard to imagine a computer without a printer. In fact, many computer stores sell a desktop computer and an inexpensive printer as a package deal. On most computers, printer-output ports are standard, and after all, a word-processing program without a printer has little value.

The laser printer has exacerbated the situation. I receive countless newsletters that bear the mark of a laser printer. With a laser printer and a desktop-publishing program, anyone can become a publisher, editor, and promoter. The equipment can exalt even the most modest communications. Instead of just telling people about a meeting, for example, you can handcraft a beautiful, printed meeting notice.

Other computer software also beckons me to use more paper. I see ads for form-designing software that lets you create a form for every possible use. Once hooked on such software, you'll find that you need many more forms than you use today-and it becomes easy to create them. Now there's no excuse to use a small stick-on note when you can design and use a 1-page form instead. We editors are guilty, too: It's easy to make profligate use of printouts for revision upon revision of an article.

Now that I have all this wisdom about the paper-full office, I'm sorry I didn't invest in paper futures a few years ago. But I'm playing it smart by looking for investment bargains among trashhauling and paper-recycling companies. It pays to be ready.



\section*{Plug new sales appeal into your system with DuPont Latch-N-Lok"Shielded Assemblies.}

Improve your system's sales appeal with Du Pont Latch-N-Lok modular interconnections.

Choose from the industry's widest variety of compact plugs and receptacles. Straight, right-angle or combination designs. Latch on top, bottom or side. Panels, chassis or board mounted receptacles. And cords-coiled or straight-in any length, with any number of conductors, and foil, serve or braid shielding. What's more, Latch-N-Lok can use a variety of wire gauges ( 22 to 30 AWG ) within an assembly.

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And Latch-N-Lok is the only quick disconnect pin-and-socket system. With a contact design proven over millions of applications.

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With our Universal Installation Kit you can easily fit our \(3.5^{\prime \prime}, 1\) or 2 megabyte floppy in the slot where a \(5.25^{\prime \prime}\) drive used to go. But if a \(5.25^{\prime \prime}\) is what you really want, then you've got plenty of options there too.


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\section*{QUALITY PRODUCTS WITH QUALITY SUPPORT. \\ Floppy, optical or Winchester, you'll} find the quality drives you need at Toshiba. Along with J.I.T. delivery, complete technical support and full service. Because, it's doing it all that's made us the best single source drive supplier around. To find out how good a source we can be for you, call us at 1-800-456-DISK.

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\section*{V33: The world's fastest 16-bi}

The V33 runs fast. 16MHz fast. And it offers a 16M-byte address space in the extended mode.
High-speed processing makes the V33 ideal for your instrumentation or control equipment design. And the ample address space gives you an edge in advanced PC applications.

Other top-class features of NEC's new V33 include separate address and data buses for the 2-clock bus cycle, and dynamic bus sizing for \(8 / 16\)-bit data bus systems. The V33 is software-compatible with the V20 and V30.

To find out how the V33 can speed your system to success, call NEC today.

\section*{microprocessor.}


\title{
VME + FCC =Electronic Solutions
}


\section*{The only VME System Enclosures with EMI/RFI Compliance}

With Electronic Solutions VME enclosures you don't have to take any static about FCC compliance. Because your VME system can meet or exceed FCC Class A Part 15 EMI/RFI standards. No other enclosure manufacturer can make that statement.
Here's why: Only Electronic Solutions puts a new face on VME, an outer front panel that-with other design fea-tures-keeps your EMI/RFI signals from straying. What's more, it hides those I/O connectors and dangling cables so your system looks a lot cleaner and more attractive.
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cage, backplane, power supply, room for peripherals and more. You can get slim enclosures with 3 slots all the way up to multi-system enclosures with 40-count 'em-40 slots. And you buy them ready for your system at a tiny fraction of what it would cost to develop your own packaging.

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 Electronic Solutions
UNIT OF ZERO CORPORATION

\section*{Microcontroller chip features analog and digital I/O lines}

Aimed specifically at low-cost applications in consumer-electronic equipment and handheld devices, the Z86C08 microcontroller chip provides 11 digital I/O lines and 2 k bytes of internal ROM. Three additional signal inputs accept either digital or analog signals, depending on how your software configures the chip. The \(\mu \mathrm{C}\) chip furnishes 124 general-purpose 8-bit registers in addition to 15 control and status registers that control the I/O ports, timers, register pointer, and interrupts. The chip also provides a stack that saves addresses and data in the general-purpose registers.

The CMOS chip operates with either a \(12-\mathrm{MHz}-\mathrm{max}\) crystal or an external oscillator. By turning off the clock signals, you'll save power, a requirement for low-power applications that depend on batteries. The software instructions let you select one of two power-saving standby modes: stop or halt. The halt mode stops the chip's internal clock, but not the crystal-oscillator circuit. The stop mode, however, stops both the chip's internal clock and its oscillator circuit. You can release the \(\mu \mathrm{C}\) from either the stop or halt mode by generating an external interrupt signal.

The chip furnishes three inputs that you can use to measure analog voltages. One input connects an external reference voltage to two comparators that monitor two unknown voltages-one at each of the two remaining pins. When an unknown voltage exceeds the reference level, the corresponding comparator interrupts the CPU. The CPU provides as many as six interrupts; four from I/O lines, which include the comparators, and two from internal counter/timer registers.

If you already program the manufacturer's Z8 \(\mu\) P-family chip, you'll be able to transfer software to the Z86C08 directly, because the new chip maintains instruction compatibility. The instruction set includes a watchdog timer code that refreshes a 15 -msec timer within the chip. If the timer is not refreshed within the 15 -msec period, it resets the CPU. The watchdog timer is a maskable option, so you must specify whether or not you want it when you order the chip and specify your masked ROM pattern. The \(\mu \mathrm{C}\) chip is available now and costs \(\$ 1.62\) in 25,000 quantities.-Jon Titus
Zilog Inc, 210 Hacienda Ave, Campbell, CA 95006. Phone (408) 370-8000. TWX 910-338-7621.

Circle No 745

\section*{Carborundum \({ }^{\circledR}\) noninductive ceramic power resistors solve tough problems.}

\author{
We make three types of noninductive ceramic resistors that can solve tough resistance problems, save money and space.
}


Regardless of the pulse shape, we have the resistor. Our Type SP handles large amounts of power from 60 cycles through VHF. Type AS can absorb huge amounts of energy in millisecond pulses. Type A solves high resistance problems in high voltage situations.

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\section*{STANDARD OIL}

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\title{
Sequential-sampling digital scope has \(2-\mathrm{GHz}\) bandwidth and \(1-\mathrm{mV} /\) div sensitivity
}

The PM 3340 2-channel digital oscilloscope uses sequential equivalenttime sampling to achieve a \(2-\mathrm{GHz}\) repetitive-signal bandwidth. Many engineers think that this technique combines extraordinary bandwidth with high resolution, but that it takes longer to reconstruct waveforms than other data-capture techniques such as random sampling. "Not so," says Hans Toorens, product marketing manager at John Fluke Manufacturing Co, exclusive US distributor of Philips' test and measurement products. "For certain simple waveforms, a randomsampling scope might give you an idea of what the picture will look like a little sooner but, in general, the sequential-sampling instrument will present a complete display more rapidly." He further points out that the economies inherent in sequential sampling permit his company to offer products with a superior price/performance ratio.
In the PM 3340, the sampling gates are right at the input. After the gates, the scope's circuits process medium-frequency information. Most of the instrument does not need-or use-costly, high-frequency design techniques. According to Toorens, competitive instruments contain a substantial amount of wideband electronics. Whereas the PM 3340 delivers a combination of \(2-\mathrm{GHz}\) signal bandwidth, \(2-\mathrm{GHz}\) trigger bandwidth, \(20-\mathrm{psec} / \mathrm{div}\) sweep speed, and \(1-\mathrm{mV} / \mathrm{div}\) sensitivity, several instruments in the same class only offer half the bandwidth and \(1 / 10\) th the sensitivity. Furthermore, the PM 3340 digitizes with 10-bit resolution-four times what some of the competitive units provide.


The front panel of the PM 3340 resembles that of an analog scope, but the row of softkeys to the right of the screen and the keypad reveal its digital heritage.

The engineers who designed the scope did not overlook the user interface in their effort to provide maximum performance. Insofar as possible, the scope's controls behave like their counterparts on analog scopes, which all potential users are familiar with. But there are also many convenience features that users of state-of-the-art digital scopes have come to expect. For example, menus appear on screen to the right of the waveforms; you make your selections using a row of softkeys to the right of the screen. Pressing an "autoset" button enables you to quickly obtain a trace. Nonvolatile memory can store 250 control settings. For automatedtest applications, the unit includes both RS-232C and IEEE-488 interfaces.
The scope also offers many signalprocessing capabilities. Signal averaging allows enhancement of \(\mathrm{S} / \mathrm{N}\)
ratios; four memory registers permit waveform comparisons (each register stores 4 k words-eight times the number of samples that appear on a full screen); an "eye" display mode facilitates jitter measurements; and firmware-based calculation functions let you measure amplitude, time, frequency, and phase. The instrument can determine FFTs and amplitude histograms. Moreover, it can add, subtract, multiply, divide, integrate, and differentiate.

Prices for the PM 3340 start at \(\$ 16,000\). Delivery is six weeks ARO.
-Dan Strassberg
John Fluke Mfg Co Inc, Box C9090, Everett, WA 98206. Phone (800) 443-5853.

Circle No 785
Philips Test and Measurement, Bldg HKF , 5600 MD Eindoven, The Netherlands. Phone local office.

Circle No 786


The Z84C90: Two serial, three parallel ports and a counter/timer on one chip. Just think what you can do with it.

Zilog's Z80 SPCT, Killer I/O," gives you a true "System on Silicon."'" With all the advantages of CMOS technology, Superintegration," and proven Z80 performance. Think of it as the door to a whole lot of new opportunities.

The Z80 Family:
Still growing strong.
The \(Z 80\) remains the most commonly used 8 -bit microprocessor in the industry. No wonder. As the family bas continued to develop, so have the adv antages: the familiarity of working with devices you know and trust, the tremendous value of being able to use softuare you've already developed, and, of course, there's the impressive Z80 performance.

As the Z80 Family bas evolved through NMOS. CMOS, bigh-performance and high-integration, our commitment to \(Z 80\) has never wavered. Neu products bave contimued to be developed. Besides the 16-bit Z280 and the new Z84C90-the Killer 1/O-there are a feu more you really ought to look at:
Z84C80/81 Z80-based systems GLU logic that can be used in every \(Z 80\) application

Z80180 the Z180 8-bit MPU combines a \(Z 80\) CPU, MMU, 2 UARTE, DMA, a C/T and more, uith no extra logic needed for \(Z 80\) peripherals combines a Z80 CPU with an on-board oscillator

Lots of I/O.
You're simply not going to get more serial/parallel I/0 anywhere. We've put together the most popular combination of discrete devices . . . two independent synch/asynch serial channels, two independent parallel ports, an 8-bit programmable port and four counter/ timers. And, since they're all fully compatible with PIO, SIO and CTC devices, you have the advantage of "commonality."
Lots of performance.
Superintegration and CMOS technology mean the Z84C90 provides plenty of performance and flexibility. 8 MHz speed for instance. Plus you've got four independent counter/timers and on-chip oscillator to work with. And the peripherals can be used in any combination you need. Lots of benefits.

You're designing with a highly integrated chip. And you're working with the familiar software and proven performance of the Z80 Family. That's enough to make the Killer I/0 the best choice. But think about the lower cost you get from less real estate, lower manufacturing cost and reduced inventory. Think about improved time to market. Or the higher performance and reliability that come with super integration. And it's all off the shelf and backed by Zilog's proven quality.

So whether you're upgrading existing designs or looking for solutions in new applications like cellular phones, personal computers, industrial control, or data communications, you owe it to yourself to contact your local Zilog sales office or your authorized distributor today. Zilog, Inc., 210 Hacienda Ave., Campbell, CA 95008, (408) 370-8000.

Right product. Right price. Right away. Zilog
ZILOG SALES OFFICES: CA (408) 370-8120, (714) 838-7800, (818) 707-2160, CO (303) 494-2905, FL (813) 585-2533, GA (404) 923-8500, IL (312) 885-8080, MA (617) 273-4222, MN (612) 831-7611, NJ (201) 288-3737, OH (216) 447-1480, PA (215) 653-0230, TX (214) 231-9090, CANADA Toronto (416) 673-0634, ENGLAND Maidenhead (44) (628) 39200, WEST GERMANY Munich (49) (89) 612-6046, JAPAN Tokyo (81) (3) 587-0528, HONG KONG Kowloon (852) (3) 723-8979. R.0.C.: Taiwan (886) (2) 731-2420, U.S. AND CANADA DISTRIBUTORS: Anthem Electric, Bell Indus., Hall-Mark Elec., JAN Devices, Inc., Lionex Corp., Schweber Elec., Western Microtech., CANADA Future Elec., SEMAD.

Making mistakes is not the problem.
It's where you make them that matters. The most inexpensive and efficient place is on a computer, where you can change a design in minutes. Building a prototype, on the other hand, can take days.

You can streamline the process even more by integrating the design steps with the rest of the manufacturing operation. So everyone, from engineers and draftsmen, to machinists and floor managers, works together more efficiently.

\section*{Stop manufacturing your mistakes.}


This approach to manufacturing is called Computer Integrated Manufacturing (CIM). And no one is better able to help you start CIM than IBM. In fact, we have the hardware, software, support and experience it takes to tie your entire business together.

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Of all the new products covered in EDN's April 14, 1988, issue, the ones reprinted here generated the most reader requests for additional information. If you missed them the first time, find out what makes them special: Just circle the appropriate numbers on the Information Retrieval Service card, refer to the indicated pages in our April 14, 1988, issue, or use EDN's Express Request service.


\section*{A DSP IN-CIRCUIT EMULATOR}

The 320C25 ICE Pak is a low-cost in-circuit emulator for Texas Instruments' TMS320C25 DSP chip (pg 98). You can use it with any host computer or terminal by plugging it into the host's RS-232C port.

\section*{Memocom}

Circle No 605

\section*{VOLTAGE TRIPLER}

The SL6670 provides voltage tripling without the use of external inductors or transformers by employing charge-pump techniques (pg 255). It is programmable via an external resistor.
Plessey Semiconductors Ltd Circle No 601
Plessey Semiconductors
Circle No 602

- CAMERA ASSEMBLY

You only need to add a chassis and lens to this solidstate image sensor assembly to produce a black-andwhite video camera suitable for surveillance or ma-chine-vision systems (pg 270).
Philips
Circle No 603
Amperex Sales Corp
Circle No 604

\section*{SCIENCE TOOLS}

The Science and Engineering Tools package consists of a set of C functions for general statistics, multiple regression, curve fitting, integration, FFTs, differential and simultaneous equations, matrix math, complex math, and special functions (pg 305).

\section*{Quinn-Curtis}

Circle No 607


\section*{- TAPE SYSTEM}

The Jumbo 40M-byte tape backup system for IBM PC, PC/XT, PC/AT, PS/2, and compatible computers can store 10 M bytes of data for an IBM PC/AT within five to six minutes, according to the vendor (pg 280).
Colorado Memory Systems Inc Circle No 606

Conservative thermal design massive heatsinks and substantial component derating keep temperatures low and reliability high

Shindengen designed and manufactured power semiconductors and control
hybrids

Full-load operation to \(50^{\circ} \mathrm{C}\) without fan cooling

Output "On" LED indicator for simplified system fault-finding

Input line transient and electrostatic discharge protection

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\title{
WITH ALL THIS GOING FOR YOU,
}


Visual orientation guides speed installation.


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CHIPAK \({ }^{\mathrm{TM}}\) - it's the finest chip carrier socket on the market today. And the only one that combines all these important features in a single design.

In fact, everything about our CHIPAK sockets has been engineered for maximum reliability and installation ease.

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In short, CHIPAK combines all the
features you'll ever want or need in a chip carrier socket. Low-profile design to save space. Convenient standoffs to aid cleaning and heat dissipation. Easy-access probe holes and easy-to-spot orientation guides to speed installation and testing. And finally, an easy-to-use chip extraction tool that, unlike others, accommodates every socket size-quickly and safely.
With all this going for you, why settle for anything less? Available in sizes from
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Foolproof. Easy to use. Convenient. One tool does it all. Accommodates full range of chip sizes.



WITH VLSIS ASIC TOOLS, YOUCAN

VLSI Technology's software is the fastest, surest way to create successful ASIC chips.

Our tools accelerate each and every step of the design process.

And our years of experience insure the
success of your finished ASIC chip. EVERYTHING'S FASTER WHEN YOU HAVE AN ASSISTANT.
Our Design Assistant'" tool partitions your chip and estimates chip size, power consump-


\section*{BLOW RIGHT BY THECOMPETITION.}
tion, and packaging possibilities. In short, it tells you the best silicon solution to your problem.

Simply enter your design in block diagrams and global interconnect forms. Design Assistant shows you ways to implement it.

Gate array, standard cell or cell-based.
You can run dozens of "what if" configurations in a few hours. In no time, you'll have the crucial information you used to wait forever for vendors to supply.

\section*{STREAMLINED DESIGN, COMPLIMENTS OF OUR COMPILERS.}

Our logic and memory compilers create multipliers, PLAs, or super fast SRAM memory blocks as fast as 8 ns.

And wéve got a couple of "expert" compilers for cell-based or gate array designs that think exactly the way you do.

When you design a datapath, you think of it as a linear schematic, right?

Well, our Datapath Compiler just happens to use schematics as input.Complex multi-bit datapaths practically pop right out of it.

When you design state machine and other logic blocks, you think in terms of equations. Our State Machine Compiler does, too.

And it even optimizes your equations.

\section*{GET IT TOGETHER. FAST.}

Ready to place and route your cell-based design? Just call up our Chip Compiler.

It provides floor-planning and auto-routing. It works with our compilers and standard cell libraries to produce an optimized layout.

It even pours standard cells into the gaps
between blocks to make sure you use the minimum chip area required.

How fast does it work? Glad you asked.
A company came to us with a layout that took them three months. We did it in two days.

\section*{TO BE FAST, \\ YOU HAVE TO BE FLEXIBLE.}

Only our Portable Library lets you choose gate array or cell-based, \(1.5 \mu\) or \(2 \mu\), after you've completed your logic design.

It can allow you to do that because your library always remains stable. No matter what process you use.

Process obsolescence is now obsolete.

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HEWLETT PACKARD

\title{
Going any less than five in our measurements th of the rest would seem
}


\section*{times faster \\ an the best}
us second-rate.

\title{
Current-feedback
}

\section*{op amps ease high-speed circuit design}

> Because they're free from gain-bandwidth limitations, current-feedback op amps excel in high-frequency and fastsettling circuitry. The latest models also achieve a fair amount of dc accuracy without sacrificing speed-an important consideration when you're designing high-frequency circuits such as flash A/D converters.

Peter Harold, European Editor

For high-frequency applications-for instance, high-speed A/D and D/A converters, video drivers, pulse amplifiers, and radar and IF pro-cessors-current-feedback op amps offer particular advantages, such as a very wide bandwidth that's independent of closed-loop gain. They also provide a fair degree of precision, sparing you the painful choice between speed and precision that conventional voltagefeedback op amps usually require you to make. Cur-rent-feedback op amps (or transimpedance amps) settle quickly, provide a linear phase response, and are generally free from stability problems. And although they might at first glance seem more expensive, they could in fact save you money: Unlike a conventional op amp, a current-feedback type doesn't require any individual tweaking to meet performance requirements in highspeed or high-frequency applications.

Unlike a conventional op amp, a current-feedback op amp can achieve high speed without a huge sacrifice in dc performance. Models with bandwidths of over 100 MHz are available with input offset voltages as low as 0.3 mV typ and input-offset-voltage temperature coefficients of \(5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\). In addition, because current-feedback op amps aren't subject to the gain-bandwidthproduct restrictions of conventional op amps, you can alter the closed-loop gain of the current-feedback de-

\(100-\mathrm{MHz}\), low-offset-voltage op amp (Analog Devices Inc)
vices without affecting their frequency response.
Although many analog-signal-processing applications, such as video- and IF-signal amplifiers, require high-bandwidth op amps, it's today's trend towards processing analog signals in the digital domain that places the greatest demands on op-amp performance. Almost by definition, digital-signal-processing (DSP) systems start off by capturing a real-world analog input, and more often than not, they end up with an analog output in the form of drive waveforms for a CRT display. Therefore, you'll almost certainly find an A/D converter at the front end of a DSP system, and a D/A converter at the back end. To accurately model the system's analog input in the digital domain, this A/D converter must satisfy the requirement of Nyquist's sampling theorem. This theorem requires that the A/D converter sample the input signal at twice the frequency of the signal's highest frequency component that has an amplitude greater than the ADC's least significant bit.
To digitize the output of a high-resolution video camera, for example, you'll need to digitize the camera's analog output signal at around 15 M to 20 M samples \(/ \mathrm{sec}\). To buffer and scale this signal to suit the input of an 8 -bit A/D converter, you'll need an op amp that can settle to within \(0.5 \mathrm{LSB}(0.2 \%)\) in well under 50 nsec.

This settling time is not outside the capabilities of a conventional high-speed op amp (especially if you can operate it at a gain close to 1 , where it has optimum settling performance), but it would require some fairly clever compensation circuitry. That settling time is, however, well within the capabilities of a low-cost, general-purpose current-feedback op amp. By using a suitable current-feedback op amp, you can also be sure that you won't need to tweak the individual \(\mathrm{A} / \mathrm{D}\) converters during production to ensure their consistent performance.
Current-feedback op amps outperform conventional types in applications that require further increases in speed or resolution. A 12-bit flash A/D converter, for example, often uses a 2 -pass subranging technique to achieve its 12 -bit resolution. On the first pass it typically digitizes the input signal to only 7 -bit resolution. During the second pass, it feeds back this 7 -bit result to the input via a 7 -bit D/A converter whose output is accurate to 12 bits. At the input, the D/A converter's output voltage is subtracted from the input signal so that the \(\mathrm{A} / \mathrm{D}\) converter can digitize the residue to obtain five more bits of resolution. In such an A/D converter, the residue must be scaled to the full input range of the \(A / D\) converter: In this case, it must operate at a gain of 32 . To aggravate matters, the op

Settling to \(0.2 \%\) accuracy in under 50 nsec is within the capabilities of even a low-cost, general-purpose current-feedback op amp.
amp must be able to drive the highly capacitive input of the flash converter's comparator chain.

Because conventional op amps are subject to a gainbandwidth limitation, operating a conventional op amp at a gain of 32 will reduce its closed-loop bandwidth by a factor of 32 (as compared to its closed-loop bandwidth at unity gain), seriously impairing its settling time. Not so for a current-feedback op amp: They're not restricted by the gain-bandwidth product (see box, "Cur-rent-feedback defies gain-bandwidth limitations"), and their closed-loop bandwidth and settling times, there-
fore, remain relatively unchanged over the gain range of 1 to 20 . Even at higher gains, the bandwidth falls by only about one third when you increase the gain to around 40 . As a result, a current-feedback op amp is the ideal choice for the residue amplifier in subranging flash A/D converters.

Comlinear Corp had this application in mind when it designed the CLC501, its latest current-feedback op amp. The relevant specifications for this device pertain to a gain of 32 , where it has a typical bandwidth of 80 MHz for a 5 V p-p output signal, and typically settles to

\section*{Current feedback defies gain-bandwidth limitations}

A current-feedback op amp owes its ability to maintain its bandwidth over a wide range of closed-loop gain settings to its internal architecture. Unlike a conventional op amp, which relies on feedback to drive its inverting and noninverting inputs to the same potential, the cur-rent-feedback op amp's noninverting and inverting inputs are linked by a unity-gain buffer. As a result, the op amp's inverting input automatically assumes the same potential as the noninverting input and has a very low input impedance, to or from which current can flow.
The output stage of the cur-rent-feedback op amp is a transimpedance amplifier with high gain \(\left(\mathrm{A}_{(s)}\right)\) that senses the current flowing in the op amp's inverting input and generates a corresponding output voltage. As you can see from the accompanying figure, the current that flows at the op amp's inverting input ( \(\mathrm{I}_{\mathrm{INV}}\) ) is given by the equation
\[
\begin{align*}
\mathrm{I}_{\mathrm{INV}} & =\mathrm{I}_{1}-\mathrm{I}_{2} \\
& =\frac{\mathrm{V}_{2}}{\mathrm{R}_{1}}-\frac{\mathrm{V}_{0}-\mathrm{V}_{2}}{\mathrm{R}_{2}} . \tag{A}
\end{align*}
\]

However, \(\mathrm{V}_{0}=\mathrm{I}_{\mathrm{INV}} \times \mathrm{A}_{(\mathrm{s})}\) and \(\mathrm{V}_{2}=\mathrm{V}_{1}\).
When you substitute these equations into \(\mathbf{E q} \mathbf{A}\), you obtain the equation
\[
\frac{\mathrm{V}_{0}}{\mathrm{~A}_{(\mathrm{s})}}=\frac{\mathrm{V}_{1}}{\mathrm{R}_{1}}-\frac{\mathrm{V}_{0}-\mathrm{V}_{1}}{\mathrm{R}_{2}}
\]
which rearranges to
\[
\frac{\mathrm{V}_{0}}{\mathrm{~V}_{1}}=\frac{1+\frac{\mathrm{R}_{2}}{\mathrm{R}_{1}}}{1+\frac{\mathrm{R}_{2}}{\mathrm{~A}_{(\mathrm{s})}}}
\]

Letting
\[
1+\frac{\mathrm{R}_{2}}{\mathrm{R}_{1}}=\mathrm{G}
\]
as you would for a conventional op amp with the same feedback configuration, gives you
\[
\begin{equation*}
\frac{\mathrm{V}_{0}}{\mathrm{~V}_{1}}=\frac{\mathrm{G}}{1+\frac{\mathrm{R}_{2}}{\mathrm{~A}_{(\mathrm{s})}}} \tag{B}
\end{equation*}
\]

When you compare Eq B with the closed-loop gain equation for a conventional op amp, you obtain
\[
\begin{equation*}
\frac{\mathrm{V}_{0}}{\mathrm{~V}_{1}}=\frac{\mathrm{G}}{1+\frac{\mathrm{G}}{\mathrm{~A}_{(\mathrm{s})}}} \tag{C}
\end{equation*}
\]
which shows that the ideal closed-loop voltage gain for the two types of op amp (that is, when the op amp has an infinite \(\left.\mathrm{A}_{(\mathrm{s})}\right)\) is the same and is equal to G.

However, in reality \(\mathrm{A}_{(\mathrm{s})}\) is a polynomial function of frequency with zeros and poles. In other words,
\[
\mathrm{A}_{(\mathrm{s})}=\frac{\mathrm{N}_{(\mathrm{s})}}{\mathrm{D}_{(\mathrm{s})}} .
\]

Substituting \(\mathrm{N}_{(8)} / \mathrm{D}_{(\mathrm{s})}\) for \(\mathrm{A}_{(\mathrm{s})}\) in Eqs B and C yields a closed-loop gain for the current-feedback op amp:
\[
\frac{V_{0}}{V_{1}}=G \frac{N_{(s)}}{N_{(s)}+R_{2} D_{(s)}}
\]

For a conventional op-amp, the substitution yields a gain of
\[
\frac{\mathrm{V}_{0}}{\mathrm{~V}_{1}}=\mathrm{G} \frac{\mathrm{~N}_{(\mathrm{s})}}{\mathrm{N}_{(\mathrm{ss})}+\mathrm{GD}_{(\mathrm{s})}} .
\]

By comparing these two equations, you can see that in the
\(0.05 \%\) in 12 nsec . The op amp also incorporates another important feature that makes it suitable for use as a residue amplifier-it has programmable output-voltage clamping, and it recovers from the condition in which the output is clamped in 1 nsec typ ( \(<3\) nsec max). The output clamping protects the A/D converter's input during the transients that can occur when the residue amplifier is switched in and out of the signal path, or when it's subjected to the full input signal before the feedback has been applied.

The device's clamping function should suit it to other
applications that require fast overload recovery-for example, radar equipment and optical transducers. You can set separate positive and negative clamping levels by applying de voltages to two of the device's pins. A simple resistive divider is all that's necessary to establish these voltage levels. Alternatively, because the clamping-voltage inputs have their own \(3-\mathrm{dB}\) bandwidth of 50 MHz , you can drive them dynamically at high frequencies to achieve functions such as pulseheight modulation.

Text continued on pg 90

current-feedback op amp-unlike the conventional op amp-the position of the poles (contained within \(\mathrm{D}_{(s)}\) ), and hence the bandwidth of the current-feedback op amp , is not dependent on the closed-loop gain. Provided that you keep \(R_{2}\) constant, you can change the closed-loop gain of a current-feedback op amp-with-
out affecting the op amp's pole positions or the circuit band-width-by changing \(R_{1}\).

In practice, second-order effects do result in some degree of bandwidth limitation as you increase the closed-loop gain of a current-feedback op amp, but these are slight compared to the gain-bandwidth limitations of
conventional op amps. For example, the bandwidth of a typical current-feedback op amp that has a \(3-\mathrm{dB}\) bandwidth of 220 MHz only falls to around 180 MHz when you increase the closed-loop gain from 4 to 40. For the same gain change, a conventional op amp's bandwidth would fall to around 20 MHz .

The approximate, dc closedloop gain equations for a cur-rent-feedback op amp are the same as those for a conventional voltage-feedback op amp. For most current-feedback op amps, the feedback resistor, which usually has a value of between 1 and \(2 \mathrm{k} \Omega\), is integrated in the device. The device's ac specifications all assume this internal feedback resistor is being used. Therefore, unless you wish to deliberately restrict (or try to improve) the device's bandwidth, you should use the internal feedback resistor whenever possible. If you use an external feedback resistor, you must take particular care in laying out the circuit in order to to minimize the capacitance to ground at the op amp's inverting input.

\section*{REPRESENTATIVE CURRENT-FEEDBACK OP AMPS}


NOTES:
- EXCEPT WHERE SPECIFIED OTHERWISE
1. ALL VALUES QUOTED ARE TYPICAL VALUES AT \(25^{\circ} \mathrm{C}\) UNLESS OTHERWISE STATED,

AND SHOULD THEREFORE ONLY BE CONSIDERED AS A GUIDE IN COMPARING OP AMPS
2. WHERE APPLICABLE, ALL SPECIFICATIONS ASSUME THE OP AMP'S INTERNAL

FEEDBACK RESISTOR IS BEING USED
3. PRICE QUOTED IS FOR THE LEAST EXPENSIVE VERSION
4. \(A B S\) MAX=ABSOLUTE MAXIMUM VALUE
5. THD=TOTAL HARMONIC DISTORTION
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|}
\hline \multicolumn{3}{|l|}{DC PERFORMANCE} & \multirow[b]{3}{*}{QUIESCENT SUPPLY CURRENT (mA, NO LOAD)} & \multicolumn{3}{|c|}{\multirow[b]{2}{*}{TEST CONDITIONS}} & \multirow[b]{3}{*}{PACKAGE} & \multirow[b]{3}{*}{PRICE (100)} & \multirow[b]{3}{*}{COMMENTS} \\
\hline \multirow[t]{2}{*}{OFFSET VOLTAGE DRIFT ( \(\mu \mathrm{V} /{ }^{\circ} \mathrm{C} \cdot{ }^{+}\))} & \multicolumn{2}{|l|}{OUTPUT DRIVE} & & & & & & & \\
\hline & CURRENT (mA) & VOLTAGE & & \[
\begin{gathered}
\text { CLOSED-LOOP } \\
\text { GAIN }
\end{gathered}
\] & \[
\begin{array}{|c|}
\hline \text { SUPPLY } \\
\text { VOLTAGE }
\end{array}
\] & \begin{tabular}{l}
LOAD \\
(I)
\end{tabular} & & & \\
\hline \[
\begin{gathered}
(-J,-S)_{2} \\
(-K) 1
\end{gathered}
\] & \(\pm 50 \mathrm{MIN}\) & \(\pm 10 \mathrm{~V}\) MIN & 5 & -1 & \(\pm 15 \mathrm{~V}\) & 500 & 8-PIN TO-99, 8-PIN DIP & \$4.50 TO \$18.75 & HIGH PRECISION \\
\hline \(\pm 5\) & \(\pm 50 \mathrm{MIN}\) & \(\pm 10 \mathrm{~V}\) & 21 & -10 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$49.88 TO \$79 & LOW OFFSET AND DRIFT \\
\hline \(\pm 5\) & \(\pm 50\) & \(\pm 3 \mathrm{~V}\) & \[
\begin{aligned}
& 70 \text { AT } 5 \mathrm{~V} \\
& 74 \mathrm{AT}-5 \mathrm{~V}
\end{aligned}
\] & -5 & \(\pm 5 \mathrm{~V}\) & 100 & 12-PIN TO-8 & \$84 TO \$125 & LOW OFFSET AND DRIFT \\
\hline \[
\begin{aligned}
& (-01) 15 \\
& (-01 A) 10
\end{aligned}
\] & \(\pm 400\) ABS MAX & \[
\begin{aligned}
& \pm 11 \mathrm{~V} \text { AT } 0.4 \mathrm{~A} \\
& \pm 12 \mathrm{~V} \text { AT } 0.1 \mathrm{~A}
\end{aligned}
\] & 25 & - & \(\pm 15 \mathrm{~V}\) & FULL LOAD & 8-PIN TO-3 & \[
\begin{aligned}
& \text { (WA01) \$107.20 } \\
& \text { (WA01A) } \$ 139.40
\end{aligned}
\] & HIGH POWER \\
\hline 50 & \(\pm 200\) ABS MAX & \begin{tabular}{l}
\(\pm 11 \mathrm{~V}\) MIN \\
NO LOAD
\end{tabular} & 30 & 20 & \(\pm 15 \mathrm{~V}\) & 100 & 24-PIN DIP & \$128 & HIGH OUTPUT CURRENT \\
\hline 35 & \(\pm 100\) ABS MAX & \(\pm 12 \mathrm{~V}\) NO LOAD & 29 & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$92 & GENERAL PURPOSE \\
\hline 5 & \(\pm 100\) ABS MAX & \(\pm 12 \mathrm{~V}\) NO LOAD & 29 & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$99 & LOW OFFSET AND DRIFT \\
\hline 5 & \(\pm 200\) ABS MAX & \begin{tabular}{l}
\(\pm\) IIV MIN \\
NO LOAD
\end{tabular} & 30 & 20 & \(\pm 15 \mathrm{~V}\) & 100 & 24-PIN DIP & \$135 & HIGH OUTPUT CURRENT \\
\hline 11 & \(\pm 50\) & \(\pm 12 \mathrm{~V}\) NO LOAD & 19, 5.6 AT 5 V & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$56 & LOW POWER, OVERDRIVE PROTECTED \\
\hline 11 & \(\pm 100\) & \(\pm 12 \mathrm{~V}\) NO LOAD & 29, 8.7 AT 5V & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$56 & OVERDRIVE PROTECTED \\
\hline 35 & \(\pm 50\) ABS MAX & \(\pm 10 \mathrm{~V}\) MIN NO LOAD & 30 & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$105 & GENERAL PURPOSE \\
\hline 5 & \(\pm 50\) ABS MAX & \(\pm 10 \mathrm{~V}\) MIN NO LOAD & 30 & 20 & \(\pm 15 \mathrm{~V}\) & 200 & 12-PIN TO-8 & \$112 & LOW OFFSET AND DRIFT \\
\hline 10 & \(\pm 100\) ABS MAX & \(\pm 12 \mathrm{~V}\) NO LOAD & 18 & 2 & \(\pm 15 \mathrm{~V}\) & 100 & 12-PIN TO-8 & \$56 & LOW-POWER BUFFER FOR LOW-GAIN APPLICATIONS \\
\hline 25 & \(\pm 100\) ABS MAX & \(\pm 10 \mathrm{~V}\) & 24 & 20 & \(\pm 15 \mathrm{~V}\) & 100 & 24-PIN DIP & \[
\begin{gathered}
(-\mathrm{A}) \$ 39 \\
(-\mathrm{B}) \$ 29.50
\end{gathered}
\] & \\
\hline 20 & \(\pm 70\) & \(\pm 3.5 \mathrm{~V}\) AT AV \(=5\) & 15 & 2 & \(\pm 5 \mathrm{~V}\) & 100 & 8-PIN DIP, NAKED DIE & \$15.50 & LOW-POWER OP AMP FOR USE AT GAINS BETWEEN
\[
\pm 1 \mathrm{AND} \pm 8
\] \\
\hline 20 & \(\pm 70\) & \(\pm 3.5 \mathrm{~V}\) & 15 & 20 & \(\pm 5 \mathrm{~V}\) & 100 & 8-PIN DIP, NAKED DIE & \$15.50 & LOW-POWER OP AMP FOR USE AT GAINS BETWEEN \(\pm 7\) AND \(\pm 50\) \\
\hline 10 & \(\pm 70\) & \(\pm 3.5 \mathrm{~V}\) & 18 & 32 & \(\pm 5 \mathrm{~V}\) & 100 & 8-PIN DIP & \$17.10 & PROGRAMMABLE OUTPUT CLAMPING; RECOVERS FROM OUTPUT CLAMPING IN 1 NSEC TYP \\
\hline \[
\begin{gathered}
-15 \text { TO } \\
+15 \mathrm{mV} \text { OVER } \\
\text { TEMPERATURE }
\end{gathered}
\] & \[
\begin{gathered}
\pm 32.5 \mathrm{iNTO} \\
400 \mathrm{~S}
\end{gathered}
\] & \[
\begin{gathered}
\pm 13 \mathrm{~V} \text { INTO } \\
400 \Omega
\end{gathered}
\] & 9 & 10 & \(\pm 15 \mathrm{~V}\) & 100 & 8-PIN DIP,
2O-LEAD PLCC,
20-PAD LCC & \$4.95 & LOW POWER, ENABLE INPUT \\
\hline 10 & \(\pm 100\) ABS MAX & \(\pm 12 \mathrm{~V}\) NO LOAD & 18 & 2 & \(\pm 15 \mathrm{~V}\) & 100 & 12 -PIN TO-8 & \$65 TO \$97 & LOW POWER \\
\hline \begin{tabular}{l}
(-A, -E) 25 MAX \\
(-F) 30 MIN
\end{tabular} & \(\pm 20 \mathrm{MIN}\) & \[
\begin{aligned}
& \pm 12 \mathrm{~V} \text { MIN } \\
& 1-\mathrm{k} \Omega \text { LOAD }
\end{aligned}
\] & 11 MAX FOR BOTH OP AMPS & - & \(\pm 15 \mathrm{~V}\) & - & \[
\begin{aligned}
& \text { 8-PIN DIP, } \\
& \text { TO-99 }
\end{aligned}
\] & \$7.95 & DUAL OP-AMP; OFFSET VOLTAGE MATCHING TO 1 mV : LOW POWER \\
\hline
\end{tabular}

Micrograph of low-cost, \(50-\mathrm{MHz}\) op amp (Elantec Inc)


The settling-time performance of current-feedback op amps makes them equally attractive as an output stage for high-speed D/A converters. As a bonus, if you use the op amp in an inverting gain mode, the low input impedance of the op amp's inverting input effectively shorts out the output capacitance of the D/A converter, reducing the effect that this capacitance has on the overall settling time.

As you increase the resolution of high-speed \(\mathrm{A} / \mathrm{D}\) and D/A converters, you'll also need to improve the dc performance of the front and back-end amplifiers. In particular, an op amp's input-offset voltage and drift will become critical. For example, for a 12 -bit A/D converter with a 10 V input range, you'll require the op amp's output offset (input offset \(\times\) gain) to be less than 1.25 mV ( 0.5 LSB ). A number of recently introduced op amps are capable of such performance. Analog Devices' AD-9610 and -9611 and Comlinear Corp's CLC-201A, -203 A , and -221 A , for example, all spec an input offset voltage of 0.5 mV or less and an offset voltage drift of \(5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\). At the same time, they provide small-signal ( 2 to 4 V p-p depending on the device) bandwidths of 100 MHz or more at low gain values, and they settle to \(0.1 \%\) or better in under 20 nsec . (Note that these numbers are all typical values at \(25^{\circ} \mathrm{C}\), so unless you're prepared to test and select op amps for specific parameters, make sure you check the devices' absolute maximum specifications over the required temperature range before committing to a design.)

Commensurate with these op amps' low input offset voltages are their input bias currents and input-biascurrent drifts, which are both relatively low compared to those of other current-feedback types. The inverting
input of the AD9610 specs \(\pm 5 \mu \mathrm{~A}\) and \(\pm 70 \mathrm{nA} /{ }^{\circ} \mathrm{C}\); the noninverting input specs \(\pm 15 \mu \mathrm{~A}\) and \(\pm 30 \mathrm{nA} /{ }^{\circ} \mathrm{C}\). For the AD9611, those figures are \(\pm 1 \mu \mathrm{~A}\) and \(\pm 140 \mathrm{nA} /{ }^{\circ} \mathrm{C}\) and \(\pm 1 \mu \mathrm{~A}\) and \(\pm 75 \mathrm{nA} /{ }^{\circ} \mathrm{C}\), respectively. For both inputs of the Comlinear devices, the numbers are \(5 \mu \mathrm{~A}\) and \(50 \mathrm{nA} /{ }^{\circ} \mathrm{C}\). (The numbers given are all typical values at \(25^{\circ} \mathrm{C}\).) You'll note from these figures that in a current-feedback op amp, the bias currents and biascurrent drifts are not always as well matched at the inverting and noninverting inputs as they are in a conventional op amp.

What's more, for some devices you can't always guarantee that the temperature drift at both inputs will be in the same direction. As a result, you can't always mimimize bias-current errors by driving both inputs from the same source impedance as you would with a conventional op amp. Because the value of the


High-speed, 400-mA-output op amp (Apex Microtechnology Corp)

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}

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\section*{Although they can't match conventional precision op amps, current-feedback op amps are available with low offset and drift characteristics.}


Fig 1-This composite op-amp circuit combines the low offset and drift of the OP-27 op amp with the high speed of the CLC231 current-feedback op amp to produce a circuit that settles to \(0.1 \%\) in 17 nsec, yet maintains high de accuracy. The OP-27 maintains the summing node at an accurate virtual earth potential; the CLC231 takes over at high frequencies to provide good ac performance.
input and feedback resistors used with high-speed op amps is low (usually \(2 \mathrm{k} \Omega\) or less), and because the devices are usually operated at fairly low gain, biascurrent errors tend to be reduced. However, in applications that require high accuracy, you'll almost certainly have to add offset- and bias-current compensation circuitry to these op amps. One way to achieve very low input offset and drift over a wide temperature range while maintaining a high ac bandwidth is to combine a current-feedback op amp with a high-precision conventional op amp, as shown in Fig 1.

You might also consider using Analog Devices' AD846, which specs a typical \(25^{\circ} \mathrm{C}\) input offset voltage of only \(25 \mu \mathrm{~V}\) and is available in two versions having input offset voltage drifts of 2 or \(1 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\). The absolute maximum values for input offset over 0 to \(70^{\circ} \mathrm{C}\) are 400 \(\mu \mathrm{V}\) for the \(2-\mu \mathrm{V} /{ }^{\circ} \mathrm{C}\) version and \(250 \mu \mathrm{~V}\) for the \(1-\mu \mathrm{V} /{ }^{\circ} \mathrm{C}\) version. Although the input bias current at the noninverting input of the AD846 is a modest \(10 \mu \mathrm{~A}\) typ, its all-important inverting input specs a typical bias current of only 75 nA that drifts at \(6.5 \mathrm{nA} /{ }^{\circ} \mathrm{C}\). In addition, the device has an open-loop transimpedance of \(500 \mathrm{M} \Omega\)-much higher than that of most currentfeedback op amps-which reduces the closed-loop gain error due to open-loop gain to a mere \(0.0005 \%\).

The AD846's common-mode rejection ratio (CMRR), specified at 125 dB , is also considerably better than the 50 to 60 dB typical of most other current-feedback op

\(200-\mathrm{MHz}, \pm 5 \mathrm{~V}\)-supply op amps (Comlinear Corp)
amps. Inevitably, you must make some tradeoff in bandwidth to achieve this high dc accuracy, but the part specs a typical small-signal \(3-\mathrm{dB}\) bandwidth of 46 MHz , and has typical 10 V -step settling times, to \(0.1 \%\) and \(0.01 \%\), of 80 and 110 nsec , respectively. A commer-cial-grade version in a plastic DIP costs \(\$ 4.50\) (100).

\section*{Digital signals need analog performance}

There are also good reasons for using current-feedback op amps to process digital signals. As these signals propagate along transmission lines, their pulse waveforms are degraded by imperfections in the line and by external interference. To recover the signal and retransmit it to the next section of cable, therefore, you must often place repeaters at strategic points in the line. Because a current-feedback op amp's internal stages are almost entirely symmetrical, the device has very well-matched output rise and fall times. When used as comparators to detect when the input signal crosses a threshold, these op amps can generate highly symmetrical pulse waveforms. In addition, many of the current-feedback op amps on the market can deliver output currents of 100 mA or more, so they can directly drive low-impedance coaxial cables. Although Comlinear Corp's CLC501 has a recommended output current that's limited to \(\pm 50 \mathrm{~mA}\), the device has particular advantages in pulse-recovery applications, because it enables you to clamp the output at the required logic levels, even though the input signal may saturate the amplifier.

Apex Microtechnology Corp's WA01 current-feedback op amp is especially suited for applications that require a high-speed, high-current drive capability.

\section*{}


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\section*{OPTOELECTRONIC PRODUCTS GROUP}

\title{
Combining a current-feedback op amp with a conventional op amp can give you the best of both worlds - both high speed and precision.
}

Even when delivering 400 mA at 20 V p-p, the op amp achieves a typical power bandwidth of 40 MHz ; with a 4 V p-p, \(50-\mathrm{mA}\) output, its typical bandwidth is 150 MHz . The op amp is available in two grades that have input offset voltages of \(\pm 4\) and \(\pm 2 \mathrm{mV}\) and offsetvoltage drifts of 15 and \(10 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\), respectively.
The WA01 provides balance pins so that you can null the amplifier's input offset with an external potentiometer. The WA01 is pin compatible with the industrystandard 3554 wideband op amp, but offers superior performance at low gain values: It has more than four times the 3554's bandwidth at a gain of 10 , provides over three times the 3554's slew rate, and settles around six times faster. Further, it doesn't require any compensation capacitors. The WA01 targets such applications as video-signal distribution, input or output drivers for flash A/D and D/A converters, and sample/ hold-circuit drivers. Its high output-current capability also makes it suitable for use as a pin driver in highspeed ATE equipment.
Because current-feedback op amps don't require phase-degrading circuitry such as ac feedforward compensation to maintain stability, they exhibit a very linear phase response. As a result, the propagation delay through the amplifier is the same for both the fundamental and harmonic components of the input signal, leading to a high degree of waveform fidelity at

\section*{For more information . . .}

For more information on the current-feedback op amps described in this article, contact the following manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.

\author{
Analog Devices Inc Box 9106 \\ Norwood, MA 02062 \\ (617) 329-4700 \\ TWX 710-394-6577 \\ Circle No 360 \\ Apex Microtechnology Corp \\ 5980 N Shannon Rd \\ Tucson, AZ 85741 (602) 742-8600 \\ Fax (602) 888-3329 \\ Circle No 361 \\ Comlinear Corp \\ 4800 Wheaton Dr \\ Fort Collins, CO 80525 \\ (303) 226-0500 \\ TLX 450881 \\ Circle No 362 \\ \section*{Elantec Inc} \\ 1996 Tarob Ct \\ Milpitas, CA 95035 \\ (408) 945-1323 \\ TWX 910-997-0649 \\ Circle No 363 \\ Precision Monolithics Inc 1500 Space Park Dr Santa Clara, CA 95052 \\ (408) 727-9222 \\ TWX 910-338-0218 \\ Circle No 364
}
```

REF LEVEL /DIV
0.000dB 1.000dB

```


200-MHz, overdrive-protected op amps (Comlinear Corp)
the output. Most current-feedback op amps exhibit phase nonlinearity of around \(1^{\circ}\) typ (at \(25^{\circ} \mathrm{C}\), from dc to approximately half of their small-signal bandwidth). However, you'll find versions for which that figure is an absolute maximum rather than a typical value. Also, because their internal bipolar transistors operate as current amplifiers rather than voltage amplifiers, they introduce very low levels of harmonic distortion: Their second and third harmonic-distortion levels are typically between -50 and -70 dB at 20 MHz for the Comlinear op amps. At lower frequencies, that parameter translates to very low levels of distortion indeed (for its AD846 op amp, for example, Analog Devices quotes a total-harmonic-distortion figure of \(0.0002 \%\) at 100 kHz , which should make for some impressive hi-fi equipment).

\section*{Op amps for all seasons}

Finally, most of the current-feedback op amps on the market are available in versions that suit the industrial temperature range ( -25 to \(+85^{\circ} \mathrm{C}\) ) and the military temperature range ( -55 to \(+125^{\circ} \mathrm{C}\) ). You'll also find devices that are tailored for \(\pm 5 \mathrm{~V}\) supplies, low power dissipation, or low cost, as well as devices that feature output-short-circuit or input-overdrive protection. However, because current-feedback op amps rely on bipolar transistors to achieve their high-frequency performance, you won't find a CMOS version.

EDN

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CIRCLE NO 60

\section*{Integrated Circuits}

\section*{Enhanced Z80 microprocessor can address 1 M byte of memory}

The Z80180 is an enhanced version of the company's Z80 microprocessor. It incorporates an on-chip mem-ory-management unit (MMU) that can address as much as 1 M byte of memory and support 64 k bytes of logical I/O space. The \(\mu \mathrm{P}\) contains two direct-memory-access (DMA) channels for transferring data from memory to memory and memory to I/O.

In addition to supporting the Z80 instruction set, the chip uses seven additional high-level instructions, including multiply. Other features include an on-chip wait-state generator, a programmable dynamic-


RAM refresh controller, two fullduplex asynchronous serialcommunications (UART) channels, a clocked serial-I/O port, two 16 -bit
programmable timers, an on-chip clock oscillator, and an interrupt controller. The CMOS chip comes in a 64-pin plastic DIP and a \(68-\) pin plastic leaded chip carrier. When operating at 8 MHz from a 5 V supply, the chip dissipates 200 mW ; in system-standby mode, it dissipates less than 50 mW . It operates over 0 to \(70^{\circ} \mathrm{C}\). The \(8-\mathrm{MHz}\) part sells for \(\$ 11.45\) (100). A \(6-\mathrm{MHz}\) version costs \(\$ 8.55(100)\). Delivery is 10 to 12 weeks ARO.

Zilog Inc, 210 Hacienda Ave, Campbell, CA 95008. Phone (408) 370-8000. TWX 910-338-7621.

Circle No 410

\title{
1-chip, 64-bit floating-point processors achieve a peak rate of 20 M flops
}

The WTL 3164 and WTL 3364 are 64 -bit floating-point processors. The single-chip devices contain a 64 -bit floating-point multiplier, a 64 -bit floating-point ALU, a divide and square-root unit, and a register file that contains thirty-two 64-bit words. The register file is accessible through six independent, internal ports.

The processors' functions include single- and double-precision float-ing-point, integer, shift, logical, and \(\mathrm{min} / \max\) instructions that operate at a peak rate of 20 M flops for the 100 -nsec grade. You can use the 3364 in applications, such as vector and array processors, that require high throughput rates. The 3164 suits applications that are computa-

tionally intensive, but the I/O bandwidth requirement is of secondary importance. The 168 -pin 3364 has three 32 -bit ports; a bidirectional,
an input, and an output port. You can use the ports in that configuration, or you can combine the input and output ports to connect to a 64 -bit bidirectional bus. The 144 -pin 3164 is functionally identical to the 3364 except that it has a single 32-bit bidirectional bus. Both chips have a single-cycle throughput for all multiplier and ALU operations. Register-to-register operations take two cycles. Double-precision divide and square-root operations take 17 and 30 cycles, respectively. The \(100-\) nsec 3164 costs \(\$ 829\) (10).

Weitek Corp, 1060 E Arques Ave, Sunnyvale, CA 94086. Phone (408) 738-8400.

Circle No 411

\section*{OP-07 Performance At 1/6TH THE POWER}

\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|l|}{Check the specs for yourself:} \\
\hline & OP-97 & OP-07 & \\
\hline V \({ }_{\text {OS }}\) & 25 & 25 & \(\mu \mathrm{V}\) Max \\
\hline TCV \({ }_{\text {OS }}\) & 0.6 & 0.6 & \(\mu \mathrm{V} /{ }^{\circ} \mathrm{C}\) Max \\
\hline Avo & 300 & 300 & \(\mathrm{V} / \mathrm{mV}\) Min \\
\hline \(\mathrm{I}_{\mathrm{B}}\) @ \(25^{\circ} \mathrm{C}\) & \(\pm 100\) & \(\pm 2,000\) & pA Max \\
\hline I \(\mathrm{B}^{\text {@ }} 125^{\circ} \mathrm{C}\) & \(\pm 250\) & \(\pm 4,000\) & pA Max \\
\hline ISY @ \(\pm 15 \mathrm{~V}\) & 600 & 4,000 & \(\mu \mathrm{A}\) Max \\
\hline
\end{tabular}

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\section*{Integrated Circuits}

\title{
GaAs programmable logic device has propagation delay of 10 nsec
}

The GA22V10 is a programmable logic device built with gallium arsenide technology. It has TTL-compatible I/O lines and is functional and pin compatible with the silicon version of the 22 V 10 . The propagation delay between an input signal and a nonregistered output is 10 nsec max for the commercial version, 12 nsec max for the military version.

The commercial version of the device can run at 90 MHz max; its military counterpart can be clocked at 71 MHz . The chip provides 22

inputs and ten 3 -state outputs that you can program for registered or combinatorial data with active-high or active-low polarities. It also has a
power-up reset command, a synchronous preset command, and an asynchronous reset mode for statemachine applications. To prevent unauthorized access to the internal configuration, a security link disables the pattern-verification function and the preloading of the registered outputs. The PLD costs \(\$ 55\) (100).

Gazelle Microcircuits Inc, 2300 Owen St, Santa Clara, CA 95054. Phone (408) 982-0900. Fax 408-9820222.

Circle No 416

\section*{256k \(\times\) l-bit dynamic RAMs offer access times as low as 70 nsec}

The V53C256 and the V53C258 are families of \(256 \mathrm{k} \times 1\)-bit dynamic RAMs. Each family contains four models offering row-access times (RAS) of \(70,80,100\), and 120 nsec , respectively. All the inputs of these CMOS devices are TTL compatible. When operating in fast-page mode, the \(70-\mathrm{nsec}-\mathrm{RAS}\) version of the -256 RAM lets you randomly access as many as 512 bits within a row at a cycle time of 50 nsec. If you keep the RAS line active while applying successive column-access strobes (CAS), the device retains the row address internally, so you don't need to reapply it on each cycle. During fast page-mode operation, the RAM lets you perform Read, Write, Read-Modify-Write, or Read-Write-Read operations. The part also provides a sustained data rate of 19 MHz .

The -258 family offers a similar mode called the static column mode; for the \(70-\) nsec-RAS version of the

-258, that mode lets you randomly access 512 bits within a row at a cycle time of 45 nsec. In this mode, the device acts as a static RAM for multiple-column accesses within a row so that you can mix read and write cycles. The chip provides a sustained data rate of 22 MHz . The RAMs ara available in 16-pin DIPs,

16-pin ZIPs (zigzag in-line packages), and 18 -pin PLCCs. The \(70-\) nsec versions of both devices cost \(\$ 7.87\) (100).

Vitelic Corp, 3910 N First St, San Jose, CA 95134. Phone (408) 4336000. TLX 3719461. Fax 408-4330331.


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For fast EPROMs, or cost-effective packaging options, call or write: Mitsubishi Electronics America, Inc., Semiconductor Division, 1050 East Arques Avenue, Sunnyvale, CA 94086. (408) 730-5900.


Products subject to availability.

\section*{Integrated Circuits}

\section*{Power MOSFETs can carry drain currents as high as 22A}

The 23 new members of the Power MOS IV line of power MOSFETs offer a range of current-handling capabilities and come either in TO-3 packages or in die form. An example of the TO-3 devices is the APT4020AN, which has a drain-tosource voltage \(\left(\mathrm{V}_{\mathrm{DS}}\right)\) rating of 400 V , a drain-to-source on-resistance\(\mathrm{R}_{\mathrm{DS}}\) (on)—of \(0.20 \Omega\), drain-to-source current \(\left(\mathrm{I}_{\mathrm{D}}\right)\) rating of 22.0 A , and \(2400-\mathrm{pF}\) input capacitance ( \(\mathrm{C}_{\text {ISS }}\) ). Another example, the APT6035AN, has a \(V_{D S}\) rating of 600 V , an \(\mathrm{R}_{\mathrm{DS}}\) (on) of \(0.35 \Omega\), an \(I_{D}\) rating of 17.0 A , and

a \(\mathrm{C}_{\text {ISS }}\) of 2400 pF . You can also obtain units with \(800-\mathrm{pF} \quad \mathrm{C}_{\text {ISS }}\), 180-pF output capacitance ( \(\mathrm{C}_{\text {OSS }}\) ), or \(60-\mathrm{pF}\) transfer capacitance ( \(\mathrm{C}_{\mathrm{RSS}}\) ). The dies for hybrid applications
have similar specifications; however, their drain-current specifications depend on the package you use. The devices' large die sizes lower their thermal resistance and increase their thermal capacity. The APT6035AN costs \(\$ 28.86\) (100); the APT4020AN is \(\$ 22.65\) (100).

Advanced Power Technology, 405 SW Columbia St, Bend, OR 97702. Phone (503) 382-8028. Fax 503-388-0364.

Circle No 417

\section*{Multiple-bus exchangers act as digital crossbar switches}

The Am29C982 and the Am29C983 are multiple-bus exchangers that act as digital crossbar switches. Both chips are fabricated with CMOS technology. You can select each of the four bidirectional, 4-bit ports on the Am29C982 independently to connect to any of the other ports. This organization permits data routing and multiple-bus communications. The device comes in a 28 -pin DIP, leadless chip carrier (LCC), or plastic leaded chip carrier (PLCC), and it exhibits typical port-to-port delays of 7 nsec.
The Am29C983 is a similar device except that the four ports are each 9 bits wide. In addition, each I/O port has an input latch and an output latch to capture incoming and outgoing data, respectively. Each latch has an independent latch-enable input. The chip has independent output-enable lines that can place the output drivers in a high-imped-

ance state. All the I/O lines go to a high-impedance state upon powerdown. The device is available in a 68 -pin PLCC or LCC, and it exhibits typical port-to-port delays of 9 nsec. The ports for each of these devices are TTL compatible and can sink 48 mA of output drive current.

The 28-pin DIP version of the Am29C982 costs \(\$ 8.50\) (100); the 68-pin PLCC version of the Am29C983 costs \(\$ 24.50\) (100).

Advanced Micro Devices Inc, Box 3453, Sunnyvale, CA 94088. Phone (408) 732-2400. TLX 346306. Circle No 413

\title{
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Conversion times: Model HI-574A, \(20 \mu \mathrm{~s} \ldots \mathrm{HI}-674 \mathrm{~A}, 12 \mu \mathrm{~s} . . . \mathrm{HI}-774,8 \mu \mathrm{~S}\).

The HI-774 features a smart successive approximation register; its digital error correction circuitry improves dynamic accuracy and throughput rate. All three ADCs can operate under control of the processor, or in a standalone mode. Models come in commercial and military temperature ranges - including MIL-STD-883. Packages: 28-pin Cerdip and leadless chip carriers (LCCs).

For information call 1-800-4-HARRIS, Ext. 1405. In Canada, 1-800-344-2444, Ext. 1405. Or write: Harris Semiconductor Products Division, P.O. Box 883, MS 53-035, Melbourne, Florida 32902-0883.


\section*{Integrated Circuits}

\title{
16-bit microcontroller executes within 400 nsec after an interrupt
}

The RTX 2000 is a 16 -bit microcontroller for embedded real-time applications. The chip, which is designed around the company's Forth-based RISC (reduced-in-struction-set computer) core processor, can begin an instruction sequence within 400 nsec after an interrupt. While running at 10 MHz , the chip can execute a maximum of 40 million Forth operations/ sec and can typically operate at a sustained rate of 10 million operations/sec. It also features a dualstack architecture; each stack is 256

words long. The parameter stack stores data temporarily and passes data between subroutines. The return stack stores the return address when the microcontroller calls sub-
routines. The chip has an on-chip \(16 \times 16\) multiplier that can execute in a single cycle; an on-chip interrupt controller; and three general-purpose, 16 -bit timers. An ASIC bus lets you attach ASIC devices to the chip. The CMOS device comes in an 84-pin PGA (pin-grid array) package and costs \(\$ 190\) (1000).
Harris Corp, Semiconductor Sector, Box 883, Melbourne, FL 32901. Phone (305) 724-7418.

Circle No 418

\section*{Dual switched-capacitor filters operate to 150 kHz}

The ML2111 consists of two independent switched-capacitor filters that operate to 150 kHz and have a Q of 20 . Each filter can approximate second-order transfer functions for lowpass, bandpass, highpass, notch, and allpass filters. The device requires an external clock and resistors that let you build Butterworth, Chebyshev, Bessel, and Cauer filters. The product of the center frequency and the \(Q\) can range as high as 5 MHz . A control pin allows you to select a clock-to-frequency ratio of \(50: 1\) or \(100: 1\) or to hold the last output sample. The vendor offers two versions of the ML2111: The -B version has a center-frequency accuracy of \(\pm 0.4 \%\) and a \(Q\) accuracy of 0 to \(-4 \% \mathrm{max}\); the -C version's cen-ter-frequency accuracy is \(\pm 0.8 \%\) \(\max\), and its \(Q\) accuracy is 0 to \(-8.0 \%\) max. Each filter consists of a

low-offset ( \(15-\mathrm{mV}\) max) voltage amplifier, a voltage summer, two integrators, and control logic. The filters maintain their specifications when operating from \(\pm 5 \mathrm{~V}\) supplies with \(\pm 10.0 \%\) tolerances. The device can also operate from a single \(5 \mathrm{~V} \pm 10 \%\) supply at a maximum operating frequency of 100 kHz . An

ML2111 in a 20 -pin DIP costs \(\$ 6.95\); in a 20 -pin SOIC package, it's \(\$ 7.15\).

Micro Linear Corp, 2092 Concourse Dr, San Jose, CA 95131. Phone (408) 433-5200. TLX 275906. Circle No 415

\section*{FEATURES}

\section*{Fast Settling:}

500 ns to \(0.01 \%\) for 10 V Step
\(1.5 \mu \mathrm{~s}\) to \(0.0025 \%\) for 10 V Step
Slew Rate: \(75 \mathrm{~V} / \mu \mathrm{s}\)
Total Harmonic Distortion (THD): 0.0003\%
\(>1000 \mathrm{pF}\) Capacitive Load Drive Capability with \(10 \mathrm{~V} / \mu \mathrm{s}\) Slew Rate

Input Offset Voltage: 0.25 mV max
Input Offset Drift: \(3 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\) max
Open Loop Gain: \(\mathbf{2 5 0 V} / \mathbf{m V}\) min
Noise: \(4 \mu \vee\) p-p max, 0.1 Hz to 10 Hz


AD744 Settling Characteristics Oto +10 V Step
PRODUCT HIGHLIGHTS
1. The AD744 offers exceptional dynamic response. It settles to \(0.01 \%\) in 500 ns and has a \(100 \%\) tested minimum slew rate of \(50 \mathrm{~V} / \mu \mathrm{s}\).
2. The combination of Analog Devices' advanced processing technology, laser wafer drift trimming and well-matched ionimplanted JFETs provide outstanding de precision. Input offset voltage, input bias current and input offset current are specified in the warmed-up condition and are \(100 \%\) tested.
3. The AD744 has a guaranteed and tested maximum voltage noise of \(4 \mu \mathrm{~V}\) p-p, 0.1 to 10 Hz .
4. The AD744 is a high speed BiFET op amp that offers excellent performance at competitive prices. It outperforms the OP42/44, OPA602, LF356 and LF400.

\section*{PRODUCT DESCRIPTION}

The AD744 makes a breakthrough in the high speed BiFET market by offering guaranteed maximum settling to \(0.01 \%\) in 750 ns . It also offers the excellent dc characteristics of the AD711 BiFET family with enhanced slew rate, bandwidth and load driving capability.

The single-pole response of the AD744 provides fast settling: 500 ns to \(0.01 \%\) typically, and 750 ns maximum. This feature, combined with high dc precision, makes the AD744 suitable for use as a buffer amplifier for 12-, 14- and 16-bit DACs and ADCs. Furthermore, the AD744's low total harmonic distortion (THD) level of \(0.0003 \%\), low noise and gain bandwidth product of 13 MHz make it an ideal amplifier for demanding audio applications. It is also an excellent choice for high speed instrumentation amplifiers and for use in active filters.

The AD744 offers optional custom compensation for additional design flexibility. This external compensation allows the AD744 to drive capacitive loads up to 2000 pF and greater with full stability, making the AD744 outstanding for use as a coaxial cable driver. Alternatively, external decompensation may be used to increase the gain bandwidth of the AD744 to over 200 MHz at high gains. This makes the AD744 ideal for use as an ac preamp in digital signal processing (DSP) front ends.

\title{
THIS PAGE THE QUESTION OF WHO

}


A lot of companies say they have fastsettling, high-performance BiFETamps. But our AD744 settles to \(0.01 \%\) in 500 ns and to \(0.0025 \%\) in \(1.5 \mu \mathrm{~s}\)-making it the world's fastest-settling, highest-performance BiFET. This superior settling, combined with excellent dc performance, makes the AD744 unbeatable for active filters, and for buffering DACs and ADCs up to 16 bits.

The AD744, with a total harmonic distortion of just \(0.0003 \%\), low noise, a clean pulse response, and a gain bandwidth product of up to 200 MHz , is also ideal for digital signal processing and audio applications.

If you work in communications, you'll appreciate the AD744's ability to drive loads greater than 2000 pF with full stability. And you'll also appreciate the fact that the AD 744 can drive 1000 pF
cap load while maintaining a slew rate of \(10 \mathrm{~V} / \mu \mathrm{s}\).
The AD744 isn't our only outstanding BiFET, either. The AD711 single, AD 712 dual, and soon the AD 713 quad, settle in \(1 \mu \mathrm{~s}\) with the same high resolution as the AD744. If low power with precision is critical, try our AD548 single or AD648 dual.

Whichever BiFET your application requires, you'll find our products deliver excellent performance at an excellent price. For example, the AD744 starts at only \(\$ 2.25\); the AD711 at \(\$ .80\); and the AD548 at \(\$ .75(100 \mathrm{~s})\).

If you'd like to see more proof on why we can say we make the best BiFETs, call Applications Engineering at (617) 935-5565, ext. 2628 or 2629 . Or write to Analog Devices, P.O. Box 9106, Norwood, MA 02062-9106.


\section*{Integrated Circuits}

\title{
Op amp offers a \(1000 \mathrm{~V} / \mu \mathrm{sec}\) slew rate and a l-GHz gain-bandwidth product
}

The EL2038 is a monolithic op amp with a gain-bandwidth product of 1 GHz and a \(1000 \mathrm{~V} / \mu\) sec slew rate. The device is stable for gains of \(20 \mathrm{~V} / \mathrm{V}\) or greater and is pin-compatible with the HA2539 and NE5539 op amps. Operating at a gain of \(20 \mathrm{~V} / \mathrm{V}\), it provides a \(50-\mathrm{MHz}\) bandwidth when driving a \(400 \Omega\) load. When operating from \(\pm 15 \mathrm{~V}\) supplies, it can deliver a \(\pm 11 \mathrm{~V}\) output signal and can source and sink 25 mA into a \(400 \Omega\) load. This outputdrive capability lets the device drive capacitive loads as high as 25 pF at

the \(1000 \mathrm{~V} / \mu \mathrm{sec}\) slew rate. You can power the device from supplies ranging from \(\pm 5\) to \(\pm 17 \mathrm{~V}\). Other
specifications include a \(2-\mathrm{mV}\) max offset voltage, a \(17-\mathrm{mA}\) max powersupply current, an \(80-\mathrm{dB}\) open-loop gain, and a high-power bandwidth of 15.9 MHz typ. The military version complies with MIL-STD-883 revision C; the vendor performs a burnin test at \(125^{\circ} \mathrm{C}\) before delivery. The commercial-grade device costs \(\$ 3.90\) (100); the MIL-STD-883 device sells for \(\$ 32.40\) (100).
Élantec Inc, 1996 Tarob Ct, Milpitas, CA 95035. Phone (800) 821-7429; in CA, (408) 945-1323.

Circle No 419

\section*{High-speed op amps offer overdrive and short-circuit protection}

The CL205 and CL206 are op amps that employ current-feedback technology. When you set its voltage gain to \(20 \mathrm{~V} / \mathrm{V}\), the CLC205 has a \(3-\mathrm{dB}\) bandwidth of 170 MHz . It settles to \(0.05 \%\) of final value in 24 nsec. It consumes 570 mW when operating from \(\pm 15 \mathrm{~V}\), and 56 mW when operating from \(\pm 5 \mathrm{~V}\). The op amp's output swing can be \(\pm 12 \mathrm{~V}\) with a \(50-\mathrm{mA}\) output current. For a voltage gain of \(20 \mathrm{~V} / \mathrm{V}\), the CL206 has a \(3-\mathrm{dB}\) bandwidth of 180 MHz . It settles to \(0.1 \%\) of its final value in 19 nsec , and its output drive current is 100 mA . The part features a slew rate of \(3400 \mathrm{~V} / \mu \mathrm{sec}\), which provides a full-power \(3-\mathrm{dB}\) bandwidth of 70 MHz for a 20 V p-p output swing when the op amp is operating at a voltage gain of \(20 \mathrm{~V} / \mathrm{V}\).
The op amps self-limit the input currents when the outputs are saturated. You can use diode-clamp circuits on the inputs when the input signal level exceeds the maximum

differential input voltage \(( \pm 3 \mathrm{~V})\) or common-mode input voltage \(\left( \pm V_{C C}-1 V\right)\). You can protect the outputs from the effects of short circuits by using output-currentlimiting techniques. Industrial versions of the op amp cost \(\$ 56\) (100) and are available from stock. Mili-
tary versions are \(\$ 138\) (100); they're available for delivery within six weeks ARO.
Comlinear Corp, 4800 Wheaton Dr, Fort Collins, CO 80525. Phone (303) 226-0500. TLX 450881. Fax 303-226-0564.

Circle No 412

\section*{NCR keeps raising the standards for SCSI.}

\section*{Finally, a cure for SCSI overheadaches.}

NCR's 53C90 is the only chip that can give you fast, fast, fast relief from overheadaches. Using combination commands, dedicated sequential logic and dual-ranked registers for command pipelining, the 53C90 is quickest on and off the bus. Plus NCR implements complex bus sequencing in hardware, not time-wasting software.

Transfer rates? NCR's 53C90 delivers the SCSI bus maximum of 5.0 MBytes/sec synchronous and 3 MBytes asynchronous at 25 MHz for the full length of the bus.

How to get zapresistance, latch-up protection and the blessings of the FCC.

\section*{A big, wellconnected family.}

NCR's family goes back to the "Mayflower" of SCSI controllers with the 5385 in 1982. The most recent offshoot-the high-performance 53 C 90 A . Other family members include a single chip host bus adapter (53C400), an integrated buffer controller (53C300) and even an ASIC supercell for circuit designers. Plus we'll be there with SCSI II.

\title{
Integrated Circuits
}

\title{
Dual op amp offers low noise and low input offset voltage
}

The OP-270 is a dual op amp that boasts a voltage-noise specification of \(5 \mathrm{nV} / \sqrt{\overline{\mathrm{Hz}}} \max\) at 1 kHz . Its input offset voltage is \(74 \mu \mathrm{~V}\) max; its offset-drift spec is \(<1 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\) over the military and industrial temperature ranges. The device is stable at unity gain; its gain-bandwidth product is 5 MHz typ. When operating at
unity voltage gain, it settles to \(0.01 \%\) of final value in \(5 \mu\) sec typ. The op amp also has a typical output slew rate of \(2.4 \mathrm{~V} / \mu \mathrm{sec}\). Its open-loop gain is \(1500 \mathrm{~V} / \mathrm{mV}\) min into a \(10-\mathrm{k} \Omega\) load, and its CMRR is 106 dB min. The device draws a maximum of 6.5 mA when operating from \(\pm 15 \mathrm{~V}\) supplies. The available packages in-
clude an 8-pin DIP and a 20 -pin LCC package. Industrial-grade devices start at \(\$ 4.50\) (100); military-grade parts cost \(\$ 10.50\).
Precision Monolithics Inc, Box 58020, Santa Clara, CA 95052. Phone (408) 562-7384. TLX 713719541. Fax 408-727-1550.

Circle No 420

\title{
Analog-digital arrays include high- \(\mathrm{f}_{\mathrm{T}}\) transistors, ECL, and \(\mathrm{I}^{2} \mathrm{~L}\)
}

The five members of the Polyuse-J family of analog-digital semicustom arrays range in complexity from 400 to 2300 components. Each array provides small-signal npn and pnp transistors, power Darlingtons, and \(I^{2} \mathrm{~L}\) and ECL elements. In addition to their customizable components, the arrays have standard analogfunction circuits, including an RCor quartz-controlled oscillator, a bandgap reference, a high-frequency op amp, a 6 -bit D/A converter, a voltage reference, and an ECL reference.
The array's npn transistors have an \(f_{T}\) (transition frequency) of 3 GHz , so they let you implement high-performance, small-signal circuitry. You can design power stages by using the array's \(200-\mathrm{mA} \mathrm{npn}\) Darlingtons and \(30-\mathrm{mA} \mathrm{pnp}\) transistors. The ECL elements have multiple emitters, which let you implement stacked logic. You can operate the ECL circuitry at speeds as high as 200 MHz , and you can configure its inputs and outputs to be either ECL- or TTL-compatible. The I \({ }^{2}\) L gates operate at frequencies as high as 2 MHz .

You can customize the arrays with two metal layers, using the

first level to route macrocells and the second to interconnect them. The vendor offers VAX-based and IBM PC-based CAD tools to support IC design with the arrays; it also offers kits of individual components and macrofunctions that you can use to make breadboards of your design. Nonrecurring engineering charges for prototyping are around \(\$ 10,000\), and engineering for volume production costs about \(\$ 10,000\). The
arrays cost \(\$ 2\) to \(\$ 4\), depending on quantity.

SGS-Thomson Microelectronics, Via C Olivetti 2, 20041 Agrate Brianza, Italy. Phone (039) 65551. TLX 330131.

Circle No 430
SGS-Thomson Microelectronics, 1000 E Bell Rd, Phoenix, AZ 85022. Phone (602) 867-6100. TLX 249976.

Circle No 431

\section*{SIEMENS}

\section*{Announcing a 27 Billion Dollar backer for your Siemens ASIC team.}

Siemens, a proven winner in electronics with \(\$ 27\) billion in sales, just entered the U.S. ASIC market. And our team is geared for the ASIC circuit.
With our first effort, we've combined ECL and CML technology in one gate array family. This means you can now design-in the ideal combination of super ECL speed with economical, high-density CML performance on one chip...everytime. You no longer must compromise the speed you need for the power you don't. In addition, speed/power programming, as well as I/Os designed for both ECL \(10 \mathrm{~K} / 100 \mathrm{~K}\) and TTL interfaces give you the flexibility you need.
And that's just the start. Coming down the home stretch are more Siemens entries... 1.5/1.2 micron CMOS standard cells and 1.2/1.0 micron CMOS sea of gates arrays. Use our sea of gates CMOS family for quick turn logic and memory on one chip or design your high-performance, cell-based ASICs utilizing the common

ADVANCELL \({ }^{\text {TM }}\) ASIC library and guarantee yourself compatible second sources.
The Siemens ASIC team's responsive service and technical innovation provides you with the winning edge. For more information on how to put your design into high gear write to Siemens Components, Inc., ASIC Marketing, 2191 Laurelwood Road, Santa Clara, California 95054. Or call ASIC Marketing at 408/980-4568, and see for yourself how Siemens is making the difference on the ASIC circuit.

\section*{Siemens... your partner for the future.}
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in certain countries.
CG/2000-460 WLM 795
CIRCLE NO 67

\section*{Integrated Circuits}

\section*{MIL-STD-1553 bus controller supports multiple protocols}

The UT1553 BCRTMP is a bus controller that supports the protocols for MIL-STD-1553A and -1553B. In addition, it supports the protocols for the McDonnell-Douglas A3818B, A5232, and A5690 and the Grumman Aerospace SP-G-151A specifications for the 1553 bus. The monolithic chip stores and retrieves 1553 messages through a host interface. The interface is either a DMA channel that must arbitrate for the host's bus or a pseudo-dual-port handshake that uses buffer registers to separate the message RAM from the host bus. The host inter-

face has an address range of 64 k words. When operating in the re-mote-terminal mode, the chip can
store as many as 128 messages and can relocate data blocks to any subaddress. When operating in the buscontroller mode, it uses a memory scheme that allows the host to chain messages. The device comes in either a 144 -pin PGA or a 132 -lead flatpack, and it complies with MIL-STD-883 revision C. It costs \(\$ 1047\).

United Technologies Microelectronics Center, Military Standard Products Dept, 1575 Garden of the Gods Rd, Colorado Springs, CO 80907. Phone (800) 645-8862 or (303) 594-8259.

Circle No 421

\section*{On-chip write-pulse generator improves ECL RAMs' cycle time}

The members of this family of ECL Self-Timed Rams (STRAMs) feature latched or registered I/O lines along with an on-chip write-pulse generator. A clock signal triggers the write-pulse generator. The timing of the internal write pulse to the memory cells minimizes the setup and hold times. As a result, the RAM automatically controls the critical relationship between the clock-pulse width and the write-enable timing. One member of the family, the MBM10486LL-13, is a \(4 \mathrm{k} \times 4\)-bit RAM having a cycle time of 13 nsec min and an address-access time of 10 nsec max. It is compatible with 10 K ECL devices. Another member, the MBM100476LL-9, is organized as a \(1024 \times 4\)-bit RAM; it has a 9 -nsec min cycle time and a 1-nsec max address-cycle time, and

it's compatible with 100 K ECL products. The devices come in 28 -pin ceramic DIPs. The \(1 \mathrm{k} \times 4\)-bit units cost \(\$ 55\) (1000); the \(4 \mathrm{k} \times 4\)-bit devices sell for \(\$ 75\) (1000).

Fujitsu Microelectronics Inc, Marketing Communications, 3545 N First St, San Jose, CA 95134. Phone (408) 922-9000.

Circle No 422


\section*{HP humbly introduces the highest performance multimeter in the world.}

\section*{100,000 readings per second and Cal Lab accuracy.}

Lots of people claim they have the best multimeter, but we're just going to let the specs do the talking. You'll see we've created a multimeter that doesn't present you with a bunch of trade-offs-you can have speed and accuracy.
- 100,000 readings per second at \(41 / 2\) digit resolution. If you need \(51 / 2\) digit resolution, you'll get it at 50,000 readings per second.
- Remarkable throughput rate-change a function and change a range, take
a measurement and output to the bus 200 to 300 times per second
- Calibration standard accuracy and \(81 / 2\) digits.
- Modest price: \(\$ 5,900.00\) *

To get complete technical specifications before you order an HP 3458A DMM for your system, call 1-800-752-0900, Dept. A215.

\section*{Integrated Circuits}

\section*{Phone IC automatically compensates for loudness of phone user's voice}

The TEA1064 speech/line-interface IC, for use in telephone sets, features dynamic gain limiting, which reduces distortion in the transmitted speech signal and reduces the transmitted side tone to a comfortable level. The device meets the performance requirements of all the major PTTs (post, telephone, and telecommunications authorities) in Europe, the USA, and the Far East. Because it can operate from a line voltage of 1.7 V , it allows several phones to operate in parallel on the same subscriber line.
The speech circuit amplifies the voice signal received over the telephone line and provides a singleended or differential drive for a variety of earpiece types. The earpiece amplifier generates only -85 dB of noise, even when operated at a gain of 31 dB . The microphone circuit is suitable for use with dynamic, magnetic, piezoelectric and electret microphones.


The IC has an interface for a dialer, and it produces an earpiececonfidence tone during DTMF dialing. A Mute input allows you to disable speech transmission during DTMF or pulse dialing. The DTMF amplifier has a typical gain stability of \(\pm 0.2 \mathrm{~dB}\) over the device's -25 to \(+75^{\circ} \mathrm{C}\) operating-temperature range. It also has automatic gain control, which allows you to operate
the device with various values of exchange supply resistor and various voltages. The IC also generates a \(3-\mathrm{mA}\) supply from the line voltage for peripheral circuitry.
During normal operation, the TEA1064 works with line currents of between 11 and 140 mA . At a reduced-performance level, it operates at line currents as low as 2 mA and at a line voltage of 1.7 V . Its maximum continuous line voltage is 12 V . The device comes in a 20 -pin DIP or a small-outline surfacemount package. It costs around gld 3 in high volume and is available for delivery within eight weeks ARO.
Philips, Components Div, Box 523, 5600 AM Eindhoven, The Netherlands. Phone (040) 757189. TLX 51573.

Circle No 428
Signetics Corp, 811 E Arques Ave, Sunnyvale, CA 94088. Phone (408) 991-4571.

Circle No 429

\section*{Error-detection and -correction units come in four speed grades}

The P74PCT632CC is a family of CMOS error-detection and correction units (EDACs). The EDACs can detect and correct single-bit errors in a 39 -bit word that consists of 32 data bits and 7 modified-Hammingcode check bits. In a typical application, when the system writes to memory, the EDAC reads the 32 data bits on the data bus and generates 7 check bits to produce a 39 -bitwide word to be stored in the system memory.

During a read operation, the chip reads all 39 bits from memory in order to detect any errors. If the chip detects a single-bit error, it corrects the data and sends the corrected data to the host and memory. If the chip detects a double-bit error, it notifies the host processor of the existence of the error. It can also detect and flag gross error conditions, such as all-zeros or all-ones conditions. The family comprises four models, which have error-de-
tection time delays of \(30 \mathrm{nsec}, 25\) nsec, 20 nsec, and 16 nsec max, respectively. Each chip operates from a -5 V supply and comes in a 52 -pin DIP. The 16 -nsec version costs \(\$ 138\) (100).

Performance Semiconductor Corp, 610 E Weddell Dr, Sunnyvale, CA 94089. Phone (408) 7348200. TWX 650-271-5784. Fax 408-734-0258.

Circle No 423

\title{
L/S Band Power GaAs FEIs: A New Era In Power Amplifiers.
}

Power amplifier technology has come a long way. Just consider NEC's new L/S band power GaAs FETs.

There's the NE345L-10B L/S-band GaAs FET with 10W of linear power or the NE345L-20B L-band GaAs FET with 20W of linear power.
\begin{tabular}{lcrc} 
Part & \begin{tabular}{c} 
P1db \\
(TYP)
\end{tabular} & \begin{tabular}{c} 
GL \\
(TYP)
\end{tabular} & \begin{tabular}{c} 
Eff. \\
(TYP)
\end{tabular} \\
\hline NE345L-10B & \(\mathbf{4 0} \mathbf{d b m}\) & \(\mathbf{9 d b}\) & \(\mathbf{4 0 \%} @ 2.3 \mathbf{~ G H z}\) \\
NE345L-20B & \(\mathbf{4 3 d b m}\) & \(\mathbf{1 0 ~ d b}\) & \(\mathbf{4 0 \%} @ \mathbf{1 . 5} \mathbf{~ G H z}\)
\end{tabular}

With MTBF's that are orders of magnitude better than TWT's, no warm-up time, and no heavy power supplies, these parts are ideal replacements for TWT's in existing systems.

The NE345L series' excellent linear gain, high performance, and hermetically sealed ceramic packaging also make them the perfect choice for many applications: such as phased array radars, airborne navigation systems, studio/transmitter links, educational TV, and mobile satellites.

Contact CEL for more information, data sheets, or application support. Then see how your power amplifiers can truly come of age.

California Eastern Laboratories
3260 Jay Street, Santa Clara, CA 95054 (408) 988-3500
Western (408) 988-3500 Eastern (301) 667-1310
Canada (613) 726-0626
Europe NEC Electronics GmbH 0211/650301
CIRCLE NO 71

State of the art 1959.


State of the art 1987.


\title{
Hard-disk-controller chip integrates formatter, buffer manager, and bus controller
}

The CL-SH260 is a hard-disk-controller chip for the IBM PC, PC/XT, and PC/AT and compatible computers. The single CMOS chip integrates formatter, buffer-manager, and bus-controller functions. The formatter can handle NRZ disk data rates as high as 20 M bps. It consists of a serializer/deserializer, a RAMbased track sequencer, and selectable 16 -bit CCITT CRC (cyclic redundancy check) or 32-bit AT ECC (error-correction code) circuitry. The hardware also generates a 56 -bit ECC polynomial. The formatter supports 1:1 interleaves and can be used with interfaces to ST506/ 412, ST412HP, ESDI, and SMD disks. The buffer manager can address as much as 64 k bytes of RAM and can support a throughput rate as high as 6 M bytes \(/ \mathrm{sec}\) to or from the buffer memory. The bus controller has on-chip registers that emu-

late the IBM Task File for the PC/AT and the IBM Command Descriptor Block for the PC/XT. The chip can sustain a 2 M -byte/sec transfer rate to the disk and a 4 M byte/sec transfer rate on the host
bus. It sells for \(\$ 30\) (OEM qty).
Cirrus Logic Inc, 1463 Centre Pointe Dr, Milpitas, CA 95035. Phone (408) 945-8300. TLX 171918. Fax 408-263-5682.

Circle No 425

\title{
Family of 8-bit microcontrollers offers EEPROMs as an on-chip option
}

The TMS370 is a family of 8 -bit microcontrollers fabricated in \(1.6-\mu \mathrm{m}\) CMOS. The parts' on-chip options include 256 bytes of EEPROM for data memory, 4 k bytes of ROM or EEPROM for program memory, an 8-bit A/D converter, a serial communications interface, and an expansion bus. Besides the six standard configurations that are currently available, the company offers 16 special-function modules for custom designs. Each of the standard \(\mu \mathrm{Cs}\) contains a

serial peripheral interface, a watchdog timer, a 16 -bit timer (on three of the units, two timers are standard),
and either 128- or 256-byte data registers. The units use a single 5 V supply and can function with a \(20-\mathrm{MHz}\) clock. They operate over -40 to \(+85^{\circ} \mathrm{C}\). Chips packaged in 28 -pin plastic DIPs or PLCCs cost \(\$ 3\) to \(\$ 7\), depending on quantity. Parts housed in 68-pin PLCCs range in price from \(\$ 4.50\) to \(\$ 10\).

Texas Instruments Inc, Semiconductor Group, SC-828, Box 809066, Dallas, TX 75380. Phone (800) 232-3200 ext 700.

Circle No 426


\section*{THE ONLY THING FASTER THAN ROCKWELI'S MODEMS IS MARSHAL'S SERVICE.}

At 14,400 bits-per-second, Rockwell's high-speed modems are a full generation ahead of the competition.

This new 14.4 Kbps modem family is the latest in Rockwell's leading R-series modems. The R144DP is the V.33/V. 29 product offering which complements their R96DP and R48DP/208 high-speed modems. And the R144HD features are compatible to the R96F, the standard for facsimile modems.

Which is why Rockwell modems can be found in high-speed network controls and multiplexers, personal computers and terminals, custom modems, facsimile, and desktop publishing equipment around the world. In fact, Rockwell is the world's leading supplier of original equipment manufacturer modems.

And for the fastest delivery of the world's fastest modems, call Marshall Industries. At Marshall, we're dedicated to customer service. When you call, we'll quickly find
the part you need with our extensive inventory tracking system, then speed your order to you by shipping same day.

So if you need to communicate at rates up to 14,400 bits-persecond, and need to do it in a hurry, call Marshall today.

Where you'll always get speedy service, but not a lot of fast talk.

\section*{Marshal}

\section*{CIRCLE NO 75}

MN Minneapolis (612) 559-2211* MO St. Louis (314) 291-4650 NC Raleigh (919) 878-9882*
NJ N. New Jersey (201) 882-0320* Philadelphia (609) 234-9100*
NY Binghamton (607) 798-1611* Long Island (516) 273-2424

Brownsville (512) 542-4589* Dallas (214) 233-5200 El Paso (915) 593-0706* Houston (713) 895-9200 San Antonio (512) 734-5100* UT Salt Lake City (801) 485-1551 WA Seattle (206) 747-9100*

\section*{Integrated Circuits}

\section*{Nonvolatile flash memories offer l-sec chip-erase times}

The 27 F 256 and 28 F 256 are 256 k bit, nonvolatile, read/write flash memories. The chips, which are based on a proprietary ETOX (EPROM tunnel oxide) process, essentially add electrical chip-erasure and reprogramming capabilities to EPROM technology. The vendor's Quick-Erase algorithm typically erases all the memory cells simultaneously in 1 sec. The Quick-Pulse Programming algorithm reprograms the flash memories in less than 4 sec . The 28 -pin 27 F 256 multiplexes the write-enable line and the A14 address line to maintain pin compatibility with standard EPROMs. The 32 -pin 28 F 256 has a nonmultiplexed pinout, which eases the interface to the CPU and accommodates as many as 2 M bytes of memory. The CMOS devices offer \(170-\) nsec max access time and

\(100-\mu \mathrm{A}\) max standby current. The devices also exhibit a failure rate of less than \(0.01 \%\) for 100 erase/program cycles. They sell for \(\$ 19.90\) \((10,000)\).

Intel Corp, Literature Dept, W-424, 3065 Bowers Ave, Santa Clara, CA 95051. Phone (800) 5484725.

\title{
Printer/display processor features \(32000 \mu \mathrm{P}\) instruction set
}

The NS32CG16 is a member of the Series 32000 processor family. This 32 -bit CMOS chip, which is designed for advanced printer and display applications, runs the 32000 instruction set plus 11 specific graphics instructions. The processor can perform bit-block transfers (BitBlts) with 6 on-chip logical operators. When supported by a DP8510 BitBlt-processing unit, the chip can transfer data at 18 M bps and can transfer a matrix of \(32 \times 54\) characters at 4700 characters/sec.

A \(15-\mathrm{MHz}\) version of the processor can draw a horizontal line at 9 M bps , a long horizontal line at

60 M bps , and a vertical line at 400 k bps. It can clear memory at 60 M bps. The processor also features a 16M-byte linear-address range; support for high-level languages, such as C, Pascal, Fortran, and Ada; floating-point support for outline fonts; and built-in binary compression or expansion instructions for font storage in RLL format. The chip comes in a 68 -pin PLCC. A \(10-\mathrm{MHz}\) version costs \(\$ 20(10,000)\).
National Semiconductor Corp, Box 58090, Santa Clara, CA 95052. Phone (408) 721-3219. TLX 346353.

Circle No 427


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* Coming soon:

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* Planning

Z8, Super 8, 68HC05, Z280, 80286-16, 68030, V60

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\title{
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}

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The real beauty of the SPI is that it eliminates limitations imposed by microcomputers.

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Right now, we can offer you a versatile family of peripheral devices, including 128 -byte and 256 -byte static RAMs, a real-time clock with RAM, an 8-bit programmable I/O port, and a 10 -bit 8 -channel A/D converter. And more parts are coming soon, including a digital pulse-width modulator and a serial bus interface chip for networking microcomputers. These serial peripherals are also compatible with other microcomputer types.

\section*{Powerful family of micros.}

We can provide 6805 microprocessors for external memory address, but the heart of our SPI system is the 6805 Series high-speed CMOS microcomputers:

\section*{68HC05 Microcomputers}
\begin{tabular}{lllll}
\hline Features & 68HC05C4 & \(68 \mathrm{HCO5C8}\) & \(68 \mathrm{HCO5D2}\) & 68HC05D2A \\
Pins & 40 & 40 & 40 & 28 \\
On-Chip RAM (bytes) & 176 & 176 & 96 & 96 \\
On-Chip User ROM (bytes) & 4160 & 7744 & 2176 & 2176 \\
Bidirectional I/O Lines & 24 & 24 & 28 & 16 \\
Unidirectional I/O Lines & 7 inputs & 7 inputs & 3 inputs & 3 inputs \\
Timer size (bits) & 16 & 16 & 16 & 16 \\
Prescaler size (bits) & \(*\) & \(*\) & \(*\) & \(*\) \\
\begin{tabular}{l} 
External timer oscillator \\
Serial peripheral interface
\end{tabular} & no & yes & no & yes \\
\begin{tabular}{l} 
Serial communications \\
interface
\end{tabular} & yes & yes & yes & no \\
\begin{tabular}{l} 
*prescaler fixed as \(\div 4\)
\end{tabular} & & & no \\
\hline
\end{tabular}

Easy to prototype, too.
If you need another reason to choose our 6805 family, here it is: they're so easy to prototype with our Piggyback! We have the 68EM05C4 and 68EM05D2 Emulators, custom 40-pin packages that contain the C4/C8 or D2 micros with a Piggyback EPROM socket.

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\section*{8k \(\times 8\)-BIT EEPROM}

The XL46HC64 SpeedPROM, an \(8 \mathrm{k} \times 8\)-bit EEPROM with a read access time of 35 nsec, is pin compatible with many bipolar PROMs and EPROMs. When operating at 10 MHz , the chip typically draws 40 mA . You can erase the entire chip in 10 msec and reprogram each byte in 1 msec ; reprogramming the entire 64 k -bit memory takes 8 sec . Users can read and verify each memory location, as well as program a data subset and then test the entire memory. You can also have a signature row of bits coded before shipment with identification data such as product codes or manufacturing lot number. This row does not take up any usable data space. The IOL (low-level output current) specification is 16 mA for TTL compatibility. \(\$ 33\) (100).
Exel Microelectronics Inc, Box 49007, San Jose, CA 95161. Phone (408) 432-0500. FAX 408-434-6444. TWX 910-338-2116.

Circle No 610


\section*{P-CHANNEL FETS}

The VP0610 and TP0610 p-channel models complement the company's 2N700 and 2N7002 enhancementmode MOS FETlington transistors. They allow you to connect complementary pairs of \(n\) - and p-channel

FETlington transistors to drive CMOS logic devices. The transistors have a maximum drain-to-source voltage of 60 V dc and a channel on-resistance of \(10 \Omega\). The VP0610 and TP0610 have thresholds of 3.5 and 2.4 V , respectively, and operate directly from low voltage supplies. Both devices are available in sur-face-mount SOT-23, and hermetically sealed TO-52 and TO-92 packages. From \(\$ 0.44\) to \(\$ 1.61\) (100). Delivery, four to eight weeks ARO.

Siliconix Inc, 2201 Laurelwood Rd, Santa Clara, CA 95054. Phone (800) 554-5565; in CA, (408) 988 8000.

Circle No 611


80C31 HYBRID \(\mu \mathrm{C}\)
The Model C8-P31 hybrid \(\mu \mathrm{C}\) combines an \(80 \mathrm{C} 31 \mu \mathrm{C}, 32 \mathrm{k}\) bytes of EEPROM, 8 k bytes of static RAM, a logical control unit, and a 7.3728 MHz crystal clock oscillator. The totally CMOS hybrid comes in a 40 -pin DIP with standard \(300-\mathrm{mil}\) centers. It operates over the temperature range of -55 to \(+150^{\circ} \mathrm{C}\) and is designed for hostile environments.

The 32 k -byte EEPROM is divided into an 8 k -byte section for code only and a user-defined 24 k -byte section for either code or data. The 8 k bytes of static RAM is for data only. The hybrid \(\mu \mathrm{C}\) can address 64 k bytes of external memory for both data and program for a total of 128 k bytes of external memory. It draws 35 mA from a 5 V supply and has a sleep mode, which consumes \(2 \mathrm{~mA} . \$ 1000\) (100). Delivery, 8 to 10 weeks ARO.

White Technology Inc, 4246 E Wood St, Phoenix, AZ 85040. Phone (602) 437-1520. TWX 910-951-4203.

Circle No 612

\(64 \mathrm{k} \times 16\)-BIT RAM
The EDI8M1664C high-speed CMOS static RAM module contains four \(32 \mathrm{k} \times 8\)-bit RAMs organized as two banks. Each bank is a \(32 \mathrm{k} \times 16\) bit unit that produces a \(64 \mathrm{k} \times 16\)-bit RAM in a single IC. The lower-byte control line selects the lower bytes (DQ0-DQ7), and the upper-byte control line selects the upper bytes (DQ8-DQ15). The unit operates from a 5 V supply, and all inputs and outputs are TTL compatible.

The access times of the commercial and military versions measure 60 and 70 nsec , respectively. A 100nsec access time is also available for both versions. The military version complies with MIL-STD-883C, test method 5004, Class B. The EDI8M1664C comes in a 600 -milwide, 40 -pin DIP. Commercial versions: \(100 \mathrm{nsec}, \$ 249 ; 60 \mathrm{nsec}\), \(\$ 445\). Military versions: 100 nsec, \(\$ 496 ; 70\) nsec, \(\$ 860\) (100). Delivery, six weeks ARO.
Electronic Designs Inc, 42 South St, Hopkinton, MA 01748. Phone (617) 435-2341. TLX 948004.

Circle No 613

\section*{MOTION CONTROLLER}

The LM628/LM629 NMOS motion controller for servo systems lets a host processor communicate with the chip through a parallel 8 -bit asynchronous I/O port. The host generates a trapezoidal velocity profile and programs an on-chip digital-

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compensation filter. An on-chip ve-locity-profile generator then calculates a 32 -bit word for the required trajectory in either a position or a velocity mode of operation.
The quadrature incremental-position feedback signal interfaces to the chip through two quadrature inputs and an index pulse input. A feedback processor calculates a 32 -bit position word and subtracts it from the profile word. The resultant error signal passes through the compensation filter to an 8 -bit output port. The output of the LM628 is designed to drive an 8 -bit DAC; the LM629 has eight PWM outputs for driving H-switches. The state of the feedback encoder can be captured at a \(750-\mathrm{kHz}\) rate. The device requires a \(6-\mathrm{MHz}\) clock and is available in a 28 -pin DIP. \(\$ 30\) ( 100 ).
National Semiconductor Corp, Box 58090, Santa Clara, CA 95052. Phone (408) 749-7421. TLX 346353.

\section*{Circle No 616}


A/D CONVERTER
Exhibiting a conversion rate of 500 kHz , the AM40016/AM40116 16-bit A/D converter has a built-in S/H amplifier. The AM40016 is available with 0 to 10 V unipolar inputs, and the AM40116 comes with \(\pm 10 \mathrm{~V}\) bipolar inputs. The unit features a \(10^{8} \Omega\) input impedance, a differential linearity of \(\pm 3 / 4 \mathrm{LSB}\), an integral linearity of \(\pm 0.003 \%\), a dc offset error of \(\pm 5 \mathrm{mV}\) max, and shielding against electromagnetic and electrostatic interference. The selectable coding formats for the digital outputs are binary, offset binary, or 2's complement. The converter requires \(\pm 15,5\), and -6 V supplies,
and operates in the 0 to \(60^{\circ} \mathrm{C}\) temperature range. It comes in a \(3 \times 5 \times 0.44-\mathrm{in}\). module that is pin compatible with the ADAM826-1 converter. \(\$ 849\) (100). Delivery, 12 weeks ARO.
Analogic Data Conversion Products Group, 360 Audubon Rd, Wakefield, MA 01880. Phone (617) 246-0300. TLX 949307.

Circle No 615

\section*{LASER DRIVER}

The ALD30010 laser-diode current driver has a frequency response from 10 kHz to 3 GHz . It features an on-chip \(180^{\circ}\) phase splitter that permits single input drives. The chip can deliver 60 mA min of peak bias current. The voltage-controlled input can modulate the current over a 0 - to \(30-\mathrm{mA}-\mathrm{min}\) range. An input signal level of 0 dBm achieves the maximum modulation current. An input amplifier has an 8-dBM thirdorder intercept point and a \(1-\mathrm{dB}\) compression point of 1 dBm . The large-signal rise and fall time is 100 psec typ with a typical propagation delay of 100 psec . The input return loss is 20 dB typ, and the reverse isolation is 30 dB typ. The device requires a -7 V supply and a \(60-\mathrm{mA}\) supply current. 8-pin metal-ceramic flatpack, \(\$ 65\); die, \(\$ 43.50\) (1000). Delivery, stock to 90 days.

Anadigics Inc, 35 Technology Dr, Warren, NJ 07060. Phone (201) 668-5000. TWX 510-600-5741. FAX 201-668-5068.

Circle No 622

\section*{6-BIT ATTENUATOR}

The CDG4460J 6-bit digitally controlled attenuator can attenuate analog signals with frequencies as high as 40 MHz . It consists of 3 CMOS/D-MOS integrated circuits and 11 laser-trimmed resistors mounted on a thick-film ceramic substrate. The attenuation range is 0 to 15.75 dB in \(0.25-\mathrm{db}\) increments. When operating in a \(75 \Omega\) system, the unit displays constant input and

output resistance over the entire resistance range. An onboard data latch lets you set predetermined attenuation settings. It operates with power supplies ranging from \(\pm 6\) to \(\pm 15 \mathrm{~V}\) and dissipates \(0.5 \mu \mathrm{~W}\) typ. The 16 -pin hybrid package occupies \(1 \mathrm{sq} \mathrm{in} .\mathrm{of} \mathrm{board} \mathrm{space} \$\).39.50 .
Topaz Semiconductor, 1971 N Capitol Ave, San Jose, CA 95132. Phone (408) 942-9100. TWX 910-338-0025.

Circle No 614

\section*{CMOS GATE ARRAYS}

RVG CMOS gate arrays incorporate rad hardening and have 5670 to 20,440 2-input gates. Representative arrays include the 5670 -gate RVG5, the 10,360 -gate RVG10, the 14,640-gate RVG15, and the 20,440gate RVG20. The 2-input NAND gate has a delay of 0.95 nsec with a fan-out of 2; its typical power dissipation is only \(8 \mu \mathrm{~W} / \mathrm{MHz}\). The gate arrays feature symmetrical switching and edge delays, operate at 250 MHz flip-flop frequencies, and are TTL/CMOS compatible. Each I/O interface includes protection circuitry for a 2000 V electrostatic discharge and is user programmable as an input, output, or bidirectional signal connection. You can select from an extensive macrocell library of SSI, MSI, and LSI functions. Military and commercial NRE (nonrecurring engineering) costs, from \(\$ 35,000\); military devices, from \(\$ 150\) (1000/year); commercial devices, from \(\$ 65\) ( \(10,000 /\) year).
Raytheon Co, Semiconductor Div, 350 Ellis St, Mountain View, CA 94043. Phone (415) 968-9211.

Circle No 636


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\section*{DIGITAL CONVERTERS}

The 278A300 series digital converters translate the outputs of LVDT (linear variable differential transformer) and RVDT (rotary variable differential transformer) transducers into an 11-bit, offset-binary digital code. They employ a type II servo loop to generate an error signal between the reference input and transducer inputs. The reference frequency ranges from 400 Hz to 10 kHz .

A change equivalent to 1 LSB in the input signal initiates a conversion; the converters then issue a \(1-\mu\) sec output busy command when the output code changes. An inhibit line provides a simple method of transferring the data to a computer. The inhibit command locks an internal up-down counter for reading. During a busy interval, the converters ignore the inhibit command. An overrange signal indicates whether the input signal exceeds the normal range. The units come in \(2 \times 2 \times 0.395-\mathrm{in}\). modules and require \(\pm 15\) and 5 V . \(\$ 230\). Delivery, stock to 10 weeks ARO.
Control Sciences Inc, 9509 Vassar Ave, Chatsworth, CA 91311. Phone (818) 709-5510.

Circle No 617

\section*{MOS IGBT MODULES}

Based on the company's high-voltage MOS IGBTs (MOS Insulated Gate Bipolar Transistors), the MOSBlock IGBT-module family consists of dual IGBTs in a phase-leg configuration with internal fast-recovery diodes across each IGBT. The continuous current ratings for the modules range from 25 to 100 A in 600 and 1000 V versions. The peak
pulsed ( 1 msec ) current ratings range as high as 200 A . They are capable of switching rates as high as 30 kHz . The modules come in standard Japanese-style bipolar Darlington packages in two outlines: \(94 \times 34\) and \(95 \times 62 \mathrm{~mm}\). They exhibit a typical \(\mathrm{V}_{\text {CE(SAT) }}\) of 3.2 volts at a current of 50 A . The current fall time is 800 nsec max for all devices. \(25 \mathrm{~A}, 600 \mathrm{~V}\) device, \(\$ 45.04\); 100A, 1000 V device, \(\$ 129.41\) (100).
IXYS Corp, 2355 Zanker Rd, San Jose, CA 95131. Phone (408) 4351900.

\section*{Circle No 623}

\section*{RF AMPLIFIER}

The LH4200 RF amplifier can operate over the \(500-\mathrm{kHz}\) to \(1-\mathrm{GHz}\) frequency range. It utilizes a GaAs input stage for high-frequency performance followed by two bipolar transistor stages. You can use series feedback for gain stabilization and control the gain via a control input. The amplifier features a \(38-\mathrm{dB}\) gain at 100 MHz , an AGC range of 60 dB at 100 MHz , an input impedance of 1 \(\mathrm{M} \Omega\), and a noise figure of 3 dB for a \(50 \Omega\) source impedance. The gain falls to 0 dB at 1 GHz . It has a typical 1-dB gain compression point of 14 dBm when operating at 100 MHz . The amplifier has internal bypass capacitors; however, it is recommended that an external capacitor of \(10 \mu \mathrm{~F}\) be used to prevent low-frequency instabilities. The device comes in a 24 -pin ceramic DIP. \(\$ 54\) (100).

National Semiconductor Corp, Box 58090, Santa Clara, CA 95052. Phone (408) 749-7421. TLX 346353.

Circle No 627

\section*{FLASH CONVERTER}

The ADC-303 \(100-\mathrm{MHz}\) flash A/D converter has an input-signal bandwidth of 40 MHz and operates on analog input signals in the 0 to -2 V range when the reference voltage is set at -2 V . The digital inputs and outputs are ECL compatible. The


8-bit parallel outputs are buffered and have open-emitter arrangements. The chip dissipates 1.2 W and has an integral and differential linearity of \(\pm 1 / 2\) LSB over its operating range. You can select one of the following output codes: binary, complementary binary, 2's complement, and complementary 2's complement. The chip operates with an external clock and reference source, and comes in a 42 -pin DIP. Its operating temperature ranges from -20 to \(+100^{\circ} \mathrm{C} . \$ 550\).
Datel Inc, 11 Cabot Blvd, Mansfield, MA 02048. Phone (617) 3393000. TLX 951340.

Circle No 618

\section*{MOTOR-CONTROL ICs}

Three custom ICs are available for designing proprietary motion-control systems. The MC3A is a 16 -bit internal, 8 -bit external, data-bus \(\mu \mathrm{P}\). It has 8 k bytes of mask-programmed ROM that holds the servo algorithms and an abbreviated command set for minimum-performance systems. Working with the MC3A is the Motion-LSI chip, which has four channels of \(\mathrm{A} / \mathrm{B}\) quadrature encoder decoding. Internal registers permit software selection of various encoder operating modes.
Six digital outputs from the Mo-tion-LSI chip provide PWM signals for the control of 2-, 3-, or 4-phase brushless dc, ac-induction, or stepper motors. The SMCC chip is a 27256 EPROM with a 12 k -byte program that extends the command set beyond that contained in the MC3A \(\mu \mathrm{P}\). The 27256 chip comes with the first trial chip set. Thereafter, you must make your own duplications.

\title{
NEW 500 MEGASAMPLES/SEC 300 MHz DIGITAL PORTABLE
}
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1-800-426-2200.


MC3A \(\mu \mathrm{P}\) and Motion-LSI chip, \(\$ 210\) (100).

Delta Tau Data Systems, 21119 Osborne St, Canoga Park, CA 91304. Phone (818) 998-2095.

Circle No 648


\section*{AMPLIFIER}

Operating over the 2 - to \(6-\mathrm{GHz}\) frequency range, the HMM-10620 GaAs monolithic microwave IC amplifier runs from an 8.5 V supply and draws a maximum of \(100 \mathrm{~mA}(65 \mathrm{~mA}\) typ). Two FET gain stages with negative feedback provide 11.5 db of gain and a gain flatness specification of \(\pm 0.5 \mathrm{~dB}\) over the frequency range. The 1-dB gain compression point is +13 dB typ, and it has a typical noise figure of 5.5 dB . Input and output VSWRs are specified at 1.75:1 max. The device's internal dc blocking lets you cascade it with other units. The chip measures \(37 \times 70\) mils and thermal-compression wedge bonding is recommended. \(\$ 45\) (1000).
Harris Microwave Semiconductor Inc, 1530 McCarthy Blvd, Milpitas, CA 95035. Phone (408) 433-2222. TWX 910-338-2247. FAX 408-432-3268.
\[
\text { Circle No } 619
\]

\section*{450V OP AMP}

The PA85 op amp operates with power supplies ranging from \(\pm 15\) to \(\pm 225 \mathrm{~V}\). It can also operate from a single supply ranging as high as 450 V . The output can swing \(430 \mathrm{~V}_{\text {p-p }}\) with a slew rate of \(1000 \mathrm{~V} / \mu \mathrm{sec}\). It can deliver as much as \(\pm 200 \mathrm{~mA}\) of continuous current. A built-in thermal shut-off circuit prevents overheating and the safe operating area
has no secondary breakdown limitations. A dual J-FET input stage provides a differential offset voltage of \(\pm 2 \mathrm{mV}\) max and a differential offset current of \(\pm 100 \mathrm{pA}\) max. The device achieves a full power bandwidth of 500 kHz with a \(3.3-\mathrm{pF}\) compensation capacitor. It comes in an 8-pin TO-3 package and dissipates 35 W at the current rating. PA85, \$99.50; PA85A improved offset version, \(\$ 129.35\) (100). Delivery, 10 weeks ARO.
Apex Microtechnology Corp, 5980 N Shannnon Rd, Tucson, AZ 85741. Phone (800) 421-1865.

Circle No 629

\section*{VOLTAGE REFERENCE}

The Ref-43 precision voltage reference provides a 2.5 V output with a maximum tolerance of \(\pm 0.05 \%\). It operates from a supply voltage of 4.5 to 40 V , and the output voltage changes less than \(178 \mu \mathrm{~V}\) over the supply range. It can provide 10 mA of output current with more than a \(10 \mathrm{ppm} / \mathrm{mA}\) load regulation. It requires \(450-\mu \mathrm{A}\)-max quiescent current, thus minimizing drift due to self-heating. The temperature coefficient is specified as \(10 \mathrm{ppm} /{ }^{\circ} \mathrm{C}\) max. The unit also has a temperature output that provides an analog output sensitivity of \(1.9 \mathrm{mV} /{ }^{\circ} \mathrm{C}\). It comes in an extended industrial temperature range of -40 to \(+85^{\circ} \mathrm{C}\), as well as military versions. MIL-STD-883 versions will be available in the 3rd quarter of 1988. Ref-43GP, from \(\$ 3.75\) (100).
Precision Monolithics Inc, Box 58020, Santa Clara, CA 95052. Phone (408) 562-7346. TLX 713719541.

Circle No 630

\section*{OP AMPS}

The LT1078 dual and LT1079 quad op amps can run from a 5 V supply or from dual supplies ranging as high as \(\pm 15 \mathrm{~V}\). They feature a \(50-\mu \mathrm{A}\)-max supply current for each amplifier, an offset voltage of \(70 \mu \mathrm{~V} \max\), a


250-pA-max offset current, 0.4 \(\mu \mathrm{V} /{ }^{\circ} \mathrm{C}\) offset voltage drift, a 200 kHz gain-bandwidth product, and a \(0.07 \mathrm{~V} / \mu \mathrm{sec}\) slew rate. The \(1 / \mathrm{f}\) corner of the voltage noise spectrum is 0.7 Hz . This low frequency corner permits a voltage noise specification of \(0.6 \mu \mathrm{~V}_{p-\mathrm{p}}\) and a current noise specification of \(3 \mathrm{pA}_{p-p}\) over the \(0.1-\mathrm{Hz}\) to \(10-\mathrm{Hz}\) frequency range. The outputs can source and sink 5 mA of load current. LT1078, \$2.80; LT1079, \(\$ 3.50\) (100).

Linear Technology Corp, 1630 McCarthy Blvd, Milpitas, CA 95035. Phone (800) 637-5545; in CA, (408) 432-1900.

Circle No 620


\section*{SWITCH DEBOUNCER}

The PSI-01 IC interfaces electromechanical switches with digital circuits. It provides switch debouncing without any external components for as many as eight independent spdt switches. The reading is 32 msec both for the typical output delay after activating a switch, and for the maximum output delay after releasing a switch. The chip provides either a strobe output pulse or a toggled output that are user selectable. The outputs are TTL compatible and can source and sink 4 mA . The chip operates on voltage supplies ranging from 3 to 25 V with a supply current drain of 5 mA . The

\section*{Who offers you the broadest line of power Darlington FIMCD MOSBP SCRAliode modules, rectiliers and thyristors?}

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Recent additions to the product line are a family of Center Tap fast recovery diodes rated at 20-100 A up to 1200 V , a new compact \(150 \mathrm{~A} / 1600 \mathrm{~V}\) Dual thyristor module, and a new \(1200 \mathrm{~V} / 300\) A GTO thyristor module.

Crily Pow Inazk offers you all this. For product literature, call POWEREX at 1-800-451-1415, Ext. 300. (In New York, 315-4579334.) For application assistance, call 412-925-7272, or write POWEREX, Inc., Hillis Street, Youngwood, PA 15697.
chip comes in a 20 -pin DIP. \(\$ 2\) (1000).

Precision MicroDevices Inc, 711 Charcot Ave, San Jose, CA 95131. Phone (408) 432-3077.

Circle No 621


RISC \(\mu \mathbf{P}\)
The S-25, a member of the Sparc 32-bit RISC \(\mu\) P family, achieves an average sustained processing rate of 15 MIPS at a clock rate of 25 MHz . A register file, consisting of 120 32 -bit registers, stores frequently used variables to reduce program execution overhead. This triple-port register file can fetch two operands and write to a destination simultaneously in a single cycle. The chip has separate 32 -bit address and data buses; it can directly address 4G bytes and can indirectly address 256 pages of 4 G bytes.

The instruction set has five operator categories: load and store, arithmetic/logical/shift, control transfer, read/write control registers, and floating-point and coprocessor functions. It uses a single-length format and fixed field positions. Optimizing C, Fortran, and Pascal compilers are available for software development. The chip comes in a 179 -pin pin-grid-array package. \(\$ 325\) (5000).

Fujitsu Microelectronics Inc, Advanced Products Div, 50 Rio Robles, San Jose, CA 95134. Phone (408) 922-9649.

Circle No 624

\section*{SMPS ICs}

Operating on a master-slave principle, the TEA 2170 and TEA 2164 ICs provide all the regulation and
protection functions required in isolated switchmode power supplies with output powers as high as 200 W . They are optimized for use in low-cost power supplies for consumer equipment such as TV receivers, video recorders, and audio equipment. The devices allow you to isolate the secondary side of the supply from the line input with a low-cost transformer.

The TEA 2164 's \(\pm 1.5 \mathrm{~A}\) output can directly drive power transistors at switching frequencies as high as 60 kHz . In critical applications, you can synchronize the switching to an external frequency. The TEA 2164 contains oscillators, a start/stop pulse shaper, a switching transistor driver, and a safety processor, as well as current-limit \(V_{C C}\) supervision and repeated overcurrent detection circuitry. The TEA 2170 contains an error amplifier, a voltage reference, a master oscillator with synchronization input, a pulsewidth modulator, and a pulse driver for the isolation transformer. Softstart circuitry is also included. Approximately \(\$ 1.46\) (1000).

SGS-Thomson Microelectronics, Via C Olivetti 2, 20041 Agrate Brianza, Italy. Phone (039) 65551. TLX 330131.

Circle No 650
SGS-Thomson Microelectronics, 1000 E Bell Rd, Phoenix, AZ 85022. Phone (602) 867-6100. TLX 249976.

Circle No 651

\section*{OPTOCOUPLER}

The SFH620 optocoupler contains two back-to-back GaAs infrared emitters so that it can remain in the On state during both the positive and negative half-cycles of an acinput waveform. The device is tested to an isolation voltage of 5.3 kV and incorporates a special ion screen that provides added protection against electrostatic field effects. The optocoupler's emitter and detector have an internal separation of 0.8 mm , allowing the company to

seek approval for the device to VDE-0884 safe electrical-operation specifications. The SFH620 is subjected to in-process testing and burn-in before being shipped. DM \(0.79(10,000)\).

Siemens AG, Zentralstelle für Information, Postfach 103, 8000 Munich 1, West Germany. Phone (089) 2340. TLX 5210025.

Circle No 652
Siemens Components Inc, 2191 Laurelwood Rd, Santa Clara, CA 95054. Phone (408) 980-4500.

Circle No 653

\section*{ANALOG MULTIPLEXERS}

Fabricated in CMOS, the HI-506A, HI-507A, HI-508A, and HI-509A are analog muliplexers that have transfer accuracies of more than \(0.1 \%\) at sampling rates to 200 kHz . The analog inputs of the devices can withstand an overvoltage to \(70 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}\), and feature break-before-make switching. The four multiplexer models each provide the following number of channels: HI-506A, 16 single-ended; HI-507A, 8 differential; HI-508A, 8 single-ended; and HI-509A, 4 differential channels.

The input signal range for all devices is \(\pm 15 \mathrm{~V}\), and crosstalk is limited to \(0.005 \%\) of the voltage level when the channel is off. The multiplexers are packaged in ceramic or plastic DIPs, specified for 0 to \(75^{\circ} \mathrm{C}\) or -55 to \(+125^{\circ} \mathrm{C}\). Prices start at \(\$ 13.50\) for the HI-506A/HI-507A, and \(\$ 7.35\) for the HI-508A/HI-509A (100).

Burr-Brown, Box 11400, Tucson, AZ 85734. Phone (602) 746-1111. TWX 910-952-1111.

Circle No 649

\section*{look, listen.}


FFT: Detailed distribution of noise and harmonic distortion vs. input signal. Measurement of sample and hold and ZAD2716 A/D converter only. Computer analysis receives digital information from A/D converter (to avoid DAC reconstruction distortion).


Using an Audio Precision system, a sine wave is digitized by the ZAD-2716, then reconstructed by the ZDA-1801, and filtered by Apogee's 944 S linear phase filter. Graph shows combined total harmonic distortion and noise vs. frequency.

\section*{Introducing the world's finest professional 16-bit stereo digitizing subsystem and 18-bit deglitched reconstruction DAC.}

Want to position yourself at the leading edge of digital audio? Then hear this.
Analog Solutions gives you a new family of professional audio products with confirmed "Golden Ear" performance using the industry's two most powerful test criteria. The eye. And the ear.

If seeing is believing, you'll be impressed by the harmonic distortion and noise plots of our ZAD2716/ZSH202 digitizing subsystem (left) and ZDA1801 deglitched reconstruction DAC (right). Nobody's ever seen plots like those before.

But there's more to the story than a good plot.
The ZAD2716-2 coupled with the ZSH202 Dual Sample-Hold can provide stereo digitizing at a 50 kHz sampling rate per channel. The ZSH202 allows simultaneous sampling of both channels within \(\pm 2\) nsec with guaranteed \(-96 \mathrm{db} /\) channel separation. And there's no sacrifice of phase linearity or \(\sin x / x\). In fact, we guarantee \(\sin x / x\) performance of 0.01 dB and phase linearity of 0.1 (degree) over 0 to 20 kHz . (Performance that brings a smile to any Golden Ear.) Cost savings are significant on a per channel basis, because you only need one converter for two channels.

Our ADC is also available in a ZAD2716-1 model with integral sample and hold amplifier and 100 kHz sampling speed for 2X oversampling applications.


Your needs extend over the entire audio spectrum. And that's how our products are tested. We test the way they're used. So we can guarantee signal-to-noise plus harmonic distortion of 92 dB over the entire \(0-20 \mathrm{kHz}\) audio range.

We maintain the same outstanding performance at the back end of your system with the ZDA1801. This 18-bit DAC, with a built-in deglitcher operating up to 200 kHz for up to 4 X oversampling, offers guaranteed signal-to-noise plus harmonic distortion of 94 dB over the full audio range.

In short, we've raised digital audio to new levels.
Which should be music to your ears.
And to every Golden Ear in the business.
Find out more about our growing line of professional audio products. And learn how we can help plot your success in a growing industry.

Please address Analog Solutions, 85 West Tasman Drive, San Jose, CA 95134.

A subsidiary of Silicon General, Inc. Phone (408) 433-1900. FAX (408) 433-9308.

CIRCLE NO 84

\section*{Integrated Circuits}


AUDIO PREAMP
Suitable for consumer audio equipment, the TEA 6300 audio preamplifier incorporates a signal source selector; volume, bass, treble, and stereo balance controls; and a quad fader to control the front and rear speaker amplifiers in surroundsound systems. A break in the signal path between the signal selector and the preamplifier's control circuits allows you to insert additional signal conditioning circuitry-that
is, a compander, a booster amplifier, an equalizer, or a noise-reduction circuit.

The preamplifier's built-in IIC Bus interface allows you to digitally control the preamplifier from a microcontroller. You can adjust the volume and fader in \(2-\mathrm{dB}\) steps, and the bass and treble controls in \(3-\mathrm{dB}\) steps. The device doesn't generate any audible noise when you adjust its controls. The preamplifier also incorporates a mute function.

The preamplifier has a frequency response from de to more than 20 kHz , an overall gain of 20 dB , and an input sensitivity for full output of 50 mV typ. It has a signal-to-noise ratio of 80 dB , a typical channel separation of 70 dB , and typically introduces \(0.05 \%\) total harmonic distortion at full output power. The TEA 6300 operates from a 7 to 13.2 V supply and is housed in a 28 -pin DIP or a surface-mount minipack. Approximately gld 8.50
\((10,000)\). The TEA 6310 , a similar device without the source selector, but with a fader disable input, is also available.
Philips, Components Division, Box 523, 5600 AM Eindhoven, The Netherlands. Phone (040) 757189. TLX 51573.

Circle No 654
Signetics Corp, 811 E Arques Ave, Sunnyvale, CA 94088. Phone (408) 991-4571.

Circle No 655

\section*{A/D ASICs}

A library of 66 analog cells, 66 digital macrocells, and 28 I/O cells supports the TSGSM Series mixed A/D standard-cell ASICs. High-level analog cells in the library include 8and 12 -bit A/D converters, 8 - and 12-bit D/A converters, twelfth-order switched-capacitor filters, \(120-\) nsec comparators, and LCD driver circuitry. Where necessary, you can

II do for color computing what the original Macintosh did for black \& white. Our RAMDAC enables Macintosh II to display some of the finest quality graphics available in a personal computer.

\section*{Integrated Circuits}
bias circuits from current or voltage bias generators, and include programmable current mirrors. You can also introduce grounded shielding around analog circuitry to improve its power-supply rejection ratio.

The digital cell library contains hard macros, but you can use them to generate your own soft macros for compiling more complex digital devices such as counters, shift registers, or dividers. All the library cells are fully characterized for operation from 3 to 10 V supplies. CAD support for the series comprises the company's TCAD2A software. This CAD package includes H3Spice for analog simulation and allows you to perform mixed A/D simulation by modeling the analog functions. To design switched-capacitor filters, you can use the company's FilCAD software, and then import these filter designs into TCAD2A. The typical NRE (nonrecurring engineer-
ing) charges are \(\$ 55,000\). Depending on die size, from \(\$ 1\) to \(\$ 8\) (OEM qty).

SGS-Thomson Microelectronics, Via C Olivetti 2, 20041 Agrate Brianza, Italy. Phone (039) 65551. TLX 330131.

Circle No 659
SGS-Thomson Microelectronics, 1000 E Bell Rd, Phoenix, AZ 85022. Phone (602) 867-6100. TLX 249976.

Circle No 660


\section*{QUAD OP AMP}

The CA5470 BiMOS-E general-purpose quad op amp combines highspeed CMOS and bipolar transistors
on a single chip. It is specified for 5 V single-supply and \(\pm 7.5 \mathrm{~V}\) dualsupply operation over the military temperature range of -55 to \(+125^{\circ} \mathrm{C}\). You can operate the unit from supplies ranging from 3 to 16 V unipolar or from \(\pm 1.5\) to \(\pm 8 \mathrm{~V}\) bipolar. It has a unity-gain-bandwidth product of 12 MHz and a slew rate of \(5 \mathrm{~V} / \mu \mathrm{sec}\). The amplifier has protection against ESD voltages as high as 2000 V . The unit comes in a 14 -lead plastic surface-mount package or plastic DIP. Plastic DIP, \(\$ 1.21\) (1000).

GE Solid State, Rte 202, Somerville, NJ 08876. Phone (201) 6856652.

\section*{INQUIRE DIRECT}

\section*{DSP CHIP}

The PDSP16116 complex multiplier can multiply together two 16 -bit complex words to produce a 32 -bit complex-word result every 100 nsec.

Text continued on pg 145

\title{
Brooktree
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\section*{SONAR}

\section*{THE IMS AIOO CASCADEABLE SIGNAL PROCESSOR}

Using pulse compression to improve range resolution, a modern radar system compares signals transmitted to and reflected from a target.

A typical system may use a pattern 512 samples long, collected at a data rate of 2.5 MHz .

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Yet the INMOS A100 Cascadeable Signal Processor enables a pulse compressor with this capability to fit a circuit board half the size of a typical PC card and dissipates less than 30Watts.

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Ordinary programmable logic devices (PLDs) solve problems, but they create some as well. That's why ICT combined CMOS and \(\mathrm{E}^{2}\) technologies to make PEEL \({ }^{\text {m }}\) (Programmable Electrically Erasable Logic) devices.PEEL devices elude the perils of PLD.

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\begin{tabular}{lllll}
\hline PEEL Device & Pins & TPD & ICC \(^{* 2}\) & Supersets \\
\hline PEEL18CV8*1 & 20 & \(25 \mathrm{~ns}, 15 \mathrm{~ns}\) & \(20 \mathrm{~mA}, 60 \mathrm{~mA}\) & PAL, EPLD \\
\hline PEEL20CG10 & 24 & 25 ns & 45 mA & PAL, EPLD \\
\hline PEEL22CV10*1 & 24 & 25 ns & 45 mA & PAL, EPLD \\
\hline PEEL153/253 & 20 & 30 ns & 35 mA & FPLA \\
\hline PEEL173/273 & 24 & 30 ns & 35 mA & FPLA \\
\hline
\end{tabular}
\({ }^{*}\) Z Zero Power versions also \(\quad{ }^{*}\) Standby, add \(\sim 0.7 \mathrm{~mA} / \mathrm{MHz}\) for active

All ICT products are exclusively distributed through Marshall Industries.

\section*{"Horror-Story Inventory"}

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INTERNATIONAL CMOS TECHNOLOGY, INC.

\section*{Integrated Circuits}

You can use the device in conjunction with two of the company's PDSP1601 ALUs, and two of its PDSP16316 complex accumulators to produce a block floating-point butterfly processor for evaluating FFTs. In combination with a single PDSP16316, you can create a \(10-\mathrm{MHz} 16 \times 16\)-bit complex multipli-er-accumulator.

Internally, the PDSP16116 contains four \(16 \times 16\) array multipliers, two 32 -bit adder/subtractors, and control logic that provides fully automatic block floating-point operation. The device also includes trap logic to prevent \(-1 \times-1\) products from leading to erroneous results. In addition, you can cause the device to complex conjugate either input to ease the implementation of complex correlators. A single pipeline delay through the device makes it suitable for use with recursive algorithms. The PDSP16116 is available in both military and industrial temperature ranges. Industrial version, £331.43 (100).

Plessey Semiconductors Ltd, Cheney Manor, Swindon, Wiltshire SN2 2QW, UK. Phone (0793) 36251. TLX 449637.

Circle No 656
Plessey Semiconductors, 9 Parker, Irvine, CA 92718. Phone (714) 472-0303.

Circle No 657


LOGIC ARRAY
The GAL39V18 program-mable-logic array is designed with electrically erasable CMOS technology. It provides 10 output-logic macro cells along with 8 internal
state-logic macro cells. In addition, it has 10 input logic cells and 10 I/O logic macro cells. The true or complement macro-cell outputs feed an AND array consisting of 75 product terms. Sixty-four of these terms serve as inputs to an OR array with 36 sum-term outputs. You can program all of the input and I/O signals into the array as registered, latched, or direct inputs. The regis-
ters are D-type flip-flops with separate clock enables and common or separate clocks. Total delay through the device is 30 nsec max and the clock-to-output delay is 15 nsec max. 24-pin plastic DIP, \(\$ 21.10\) (100).

Lattice Semiconductor Corp, 5555 NE Moore Ct, Hillsboro, OR 97124. Phone (505) 681-0118.

Circle No 631


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Amperex Electronic Company, A Division of North American Philips Corporation, George Washington Highway, Smithfield, RI 02917, (401) 232-0500. In Canada: Philips Electronics Ltd., 601 Milner Ave., Scarborough, M1B 1MB, (416) 292-5161.
}

\section*{Integrated Circuits}


\section*{DATA MANAGER}

The Am95C85 Content Addressable Data Manager (CADM) is a CMOS peripheral chip for sorting and manipulating data, and stores data in an internal 1 k -byte memory. The stored data is collated into records consisting of a key field and a pointer field. The chip uses the values in the key field to sort and search records. The content-addressable feature lets the host retrieve data without calculating physical addresses. The chip generates the addresses for memory access, and it
can search an 8-byte field in less than \(10 \mu \mathrm{sec}\). A stack-mode operation lets the user delete records by popping the records out of memory, and lets the user insert records by pushing the records into memory. You can cascade as many as 256 chips for applications that require a large storage area. The CMOS device comes in a 44 -pin plastic leaded chip carrier. \(12-\mathrm{MHz}\) verion, \(\$ 49.20\); \(16-\mathrm{MHz}\) version, \(\$ 66.50\) (100).
Advanced Micro Devices Inc, Box 3453 Sunnyvale, CA 94088. Phone (408) 732-2400.

Circle No 634

\section*{COPROCESSOR}

The VL82C389 message-passing coprocessor for Multibus II systems offloads the task of managing the iPSB (Parallel System Bus) arbitration from the local CPU. It also handles the bus transfers and excep-tion-cycle protocols, and supports

dual-memory systems that coexist with message-passing architectures. The chip can handle all functions of the central service module (CSM).

The CSM features slot identification, arbitration identification, busclock generation, and bus time-out (watchdog) control. The CSM must reside in slot 0 of the Multibus II backplane, and its local microcontroller assigns slot and arbitration IDs to each module. No DIP switches are required. Once the module sets the system configuration, the CSM monitors the iPSB for

is the Most


 Parylene is




\section*{Before you buy ASICs, you should be aware of some of our hang ups.}


Sure, we've got a lot of hang ups at UTMC. And we're proud of every one of them. Because each one shows you our commitment to the military and aerospace marketplace.

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Our four-member UTD \(1.5 \mu\) gatearray family is built using UTMC's patented continuous-column archi-
tecture. The UTD handles design complexities up to 11,000 equivalent gates with auto placement and routing. The rad-hard version of this family, the UTD-R, easily meets total-dose tolerance of \(1 \times 10^{6}\) rads ( Si ).
UTMC's five-member UTB \(3 \mu\) gate arrays are DESC-certified and listed on the MIL-M-38510 JAN QPL. Like the \(1.5 \mu\) family, it's designed to be more efficient than other gate array architectures by reducing the number of wasted transistors common on conventional structures.
Our VAX \({ }^{\circledR}\) based HIGHLAND \({ }^{\circledR}\) Design System is certified as part of our JAN qualification and is available to assist you with design tasks ranging from design concept through layout and test generation. Additionally, our HIGHLAND system supports ASIC design on
popular workstations such as Daisy \({ }^{\circledR}\), Mentor \({ }^{\circledR}\), and Valid \({ }^{\circledR}\). The HIGHLAND workstation tool kits give you the ability to capture and simulate your design in a familiar environment. Or, if you prefer, a UTMC applications engineer can take your concept and create a complete design for you.
So when you're looking for semi custom products to enhance your systems, look to UTMC. We manufacture products specifically for military and aerospace systems. And we've got the hang ups to prove it.

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Now, 19ns settling op amps that survive saturations and shorts.
Comlinear's two new high-speed op amps bring you built-in protection against saturation. Plus simple shortcircuit protection. That means easy solutions for fast input and output amplifiers in systems where signal level or load can't be controlled.

\section*{use as little as \(\mathbf{5 7 m W}\)...}

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\section*{or drive up to \(\pm 100 \mathrm{~mA}\).}

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error conditions such as nonresponding nodes in order to prevent blocked transfers between functional modules. \(\$ 175\) (100).
VLSI Technology Inc, 8375 S River Parkway, Tempe, AZ 85284. Phone (602) 752-6200. FAX 602-7526000.

Circle No 625

\section*{ANALOG DELAY LINE}

Suitable for television, video, radar, or sonar equipment, the WA1101 programmable CCD analog delay line allows you to select signal delays of \(4,5,10\), or 14 clock periods by using a 2 -bit control code. By selecting a suitable clock frequency, you can arrange for the device to delay an analog signal between 100 nsec and 10 msec .
The device has both a signal and reference input that normally result in an output which is inverted with respect to the signal input. If you're willing to accept a slight degradation in linearity, you can interchange the signal and reference inputs to avoid this output inversion. An on-chip S/H circuit reconstitutes the analog waveform at the output. The WA1101 has a typical dynamic range of 68 dB and a TTL-compatible clock input. It draws approximately 15 mA from a 15 V power supply. Commercial-grade device, approximately \(\$ 28\); high-reliability version, approximately \(\$ 65\) (1000).
Walmsley Microsystems Ltd, Aston Science Park, Love Lane, Birmingham B7 4BJ, UK. Phone 021-359 0981. TLX 334535.

Circle No 658

\section*{GRAPHICS CHIP}

The HD63487, known as the MIVAC (Memory Interface and Video Attribute Controller), can control graphics memory, move images to a CRT screen at a rate of 33 M pixels/ sec, and perform horizontal scrolling and zooming. You can use the chip with the company's ACRTC controller and 1 M bit of dynamic


RAM, arranged as \(256 \mathrm{k} \times 4\) bits, to build a graphics system in a space about the size of a business card. Such a system can draw images into its memory at 2 M pixels \(/ \mathrm{sec}\). Adjustable parameters include resolution, number of colors, pixel shift rate, and frame buffer size. The BiCMOS chip supports the hardware window feature of the ACRTC controller. It runs from a 5 V supply and draws 120 mA max of supply current. It comes in a 68 -pin sur-face-mount plastic leaded chip carrier or a 64 -pin plastic DIP. \(\$ 25\) (5000).

Hitachi America Ltd, Semiconductor \& IC Div, 2210 O'Toole Ave, San Jose, CA 95131. Phone (408) 435-8300. TLX 171581.

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\section*{16-BIT MULTIPLIER}

The \(\mu\) IC16MP high-speed parallel multiplier produces a 32 -bit product from two 16 -bit operands in less than 20 nsec . The operands may be either 2's complement, unsigned magnitude, or mixed mode. The device, which operates from a single clock that has separate enables for x and y registers, can generate products at a rate of 50 MHz . A proprietary pipehold mode disables the clock enables for both registers and also reduces the current to less than 30 mA at \(50 \mathrm{MHz} .50-\mathrm{MHz}\) version in plastic leaded chip carrier, \(\$ 112\); in a pin-grid-array package, \(\$ 136\) (100).

Micro Integration Corp, 2833 Junction Ave, Suite 209, San Jose, CA 95134. Phone (800) 541-3425; in CA (408) 943-0344.

Circle No 639
Text continued on pg 152
EDN July 7, 1988

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\section*{MOTOR DRIVERS}

In the output stages, the UDN2931/ 2906 series 3 -phase brushless de motor drivers can provide 15 V and \(\pm 2 \mathrm{~A}\) of continuous current. At these ratings, the maximum output saturation voltage ( \(\mathrm{V}_{\mathrm{CE} \text { (SAT }}\) ) is 1.3 V . The peak start-up currents can be as high as \(\pm 3.5 \mathrm{~A}\). The series operates from a 15 V supply to drive 12 V brushless dc motors. The chips contain high-current clamp and flyback diodes to suppress inductive transients.

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Sprague Electric Co, Semiconductor Group, Box 2036, Worcester, MA 01613. Phone (800) 247-2077; in MA, (800) 247-2076. TWX 710-3406304.

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address buses. In addition, it contains a separate 30 -bit instruction address bus and a separate 32 -bit instruction bus.

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Rifa Inc, Box 853904, Richardson, TX 75085. Phone (214) 480 8300. FAX 214-680-1059.

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\section*{CODEC/FILTER}

The M5913 CMOS codec/filter IC provides the \(\mathrm{A} / \mathrm{D}\) and \(\mathrm{D} / \mathrm{A}\) conversion and the transmit and receive filtering required to interface a fullduplex voice circuit to a time-divi-sion-mutiplexed PCM digital telephone system. The device is compatible with AT\&T's D3/D4 standard and with applicable CCITT standards. It has a powersupply rejection ratio of -40 dB from de to 150 kHz . You can operate the codec at either a fixed data-rate

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or in a variable data-rate mode. To ensure the integrity of the PCM highway, the unit contains power-on-reset circuitry and circuitry that permits detection of an interrupted clock. The device operates from \(\pm 5 \mathrm{~V}\) supplies and has a typical active power dissipation of 60 mW . Approximately \(\$ 6\) (1000).

SGS Microelectronica SpA, Via C Olivetti 2, 20041 Agrate Brianza, Italy. Phone (039) 65551. TLX 330131.

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SGS Semiconductor Corp, 1000 E Bell Rd, Phoenix, AZ 85022. Phone (602) 867-6100. TLX 249976.

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Hybrid Systems, 22 Linnell Circle, Suburban Industrial Park, Billerica, MA 01821. Phone (617) 667-8700. TWX 710-347-1575.

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\title{
High-power switching supplies stress efficiency
}

\begin{abstract}
Manufacturers of high-power supplies tend to emphasize efficiency over other performance considerations. Most of these supplies operate at frequencies well below 100 kHz and use half- or fullbridge circuits. Yet some manufacturers are shifting to higher frequencies and to the use of power MOSFETs instead of bipolar transistors.
\end{abstract}

Dave Pryce, Associate Editor

Efficiency is one of the biggest problems manufacturers of high-power switching power supplies ( 500 W and up) face. For this reason as well as others, high-power supplies almost always use half- or full-bridge switching circuits rather than the singletransistor flyback converter or forward converters normally found on low-power supplies. And most of them operate at frequencies between 20 and 80 kHz and use the traditional bipolar transistors. Furthermore, although power MOSFETs are used in some high-power switchers because of their simpler drive requirements and their ability to operate in parallel, their fast switching times offer no advantage at low frequencies. The typical 20 - to 30 -nsec fall time of a power MOSFET is better utilized at switching frequencies in the range of 300 to 500 kHz .
Basically manufacturers of high-power supplies, which are used for products like superminicomputers, telecommunications devices, and laboratory and automated industrial equipment, don't gain much by reducing the size and weight of their products because portability is far less important in such applications than it is in low-power PC applications.
 egory generally use half-bridge or fullbridge switching circuits.


A 600 W supply that uses a separate assembly for a \(150-\mathrm{kHz}\) converter (Kepco Inc)

One benefit of using bridge configurations in highpower switchers is that the corresponding transistors need only about half the blocking rating ( \(\mathrm{V}_{\mathrm{CEV}}\) ) that's required for the transistor in a flyback or forward converter. This characteristic is particularly important for supplies operating from 220 V ac lines because a flyback converter needs a transistor with a \(\mathrm{V}_{\text {CEV }}\) rating greater than 850 V . Bridge circuits, on the other hand, turn on and off from the de bus and never see more than the peak line voltage. In these latter circuits, \(\mathrm{V}_{\text {CEO }}\) is the important transistor parameter-a 450 V rating is sufficient.

Bridge circuits also offer other advantages: The primary winding of the transformer is driven in both directions, which results in better transformer-core utilization and makes the use of a full-wave output filter feasible. And, because the input-filter capacitors are located in series across the rectified 220 V line, you can use them to double the voltage on a 120 V line. In either case, the transformer continues to work from a nominal 320 V bus. Finally, a bridge circuit allows the use of diode clamps across each transistor to minimize switching transients.

Qualidyne Systems has found that bridge circuits bolster the efficiency of its 400 to 1500 W modular power supplies considerably. The bridge circuits not only enhance core utilization, they also facilitate the flow of huge amounts of current through the transformer while minimizing losses in the switching devices. The company has found it much easier to reduce these losses by


Fig 1-Most high-power switching supplies use either a half- or a full-bridge configuration. The half-bridge circuit (a) is the most popular for transistor currents to about 20A. At higher currents, designers usually switch to the full-bridge circuit (b), which nearly doubles the power output. Bridge circuits can use transistors that have about half the breakdown-voltage rating required for flyback or push-pull circuits.


500W open-frame switcher (Todd Products Corp)
operating at lower frequencies; thus, although these units contain FETs, they don't use them to reach a higher frequency. In fact, they operate at only 30 kHz . Qualidyne has also found that using a single turn of copper foil rather than several turns of copper wire for the 5 V high-current main output also reduces losses. Lower frequency combined with the single turn of foil actually decreases both resistance and heat losses.
The Qualidyne designs emphasize efficiency over and above such parameters as dynamic response, regulation, inrush current, and noise and ripple, and they do so for a very good reason. A 2000 W supply that is \(70 \%\) efficient, for example, loses 600 W in heat. An improvement of only \(5 \%\) in the efficiency rating decreases the loss by 100 W -and 100 W , which must be otherwise dissipated in the form of heat, can make a big difference to a circuit designer faced with overheating problems.

Qualidyne offers several series of high-power switching supplies ranging in power from 500 to 2000 W and packaged in various case sizes. The Case- 23 series provides power from as many as 7 fully regulated outputs. They come in low-profile \(3 \times 5 \times 14.25-\mathrm{in}\). cases and typically cost less than \(\$ 0.70 / \mathrm{W}\) (OEM qty). The maximum total power is 600 W . The efficiency rating is typically \(75 \%\). The main channel on these supplies is \(5 \mathrm{~V} / 100 \mathrm{~A}\); other channels, which are limited to a combined total of 240 W , have 12,15 , or 24 V outputs with current ratings in the 5 to 10A range. One model in this series provides a single \(48 \mathrm{~V} / 13 \mathrm{~A}\) output for use in telephone systems. Like most power supplies, all units operate from a selectable 90 to 132 V or 180 to 264 V ac input range. The line and load regulation is \(1 \%\); a typical holdup time is 16 msec . Other features include overload protection, thermal protection, and remote sensing.

Power Components, a subsidiary of Vanguard Electronics, makes custom and semicustom supplies, including a number of semicustom multi-output switchers with power ratings to 600 W for open-frame types and 1000 W for encased supplies. The mechanical design of these supplies permits a standard board to accept TO-3 or TO-220 devices, discrete postregulators, and prewound power transformers. The manufacturer establishes the output ratings during the final assembly stages when it adds the appropriate transformer, postregulators, and a few passive components. If you have an unusual mechanical configuration, the company will generate a new pe board to accommodate it. Power Components uses bipolar transistors that operate at 25 to 40 kHz in either a half-bridge (below 700 W ) or full-bridge configuration.

\section*{Common core moderates magnetic flux}

One unconventional feature of the Vanguard switchers is the use of a single, common core for all the output filter chokes. Because of this construction, the magnetic flux within the core is proportional to the combined power demand from all of the outputs, not just the main output. As a result, the switching-control circuit functions equally well with either a lightly or heavily loaded main output. Transient response also improves because sudden changes in auxiliary loading immediately are reflected to the control circuit. All Power Components switchers meet CSA, UL, and VDE isolation requirements. Instead of using the more common optocouplers, which place the control loop across the input-output isolation boundary, the manufacturer incorporates isolation transformers. A typical 4 -output, 550 W openframe switcher sells for \(\$ 385(1000)\).

Computer Products' Power Conversion Group includes two organizations that participate in the highpower market: Boschert Inc and the group's Power Products Division. Both offer switching supplies in the 500 to 1500 W range that are suitable for applications like minicomputers, CAD/CAE workstations, mainframe hard-disk systems, test equipment, industrial systems, and automated office equipment.

Although the Power Products Division uses \(100-\mathrm{kHz}\) switching frequencies and power MOSFETs for their low-power supplies, their high-power models use bipolar transistors switching at 20 kHz in either a halfbridge or full-bridge circuit. Both single-output and quad-output models are available. Typical of the singleoutput models is the 15 S , a 1500 W unit in a standard \(5 \times 8 \times 11\)-in. case; it sells for \(\$ 1250\). Available output

Bridge circuits reduce the voltage stress on the switching transistors and also provide better transformer-core utilization.


Low-profile 600W switcher (Qualidyne Systems Inc)
ratings include \(5 \mathrm{~V} / 300 \mathrm{~A}, 12 \mathrm{~V} / 125 \mathrm{~A}, 15 \mathrm{~V} / 100 \mathrm{~A}, 24 \mathrm{~V} /\) \(62 \mathrm{~A}, 28 \mathrm{~V} / 53 \mathrm{~A}\), and \(48 \mathrm{~V} / 31 \mathrm{~A}\). Standard features include protection against no load, overload, overvoltage, and thermal overload. The supply also incorporates remote sensing, current sharing, and a soft-start function.

\section*{Small package can pack power, too}

Lambda Electronics, a long-time supplier of both linear and switching power supplies to industrial users and OEMs, has a new series of high-power switchers that pack a lot of power into a relatively small package. The LFS-49, for example, comes in a \(4.875 \times 7.375 \times 11.5-\mathrm{in}\). case and can provide 1500 W . If you can accommodate a slightly longer (12.875-in.) case, the LFS- 50 model can deliver 2000 W . These singleoutput switchers are available with the standard output voltages of \(2,5,12,15,20,24,28\), and 48 V . The LFS-49 and LSF-50 cost \(\$ 900\) and \(\$ 1100\) (100), respectively.

\section*{Innovative circuit boasts high efficiency}

Instead of using bridge circuits for switching, the LFS series uses two alternately conducting forward converters. Each converter operates at 50 kHz and contains a MOSFET driver and two bipolar transistors, which are connected in parallel and which drive the primary of the switching transformer. The secondaries of the two transformers connect to a common full-wave rectifier that doubles the frequency to produce the \(100-\mathrm{kHz}\) switching rate. This dual-circuit configuration requires just a low-voltage MOSFET driver; it also


Fan-cooled 3-kW supply in a \(5 \times 8 \times 15-\mathrm{in}\). package (Powertec Inc)
reduces the operating frequency of the high-voltage bipolar transistors. The efficiency of these supplies attests to the effectiveness of this unusual circuit arrangement: Both the LFS-49 and LFS-50 have a minimum efficiency of \(75 \%\) for their 5 through 15 V models, and \(80 \%\) for their 20 through 48 V models. Line and load regulation is \(0.1 \%\), and the minimum holdup time falls between 16.7 and 20 msec (depending on model). These supplies also feature in-rush current limiting and overvoltage protection.
In what may be the most power yet stuffed into a \(5 \times 8 \times 15-\mathrm{in}\). case, the 9R5-600-381 from Powertec provides a \(5 \mathrm{~V} / 600 \mathrm{~A}\) output at \(50^{\circ} \mathrm{C}\). The \(3-\mathrm{kW}\) supply uses a combination of hybrid and monolithic ICs, low-noise miniature capacitors, and \(100-\mathrm{kHz}\) power MOSFET switches. Other standard models in the 9R series provide outputs of \(2,12,15,24,36\) and 48 V . The supplies operate from a nominal \(220 \mathrm{~V}(165\) to 265 V ) ac input, either single-phase or three-phase. They can also be de-to-dc converters from a 200 to 375 V de input. Standard features include current sharing, soft start, remote sensing, and thermal shutdown. Each switcher contains a ball-bearing dc fan. Four power inverters, which operate at different intervals in time, reduce the ripple current and minimize the stress on the input and output capacitors. Approvals are pending for UL, CSA, IEC, and VDE specifications. The 9R5-600-381 sells for \(\$ 1890\).

Another vendor who is adopting MOSFETs is Power Ten Inc. Its rack-mounted 4600 series, for example, uses MOSFETs in a full-bridge configuration that


Rack-mounted 5-kW supply (Power Ten Inc)
switches at 60 kHz . Operating from a 3 -phase ac supply, the \(5-\mathrm{kW} 10500\) model delivers 10 V at 500 A and sells for \(\$ 2700\). Other voltages from 7.5 to 80 V are available within the 4600 series as are power levels of 2.5 kW and 3.75 kW . Modular in construction, the \(5-\mathrm{kW}\) models use 4 field-replaceable units that deliver 1.25 kW each. The outputs from the modules are displaced slightly in time; this regulation of the outputs lightens the burden of filtering the rectified dc.
Power Ten has chosen to use the MOSFETs because of their relatively low prices. The drive circuitry is also quite simple in comparison to bipolar devices. Because MOSFETs are simpler and because they can function efficiently at frequencies of 500 kHz (and higher), the company projects power densities of 15 to \(20 \mathrm{~W} / \mathrm{in}^{3}\) within the next several years. Right now, 3 to \(5 \mathrm{~W} / \mathrm{in}^{3}\) densities are available.

\section*{Higher frequencies create losses}

Operating at a frequency of 150 kHz , the RBX (600W) series from Kepco Inc uses MOSFET power switches in a half-bridge configuration. Because of the high-frequency operation, Kepco manages to pack the 600 W of power into a package measuring \(3.74 \times 7.87 \times 8.66 \mathrm{in}\). Switching frequencies in the range of 150 kHz and higher, however, produce three kinds of major losses. The first two involve power losses in the choke and the transformer cores, which increase with frequency, and copper losses in the transformer windings from the skin effect. The company solved the first problem with a low-loss ferrite core; it corrected the second by using separately insulated strands of wire called litz wire to increase the surface area of the windings.
The third problem arising at higher frequencies is the
loss of filtering efficiency caused by the equivalent series resistance (ESR) and the equivalent series inductance (ESL) of the output filter capacitors. To overcome this problem, Kepco stacks 39 capacitors (all in parallel) on either side of the converter's output chokes. For a 5 V output, each of these capacitors has a value of about 220 \(\mu \mathrm{F}\). Capacitors of this value have much lower ESR than do \(5000-\mu \mathrm{F}\) devices, for example. The short parallel connections also reduce the ESL. As a result of these compensations, the efficiency of the RBX 600 W supplies approaches the state of the art: \(80 \%\) for the 5 and 12 V models and \(85 \%\) for the 24 and 48 V units.
The RBX series uses heavy-duty, chrome-plated bus bars to carry the 120 A of current available at 5 V . The bus bars not only handle output currents to 120A; they are also part of the mechanical framework that holds the main power section together. Other features of these supplies, which sell for \(\$ 675\), include remote voltage control, current limiting, remote on/off control, remote error sensing, and a selectable \(115 / 230 \mathrm{~V}\) ac input.
Todd Products Corp offers a variety of power supplies targeted for specific market segments; it sells different supplies for the computer market, for example, and for the telecommunications segment. The computer series B MAX-503-0512 provides a maximum of 500 W from outputs of \(5 \mathrm{~V} / 80 \mathrm{~A}, 12 \mathrm{~V} / 10 \mathrm{~A}\), and \(-12 \mathrm{~V} /\) 10 A . The MAX- 500 series comes in a \(2.5 \times 5 \times 11.5-\mathrm{in}\). package, which is substantially smaller than the \(5 \times 8 \times 11-\mathrm{in}\). package usually used for 500 W supplies. The extensive use of surface-mount components contributes to the small size.
Todd also improved its current-fed inverter design, which now allows the supply to handle 500 W of power with the same amount of cooling as a 400 W switcher. The open-frame design of the MAX-500 uses system air for cooling, eliminating the need for a power supply fan. Priced at \(\$ 431\) (100), the MAX-500 series meets international safety standards for SELV applications, and the Class A RFI requirements of the FCC and VDE0871.

\section*{Battery-backup saves data}

If your computer system needs both a significant holdup time to save data and an orderly shutdown in the event of an ac power-line failure, consider the 750 W H754 from Jeta Power Systems. The output rating for the main channel is 2 to 28 V at 120 A ( 600 W max). Auxiliary channels 1, 2, and 3 are respectively rated at 2 to 28 V at 20 A ( 240 W max), 2 to 28 V at 5 A ( 75 W max),

Many high-power supplies still use bipolar transistors in converter circuits that switch at only 20 to 30 kHz .


A 1500W modular supply (Power-One)
and 2 to 48 V at 1 A ( \(48 \mathrm{~W} \max\) ).
A 48 V backup battery provides full power during a line failure, making it an uninterruptible power system. The holdup time depends upon the battery rating-a 5-Ahr battery provides 4 minutes of full-power operation. If the line power fails, the switch to battery power is automatic. A built-in, constant-current battery charger maintains battery power. During battery operation, a \(100-\mathrm{kHz}\) converter provides 300 V dc for the main power stage. The dc/dc converter uses power MOSFETs and current-mode control for protection against transformer saturation. The main power supply is a half-bridge circuit operating at 20 kHz and using bipolar transistors. The outputs and the backup battery are fully isolated from the input. The entire power supply (including the battery) is contained in a \(5 \times 8 \times 13\) in. package. The H754 design meets UL, CSA, and VDE regulations. It costs \(\$ 981\) (100).

One modular power system that you can configure for your specific needs comes from Power-One. The SPM5 series has 5 slots that can hold 5 plug-in modules. The 12 available single-output modules range in power from 70 to 780 W ; voltages extend from 2 to 48 V (with the usual steps at \(5,12,15\), and 24 V ). You can plug in 2 modules with power ratings of 300 W and higher. Dualoutput modules rated at 240 W are available with voltage ratings of \(\pm 12 \mathrm{~V}, \pm 15 \mathrm{~V}\), and \(\pm 24 \mathrm{~V}\). A 1000 W uninterruptible power supply battery-backup model operates from a 48 V battery.

Designed for international use, the SPM5 can supply 1500 W of de power from a selectable ac input of 90 to 132 V or 180 to 264 V at line frequencies from 47 to 440

Hz . You can also operate the supply from a 300 V dc source. The SPM5 modules use power MOSFETs in a current-mode switching circuit that allows parallel operation of the modules for higher output current, or series operation for higher output voltage. The 70 to 240 W modules use half-bridge circuits switching at 120 kHz ; the 750 W modules use full-bridge circuits switching at 60 kHz .

The SPM5 comes in an industry-standard \(5 \times 8 \times 11-\mathrm{in}\). case. The supply includes protection against short circuits, reverse voltages and overtemperature conditions. Standard control functions include remote sensing, current sharing, and remote voltage adjusting. The supply signals when the input power fails. The SPM5 meets VDE, IEC, CSA, and UL safety specifications, as well as FCC and VDE EMI specifications. Pricing varies according to the modules selected. A system configured with one 750 W and two 240 W modules costs \(\$ 922\); a system with two 750 W modules costs \(\$ 857\) (100).

\section*{Multi-output units extend to 2000 W}

HC Power Inc offers switching power supplies in the 500 to 2000 W range with from 1 to 5 outputs. The units use a modified forward-converter construction that eliminates saturation of the inverter transformer core and also permits the use of lower-voltage MOSFETs. The converter switches at 147 kHz and uses currentmode control to achieve a transient response to load changes of under \(50 \mu \mathrm{sec}\). Models are available in ratings of \(500,750,1000,1500\) and 2000 W . Output voltages for multi-output models range from 2 to 24 V ; for single-output models, the range is 2 to 48 V . The supplies meet all specifications for UL, CSA, IEC and VDE. They are available with 1 - or 3 -phase ac inputs or with a 48 V de input. The base price for single-output switchers is \(\$ 890(1000 \mathrm{~W}), \$ 1277(1500 \mathrm{~W})\), and \(\$ 1550\) (2000W). Quad-output supplies cost \(\$ 1205(1000 \mathrm{~W})\) and \$1661 (1500W).

A line of high-power switches from Deltron Inc provides power outputs from 500 to 1750 W . The VF series is available with from 1 to 5 outputs at voltages of \(5,12,15,18,24\), or 48 V . The VF series of switchers operate from 90 to 132 or from 180 to 264 V with a 47 - to \(63-\mathrm{Hz}\) ac line. A factory option for \(400-\mathrm{Hz}\) line operation is available. These units convert at 80 kHz and have a typical efficiency of \(80 \%\). The triple-output VF3C 1000 W model provides outputs of 200,300 , and 500 W and comes in a \(5 \times 6 \times 13\)-in. case and costs \(\$ 608(1000)\).

Introduced at Electro/88, Deltron's DVF series is

unit still provided a steady .8 amp of

In a Navy test, a Tomahawk cruise missile exploded into a concrete building. When the dust settled, little remained but gravel and fragments of casing.

And the Abbott model C28D0.8 you see here.

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MILITARY POWER SUPPLIES

\section*{Transformer design becomes critical at high currents. High operating frequencies only add to this problem.}
similar to the VF series except that the newer supplies operate from a 40 to 60 V dc source and have slightly different ratings. The available output voltages for the DVF series are \(2,3.3,5,12,15,18,24,28\) and 48 V . Efficiency is 75\% typ. The triple-output DVF3C 1000W units with output of 200,300 , and 500 W costs \(\$ 672\).

\section*{500 W open-frame units have 3 outputs}

Valor Electronics Inc makes 500 W open-frame switchers with output voltages of \(5,12,15\), and 24 V . The 5 K series is convection cooled. It's designed for use in workstations, superminicomputers, and test systems. All units use a half-bridge configuration with bipolar transistors. The supplies have a minimum efficiency of \(75 \%\). Other specifications include load and line regulation of \(0.15 \%\), a transient response of \(500 \mu \mathrm{sec}\), and maximum ripple and noise of 75 mV ( \(\mathrm{p}-\mathrm{p}\) at full load). Standard features include overvoltage protection, foldback current limiting, and remote sensing. All 5 K models meet UL, CSA, and VDE safety requirements; they also comply with FCC and VDE Class-A
specifications for conducted noise. They cost \(\$ 510\) in small quantities and \(\$ 300\) in quantities of 1000 .

Numerous other manufacturers offer lines of power supplies with a variety of options. Among the most significant are ACDC Electronics, Hewlett-Packard, LH Research, and Pioneer Magnetics. Each offers a variety of products in the 500 W and up range.

ACDC Electronics, for example, has four different series in the high-power range: The RSF series has 500 W models with 1 to 4 outputs and costs \(\$ 598\). The REV line has 800 and 1000 W models with 2 to 4 outputs; the 800 W models start at \(\$ 1014\). The JF series includes 750, 1000, and 1500 W single-output versions; a 1000 W model costs \(\$ 751\). The JFM line has 1600 W models with 3 to 5 outputs and starts at \(\$ 1595\).

Hewlett-Packard's HP6030 series is a group of five power supplies in the 840 to 1200 W range; they are designed for laboratory applications. These autoranging power supplies have DVMs and offer 12 -bit programming and readback of both output voltage and output current, self-documenting programming codes,

\section*{For more information . . .}

For more information on the high-power switching supplies described in this article, contact the manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request Service.

ACDC Electronics
401 Jones Rd
Oceanside, CA 92054
(619) 757-1880

Circle No 391
Computer Products
Boschert Inc
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Milpitas, CA 95035
(408) 942-1600

Circle No 392
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Power Products Division
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Circle No 394
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Circle No 395

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Jeta Power Systems Inc 2675 Junipero Ave Signal Hill, CA 90806
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Flushing, NY 11352
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Circle No 397
Lambda Electronics
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Melville, NY 11747
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Circle No 398
LH Research
14402 Franklin Ave
Tustin, CA 92680
(800) 547-2537

Circle No 399

Pioneer Magnetics
1745 Berkeley St Santa Monica, CA 90404 (213) 829-6751

Circle No 400
Power Components
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Gardena, CA 90248
(213) 323-8120

Circle No 401
Power-One
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Camarillo, CA 93010
(805) 987-8741

Circle No 402
Powertec Inc
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Chatsworth, CA 91311
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Power Ten Inc
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Qualidyne Systems Inc 3055 Del Sol Blvd
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programming error detection, and readback of all programmable functions. The series offers a complete solution to automated-system power requirements. Base prices range from \(\$ 2575\) to \(\$ 3400\).

Four different lines of high-power switchers from LH Research give you several options: The MMA series provides 500 to 1500 W with 1 to 9 outputs and sells for \(\$ 630\) to \(\$ 1295\). The SM line has 750 to 1500 W supplies with 1 to 4 outputs and costs \(\$ 680\) to \(\$ 1295\). The SMA supplies provide a single output from 600 to 2000 W ; prices range from \(\$ 1325\) to \(\$ 1600\). The MGA switchers include 500 and 1000 W models with single or multiple outputs; they are part of a high-reliability series and are guaranteed for five years; prices begin at \(\$ 945\) and run to \(\$ 1613\).

Pioneer Magnetics has more than a dozen kinds of high-power switchers at various levels in the 500 to 2000W range. Both single- and multi-output models are available. You can choose from models that operate from a 48 V de source. A battery-charging option is available for one of the 1400 W models. Single-output versions cost from \(\$ 500\) to \(\$ 1250\). Multi-output models run from \(\$ 695\) to \(\$ 1450\) (OEM qty).

Power-hungry applications such as those in superminicomputers, large-scale computerized test systems, and telecommunications equipment have created a large and various market for high-power switching supplies that are required to perfrom as efficiently as possible. Beyond efficiency, these supplies reveal a trend, albeit a gradual one, toward switchers with higher frequencies and the consequently reduced size and higher component density. This trend promises to soon produce high-power switchers that will offer a three- to fourfold improvement in power density-at higher efficiency ratings and lower costs.

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\title{
Precision board-mountable module delivers 10W of localized power
}

The 910AP1 power module precisely regulates its output to within \(1.3 \%\) of nominal voltage despite changes in load, line, and temperature. This miniature de/dc converter delivers a \(5 \mathrm{~V} / 2 \mathrm{~A}\) output from a 40 to 60 V dc input. An overvoltage-protection circuit that uses a totally independent control loop limits the module's output to 7 V dc.
Surface-mount technology keeps the module's parts count low and improves its reliability. The 910AP1 operates over 0 to \(70^{\circ} \mathrm{C}\) without requiring forced-air cooling or a heat sink. Its typical conversion efficiency is specified at \(80 \%\).

Complete input- and output-filter

circuits located inside the module maintain the module's output ripple (from de to 20 MHz ) at 100 mV p-p max. The product needs no external components to attain this performance. The input and output sections of the 910 AP 1 are completely
isolated from each other, so users can choose any polarity and grounding configuration. The converter's current-limiting circuitry automatically returns the output to normal levels when any fault condition is removed.
The 910AP1 meets UL specifications. It comes in a nonconductive \(1.7 \times 2.7\)-in. case that's ready for pcboard mounting and costs \(\$ 37\) (1000).

AT\&T, Dept 51AL23030, 555 Union Blvd, Allentown, PA 18103. Phone (800) 372-2447.

Circle No 452

\title{
Low-profile, enclosed switchers offer parallel capability for all outputs
}

The enclosed switchers in the 250 W PPM Series and the 400 W PFS Series offer parallel capability (current sharing) for both main and auxiliary outputs. The vendor claims this feature meets stringent requirements for single-point failures in redundant-power-system applications.

The supplies offer a host of standard features, including a remote on/off control, margining, a powergood signal, thermal warning and protection, overload and overvoltage protection, an input line filter, inrush-current limiting, soft-start capability, and a switch-selectable \(115 / 230 \mathrm{~V}\) ac input. Various options are also available to satisfy almost any power-supply requirement.

All units in both series are UL recognized, CSA certified, and TUV approved. The supplies fea-

ture a modular design that provides one to seven outputs. Their line- and load-regulation performance is \(0.2 \%\) or 10 mV max, and their switching frequency measures 100 kHz . The units' efficiency figures range from 70 to \(85 \%\). The maximum ripple and noise specs (for a \(20-\mathrm{MHz}\) bandwidth) are \(1 \%\) or 50 mV for the main
output and \(1.5 \%\) or 75 mV for auxiliary outputs. A 5V/80A model sells for \(\$ 432\) (100).

Computer Products, Power Conversion Group, 2900 Gateway Dr, Pompano Beach, FL 33069. Phone (305) 974-5500. TWX 510-956-3098.

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\section*{Fully isolated, \(100 \mathrm{~W} \mathrm{dc} / \mathrm{dc}\) converters feature wide operating range}

The KZ 400 Series 100 W de/dc converters offer as many as three outputs. The converters are protected against a variety of fault conditions, and they operate over -40 to \(+85^{\circ} \mathrm{C}\).
The converters are available in models that accommodate input ranges of 20 to 60 V dc or 36 to 72 V dc. Models KZ 431, 432, and 433 provide \(5 \mathrm{~V} / 20 \mathrm{~A}, 12 \mathrm{~V} / 8.3 \mathrm{~A}\), and \(15 \mathrm{~V} / 6.7 \mathrm{~A}\) outputs, respectively. The triple-output converters in the series supply 5 V and \(\pm 12\) or \(\pm 15 \mathrm{~V}\), and they provide 15 A on the main output and 1 A on the auxiliary outputs. On the auxiliary outputs, 1.5 A capability is optional.

The converters' accuracy is specified at \(\pm 1 \%\) for the main output and \(\pm 2 \%\) for auxiliary outputs. You can adjust the main output over a \(\pm 5 \%\) range. The ripple and noise specs

(over a \(20-\mathrm{MHz}\) bandwidth) are \(1 \%\) and \(2 \%\), respectively. The switching frequency is 200 kHz , and the converters' typical efficiency is \(80 \%\). The line- and load-regulation specs are \(\pm 1 \%\) for the main output and \(\pm 5 \%\) for the auxiliary outputs.

Overvoltage protection for the main output is standard, as is re-verse-polarity and short-circuit protection, which features automatic recovery. Thermal protection and
input-overvoltage protection are optional.
The KZ 400 Series converters sell for \(\$ 140\) to \(\$ 175\) (100) and are housed in a package that features 6 -sided shielding. A side-mount model is available to satisfy low-profile applications.
Intronics Inc, 57 Chapel St, Newton, MA 02158. Phone (617) 964-4000. TLX 200095.

Circle No 454

\section*{Single-output, current-mode power supply packs 3000 W into a \(5 \times 8 \times 15-\mathrm{in}\). package}

Series 9R power supplies include 2, \(3,5,12,15,24,28,36\), and 48 V units. Each of the current-mode supplies generates 3000 W and fits into a \(5 \times 8 \times 15\)-in. package, which is only 4 in . longer than the industrystandard package for a 1500 W supply.
The supplies typically operate at \(75 \%\) efficiency when running from a 230 V ac source. When operating in the 0 to \(50^{\circ} \mathrm{C}\) range, they supply \(100 \%\) of their rated load, although the 2 V unit can supply only 1400 W because of output-bus-bar and recti-fier-diode current-carrying limitations.

Other key specs include 5 mV or

\(0.1 \%\) (whichever is greater) line regulation, and 10 mV or \(0.2 \%\) (whichever is greater) load regulation; dynamic response of \(4 \% \max\) for a \(25 \%\)-step load change; \(200-\mu\) sec output recovery to within \(1 \%\) of nomi-
nal; and thermal regulation of \(\pm 0.02 \% /{ }^{\circ} \mathrm{C}\).
Current limiting and overvoltage protection are standard features of the supplies. You can also specify an SCR-crowbar feature for additional overvoltage protection. The units have terminals for remote sensing, and you can remotely control their outputs to within \(\pm 10 \%\) of nominal. The supplies also include indicators for valid outputs and for input-power failure. They cost \(\$ 1800\).

Bonar Powertec, 20550 Nordhoff St, Chatsworth, CA 91311. Phone (818) 882-0004. TLX 277483.

Circle No 450


Here's the shock-safe fuseholder family that puts the right replacement fuse at your fingertips. That eliminates the aggravating search for a misplaced fuseholder cap. And that ends the risk of accidental live terminal shock.

That's good news for everyone from design and production engineers, to the consumers of your products.

Because only the Littelfuse family of \(2 \mathrm{AG}, 5 \times 20 \mathrm{~mm}\), and 3AG Flip-top \({ }^{m}\) Shock-safe fuseholders are UL Recognized, CSA Certified and have been designed to meet or exceed IEC Shocksafe Standards 65 and 257.
But best of all, these low profile, panel-mount fuseholders are affordable, easy to install, and are as quick and safe to service as
push, pull, flip and snap. For your free product bulletin, samples, pricing and delivery dates, talk to your local Littelfuse representative or distributor.

\section*{Littelfuse Tracor} a WESTMARK company
800 E. Northwest Highway - Des Plaines, IL 60016 (312) \(824-1188\)



\section*{Power Sources}

\title{
Open-frame switching supplies handle drive-surge requirements
}

Among the SQM International Series switchers are models that offer a 150 to 350 W continuous-output capability. All the units feature peak-power capabilities to accommodate the initial turn-up/spin-up loads required by peripherals such as disk or tape drives and printers.

The switchers feature main 5 V outputs with current capabilities of 20 to 50 A . Auxiliary 5, 12, 15, and 24 V outputs are available with as much as 16 A of peak current. All the units are available with as many as four outputs.

The supplies accept a variety of input voltages and offer overvoltage protection for their transformers' primary windings. Other standard

features of the switchers are built-in line filtering, 3750 V ac safety isolation, shielded power transformers,
built-in overload and overvoltage protection, overtemperature shutdown, remote-sensing capability, vacuum-impregnated magnetics, and fixed-frequency pulse-width modulation. You can opt for a powervalid indicator and for TTL-high or TTL-low power-fail-detection circuitry.

All the units meet VDE, IEC, UL, CSA, FCC, and VDE specifications for safety isolation and EMI/ RFI emissions. The supplies cost \(\$ 266\) to \(\$ 358\) and are available for delivery from stock to 10 weeks.

Switching Systems International, Box 1599, Placentia, CA 92670. Phone (714) 996-0909.

Circle No 455

\section*{Modular system produces semicustom power supplies}

The AMPSS modular power-supply system provides a semicustom approach to power-supply design. Each supply consists of an \(8 \times 5 \times 11\) in. mainframe that contains lineinput rectification and smoothing circuitry and accepts de/dc-converter modules. The mainframe operates from 90 to 135 V or 180 to 270 V line-input supplies and accepts pow-er-supply modules having outputs as high as 1600 W . It comes in versions that accommodate as many as eight 100 W de/dc-converter modules, four 100 W modules and one 800 W module, or two 800 W modules. Smaller mainframes are available to house 600 W max supplies.

The dc/dc-converter modules currently available include 100 W units with output voltages of \(5,12,15\), or 24 V ; an 800 W unit that generates a 5 V output; and a 50 W unit with -5

and -12 V outputs. Except for the 2 -output version, whose line- and load-regulation spec is \(1 \%\), the modules feature \(0.3 \%\) typ line and load regulation. The units' output ripple is \(1 \%\) of output (max) over a \(20-\mathrm{MHz}\) bandwidth. The single-output modules can share current when operating in parallel.

All the outputs feature short-cir-
cuit and overvoltage protection. An optional housekeeping module provides remote on-off facilities, overtemperature and input-undervoltage protection, ac-input-failure indicators, a 5 V standby supply, a 12 V fan supply, and semiregulated supplies for external circuitry. The supplies are designed to meet the relevant UL, CSA, and VDE requirements. A typical 400 W version with a housekeeping module costs around \(£ 430\) (100).

Astec Europe Ltd, 8 B Portman Rd, Reading, Berkshire RG3 1EA, UK. Phone (0734) 509411. TLX 848047.

Circle No 456
Astec USA (HK) Ltd, 2880 San Tomas Expressway, Suite 200, Santa Clara, CA 95051. Phone (408) 748-1200. TLX 6839191.

Circle No 457

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Save Space: Completely protected AC-DC systems with rugged high density packaging to 8 watts \(/ \mathrm{in}^{3}\).
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\section*{DOTMET NOTHCES}

\section*{High-current, low-profile switcher offers significant size advantage}

The quad-output MAX-750 provides 750 W of power in a package measuring \(2.6 \times 5 \times 13.5\) in., which is significantly smaller than the typical \(5 \times 8 \times 11\)-in. shoebox switcher. The supply comes with a cover/fan assembly for self-cooling and features a thermal-cutoff system that protects the supply should the cooling system fail.

The supply's main output provides 5 V at 120 A . The three auxiliary outputs provide \(12,-12,-5.2\), and 24 V in various combinations. A magnetic, switching postregulator holds the line and load regulation to \(1 \%\) on the auxiliary outputs. A cur-rent-fed inverter that operates at 50 kHz achieves an efficiency rating of

\(80 \% \mathrm{~min}\), the vendor claims. The unit specs \(0.2 \% \mathrm{rms}\) max ripple and noise.

The switcher's standard features include user-adjustable outputs ( \(\pm 5 \%\) ), remote sensing on the 5 V
output, overload protection on all outputs, overvoltage protection on the main output, and the capability to inhibit the output remotely. An ac-power-failure indicator is optional.

According to the vendor, the MAX-750 meets international safety standards, including UL 478, CSA 22.2, IEC 380, IEC 435, and the Class A RFI requirements of FCC and VDE 0871. It also meets SELV requirements for creepage and clearance. It costs \(\$ 589\) (100).

Todd Products Corp, 50 Emjay Blvd, Brentwood, NY 11717. Phone (516) 231-3366. TWX 510-227-4905.

Circle No 451



\section*{WeHave Them All, Why Should We Be Biased?}


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\title{
High performance power, in a low profile package
}

\section*{Ericsson's new triple output, 15 Watt DC/DC converter is only \(8 \cdot 5 \mathrm{~mm}\) high with In-Card ()\(^{\text {mounting! }}\)}

When you need high performance and a low profile in a DC/DC converter, Ericsson's new 15W PKC series has the answer.

The PKC's advanced mechanical design offers two mounting options, Using conventional "On-Card" pcb mounting results in a height of \(0 \cdot 42^{\prime \prime}(10 \cdot 7 \mathrm{~mm})\).

Where profile is critical our In-Card © mounting provides the unique capability of recessing the supply into the pcb providing a height of \(0 \cdot 33^{\prime \prime}(8 \cdot 5 \mathrm{~mm})\) !

Allowing inputs of \(24 \mathrm{~V}(18-36 \mathrm{~V})\) and 48 V ( \(36-72 \mathrm{~V}\) ) the PKC series provides 5,12 and 15 V dc outputs in single, dual and triple output configurations. A high reliability magnetic feed back loop ensures that all outputs are regulated to within \(2 \cdot 5 \%\). Each model is fully isolated to 500 V dc and all outputs are overvoltage protected.

Advanced technology, including the use of highly automated surface mount manufacturing and 300 kHz switching frequency results in the PKC's exceptional MTBF of over 200 years.

SMD components are assembled onto a thick film ceramic substrate to give excellent thermal characteristics. As a result, efficiency is \(85 \%\) and the unit operates without derating over the temperature range of \(-45^{\circ} \mathrm{C}\) to \(+85^{\circ} \mathrm{C}\) !


No extra components, filters or heatsinks are required to meet VDE, FCC or CISPR regulations with respect to RFI specifications or high temperature environments.

Paralleling units for higher output requirements or redundancy configurations is easily accomplished by a simple connection.

To find out more about the world's most advanced 15W DCIDC converter, contact Ericsson today.

The PKC takes up no more pcb space than a credit card.

\section*{ERICSSON}


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\section*{Ericsson Components AB}

Power Products, S-16481 KISTA-STOCKHOLM, SWEDEN
Tel: +4687575000 Fax: +4687574884 Telex: 10948 POWERI S


\section*{COMPACT 250 \& 400 WATT SWITCHERS FEATURE FOUR-OUTPUT PARALLELING}

The PPM and PFS Series 250 and 400 Watt cased switchers incorporate a unique closed-loop current sharing control circuit which permits simultaneous paralleling of up to four outputs, each with remote sensing. In addition, a single-point failure protection circuit assures that a single failure in a redundant power system will not cause the power bus to fail.
The series can provide as many as seven outputs.
incorporated in three plug-in modules. Control circuitry is fabricated on a single circuit board using surface-mount technology. Standard features include switch selectable 115/230 VAC input, input line filter, remote on/off, power good signal, remote sensing, thermal protection with warning signal, overload protection, overvoltage protection, soft start and inrush current limiting. Optional features include current sharing, OVP crowbar and DC OK signal on main output, current monitor signal and others.

Low profile cases only 5 " wide by 2.75 " high permit convenient stacking of the supplies. Power density is up to 2.4 Watts per cubic inch, and all models are efficiently cooled by a miniature internal ball bearing DC fan.
These switchers employ a bridgedriven forward converter using 100 kHz MOSFET switching. The auxiliary outputs have either linear or magnetic amplifier regulation and are

There are 25 stocked standard models in the 250 W and 400 W series. The models are UL recognized, CSA certified, and TUV approved.
Key Specifications:
Line Regulation .......... 0.2\%
Load Regulation .......... 0.2\%
Holdup Time . ... 30 msec . min.
Efficiency . . . . . . . . . . 70 to 85\%
Oper. Temp. Range . . . . \(0^{\circ}\) to \(65^{\circ} \mathrm{C}\) From Computer Products/Power Products

\section*{NEW 4.5 WATT DC/DCS NEED ONLY HALF THE SPACE}

The AF Series DC/DC converters produce 4.5 Watts of DC power in half the space of previous industrystandard units.

With
a case size
of only \(1.0 \times 2.0 \times\)
0.375 inches, they
replace standard
\(2 \times 2\) inch converters, yet
have the same pinouts.
This series is fabricated with
surface-mounted components on a miniature thick-film substrate to achieve 6 watts per cubic inch power density and efficiencies up to \(66 \%\). Available in single ( +5 ,
+12 or +15 V ) and dual ( \(\pm 12 \mathrm{~V}\) or \(\pm 15 \mathrm{~V}\) ) outputs, the input voltage can be either +5 or +12 VDC Other key specifications include \(\pm 1.0 \%\) line regulation, \(0.5 \%\) load regulation, and 40 mV p-p ripple and noise. These converters are ideal for board-mounted applications in computers, I/O boards, telecommunications, process control and scientific instruments. The AF Series is on distributor shelves now.
From Computer Products/
Stevens-Arnold
Cirde No. 2

\section*{UNIVERSAL INPUT SWITCHERS OPERATE FROM 85 to 264 VAC}

Operate from any line voltage from 85 to 264 VAC without changing jumper wires or switches. Computer Products/Boschert's new NFS40,

line voltage
selection problems and offer system design flexibility.
On the low end, the NFS40
series of single and multiple output 40 watt switchers measures a tiny 5 " \(\times 3\) " \(\times 1.2\) ". These switchers provide more than 2.2 watts/cubic inch, and will fit into very small spaces.
The NFS50-7608 50 watt switcher directly replaces the industry standard \(6.3^{\prime \prime} \times 3.9^{\prime \prime} 40\) watt unit. No need to mechanically redesign. Just drop it in and enjoy the benefits of universal input, plus 10 additional watts for


For small to medium systems, check out the NFS110 series 110 watt switchers. The +12 V output will deliver up to 9 amps to start disk drives. The small \(7.0^{\prime \prime} \times 4.25^{\prime \prime} \times 1.8^{\prime \prime}\) size delivers more than 2.0 watts per cubic inch.
All of these switchers will operate reliably with no load on the output, making them ideal for expansion systems. Also, each output is fully regulated. NFS40 and NFS110 evaluation units are now available. NFS50 is in distribution. From Computer Products/Boschert Circle No. 3

\section*{POPLLAR DC/DC CONVERTERS HAVE NEW LOW PRICES} Design in high performance and reliable operation at new low prices. The ES, EA and H series have filled a lot of PC boards to date but never so economically. Prices have been reduced by \(10 \%\) for the ES series and \(15 \%\) for the \(H\) and EA series. Also, the H and EA series are available in new \(1.0 \times 2.0 \times 0.38\) inch nonconductive packages. Check the table below and send for the latest data on our top performers. Or better yet, call your distributor for these low cost solutions.
\begin{tabular}{lcccc} 
Series & \begin{tabular}{c} 
Output \\
Power
\end{tabular} & \begin{tabular}{c} 
Output \\
Type
\end{tabular} & \begin{tabular}{c} 
Input \\
Voliage
\end{tabular} & Output \\
\hline H & 1 Watt & Single & \(5,24,48 \mathrm{~V}\) & \(\pm 5 \mathrm{~V} @ 200 \mathrm{~mA}\) \\
EA & 1.8 Watts & Dual & 5 V & \(\pm 12\) or \(\pm 15 \mathrm{~V} @ 60 \mathrm{~mA}\) \\
ES & 15 Watts & Triple & \(12,24,48 \mathrm{~V}\) & \begin{tabular}{c}
\(+5 \mathrm{~V} \& \pm 12 \mathrm{~V}\) \\
\(+5 \mathrm{~V} \& \pm 15 \mathrm{~V}\) \\
or \\
\end{tabular} \\
& & & \(\pm 5 \mathrm{~V} \&+12 \mathrm{~V}\)
\end{tabular}

From Computer Products/Stevens-Arnold Circle No. 4

\section*{TAKING THE RISK OUT OF CUSTOM OPEN FRAME SWITCHERS}

Computer Products/Boschert can take the risk out of custom switcher development. You get a predictable, highly reliable switcher, based on our well-known standard circuits.

Here's why:

\section*{Field-Proven Building Blocks. The quickest,} least expensive way to build a custom switcher is using existing circuits. Computer Product/Boschert has hundreds of building blocks, field proven over our 15 year history. We understand their use and limitations. Your custom design is predictable, economical and virtually risk-free.

\section*{Appropriafe Topology}

Using the wrong topology is either costly or unreliable. Since we understand and build virtually every type of switcher, there is no need to squeeze your custom needs into our favorite topology. You get the correct topology for your power requirement.
Experience. Your supply is built by a company exclusively dedicated to switchers for 15 years. Our ability to integrate electrical, mechanical and thermal designs is exceptional. Our new medium power designs deliver more than 2 watts/cubic inch. The result is a highly compact switcher that meets or exceeds your system requirements.
Safety Approval. UL and CSA usually take no more than eight weeks. Full VDE certification usually takes only 12 to 16 weeks after your final prototype approval. TUV approval is also available.
Manufacłurability. Every supply we make is designed with manufacturability in mind. We use a common parts base, with well characterized components that meet our conservative stress derating guidelines. Check for further information on custom switchers.
From Computer Products/Boschert
Circle No. 5

\section*{Buif Re No. 1}

\section*{LINEAR POWER MODULES MEET UL544 MEDICAL REQUIREMENTS}

For UL544 approval you can rely on Computer Products/Power Products new MED 300/500 Series of AC/DC encapsulated power modules. They meet or exceed the stringent UL544 requirements for medical equipment. These supplies are recommended for use in nonpatient contact medical, dental and laboratory applications where high isolation and low leakage are critical. The MED 300/500 series is provided in single, dual and triple output models offering 63 different output voltage and current variations. The units are linear regulated and have output power from one to 15 watts with popular output voltages of \(5,12,15\), \(24, \pm 12\),


100 WATT DC/DC CONVERTERS HAVE LOW PROFILES, HIGH EFFICIENCIES
of either 18 to 36 VDC or 36 to 72 VDC Output ranges supplied are single \((+5\), +12 or +15 V ), dual ( +5 and +12 V ), or triple ( +5 and \(\pm 12 \mathrm{~V}\), Designed primarily for telecom and computer applications, the new WS Series from Computer Products/ StevensArnold offers 100 watts with single, dual, and triple
outputs. The converters are packaged in a low profile case ( \(3.5 \times 5.5 \times\)
mounting required. \(84 \%\) minimum and a \(2: 1\) input range
0.91 inches) producing a power density of 5.7 watts per cubic inch. Available in chassis mount with screw terminations or printed circuit board mounting. The PCB mount version is supplied with a heat sink (adds 0.35 inches to height) which allow for conventional cooling with no special

The WS Series has an efficiency of

important features include 500 VDC isolation, input surge protection, reverse voltage protection and remote on/off control with idle currents down to 10 mA .
Key Specifications:
Line Regulation . . . . . \(\pm 0.5 \%\) max. Load Regulation . . 2.0\% (to no load) Ripple and Noise . . 100mV p-p max. This series is available through distribution.
From Computer Products/Stevens-Arnold
Circle No. 7

If reply card is missing, please circle reader service number.
Consult 1987/88 EEM, page 643 for local sales office or call (305) 974-5500, Ext. 7514.
\(\pm 15\) VDC,
plus others
available. Standard
protection features such as overtemperature, overload and short circuit protection are included.
The power modules incorporate a split-bobbin wound transformer which provides high isolation between primary and secondary with low coupling capacitance. This results in 2500 VAC isolation voltage and less than 10uA leakage current.
The units are available in either printed circuit mountable or chassis mountable packages. In addition to UL544 approval the series is CSA certified.
From Computer Products/ Power Products Circle No. 6

\section*{UP TO 591，000 HOURS MTBF WITH OPEN FRAME LINEARS}

Step up to higher standards of reliabilty with the World－Standard Series of open frame linears．
The aluminum frames and power
receive a four－hour burn－in before shipment．
Other features include current－ limiting short circuit protection on all outputs，overvoltage crowbar on 5 V outputs， remote sensing on single
\begin{tabular}{|c|c|c|}
\hline \multicolumn{3}{|c|}{TYPICAL CALCULATED MTBFS} \\
\hline Output Voltage \({ }^{\text {e }}\) & Output Current & MTBF \({ }^{2}\) \\
\hline 5 V & 3.0 A & 591，533 rs \\
\hline 12 V & 6.8 A & 417，240 hrs \\
\hline 15 V & 6.0 A & 420，943 hrs \\
\hline 24 V & 4.8 A & 328，798 hrs \\
\hline \(5 \mathrm{~V} \pm 12 \mathrm{~V}\) & \(3 \mathrm{~A} / \pm 1.0 \mathrm{~A}\) & 261，201 hrs \\
\hline
\end{tabular}
\({ }^{1}\) Consult factory for additional models．
\({ }^{2}\) Calculations per MIL－HDBK－217E＠ \(100 \%\) load； 115 VAC line； \(25^{\circ} \mathrm{C}\) ambient temperature；ground，benign outputs and the 5 V output of triples， reverse voltage protection on all out－ puts，and operation from \(0^{\circ} \mathrm{C}\) to \(50^{\circ} \mathrm{C}\) with no derating．The World－ Standard Series （PL Series）are UL
ratings of the World－Standard Series are form，fit，and function replacements for all other open frame linear power supply manufacturers． With four different AC input line voltage ranges，these supplies can be connected for use in any country of the world．
These units have a VDE construction power transformer with enclosed split－bobbin windings and 3750 VAC minimum isolation．
The designs are conservative with efficiencies up to \(60 \%\) and all power supplies

\section*{QUICK ACTION REPLY CARD}

\section*{PLEASE SEND：}
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7．WS Series
11．Have an Applications Engineer Call
4．ES／EA／H Series
8．World－Standard Series
12．Have a Sales Person Call

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Specialists in Mil－Spec AC／DC power supplies and DC／DC converters，Computer Products／ Tecnetics has participated in many major military programs including F－16，Tomahawk， EA－6B，MSE and E2－C． From state－of－the－art topologies to advanced package design， every attention is paid to cost－ effective custom designs for demanding environmental requirements．Let us modify an existing standard product or develop a totally new design to meet your system specifications．We are certified to manufacture to Mil－Q－9858A and conform to the guidelines of NAVMAT P4855－1．
Send for you free copy of our Mil－Spec Power Supply catalog． From Computer Products／ Tecnetics Circle No． 9

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ions for our complete line of power supplies with power ranges from \(1 / 2\) Watt to 1500 Watts．
Cirde No． 10

\section*{Power Sources}


\section*{DC/DC CONVERTERS}

Housed in a \(1 \times 2 \times 0.4\)-in. package, ITS Series isolated 2 W de/dc converters accommodate input voltages of 18 to 70 V de to cover \(24,36,48\), and 60 V batteries or to compensate for long input-line drops in telecommunications applications. The converters are available with single or dual outputs of 5,12 , or 15 V .
Each converter features a 500 V isolation spec, protection against continuous short circuits and output overvoltage, and a typical efficiency of \(80 \%\). An integral input \(\pi\) filter suppresses all noise from the source -RFI input current measures a low 30 mA p-p.
Full power is available over a -25 to \(+71^{\circ} \mathrm{C}\) ambient range ( -25 to \(+91^{\circ} \mathrm{C}\) case). Each unit features a shutdown pin for remote control of the output. \(\$ 68.50\) (100). Delivery, stock to eight weeks ARO.
Melcher Inc, 10 Cochituate St, Natick, MA 01760. Phone (617) 653-9979. TWX 510-100-3830.

Circle No 712

\section*{DC/DC CONVERTERS}

The vendor offers WS Series 100 W de/de converters in chassis-mount and pc-board versions. Housed in a \(3.5 \times 5.5 \times 0.91-\mathrm{in}\). case, they feature a \(5.7 \mathrm{~W} / \mathrm{in} .{ }^{3}\) power density.

The converters accommodate input voltages of 18 to 36 V and 36 to 72 V dc. They come with a 5,12 , or 15 V output; 5 and 12 V dual output; and 5 and \(\pm 12\) or \(\pm 15 \mathrm{~V}\) triple output. All models spec \(82 \%\) efficiency \(\min\).

The key features include \(\pm 0.5 \%\) max line regulation and \(1 \%\) load regulation on the primary outputs for a full to \(0.25 \%\) load change and \(2 \%\) load change from full to no load. The ripple and noise measures, for a \(20-\mathrm{MHz}\) bandwidth, are 75 mV p-p. The devices also provide input surge, output short-circuit, and overvoltage protection. They also offer remote sensing on the primary outputs and remote on/off control. \(\$ 260\) to \(\$ 295\).

Computer Products Inc, Power Conversion Group, 2900 Gateway Dr, Pompano Beach, FL 33069. Phone (305) 974-5500. TWX 510-956-3098.

Circle No 714


\section*{SWITCHING SUPPLY}

The F350 single-output switching power supply provides 350 W of continuous power at \(82 \%\) typ efficiency. In order to satisfy the needs of drives, printers, and other peripherals that have high peak-power demands, the supply features as much as 450 W .
The supply provides 5 V at 70 A , \(12 / 15 \mathrm{~V}\) at \(30 \mathrm{~A}, 24 / 30 \mathrm{~V}\) at 15 A , or \(48 / 60 \mathrm{~V}\) at 7.5 A output. The selectable inputs are 95 to 132,180 to 264 , or 250 to 370 V ac.
Some of its features include \(\pm 0.02 \%\) line regulation for 5 V models and \(\pm 0.05 \%\) for other versions. Load regulation, for a 10 to \(100 \%\) load change, ranges from \(\pm 1\) to \(\pm 4 \%\). Overcurrent, overvoltage, thermal protection, and a power-fail indication are standard. The supply's conducted RFI is specified with VDE0871 Curve A. You can also obtain a version of the device
that provides Curve B. \(\$ 343\) (500).
Powerline Inc, 10 Cochituate St, Natick, MA 01760. Phone (617) 655-7987. TWX 510-100-3630.

Circle No 715


\section*{LITHIUM BATTERIES}

Units in the CSC line of lithium power cells feature no voltage delay and are suited for applications requiring immediate power on demand. With an open-circuit voltage of 3.395 V , the cells provide an operating range of -32 to \(+93^{\circ} \mathrm{C}\). Their capacity ranges from \(0.75 \mathrm{~A} /\) hour for the \(1 / 2 \mathrm{AA}\) size to \(30 \mathrm{~A} /\) hour for the DD size.
The cells have a 304L stainlesssteel case and a header assembly that features a glass-to-metal hermetic seal for high reliability and safety. You can order the batteries in custom versions that configure the cells in series or in series-parallel to provide the energy density necessary to satisfy a range of applications. Some of the options are internal fusing, protection circuitry, and insulation. \(\$ 14.95\) to \(\$ 42.50\) (100).

Electrochem Industries, 10000 Wehrle Dr, Clarence, NY 14031. Phone (716) 759-2828. TLX 91386.

Circle No 713

\section*{DC/DC CONVERTERS}

Available in chassis-mount packages, DCE and DCF Series dc/dc converters output 100 and 150 W , respectively. Their efficiency ranges to \(90 \%\).
You can order units that provide outputs of \(5,12,15,24\), and 48 V dc from nominal inputs of \(12,24,48\), and 110 V dc. These convection-

Single Output Up To 400A
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\section*{Case 10 \\ 750 to 2000 Watts 5 " \(\times 8^{\prime \prime} \times 11^{\prime \prime}\) \(\mathrm{N}+1\) Redundancy AC and DC Inputs \\ 1 to 5 Outputs 50A Auxiliary Mag Amp Output Ch 2}

For \(5 \times 8 \times 11\) "slot" switching power supplies from 750 to 2000 Watts, the Qualidyne Case 10 is all you need to know. MTBF of 150,000 hours. Single or multiple (up to 5) fully regulated outputs from 2 to 48 VDC. Precision paralleling for \(\mathrm{N}+1\) redundancy. AC and DC input voltages. Safety listings from UL, CSA and TUV. Compliance with IEC 380 \& VDE 0806. FCC 20780 Class A filtering. Nothing fancy, just reliable slot power-period.


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CIRCLE NO 117

\section*{POWVRLINE—A FULL LINE OF SWITCHING POWER SUPPLIES FROM 30-1500 WATTS}

\section*{PX Series}
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- FCC Approved
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LOW COST CUSTOMS WELCOMED

cooled units are housed in brushedaluminum cases and are specified for full output at temperatures to \(50^{\circ} \mathrm{C}\). The line and load regulation figures spec at 0.8 and \(0.9 \%\), respectively.
All models feature adjustable outputs \(- \pm 10 \%\), input filter, overvoltage protection, and short-circuit protection. Also, remote-sense and disable features are standard. DCE Series, \(\$ 219\); DCF Series, \(\$ 319\). Delivery, stock to eight weeks ARO.
International Power Sources Inc, 10 Cochituate St, Natick, MA 01760. Phone (617) 651-1818. TWX 510-100-3630.

Circle No 716


\section*{SUPPLIES}

The MK line of 75W high-voltage de modules features two models with outputs that range from 0 to 3 kV at 0 to 25 mA , to 0 to 60 kV at 0 to 1.2 mA . These air-insulated units are housed in a package that measures \(53 / 16 \times 4^{3 / 4} \times 11\) in.

Using a circuit, you can monitor true current accurately, \(1 \%\) of reading plus \(0.05 \%\) of rated current, with respect to ground without breaking the common-ground connection. The laboratory-grade fea-

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The only power supplies that: source and sink power in both directions; produce stabilized voltage or current, positive or negative, a-c or d-c; and get on the bus without fuss... \\ \\ Kepco BOP Power Managers now \\ \\ Kepco BOP Power Managers now have dazzling new speed!
} have dazzling new speed!
}

For years, now, people who have wanted a voltage that was equally agile above zero, below zero, and above and below zero, have counted on Kepco BOP (bipolar) Power Managers.

There's just nothing else like a BOP. Nothing else that can jump from, say, plus 100 V to minus 100 V , or from minus 20 Amps to plus 20 Amps in microseconds, passing through zero without a glitch. There's also nothing else that can produce rock-steady positive or negative voltages in the face of wildly fluctuating current-or vice versa. There's certainly nothing else that can operate not only as a source, but also as a sink, i.e., absorb \(100 \%\) of its rated voltage or current when its voltage polarity and current direction are out of phase, or when its load suddenly turns around and becomes a source.

On top of all that, the BOPs have also been pretty fast performers. Their bandwidths have ranged from 1.8 to 4.5 KHz , their slewing rates from 1 to \(2 \mathrm{~V} / \mu \mathrm{sec}\). Our model BOP \(100-2 \mathrm{M}\), for example, had a bandwidth of 3 KHz in both the voltage and the current mode, and could jump from -100 V to +100 V in \(100 \mu \mathrm{sec}\).

0


Two groups of BOP Power Managers are available: \(3 / 4\)-rack 100 - and 200 -Watt models (top), and full-rack 400 -Watt models.

Well, now Kepco has improved the BOP. That same model can now make that same leap in \(20 \mu \mathrm{sec}\)-a fivefold gain, and its bandwidth is 22 KHz in the voltage mode, 15 KHz in the current mode.

Furthermore Kepco has been able to speed up the BOPs without sacrificing stability-or anything else. They still do all the wonderful things they've always done. They just do it faster. Incidentally, acquiring dazzling new speed isn't the only thing that's been happening to the BOP series. It's also acquired a new model, the \(\pm 200 \mathrm{~V}, \pm 1 \mathrm{Amp}\) BOP \(200-1 \mathrm{M}\). Also, models BOP 500 M \(( \pm 500 \mathrm{~V} / 80 \mathrm{~mA})\) and BOP \(1000 \mathrm{M}( \pm 1000 \mathrm{~V} / 40 \mathrm{~mA})\) have had their noise levels significantly improved. Maximum p-p noise on the BOP 500 M was 500 mV , is now 100 mV . The total range of the BOP 500 M is 1000 Volts-from minus 500 V to plus 500 V and

100 mV noise in a 1000 V p-p signal is 100 ppm . That's an 80 dB signal-to-noise ratio!

BOP can interface with the IEEE-488 bus using a selection
0 of external controllers, or an optional built-in interface card.

To find out more, call or write Dept. JYF-12.

Reproduction of a 10 KHz square wave by a Kepco bipolar Power Manager Model BOP \(50-2 \mathrm{M}\), which actually has a 20 KHz bandwidth in the voltage mode.



\section*{Plug-in power for VME!}

Here's a fully featured 400-watt, tripleoutput power system that's configured for direct connection to the motherboard in your VME bus system. It's a standard MOSFET switcher that includes all the features you've come to expect from NCR Power Systems.

Switching frequency is 80 kHz , and the unit has protection against overvoltage, overcurrent, and input surges. It meets the stringent safety and EMI requirements established by UL, CSA and TUV (VDE). Packaging complies with the Eurocard standards defined in DIN-41494 and IEC297 for plug-in attachment to the motherboard. Precise ( \(\pm 0.4 \%\) ) line and load
regulation and 75\% efficiency make the unit ideal for data communications and processing applications.

Standard units can be modified by NCR for precise conformance to customer requirements.

For detailed specifications and price quotation, contact NCR Power Systems, 3200 Lake Emma Road, Lake Mary, FL 32746-3393; Telephone 800/327-7612 or in Florida, call 407/323-9250.
tures include a voltage regulation of better than \(0.005 \%\) for line and load, \(0.05 \%\) max current regulation from short circuit to rated voltage, and a ripple of less than \(0.03 \% \mathrm{rms}\) at rated voltage for a full load.
You also get TTL enable/disable capability, interlock capacity, a 10 V reference source, constant-current/ constant-voltage operation, voltage and current monitoring and programming, and a shielded output cable. From \(\$ 850\).
Glassman High Voltage Inc, Box 551, Whitehouse Station, NJ 08889. Phone (201) 534-9007. TWX 710-480-2839.

Circle No 717


\section*{150W SUPPLY}

MK-150 Series 150W switching power supplies provide a \(2.45 \mathrm{~W} / \mathrm{in}\). \({ }^{\text {² }}\) power density and efficiencies ranging from 77 to \(84 \%\), depending on output voltage.

The supplies have outputs of 5 , 12,15 , and 24 V . The input voltage is either 85 to 132 V ac or 110 to 175 V dc. The total regulation equals \(\pm 3 \%\). The switching frequency is 250 kHz .

These convection-cooled units operate over a 0 to \(60^{\circ} \mathrm{C}\) range. They use \(105^{\circ} \mathrm{C}\) electrolytic capacitors to ensure reliability. Other components are derated by \(20 \%\) to provide an added safety margin.

All supplies feature overcurrent and overvoltage protection and are available in open- or closed-frame designs. A 5 V open-frame model, \$188.

Toko America Inc, 1250 Feehan-

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Extreme reliability and extended life are the pedigree of every Christie product.

\section*{HEAVY DUTY DC SUPPLIE} HEAVY DUTY Our MAGAMP Series of power supplies offers a good example. These workhorses meet and exceed Mil and industry specs. Outputs range from 40A to 1500 A, with up to \(500 \%\) intermittent load current capability.
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CIRCLE NO 121


\section*{Power Sources}
ville Dr, Mount Prospect, IL 60056. Phone (312) 297-0070. TLX 724372.

Circle No 718

\section*{AC/DC SUPPLIES}

Although they measure \(1.5 \times 0.75 \times 0.375\) in., these miniature PS Series ac/dc supplies are suited
for pc-board mounting. Input voltage capabilities of 90 to 264 V ac lift the units to worldwide requirements.

The single-output power modules are available in 5,12 , and 15 V models. The current capabilities are 150, 65 , and 50 mA , respectively. The line regulation equals \(0.05 \%\) for all versions; the load regulation is \(0.8 \%\)

for the 5 V versions and \(0.5 \%\) for 12 and 15 V models.
'The supplies' metal cases provide 6 -sided shielding. The operating range, with no derating, spans -25 to \(+70^{\circ} \mathrm{C} . \$ 39.50\) to \(\$ 43.50\) (100).

Knightfire Technology, Box 8761, Coral Gables, FL 33075. Phone (305) 344-3682.

Circle No 719


\section*{HV SUPPLIES}

The PS310, PS325, and PS350 provide \(1.25,2.5\), and 5 kV , respectively, with 25 W of output power. Their voltage regulation equals \(0.001 \%\) for a \(\pm 10 \%\) line-voltage change, and their ripple is less than \(0.002 \%\) of FS.

The units have two 4-digit displays that give continuous readout of current and voltage. A third readout displays the parameter being entered. You can set hard and soft current limits, and output voltages to 1 V resolution.

Arc and short-circuit protection are standard. The devices feature connectors for remote voltage setting/ramping, and for current and voltage monitoring. The supplies will store and recall as many as 10 instrument settings. An optional GPIB port allows you to program input settings, and read output and


\section*{Power Sources}
instrument settings. \$995.
Stanford Research Systems Inc, 1290D Reamwood Ave, Sunnyvale, CA 94089. Phone (408) 744-9040. TLX 706891.

Circle No 720

\section*{DC/DC CONVERTERS}

NWD Series 15 W dual-output de/de converters have no min load requirements. The converters use current mode-control topology to accommodate inputs with 2:1 voltage ratios- 9 to 18,18 to 36 , and 36 to 72 V .

The modules operate at 100 kHz and provide efficiencies of 80 to \(83 \%\). Each features remote on/off capability and an LC input filter. Outputs of \(5 / 12 \mathrm{~V}, 12 / 12 \mathrm{~V}\), and \(12 / 15 \mathrm{~V}\) are available. Their line and load regulations are \(1 \%\). All units will withstand long-term output short-circuit conditions.

They offer voltage accuracy of

better than \(2 \%\). For a \(20-\mathrm{MHz}\) bandwidth, the ripple and noise equal 100 mV p-p. And for a \(50 \%\) load change, the transient response is less than \(500 \mu \mathrm{sec}\). The temperature coefficient measures \(\pm 0.02 \% /{ }^{\circ} \mathrm{C}\).

The converters are encapsulated in 6 -sided metal cases to eliminate noise radiation. \(\$ 111\). Delivery, stock to six weeks ARO.

International Power Devices Inc, 155 North Beacon St, Brighton, MA 02135. Phone (617) 7823331. TLX 989752.

Circle No 721


DC/DC CONVERTERS
VMEC Series dc/dc converters are designed for VME Bus applications. They come in 40 and 80 W versions. Both are housed in 3U-high packages that are 1 and \(1.4-\mathrm{in}\). wide, respectively.

Models are available with single and dual outputs ranging from 2 to 48 V dc. The temperature coefficient measures \(0.02 \% /^{\circ} \mathrm{C}\), the switching frequency equals 125 kHz , and the ripple and noise spec at \(1 \%\) or 50 mV , whichever is greater.

Their standard features include

\section*{If You're Fed Up With Outdated DC/DC Converter Technology, Get It Out Of Your System.}


For more information on these superior units, just give us a call. Because when it comes to P.C.B. mounted converters, Melcher has the answer.
- ISR 5 \& ISR 10 Series (5 \& 10 watts)
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-2:I input range
\(-75 \%\) efficiency
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-Reliable to 350,000 hours
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\(-18-70 \mathrm{~V}\) input range
\(-1 " \times 22^{\prime \prime}\) package P.C.B. mounted converters. Because the market is filled with old technologies that are inefficient, unreliable, bulky, and lack the kind of flexibility you need.
Melcher has the solution: four lines of state-of-the-art units that offer you superior performance in a smaller package size. These units have higher isolation, lower noise, a wider input range, and the widest temperature range compared to the "competition.' Consider these facts:

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\title{
Pit Our DE-DG Converters To the OFFLINE Teat
}


\section*{NEW AC INPUT / MODULE EVALUATION BOARD}

Incorporating an on-board 750 watt AC front end, the VI-MEBAC Module Evaluation Board is designed specifically for evaluation of Vicor converters in offline applications. The front end features input surge protection, removable line cord, a visible power indicator, and provides easily accessible ON/OFF and \(110 / 220\) VAC range select switches.

Up to three 300 volt-input Vicor converters can be plugged into the assembly for evaluation of single, dual or triple output applications. Independent access to all converter interface
pins allows for easy configuration of virtually any multiple-output or array application with a total power output of up to 600 watts. The \(10.5^{\prime \prime}\) \(\times 12^{\prime \prime}\) assembly comes complete with module sockets, strappable heavy-duty output lugs, output measurement jacks, and provisions for accessing the Gate, trim and sense connections on each converter. Module inputs are individually fused and on-board sockets are provided for adding Vicor Phased Array controllers. A detailed user's manual is provided which covers measurement techniques and useful applications information.

Vicor's "component level" megahertz converters allow power systems designers to focus on system solutions instead of circuit details. The benefits: fast, predictable designcycles; predictable field performance; and smaller, simpler, highreliability power assemblies with significantly reduced component count.
Call Vicortoday at (617) 470-2900, 23Frontage Road, Andover, MA 01810 for the Module Evaluation Board data sheet and put Vicor modules to the test!


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*Study done by Venture Development Corporation.
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\section*{Power Sources}
isolated outputs, EMI filtering, remote sensing, and overvoltage and overload protection. The line regulation equals \(\pm 0.1 \%\), and the load regulation equals \(\pm 0.2 \%\). 40 W version, \(\$ 127\); 80 W version, \(\$ 169\).
Power Pac Inc, Box 777, Norwalk, CT 06856. Phone (203) 8664484.

Circle No 722

\section*{DC/DC CONVERTER}

The PWS726 de/dc converter features an oscillator, a driver circuit, de switches, a transformer, internal filter capacitors, and a rectifier housed in a 32 -pin DIP. It supplies \(\pm 7\) to \(\pm 18 \mathrm{~V}\) dc outputs at \(\pm 40 \mathrm{~mA}\).
The galvanic input/output isolation is \(100 \%\) tested at 800 V dc and guaranteed to 2500 V rms continuous, 3500 V rms momentary. Other features include a \(1.2 \mu \mathrm{~A}\) leakage current and a \(9-\mathrm{pF}\) leakage capacitance. A separate synchronous con-
nection lets you frequency-synchronize as many as eight converters, while an enable input provides flexible control over outputs for power conversion and sequencing.

In order to protect the switches and prevent high-inrush current during the turn-on stage, a soft start/driver design ensures that the oscillator is fully operational before either MOSFET driver turns on. Input current sensing protects both the converter and the load from thermal damage by limiting the output fault currents. \(\$ 27.70\) (100).

Burr-Brown Corp, Box 11400, Tucson, AZ 85734. Phone (602) 746-1111. TWX 910-952-1111.

Circle No 723

\section*{POWER SUPPLY}

The quad-output Model 524 EU power supply employs a \(50-\mathrm{kHz}\) MOSFET design with independent magnetic amplifiers in the secon-

dary windings. The supply outputs are 5 V at \(20 \mathrm{~A}, 12 \mathrm{~V}\) at \(3 \mathrm{~A},-12 \mathrm{~V}\) at 2 A , and 24 V at 2 A (for disk-drive applications).
The secondary windings are adjustable to 15 V at 2.5 A and -15 V at 1.5 A . You can select input ranges of 105 to 130 or 198 to 265 V ac. All outputs are current limited and have continuous overload, short-circuit, and overtemperature protection. Self-recovering overvoltage protection is provided on all outputs.


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\section*{Power Sources}

The supply has no minimum load requirements and has an input protection that meets the IEEE-587 voltage-transient test. Power-fail and logic-inhibit signals are available as options. From \(\$ 395\).

RO Associates Inc, 246 Caspian Dr, Sunnyvale, CA 94088. Phone (408) 744-1450. TWX 910-339-9304.

Circle No 724

\section*{DC/DC CONVERTERS}

XC Series de/dc converters accept input voltage between 24 and 72 V dc and provide one of three outputs: 5 V at \(3 \mathrm{~A}, 12 \mathrm{~V}\) at 1.25 A , or 15 V at 1 A . All three converters have an \(80 \%\) min efficiency, even for \(20 \%\) load levels.

Their key specifications include \(0.2 \%\) line and load regulations,

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\(30-\mathrm{mV}\) output noise p-p, 500 V dc isolation, and -25 to \(+80^{\circ} \mathrm{C}\) operating range. The internal circuitry supplies 8 hours min short-circuit protection, and an internal thermallimit circuit shuts the converters when the case temperature exceeds the specified limit. They restart automatically when the temperature returns to normal.

Filter circuits provide conducted noise protection for input and output. A 6 -sided shielded case minimizes RFI problems. \(\$ 120\).

Calex Mfg Co Inc, 3355 Vincent Rd, Pleasant Hill, CA 94523. Phone (415) 932-3911. TLX 338506.

Circle No 725


DC/DC CONVERTERS
Models in the NMA line of \(\mathrm{de} / \mathrm{dc}\) converters accept inputs of 5, 12, 24, or 48 V dc. The units provide an output of \(\pm 5, \pm 12\), or \(\pm 15 \mathrm{~V}\) with 750 mW of power.

The converters come in single inline packages and in DIPs-both will deliver the full rated output. Their efficiency ranges to \(80 \%\). Over the operating range of -25 to \(+80^{\circ} \mathrm{C}\), the converters require no heatsinking. With a 0.18 -in. \({ }^{2}\) real estate requirement, the converters are adaptive to applications where space constraint is a factor.
The converters' isolation figure is 500 V dc. Both the single in-line and DIP versions are encapuslated in epoxy and thereby meet UL 94V-0 requirements. \(\$ 19.50\). Delivery, stock to eight weeks ARO.

International Power Sources



\section*{Condor's SDS and SDM Series power supplies meet FCCIVDE Level B and have agency safety approvals!}

Want a tough, versatile, well-designed power supply for your next application? Try a single (SDS) or multiple (SDM) output model from Condor!

Every Condor SDS or SDM power supply meets the toughest domestic and international safety requirements, and is UL, CSA and TUVNDE certified. All units also meet FCC 20870 Level B and VDE 0871 Class B above 150 KHz .

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Inc, 10 Cochituate St, Natick, MA 01760. Phone (617) 651-1818. TWX 510-100-3630.

\author{
Circle No 726
}


\section*{DC/DC CONVERTER}

The PKA 4411 PIL isolated dc/dc converter provides a \(5 \mathrm{~V} / 8 \mathrm{~A}\) output from a pe-board-mount package that measures \(3 \times 3 \times 0.78 \mathrm{in}\). Because the package extends only \(0.78-\mathrm{in}\). above the pe board, you can mount the units on boards that have 6TE (1.2-in.) spacing.

The converter accepts de input voltages ranging from 39 to 64 V and has an input-to-output isolation to 500 V dc. Its predicted MTBF, at an ambient temperature of \(45^{\circ} \mathrm{C}\), is more than 200 years. The converter operates from -45 to \(+65^{\circ} \mathrm{C}\).

However, you can obtain the PKA-4411-PI model, which has an integral heat sink that extends the operating range to \(85^{\circ} \mathrm{C}\). This other version also has a \(3 \times 3\)-in. footprint, but it is 1.39 in . high. A chassismount version with fast-on terminals is also available. Approximately 811 Swedish Krona (100).
Rifa AB, Power Products Div, 16381 Stockholm, Sweden. Phone (8) 757-5000 TLX 10948.

Circle No 727
Rifa Inc, Greenwich Office Park 3, Greenwich, CT 06836. Phone (203) 625-7300.

Circle No 728

\section*{POWER SUPPLIES}

SMS600 Series single-output, fancooled, switch-mode power supplies provide a 600 W output. They are available with nominal output voltages of \(12,15,24\), or 50 V . The 12

and 15 V outputs are provided by a single model, which employs a potentiometer for output selection. The 24 and 50 V versions have a potentiometer, which gives you approximately \(\pm 10 \%\) control over the output voltage.

The line regulation is less than \(0.25 \%\), for a \(\pm 15 \%\) change in the input voltage, and the load regulation is better than \(0.5 \%\), for a 10 to \(100 \%\) load change. At normal operating loads, the supplies have an \(80 \%\) efficiency.

Some of the key features include remote output sensing, and signals that either indicate that the devices are fully operational or notify you that a power failure occurred in the output fan. The output is protected against overcurrent and overvoltage conditions. The supplies meet major safety and RFI standards. £275 (100).

Weir Electronics Ltd, Durban Rd, Bognor Regis, Sussex PO22 9RW, UK. Phone (0243) 865991. TLX 86543.

Circle No 729
Weir Inc, 418 3rd St, Annapolis, MD 21403. Phone (301) 268-0122.

Circle No 730

\section*{DC/DC CONVERTER}

By employing surface-mount technology on a ceramic substrate, the type 3T switch-mode de/dc converter produces an output power of 50 W from a pe-board mount single in-line package that measures \(2.0 \times 1.1 \times 0.16 \mathrm{in}\). It can provide full output power over a 0 to \(70{ }^{\circ} \mathrm{C}\) temperature range without additional heatsinking. You can program the converter with shorting links so that it produces an output voltage of


Got an "on card" power distribution problem? Not enough real estate for a big modular supply? Don't want to use two card slots for any of the readily available "high profile" converters?

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\section*{Power Sources}
\(5,12,15,18\), or 24 V . At 12 V , it achieves an operating efficiency of around \(94 \%\).

The converter accepts a dc input voltage of between 11 and 40 V ; you can also configure the device so that it operates as an ac/dc converter. Zero to full-load output regulation is \(0.5 \%\) for an output voltage of 5 V and \(1.0 \%\) for all other output voltages.

The line regulation over the 11 to 40 V input range is \(1 \%\) for a 5 V output and \(2 \%\) for other output voltages. For \(\pm 10 \%\) input changes, however, the line regulation for all output voltages is only \(0.2 \%\). The maximum output current is 3 A , but you can add external power transistors to provide greater output currents. £10 (100).

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The secondary outputs are either \(\pm 12 \mathrm{~V}\) and \(12 \mathrm{~V}, \pm 12 \mathrm{~V}\) and 24 V , \(\pm 12 \mathrm{~V}\) and -5 V , or \(\pm 15 \mathrm{~V}\) and -5 V . You can trim the output voltages by \(\pm 5 \%\). The supplies will operate with a main output load of as little as 1.2 A with all other outputs unloaded, and they can cope with the high peak current requirements of, for example, disk drives. Other features include \(75-\mathrm{kHz}\) FET switching, warm- and cold-start inrush current control, and line input failure signaling.

The load regulation for a \(40 \%\) change on a \(60 \%\) load is \(\pm 0.5 \%\) for the main output, \(\pm 2 \%\) for the split positive and negative supplies, and \(\pm 0.5 \%\) for the single supply secondary. The supplies operate from ac line input voltages of 99 to 132 V or 187 to 265 V and have a line regulation of \(\pm 0.1 \%\) for a \(\pm 15 \%\) line input change. They meet the relevant IEC, CSA, VDE, UL, and BS reliability standards as well as the re-
quirement of NATO standard AQAP4.

Coutant Electronics Ltd, Kingsley Ave, Ilfracombe, EX34 8ES, UK. Phone (0271) 65656. TLX 46310.

Circle No 741
Qualidyne Systems Inc, 3055 Del Sol Blvd, San Diego, CA 92154. Phone (619) 575-1100. TLX 709029.

Circle No 742

\section*{SWITCH-MODE SUPPLY}

The SMM1500-0000 is the part of a series of single- and multiple-output power supplies, which are capable of delivering 1500 W of dc power from an industry-standard \(11 \times 8 \times 5\)-in. package. This initial version has five outputs. Its fully floating main output can deliver 5 V at 200 A and has remote sensing that can compensate for a voltage drop of 0.25 V in the power connecting leads. You can parallel the main outputs of two or

more power supplies to increase the 5 V output's current capability. The four auxiliary outputs share a common 0 V terminal and provide outputs of \(\pm 12 \mathrm{~V},-5.2 \mathrm{~V}\), and +24 V . The combined load rating of the auxiliary outputs is 500 W .

All outputs are protected against overcurrent and overvoltage conditions, and the fan-cooled supply is protected against overtemperature conditions. The supply has a remote shutdown input. The unit's power

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\section*{Power Sources}
factor is internally corrected to a value of \(0.95 \pm 0.05\) so that it draws a sinusoidal current from the line supply.
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Weir Electronics Ltd, Durban Rd, Bognor Regis, Sussex PO22 9RW, UK. Phone (0243) 865991. TLX 86543.

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Weir Inc, 418 3rd St, Annapolis, MD 21403. Phone (301) 268-0122. TWX 510-600-7370.

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\author{
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}

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\section*{HV SUPPLY}

To meet the requirements of the latest ion implantation equipment, Series 1200 high-voltage power supplies are capable of delivering 485 kV max at currents as high as 2 mA .
The power supply features independent voltage and current control with automatic crossover between the two modes. The overcurrent protection is built into the supply; in the event of output arcing, the damage to components is minimized by the low level of stored energy in the supply's multiplier stack. The supply features a 0 - to full-load output regulation of \(0.01 \%\); its output ripple equals \(0.1 \%\) of the rated output voltage p-p to 200 kV . Above 200 kV , its output ripple is less than \(0.5 \%\).
The power supply offers a rackmounted converter that operates from 108 to 132 V or 216 to 264 V , 48 to \(62-\mathrm{Hz}\) ac line supplies. It also provides a voltage multiplier stack whose sections are rigidly joined so that you can mount the stack in any orientation. Around \(£ 19,000\) for a 485 kV model. Delivery, 20 weeks ARO.

Bonar Wallis Hivolt Ltd, Dominion Way, Worthing, Sussex BN14 8NW, UK. Phone (0903) 211241. TLX 877112.

Circle No 754


\title{
Multilayer backplanes require careful design specs
}

Dense, multilayer backplanes are necessary for connecting today's heavily populated daughter boards. But to get the highest performance out of your system, you have to keep in mind design considerations that reduce noise and prevent transmission degradation.

\author{
J D Mosley, Regional Editor
}

As electronic components continue to surpass the speed specs of previous-generation components, the need for transmission networks that can rush signals to their destination has forced the continuing evolution of backplanes. Along with these increasingly sophisticated backplanes has come denser and morecomplex board-to-board connectors. If you don't understand subtle electrical-design considerations when it comes time to specify your backplane, you may not be able to adequately control the noise that can result.

The first basic design step, of course, is to define the physical characteristics of the backplane, or mother board. Count the number of daughter boards you need to mate and the type and number of interconnections that are required. Determine the connector pitch and I/O locations. Estimate the physical size of the backplane and the card cage.

Once you've specified these physical characteristics, the remaining design steps have to do with electricaldesign considerations. Preventing excessive noise will


Backplane with modular connectors (Teradyne Connector Systems)
pose the primary electrical-design obstacle. Several parameters are critical: reference impedance, signal and power leads, system rise time, and your circuit's plated through holes, if any. Luckily, you can adjust these parameters and thereby decrease the amount of noise in your system. Bear in mind that easing the impact of one factor may aggravate the problems generated by another.

According to Michael Hayward, president of Hybricon Corp, a supplier of VME Bus backplanes, the characteristic reference impedance of a typical backplane varies between 50 and \(100 \Omega\). To ensure the fastest possible rise and fall times and the lowest possible crosstalk, he suggests lowering the signal density as much as possible and maintaining a lowimpedance ground plane to reduce capacitive crosstalk.

He also sees power distribution, or the location of the power leads, as an important contributor to noise. You should carefully assign power-entry points on your backplane to produce the most uniform current density
possible. And, if the number of voltage drops across the board seems excessive, increase the number of entry points. You may also want to consider increasing the copper thickness if the problem persists.

A common cause of reflections (and noise) is when a signal encounters an impedance level that differs from that of the originating transmission line's. Engineers at Teradyne Connector Systems know of several techniques that can help you control a mother board's impedance level. These techniques include varying factors such as the grounding of the board stiffener, the direction of the signal being transmitted, the connector tail lengths, and the density or location of the through holes to help alleviate impedance-level imbalances. You can use a time-domain reflectometer to measure and evaluate the reflection amplitudes as you tweak these parameters to attain optimal performance.

In addition, you can reduce reflection peaks by providing additional dielectric between the rows of your daughter boards. The tradeoff is that you slow the

Variations in the density of the conductors on your backplane can contribute to an unevenly distributed plating thickness across the panel's surface.


MIL-spec backplane measuring \(42 \times 22\) in. (Methode Electronics)
signal speed and increase the reflection length and propagation delay. This approach may also increase the capacitive crosstalk between signal paths, so you may prefer to provide a local ground shield instead.

Remember that multilayer backplanes require sufficient resin to fill the core regions between the signal traces and the features in the power and ground planes. You can determine the minimum dielectric thickness needed for your design by multiplying the total thickness of opposing copper layers by a factor of \(11 / 2\).

Because a backplane is thicker than a daughter board, its plated through holes offer greater capacitance on the transmission line. For this reason, signals travelling from the backplane to the daughter board tend to suffer degradation more often than signals travelling in the opposite direction. Feed-through connector tails appear to the signal as shunt transmission lines and produce successive reflections. You can control
the severity of such reflections in the daughter board by grounding the contacts to improve impedance matching in the circuit.

The changing rise time of a signal through a circuit, when combined with mismatched connector impedances along the transmission line, also contributes to connector reflection. Because the circuit acts as a filter, you can tweak the L and C values anywhere in the circuit to help alleviate reflectance.

In addition, mismatched transmission-line connectors can slow the rising edge of digital signals, because these connectors act like lowpass filters in the line. As a result, you may also encounter signal absorption or delay in the extremely high frequency components of the transmitted signals.

\section*{Consider even the board's peripherals}

Aluminum stiffeners, which you'll find on the leading edge of many daughter boards and header assemblies, can affect a backplane's electrical performance by increasing static discharge to components on adjacent boards during insertion or removal. Grounding the stiffeners will protect your backplane from transmis-sion-degradation problems. The use of mounting screws offers a convenient means of providing grounding between a stiffener and the board.

Depending on your system's I/O requirements, the tail lengths of the backplane will differ. Remember, though, that the longer the tails, the more capacitance you add to your system. By carefully considering what the impact of this capacitance has on your system's impedance levels, you may want to redesign your board and I/O specifications to optimize this factor in a highspeed system.

Don't overlook the effects of the capacitive nature of the plated through holes in your circuit. You can tune the effective impedance associated with the through holes by varying the dimensions of the through-hole pads and the clearance between the signal and ground layers. You may also consider changing the dimensions of the traces leading from the through holes.

\section*{Take advantage of experts' advice}

Backplane experts at Teradyne Connector Systems have written numerous technical bulletins and tutorials that contain suggestions for enhancing the manufacturability of your backplane and ensuring that your design is cost effective. According to Teradyne, for example, you should specify your mother-board design so that the finished product makes maximum use of the avail-

able real estate. For a final panel size of \(12 \times 12\)-in., for instance, the optimal working area you can expect is \(10 \times 10 \mathrm{in}\).; for an \(18 \times 24-\mathrm{in}\). panel, the final working area is \(15 \times 21 \mathrm{in}\).

The backplane design team at AMP Packaging Systems recommends that you specify gas-tight press-fit connectors to prevent damage to plated through holes. The company claims that its Action Pin connectors make the board manufacturing process less costly by allowing a greater tolerance ( 0.034 to 0.043 in .) in hole size. These contacts require less than 40 lbs of insertion force. Furthermore, you can remove and replace damaged press-fit connectors because the connectors never cut through the walls of the plated area, and the spring
properties of the contacts help prevent damage to the hole upon removal.

AMP also offers the advice that variations in the density of the copper-wire conductors on your backplane can contribute to uneven plating thicknesses across the surface of the panel. To rectify this problem and achieve a uniform current density of evenly electrodeposited copper and tin lead, you can add nonfunctional connectors to your design in sparsely populated sections of the panel-a technique called thieving.

You also have to keep in mind the commercial safety standards for spacing the copper-wire conductors in a printed circuit. Agencies such as Underwriters Laboratories and the International Electrotechnical Commis-

Aluminum stiffeners on the daughter boards' leading edges may increase the danger of static discharge to components on adjacent boards.
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|l|}{TRADEOFFS TO CONSIDER WHEN DESIGNING A BACKPLANE} \\
\hline Increasing the number of conductors per routing channel will. & decrease the number of signal layers, but increase crosstalk and decrease controlled-impedance-design flexibility. \\
\hline Decreasing the conductor spacing between connectors will. . & increase crosstalk. \\
\hline Reducing the conductor widths and increasing the number of conductors will. & reduce design flexibility in controlled-impedance designs. \\
\hline Increasing the thickness and width of the copper printed-circuit connectors and increasing the acceptable high-temperature limit will. & increase the current-carrying capacity of the connectors. \\
\hline Specifying a large printed-circuit panel size will. & decrease labor cost per unit-area-processed, but will increase problems regarding dimensional stability and registration. \\
\hline Specifying nonplated through holes smaller than 0.187 in. will. . . & prevent increased manufacturing costs resulting from the need to plug larger nonplated holes prior to electrodeposition, or the need to drill them a second time, which results in reduced registration accuracy. \\
\hline Using heavy-weight copper foils in your power and ground planes will. & minimize current variations in high-power and highcurrent applications. \\
\hline Specifying heat-relief pads in copper layers of 2 oz or less will. & ensure the complete fusion of the through-hole solders. \\
\hline Distributing multiple voltages (either by using multiple voltage planes, by dividing the voltagedistribution plane into voltage zones, or by utilizing externally applied bus bars) will & decrease noise. \\
\hline
\end{tabular}
sion make recommendations to which your system may need to adhere.

\section*{Backplanes with modular connectors}

Teradyne offers modular backplane and daughterboard connectors to provide you with maximum flexibility for your system design. The High Density Plus modules contain three to six rows of contacts and offer electrical compatibility with high-speed VLSI circuits. The field-repairable backplane connectors have low-insertion-force sockets and compliant, press-fit contacts. They include low-inductance grounding, multivoltage power distribution, and controlled-impedance printed-circuit panels. An integral daughter-board stiffener is standard. You can select from several modu-
lar formats of power buses. The price for a complete backplane ranges from \(\$ 0.20\) to \(\$ 0.26\) per mated-signal contact pair; delivery is 8 to 12 weeks ARO.
For even more design convenience, you can order a mother board already mounted in an industrial card cage. RLC Enterprises offers its STD Bus BP-300 with this option. A 4 -slot mother board, it carries signals from \(8-\mathrm{MHz} \mu \mathrm{Ps}\). It also contains onboard passive terminations on all signal lines and has a ground plane for noise reduction. The connectors have gold contacts and are located on \(0.6-\mathrm{in}\). centers.

The card cage is made of gold-irridited aluminum with antivibratory, shock-dampening, spring-loaded card guides. You can order it in either a side- or rear-mounting style or in a 19 -in. rack-mount version.


Monolithic backplane with press-fit construction (BiccVero Electronics)

Alone, the BP- 300 costs \(\$ 94\). If you order it mounted in a card cage, it will cost \(\$ 144\). If you order it with a 5 V at \(6 \mathrm{~A}, 12 \mathrm{~V}\) at \(2 \mathrm{~A},-12 \mathrm{~V}\) at \(200-\mathrm{mA}\) optional power supply, it will cost \(\$ 279\). The company also offers 6 -slot, 8 -slot, and 12 -slot versions of the BP-300.

\section*{Monolithic backplane reduces noise}

Exhibiting an 11 to \(56 \%\) improvement in crosstalk specs compared with earlier models, Bicc-Vero Electronics' new monolithic backplane for VME Bus systems features an 8 -layer, press-fit construction. According to the manufacturer, this type of construction guarantees controlled impedance even when the board is fully populated with connectors and components. The board has two \(\mathrm{V}_{\mathrm{CC}}\) planes and three ground planes; the outer planes contain extra copper to reduce ground shift and protect the integrity of signals travelling through the inner layers.
In one 5.5 -mm-thick board, the 21 -slot backplane accommodates both J1 and J2 standards for the VME Bus system. The board contains press-fit connectors, multiple power taps, and through-board pins that let you make bus and jumper connections from either side of the mother board. Terminations are located at the back of the unit, so you can access terminations and
jumpers from the front of the board. Among the board's five available power options is an external power-bus system. Prices range from \(\$ 1000\) to \(\$ 1100\), depending on the power option you choose.

If your design has to meet military standards, you may want to consider the MIL-C-28754 aluminum backplanes and components that Methode Electronics offers. These mother boards have drilled 6061-T6 aluminum per QQ-A-250/11, individually replaceable fork contacts of heat-treated beryllium copper per \(Q Q-C\) -


STD Bus backplane and card cage (RLC Enterprises)

533, individually replaceable nylon insulators per MIL-M-20693, and a standardized military polarization system. Another configuration, which may better counteract excessive noise, consists of an aluminum backplane with separate ground and voltage planes, provided by laminating multiple electrically isolated aluminum plates. Other Methode MIL-spec boards are also available, including a \(42 \times 22\)-in. model coated with polyurethane or acrylic and processed to MIL-I-46058.

EDN

\section*{Article Interest Quotient (Circle One) High 473 Medium 474 Low 475}

\section*{For more information . . .}

For more information on the backplanes discussed in this article, contact the following manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.

AMP Packaging Systems Inc 700-E Jeffrey Way
Round Rock, TX 78664
(512) 244-5100

FAX (512) 244-5112
Circle No 351
BICC-Vero Electronics Inc
100 Sherman Ave
Hamden, CT 06514
(203) 288-8001

TWX 510-227-8890
Circle No 352
BICC-Vero Electronics Ltd
Flanders Rd, Hedge End
Southampton SO3 3LG, UK
(0703) 266300

TLX 477984
Circle No 353
Hybricon Corp
Box 149
Ayer, MA 01432
(617) 772-5422

TWX 710-347-0654
Circle No 354

Methode Electronics Inc 7444 Wilson Ave
Chicago, IL 60656
(312) 867-9600

Circle No 355
RLC Enterprises
4800 Templeton Rd
Atascadero, CA 93422
(805) 466-9717

Circle No 356
Teradyne Connection Systems Inc 44 Simon St
Nashua, NH 03060
(603) 889-5156

Circle No 357

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The Tandem NonStop VLX \({ }^{\text {TM }}\) system. later, is Tandem still completely sold on Teradyne's High Density Plus? "We made the right choice with Teradyne. And we look forward to working with Teradyne WE MADE to meet THE RIGHT CHOICE. just future write: interconnection requirements."

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\section*{MULIFILAR}

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MULTIFILAR magnet wire is custom produced to guarantee flat, parallel construction in an array of sizes, 16-52 AWG, with up to 20 conductors in some sizes.

Available constructions include round, flat or square conductors. Individual conductors meet NEMA MW 1000, JW 1177, temperature classes \(105-220^{\circ} \mathrm{C}\); single through quadruple film builds. Colorcoded conductors are available in most insulations with up to 10 different colors in some sizes.

Call or write for your free copy of our Technical Data Booklet and Capabilities Brochure. Both contain valuable information on all wire produced and inventoried by MWS Wire Industries. Samples of MULTIFILAR are available upon request.


\title{
Hardware and Interconnect Devices
}

\section*{Closed-loop air-to-air heat exchangers offer corrosion protection}

The HE-11 and HE-15 air-to-air heat exchangers cool equipment in enclosures sealed against entry of outside air. Each unit has two blowers. One of them circulates air within the equipment enclosure and sends heat to a heat-core pipe that conducts it outside the enclosure. There, the second blower forces outside air past the heat core, dispersing heat into outside air by convection.
Two sizes are available: The HE-11 measures \(51 / 4 \times 51 / 8 \times 11^{5 / 16}\) in. ( \(55 / 8 \mathrm{in}\). outside the cabinet). Its mounting flange is \(6^{3 / 4} \mathrm{in}\). square. When this unit is operating, the temperature of the air inside an enclosure rises an average of \(1^{\circ} \mathrm{F}\) above the ambient temperature for each 7 W of power dissipated by the equipment contained in the enclo-

sure. The HE-15 measures \(71 / 8 \times 77 / 8 \times 15\) in. ( \(7^{1 / 2}\) in. outside the
cabinet). It has approximately four times as much cooling capacity.
Both heat exchangers can operate at ambient temperatures as high as \(155^{\circ} \mathrm{F}\). Quarter-turn fasteners facilitate disassembly and cleaning of the external portion of heat exchangers. You can order heat cores made of special materials to withstand specific corrosive atmospheres. Each unit has a gasket that keeps your enclosure sealed. Standard units operate from \(50-\) or \(60-\mathrm{Hz}\) power sources. You can order models for 115 or 230 V . The vendor can also supply dc-powered units. From \(\$ 273\).
McLean Midwest, 4000 83rd Ave N, Brooklyn Park, MN 55443. Phone (612) 561-9400. TLX 290883. FAX 612-569-0533.

Circle No 435

\section*{Small connector lets linked boards pivot for adjustment and service}

This hinged connector places mated pairs of 0.06 -in.-thick pe boards on 0.36 -in. centers, and lets you pivot one of the boards \(90^{\circ}\) to gain access to otherwise inaccessible components. You can also mount the connectors in a way that lets you rotate your boards into a coplanar position; when you use the connectors in this way, you can move the boards to within \(90^{\circ}\) of each other. The vendor has demonstrated that the connectors withstand 50 rotational cycles and 30 mating/unmating cycles.

The contacts are tin-plated phos-

phor bronze; they are on \(1.25-\mathrm{mm}\) centers and are rated for 125 V and 1A max/circuit. The tails are kinked
to retain the connectors during soldering; they are arranged in a staggered pattern to reduce the likelihood of solder bridging. The glass-filled polyester dielectric has a 94V-0 UL rating. The line includes units with even numbers of contacts from 4 to 20. \$0.054/circuit. Samples of 14 - and 18 -contact units are in stock; production quantities, seven to eight weeks ARO.
Molex Inc, 2222 Wellington Ct, Lisle, IL 60532. Phone (312) 9694550. TLX 254069.

Circle No 438

\section*{Hardware and Interconnect Devices}

\title{
Electrically insulated cold plates remove 1.5 kW from components
}

These liquid- and air-cooled plates have a ceramic coating with high thermal conductivity as well as good electrical insulation. The manufacturer uses a proprietary process for applying ceramic coatings to a metal base-an approach that provides lower thermal resistance than other insulation techniques do. The process also permits the application of a solderable copper coating over the ceramic layers. The solderable copper coating lets you mount several power semiconductor devices on a common heat sink. You can make connections to those devices and electrically isolate them from each other without using special hardware.


Part number 180-10-6, for example, is a \(3 \times 6\)-in. plate, with 0.01 -in.thick ceramic pads cooled by water. When the attached components are dissipating 400 W , and the water is flowing at \(1 \mathrm{gal} /\) minute, the plate's
temperature increases \(40^{\circ} \mathrm{C}\). If you increase the water-flow rate, you can lessen the temperature rise (or raise the dissipation while maintaining the temperature rise). This type of water-cooled plate is suitable for cooling devices that dissipate 200 W to 1.5 kW .

You can get some configurations from stock, but the manufacturer expects that most orders will be custom ones. From \(\$ 35\).
EG\&G Wakefield Engineering, 60 Audubon Rd, Wakefield, MA 01880. Phone (617) 245-5900. TWX 710-348-6713.

Circle No 436

\section*{PS / 2 cards speed fast prototyping of computer add-ins}

These prototyping boards are suitable for use in IBM PS/2 Model 50 and 60 computers. They are available as fully drilled pe boards etched with the company's Microboard interconnect pattern, or as boards fully populated with Speedwire terminals.
Adjacent to the Micro Channel connector, the boards have an additional edge connector that's compatible with the IBM PS/2's video expansion interface. In addition, you can mount a miniature D-connector, with as many as 37 pins, on the rear edge of the board.
The 4-layer boards contain internal power planes for \(+5 \mathrm{~V},-5 \mathrm{~V}\), +12 V , and -12 V supplies, and a full copper ground plane. To ensure that

all ground and signal connections are made before power is applied to the board, the 5 V edge-connector fingers are reduced in length so that they are the last to make contact. All the edge connectors are gold plated.
Each board is supplied with a card mounting kit that contains a plastic card pusher, a handle and
card guide, and a stainless-steel bracket with a cutout for a 37 -pin D-connector. An extender board, which allows you to work on the cards above the computer's chassis, is also available. Microboard protoyping card, around £40; Speedwire prototyping card, around \(£ 100\).

\section*{Bicc-Vero Electronics Ltd, Flan-} ders Rd, Hedge End, Southampton SO3 3LG, UK. Phone (0703) 266300. TLX 477984.

Circle No 439
Bicc-Vero Electronics Inc, 1000 Sherman Ave, Hamden, CT 06514. Phone (203) 288-8001. TWX 510-227-8890.

Circle No 440

\section*{Products of Your Environment.}
connector so tough, you could roll a 5 ton truck over it. A Parallel Interconnect that allows a gas-tight interface in 60 seconds flat. A cabling assembly enabling a single connector to handle electronic signals, fiber optics and power supply.

These are but a few innovations in interconnections from ITT Cannon. But we couldn't have done it without you. Because success throughout our company relies on a thorough knowledge of your company's environment.

Take strategic partnering, for example. We don't create a custom solution by shaking your hand and jotting down a few notes. When we design-in, we get inside your environment.

Then there's Cannon's near-zero defect rate. It got that way, and stays that way, because we test each product in a carefully simulated environment.

As for delivery, we built a dependable system by studying the needs and scheduling realities of our customers' business environments worldwide.

And Cannon stays price competitive by always asking the question, "How will this connector be used?" Considering the connector's ultimate environment has taught us that keeping quality high ends up costing our customer less.

So if you'd like a partner who will take the time to learn about your environment, take a moment to contact ITT Cannon at (714) 964-7400.

Worldwide Headquarters
10550 Talbert Ave.
Fountain Valley, CA 92708
Or call (714) 964-7400
CIRCLE NO 143
ITTCANNoN
We're making progress. Not excuses.

\section*{Hardware and Interconnect Devices}

\section*{Connectors have redundant contacts on \(0.050-\mathrm{in}\). centers}

The RN Pak 50 series is a family of 2-piece connectors for mating pc boards to one another, to input/ output connections, and to flat cables. The cables can have a conductor pitch of 0.025 in ., and you mass terminate them through IDC (insu-lation-displacement crimping). The connector's halves have identical contacts. Pairs of contacts positioned on a \(0.05-\mathrm{in}\). pitch permit high-density interconnections. You can mix them with older connectors by terminating a pair of 0.05 -in.center flat cables to a single IDC connector.
Both straight-through and rightangle styles are available to mount

on pe boards. You can order 30-, 40-, 50 -, or 60 -pin versions of either style with as many as 200 contact positions. On pe-mounted connectors, alternate pairs of tails are staggered by 0.075 in . The insulators have clips to keep the connectors on the boards before soldering.
The devices use flat-beam, cop-per-alloy contacts. To improve reliability, each mated pair touches at
two points. Selective gold-over-nickel plating covers the actual area of contact; tails are solder plated. The contacts can carry 0.5 A ; mated pairs have a maximum resistance of \(25 \mathrm{~m} \Omega\). The rated, glass-filled-nylon insulation has a \(94 \mathrm{~V}-0\) UL rating and withstands 650 V rms for \(1 \mathrm{~min}-\) ute. Insertion force is approximately \(3.4 \mathrm{oz} /\) contact position; withdrawal force is about one-fifth of insertion force. \(\$ 0.10 /\) mated pair (OEM qty).
Robinson Nugent Inc, 800 E Eighth St, New Albany, IN 47150. Phone (812) 948-0564. TWX 810-540-4082.

Circle No 437

\title{
IT COSTS NO MOR
BEST HEAT DIS
35 vears of total peatication to Cooling Electronic Circuits
}

WIDE SELECTION
Over 3,000 heat dissipators and 600 hole patterns.

PROVEN QUALITY
Quality control in conformance to Military Standard (MIL--45208).

\section*{ON-TIME DELIVERY}

A history of meeting delivery commitments - including "Just-in-time"
free heat DISSIPATOR CATALOG
36 pages featuring over 125 types of aluminum heat dissipators for popular semionductor case types. Includes photos, outline drawings and specifications.


\section*{TECHNICAL SUPPORT}

Engineering assistance available upon request
COMPETITIVE PRICING
Dedication to providing quality products at a competitive price

\section*{FREE HEAT SINK/DISSIPATOR PRODUCT AND THERMAL MANAGEMENT GUIDE}

152 pages of performance data on hundreds of heat dissipators and heat sinks. Includes a comprehensive section on practical application criteria and selection of heat dissipator devices for semiconductors.



\section*{Think SAE \\ Stanford Applied Engineering}


Fo For 25 years OEMs have selected SAE to manufacture their backplanes, PCBs, connectors, and card cages. Now they are using us as their manufacturing partner. We take total responsibility to design, fabricate, test and deliver a complete "engineered assembly." This allows the OEM to concentrate all their time and monies on marketing and developing new products - not on new and expanded manufacturing facilities and personnel.
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- Engineered Assemblies backplane, wire wrap, card cage, connectors, PCBs, wiring, filters, power supplies, fans, mounting hardware, peripherals, and accessories; complete design, fab and test - ready for your functional test.
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- Commercial VMEbus, VERSAbus, S-100, DEC, STD, Multibus backplanes; PCBs, OPL to 16 layers, UL to 4 mils.
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- Connectors commercial and military.
- Filters EMI/RFI, emission test lab.
- Magnetic Components transformers, modules and filters.
- Card Cages custom, kits, and accessories.
- Wire Wrap semi and automatic, numerous formats and test.

Call, write or FAX for our "engineered assembly" brochure and your local representative/distributor ... Think SAE!

\section*{How topleas significant portion}

\section*{easmall but of yourcustomers.}

Pushbutton-operated devices are much more marketable when the buttons respond in satisfying ways to the user's touch.

So we made our new SLP control panels a touch more responsive.

We increased button travel to a more comfortable . \(050^{\prime \prime}\).

And we incorporated an enhanced con-


Actuator travel of \(0.050^{\prime \prime}\) is more than twice that of conventional low profile keypads. ductive rubber pad, tested to ten million operations, that rewards the finger with a satisfying snap back when a button is actuated.
The result is a control panel that adds to a user's comfort and confidence, for about the same price as those that generate feelings of vagueness and uncertainty.

But that's not the only bright feature of these new controls.

You can also get them in fullface, LED lighted versions that show up unmistakably, even in broad daylight.

And each button can be made to light independently of its on/off status,
enabling the panel to guide a user through a sequence of operating steps.

These new panels are available off-the-shelf


Selective button lighting belps guide a user through a sequence of operating steps. in several standard arrays, or in custom packages with microprocessors, visual displays, special enclosures, or attachment cable included.

Making controls more responsive is just one of the many ways we can help you build competitive advantages into your products.

To make your designs easier to sell, more satisfying to use, or more costefficient to produce, call us at
815-235-6600.

Together, we can find the answers.


More flexible than rigid glass epoxy boards. More rigid than polyimide film.
BEND/flex bendable circuit board material* from Rogers. The best of two technologies with a remarkable new twist: reduced production costs of up to \(30 \%\).

BEND/flex vs. rigid glass epoxy boards.
No more difficult design problems, labor-intensive wire harnesses or costly board-to-board interconnects. With BEND/flex, you design on a single board, bend or curve to meet your space requirementseven unfold to inspect or repair components. And BEND/flex is rugged enough to handle standard processing techniques, including wave-soldering and pick-and-place automation assembly.


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BEND flex meets ANSI/IPC-FC-24/1B/6.
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\section*{आ ROGERS}

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\section*{Hardware and Interconnect Devices}


\section*{TERMINAL POSTS}

These tin-plated brass, screw-terminal posts solder to pe boards at four points to resist rocking and rotation.

You can specify a model that accepts wires inserted perpendicular to the board (you secure them using a screwdriver held parallel to the board plane) or a model whose wire entry is parallel to the board (in which case, you secure the wires with a screwdriver held perpendicular to the board).

The terminals' steel 6 -32 binding screws accept solid or stranded wire from 14 through 22 AWG; they also accept crimped-on terminals and lugs. From \(\$ 0.075\).
Keystone Electronics Corp, 49 Bleeker St, New York, NY 10012. Phone (212) 475-4600. TLX 353700.

Circle No 665


\subsection*{0.025-IN PITCH CABLE}

This "ribbon" cable contains wires spaced 0.025 in. center to centerwhich amounts to twice the density commonly found in ribbon cables.

You can use it with female-socket,
transition, and D-subminiature in-sulation-displacement-terminated connectors that have \(0.05 \times 0.05-\mathrm{in}\). grids. Such connectors provide four times the density of more common interconnection devices.
Standard versions of the cable feature from 26 to 100 single-strand, \(30-\mathrm{AWG}\), bare copper wires; the vendor can also supply the cable with stranded 32-AWG copper wire. The gray PVC insulation is marked for polarity along one edge, UL listed, and rated for operation from -20 to \(105^{\circ} \mathrm{C}\). One version of the cable includes a copper-mesh shield and drain wire; another is jacketed as well as shielded; and a third is round and features a jacket and shield. This last version includes flat sections for termination. From \(\$ 82.66 / 100 \mathrm{ft}\) for 50 conductors (in \(1000-\mathrm{ft}\) rolls).

Amphenol/Spectra-Strip, 720 Sherman Ave, Hamden, CT 06514. Phone (800) 572-2253; in CT, (203) 281-3200.

Circle No 666


\section*{POWER INLET}

The F15352 EMI filter attenuates transients generated by switchingregulated power supplies; it includes an IEC ac inlet connector, an on/off switch, a fuse, an indicator lamp, and an IEC ac outlet. The unit snaps into a \(3.10 \times 2.32\)-in. panel cutout without the use of screws, extra holes, or mounting brackets.

It is rated to carry 10A in the US (6A in Europe) at 115 V to 250 V ac 50 or 60 Hz . The line-to-ground leakage is 0.25 mA max at 115 V 60 Hz and 0.05 mA max at 250 V 50 Hz . The test voltages are 2250 V dc from line to ground and 1450 V dc from
line to line. According to the vendor, the filter complies with UL, CSA, and TUV interference requirements. \(\$ 24.00\) (1000). Delivery, two weeks ARO for samples and eight weeks ARO for production quantities.

Stanford Applied Engineering, 3520 De La Cruz Blvd, Santa Clara, CA 95050. Phone (408) 9880700. TWX 910-338-0132.

Circle No 667


EMI SHIELD
According to the vendor, this Copper Cloth material provides exceptional shielding. One version features \(80 \times 80\) wires \(/ \mathrm{in}\). and is 0.011 in. thick; the other offers \(100 \times 100\) wires \(/ \mathrm{in}\). and is 0.0046 in . thick. The vendor combines the shielding material with 12 types of zippered insulating jackets that you can apply to round and flat cables. A continuous copper strap facilitates termination and grounding of the shield. \(\$ 2\) to \(\$ 3 / \mathrm{ft}\).

The Zippertubing Co, Box 61129, Los Angeles, CA 90061. Phone (312) 321-3901.

\author{
Circle No 668
}

\section*{BURN-IN SOCKETS}

These test/burn-in sockets accept ICs in plastic-quad flatpacks with or without bumpers. One socket, which features a hinged cover, accepts "naked" devices; the other serves as a socket/carrier combination. The carrier-style socket ac-


Augat's All-Stars demonstrate real team spirit - the concept that the whole is greater than the sum of its parts. With Augat, you benefit from an integrated program that assures you a COMPLETE VME solution. Unique advantages like nationwide distribution, a comprehensive VME Engineering Design Guide and an in-depth line of quality products combine to give you value that greatly exceeds any line of "hardware only."

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Over 100 distributor locations because one or even several stocking locations can never provide the product availability and service that 100 savvy Augat distribution sites can muster.


Augat is teamed with national distributors whose business depends on their ability to supply off-the-shelf product and value-added custom assemblies "just in time." You can count on your Augat distributor and Augat's extensive VME product line to give you what you want when you want it!


\section*{In-depth technical documentation always at your desk}

Design and systems engineers require complete product documentation. Augat's VME Engineering Design Guide has been prepared to assist you in the critical design/specification stage of projects that require system-level packaging. It focuses on the whole system from backplanes to power supplies, including a detailed electrical

\section*{MEINETERCONNECTION}
characterization of backplanes, and how each variation can affect your system. So next time you're considering a VME approach, buy more than product - buy an integrated solution from Augat. Work with an All-Star Team!


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Backplanes Card Cages Wire-Wrap Panels \(\square\) Unilayer II Boards
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INTERNATIONAL OFFICES: Northern Europe Augat Ltd. (0908) 676655/Central Europe Augat Gmbh 089576085 Southern Europe Augat SA 1466830 90/Japan Augat KK 45 473-7301

\section*{Hardware and Interconnect Devices}

cepts devices with \(52,68,84,100\), and 132 leads. The hinged-lid socket accepts devices with 84,100 , and 132 leads. The insulation material is polyethersulphone/polyetherimide. The contacts are made of berylliumcopper and feature gold-over-nickel plating. 132-pin socket and carrier, \(\$ 92.11\) (100).
Nepenthe, 2471 E Bayshore Rd, Palo Alto, CA 94303. Phone (800) 637-3684; in CA, (415) 856-9332.

Circle No 669


\section*{ZIP SOCKET}

This socket accommodates devices in zigzag in-line packages. Its two rows of contacts are on a staggered, \(0.1 \times 0.1-\mathrm{in}\). grid. They accommodate recently introduced RAM devices from manufacturers such as Fujitsu and Mitsubishi. The contacts incorporate a screw-machined outer sleeve with a closed bottom that the vendor claims provides complete protection against solder wicking. You can specify 16 -, 20-, 24 -, or 28 -position versions with solder tails or 2 - or 3 -level wirewrappable terminations. 16-pin socket, \(\$ 0.83\).

Electronic Molding Corp, 96 Mill St, Woonsocket, RI 02895. Phone (401) 769-3800.

Circle No 670


\section*{LOW-PROFILE SOCKETS}

M600 DIP sockets project 0.100 in . above the board they're mounted on, which amounts to 0.070 in . less than a standard IC does. By stacking them end to end and side to side, you can create very dense packages. Their black, 94 V -0-rated thermoplastic, polyester insulators are open to facilitate visual inspection after soldering. Their berylliumcopper contacts have three tines to achieve high reliability through redundancy, and the contacts' seamless, drawn outer sleeves have hollow bases to accept longer-than-normal device leads.
You can obtain the sockets with 24 to 40 positions in three configurations, and can mount them on 0.063to 0.093 -in. thick boards. You can specify contacts with gold-overnickel or tin-over-copper plating. \(\$ 0.80\) to \(\$ 1.20\).
Mark Eyelet Co, 63 Wakelee Rd, Wolcott, CT 06716. Phone (203) 756-8847. TWX 510-600-7291.

Circle No 671


\section*{MINIATURE SHUNT}

You can use these B-type shunts as jumpers to set equipment configurations. Their \(2.44-\mathrm{mm}\) width allows you to position them on \(0.1-\mathrm{in}\). cen-
ters. The shunts are 4.5 mm high, which is nearly the same height as a standard DIP IC. Their gold-plated, phosphor-bronze, double-tuningfork contacts have a maximum resistance of \(20 \mathrm{~m} \Omega\); they carry 3 A and fit \(0.64-\mathrm{mm}\) square posts. Their black, \(94 \mathrm{~V}-0\)-rated thermoplastic PBT insulators are self-extinguishing and \(30 \%\)-glass filled, exhibit a \(100 \mathrm{M} \Omega\) resistance \(\min\), and withstand 650 V rms at sea level. \(\$ 0.06\).
Kycon Cable and Connector Inc, 1887 O'Toole Ave C103, San Jose, CA 95131. Phone (408) 435-1110. FAX 408-435-1149.

Circle No 672


\section*{ENCLOSURE}

The ZX-981 enclosure can house a Multibus, Multibus II, or a VME Bus Hbackplane. Rear-loading versions accept 15 -slot Multibus or 12 slot VME or Multibus II backplanes. Top-loading versions accept 15 - or 20 -slot Multibus, 12 - or 20 -slot Multibus II, or 12 -, 16 -, or 20 -slot VME Bus backplanes.
The enclosures feature a removable front panel; lift-up top cover; front-panel-mounted power, interrupt, and reset switches; and four dc-powered fans. The vendor can supply mounting brackets that accommodate two full-height or four half-height storage peripherals. A shielded power-supply bay accommodates off-the-shelf switching supplies with outputs to 1000 W . Troughs and numerous prepunched holes facilitate the neat routng of cables. You can attach rack slides

\section*{Pevilit' Eurocard DIN Connectors}

\section*{NOW-Assured Performance PLUS-Guaranteed Gold Thickness}

ALL AT COMPETITIVE PRICES...While others talk of meeting DIN performance levels, Panduit meets these requirements with an added plus-guaranteed gold thickness in the mating area for maximum reliability, with a minimum of 50 micro-inches for Level I, 15 (or 30) for Level II and 8 for Level III. Lets you choose the connector you need for your application.

Applicable sizes are fully intermateable to DIN 41612, MIL-C-55302, BSI9525 and IEC 603-2 specifications. And provide compatibility with VMEbus, NUBUS, STE Bus, MULTIBUS II,

FUTUREBUS, or CIMBUS. But we don't stop there,
Panduit also provides:
- A full line of connectors and accessories to meet your package requirements and to lower your total installed cost.
- Over 15 years of connector manufacturing and engineering experience, using state-of-the-art equipment to keep pricing competitive without sacrificing quality.
- Full service support and just-in-time delivery to meet your production requirements.
Buy the DIN connectors that are best for your budget AND your engineers' design require-ments-PANDUIT performance rated Eurocard Connectors.

Call 1-800-323-2428 Toll Free today and
ask for the DIN Connector Product Manager-For pricing and further information.
(In Illinois,
1-312-532-1800)

\section*{Hardware and Interconnect Devices}
using holes provided for the purpose. \(\$ 4500\).

Zendex Corp, 6700 Sierra Lane, Dublin, CA 94568. Phone (415) 8283000.

Circle No 673


\section*{TERMINAL BLOCKS}

LP and LPK Series terminal blocks mount on pe boards and permit the attachment of discrete wires from 14 to 26 AWG. You secure the wires by tightening self-locking and, according to the vendor, vibrationproof screw clamps. The vendor provides the Dekafix marking system for standard- or custom-circuit identification. The blocks are UL and CSA approved for carrying currents to 10 A at voltages to 300 V . The blocks have closed bottoms. The contacts are spaced 0.2 in . apart, on one, two, or three levels. \(\$ 0.55 /\) point.

Weidmuller Inc, 821 Southlake Blvd, Richmond, VA 23236. Phone (804) 794-2877. TLX 828376.

Circle No 674

\section*{SPEED-SENSING FANS}

These "intelligent" fans sense their rotation speed and automatically restart after a service interruption. You can obtain models that draw power from 115 V ac and from 12 and 24 V de sources. Though most models feature ball bearings, some have

sleeve bearings. Their diameters range from 60 to 164 mm , and their free-air delivery ranges from 9.5 to 180 cfm . \(\$ 15\) to \(\$ 20\) (OEM qty).

NMB Corp, 9730 Independence Ave, Chatsworth, CA 91311. Phone (818) 341-0820. TLX 651340.

Circle No 675


\section*{FUSE BLOCKS}

These fuse terminal blocks snap into DIN mounting rails. The M \(10 /\) 16.SF accepts wire sizes from 18 to 8 AWG and accommodates standard \(1 / 4 \times 1^{1 / 4}-\mathrm{in}\). cartridge fuses rated from 0.25 to 16 A . It is 16 mm thick and is UL recognized for operation at 600 V . You can specify a unit containing a blown-fuse indicatora neon lamp rated for operation from 57 to 480 V ac or an LED that operates at 24 V dc.

Other of the vendor's fuse terminal blocks accept type GMT fuses rated from 0.18 to 10A. One model,
the 13 -mm-wide T \(7 / 1\), accepts a single fuse and incorporates an auxiliary contact; the 89 -mm-wide T \(7 / 8\) accepts eight fuses and features eight auxiliary contacts. M 10/16.SF, \(\$ 8.30\); T 7/8, \(\$ 98.50\).

Entrelec, 2 Tam Ridge Rd, Spring Valley, NY 10977. Phone (800) 431-2308; in NY, (914) 4257460. TLX 996619.

Circle No 679


\section*{FLEXIBLE CABLE}

You can bend Conformable Coaxial Cable sharply without using special tools. According to the vendor, its copper-tin composite shield provides \(100 \%\) shielding. The cable is rated for use at \(200^{\circ} \mathrm{C}\). It is compatible with most noncrimp connectors usually used for cable. The standard lengths range from 100 to 1000 ft . From \(\$ 295\) ( 500 ft ).

Belden Wire and Cable, Box 1980, Richmond, IN 47375. Phone (800) 235-3364; in IN, (317) 9835200.

Circle No 676


DIN CONNECTORS
These connectors comply with DIN

\section*{It Takes An Unbeatable Combination To Score A Technical Knockout.}


\section*{ZENITH'S STANDARD SWITCHING POWER SUPPLIES-A DEVASTATING 1-2 PUNCH OF SUPERB FEATURES AT A COMPETITIVE PRICE.}


It took an industry heavyweight like Zenith to put it all together. The kinds of options and features you'd have to pay extra for with anyone else. Combined with competitive pricing. And even a smaller footprint so you save on space, too.
It all adds up to a full family of technically superior standard switching power supplies. And a value you'll find hard to match. Because we've outclassed the other contenders in every round:
- CERTIFIED. The Zenith series meets international safery requirements of

UL 478; CSA 22.2, No. 220; IEC 380 \& 950; VDE 0806; VDE 0871/B; FCC PART J, CLASS B.
- FLEXIBLE. All outputs are independently isolated, mag amp regulated, and adjustable. Outputs 2 and 3 can be widely adjusted berween 10 and 15 V . to meet different equipment demands-at no extra cost!
- TOUGH. Exceeds 50,000 MTBF for durable performance.
- RELLABLE. Comes with 2 -year warranty, fully backed by Zenith.
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For optional steel cover substitute-C for- N in model number

Test it For 90 Days-Free. To find out how you can qualify, call today:
1-312-391-8700.


\section*{Hardware and Interconnect Devices}

41612, IEC 603-2, VG 95324, MIL-C-55302/131-134, and MILC55302/ 157-158. Types BK, CK, and C can have as many as 96 contacts; types D, E, and F have 32 or 48 contacts; and types Q and R have an inverse configuration. You can obtain the devices equipped for wire wrappable, wave solderable, or press fit termination. Type BK with 48 pins, \(\$ 2\); type BK with 96 pins, \(\$ 3\) (1000).

Stanford Applied Engineering, 3520 De La Cruz Blvd, Santa Clara, CA 95050. Phone (408) 9880700. TWX 910-338-0132.

Circle No 677


\section*{¼-IN.-HIGH SOCKETS}

The FCN723 2-mm-pitch discretewire connector is 0.256 in . high and accommodates wire sizes from 26 to 34 AWG. According to the vendor, board-to-board and wire-to-board interconnections made with dis-crete-wire sockets and boardmounted headers take up less space and permit better airflow than do ribbon-cable connections. The vendor also claims that a single connector can join three pc boards. The series includes straight- and rightangle styles with 2 to 18 pins in a single row and 4 to 36 pins in 2 rows. 30 -pin header and socket, \(\$ 0.80\).

Fujitsu Component of America Inc, 3320 Scott Blvd, Santa Clara, CA 95054. Phone (408) 562-1000. TWX 910-338-0190.

Circle No 678

\section*{SOCKETS}

Designed for burn-in service, these sockets accommodate 44- and 84-pin PLCC (plastic leaded chip carrier) devices. They have a locking mechanism that facilitates manual or automated loading and unloading, prevents damage to delicate leads, and ensures positive lead contact. A simple push seats the PLCC firmly in the socket with an audible click. A second push ejects the device above the socket edge for easy removal. Guided entry and alignment ribs ease the PLCC into proper orientation within the socket. An improved socket design provides more contact area at the top and sides of the leads to improve reliability. The sockets feature quick visual polarization, and its side and bottom vents allow increased airflow for heat dissipation, as well as access for test probes. 44-pin unit, \(\$ 9.98\); 84-pin version, \(\$ 15.12\) (1000).

3M, Dept EP87-109, Box 2963, Austin, TX 78769. Phone (512) 8341803.

Circle No 680


\section*{IC SOCKETS}

Type 105 and 117 IC sockets are designed for surface-mount applications. Type 105 units have angled pins (gull type) that provide easy access for in-circuit testing and troubleshooting. Type 117 units feature a floating-contact design that compensates for the effects of unevenly dispensed solder paste. Both types can accommodate most soldering processes that are used for sur-face-mount fabrication.

The insulator body is glass-filled thermoplastic polyester with a UL \(94 \mathrm{~V}-0\) flammability rating. The contacts use a 4-finger clip made of
stamped beryllium copper, gold, or tin plate over copper and nickel. The pins are screw-machined brass with tin plating over copper and nickel. Types 105 and 117 , with 28 pins and tin plating, cost \(\$ 1.75\) and \(\$ 1.65\), (100) respectively. Delivery, four to six weeks ARO.

IEE Inc, Component Products Div, 7740 Lemona Ave, Van Nuys, CA 91409. Phone (818) 787-0311. TLX 4720556.

Circle No 681


\section*{LCC SOCKETS}

These sockets can accommodate 68-pin LCC devices in actual-use applications-not just in burn-in service. They are available with or without aluminum heat sinks, in versions that mount to either JEDEC Type A or Type B ceramic leadless chip carriers. Their body material consists of polyphenylene sulfide, and their cover material is made of stainless steel. Their beryllium copper contacts feature gold-over-nickel plating. The sockets can withstand \(150^{\circ} \mathrm{C}\) operating temperatures. \(\$ 4.01\) (1000). Delivery, stock to six weeks ARO.

Nepenthe, 2471 E Bayshore Rd, Palo Alto, CA 94303. Phone (415) 856-9332.

Circle No 682

\section*{DATA LINKS}

Series 5660 analog fiber-optic data links consist of a transmitter, a receiver, and a 50 m cable. The transmitter and receiver are housed in 24-pin DIPs. The devices feature a \(10-\mathrm{kHz}\) bandwidth, \(\pm 0.025 \%\) FS linearity, an \(80-\mathrm{dB}\) signal-to-noise

\title{
The fact that we make the highest quality, most reliable MII-C-38999 SERTIES III connectors
}

\section*{is not a militiny secret.}

Our competitors know it, too, but they're not likely to blurt it out to anyone.
Our attitude is this. Whenever we design a critical connector line, we tend to engineer it with features and quality that exceed even the most rigid specifications. It's the only way we know to give you the reliability and dependability you have come to expect from us.
Our MIL-C-38999 connectors are ideally suited for the critical performance requirements of air craft and aircraft engines, as well as the rugged field-operation demands of weapon systems, military vehicles, back-pack radios and radar systems. In addition to our full complement of shell styles, classes and configurations, we offer qualified products within the whole D38999 SERIES III family to meet new and growing industry needs. These include such things as: Space-rated connectors to meet outgassing requirements, MIL-STD-

1553B units for multiplexed data bus systems and MIL-STD-1760 lanyard release connectors.
So, for those applications in which reliability and dependability absolutely cannot be compromised, specify Pyle MIL-C-38999 SERIES III connectors,
For more information on these and our excellent field support, call us or write us. Or.just for the fun of it, ask one of our competitors.
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PYLE-NATIONAL DIVISION
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Chicago, IL 60651

\section*{Hardware and Interconnect Devices}

ratio, and a \(100-\mathrm{ppm} /{ }^{\circ} \mathrm{C}\) temperature coefficient. The transmitter and receiver FS input and output voltages are pin-selectable for 0 to 1,0 to \(2, \pm 5\), and \(\pm 1 \mathrm{~V}\). The transmitter and receiver modules operate from a single supply of 11.5 to 20 V . The receiver includes a 3 -pole, active lowpass filter, which you can bypass to suit your application. \(\$ 200\). Delivery, six to eight weeks ARO.

Dymec Inc, 8 Lowell Ave, Winchester, MA 01890. Phone (800) 225-1151; in MA, (617) 729-7870. TWX 710-348-6596.

Circle No 683

\section*{INDUSTRIAL NETWORK}

The Efiway industrial control network lets you control as many as 8064 distributed I/O lines from an IBM PC/AT or a Multibus I. The token-ring network, which employs an optical fiber cable that operates at 3.5 M bits/sec, can accommodate as many as 63 computers or I/O stations.

Communication between a computer and the network occurs via a dual-port RAM on the computer's network interface card. Automatic packet transmission over the network transfers the contents of this RAM into remote I/O locations or into the dual-port RAM of other network computers. In addition, the input states of remote input channels are transferred back into the computer's dual-port RAM. These transfers occur automatically, without software support from the control computer; consequently, the computer perceives the system as a
memory-mapped system rather than as a network. The vendor estimates the cost of a 5 -station network with one intelligent station at around FFr 150,000.

Efisysteme, 21-43 rue de la Grande-Charriere, La Boisse, 01120 Montluel, France. Phone 78062155. TLX 340821.

Circle No 708

\section*{INDUSTRIAL NETWORK}

The Signatrans-ZM50 modular industrial networking system, which operates over 2 -wire cabling, lets you transmit data and commands between as many as 256 network stations. It is suitable for use in data-acquisition and process-control applications that employ a range of sensors and actuators.

When the networking system is running in its simplest operating mode, you don't need an intelligent network controller. Rather, you can cause the network to automatically transfer data from the input to the output channel simply by setting DIP switches to allocate the same address to an input channel and an output channel. You can set up as many as 128 of these I/O pairs and can join input and output channels anywhere in the network.

You can use a handheld programming unit to program the system to perform point-to-multipoint transfers, so that the information on one input channel automatically appears at several output channels. Alternatively, you can add to the network a communications processor station, which provides terminal control of the network as well as a gateway to other networks. The network can have open-ended or ring topology and can be as long as 20 km . A typical I/O station, with modules to provide analog and digital I/O capability, costs around DM 2000.
Funke \& Huster GmbH, Langemarckstrasse 28, 4300 Essen 1, West Germany. Phone (0201) 22091. TLX 857637.

Circle No 709


\section*{BACKPLANE}

Suitable for use in 32-bit VME Bus systems, V316 Series double-height VME Bus backplanes incorporate both J1 and J2 connectors. The units currently come in 9 -, 12 -, and 20 -slot versions, and a 5 -slot version is in development. The boards feature DIN-41612 connectors and passive termination networks. They employ a tracking arrangement that provides a ground screen around every signal track to reduce crosstalk and ringing.

The units incorporate the busbusy signal routing as recommended in the latest revision of the VME Bus specification. The boards have full copper grounding and +5 V power planes that extend across the J1 and J2 sections of their backplanes. You make power connections to the board via spade connectors. £145 (100).
Dage (GB) Ltd, Rabans Lane, Aylesbury, Bucks HP19 3RG, UK. Phone (0296) 393200. TLX 83518.

Circle No 710
Dage Precision Industries Inc, 46701 Fremont Blvd, Fremont, CA 94538. Phone (415) 683-3930.

Circle No 711

\section*{BREADBOARDS}

X-tra Edge-style solderless breadboards feature an extra multiuse edge panel for organizing and mounting components that don't fit

Text continued on pg 241

\title{
Bendix Connectors tomorrow's technology today
}


\section*{LRM Surface Mount connectors with Bristle \({ }^{\circledR}\) Brush \(^{\circledR}\) contacts for integrated avionics packaging}

Designed to meet the high density needs of today's integrated electronic modules, this Straddle Mount connector uses the Bendix \({ }^{\circledR}\) Brush \({ }^{\circledR}\) contact which has been proven in military avionics packages, is qualified to MIL-C-55302 and is the consensus choice for integrated avionics systems. Low mating force, extended service life and stable electrical performance of the Brush contact provides the high level of performance demanded by today's Line Replaceable Module (LRM) applications.

The Bendix LRM connector is available with 300 or more electrical contacts and up to eight MIL-C-38999 style fiber optic termini. Optional features include first make/last break grounding for ESD protection, EMI shielding, individually replaceable contacts, polarized shells, numerous keying combinations and accomodation of a wide range of PC board/heat sink combinations.


\section*{MIL-C-38999 Series III intermateable Composites for weight savings and corrosion resistance}

Weight savings are often critical. Bendix \({ }^{\circledR}\) Tri-Start \({ }^{\text {TM }}\) Composite connectors afford \(17 \%\) to \(40 \%\) weight savings when compared to standard aluminum product. When compared to stainless steel product, these connectors provide weight savings of \(55 \%\) to \(80 \%\).

Corrosion resistant capabilities surpass that of standard aluminum connectors. Composites will withstand extended exposure to salt spray with no evidence of corrosion. EMI shielding effectiveness capabilities exceed requirements of Military Specification MIL-C-38999 Series III.
Intermateable with metal MIL-C-38999 Series III connectors, Bendix Composites can be designed into new applications or used on existing programs where Series III requirements exist. Commonality continues with the use of standard MIL-C-38999 Series III inserts, contacts and assembly procedures.
For more information on Bendix \({ }^{\circledR}\) connectors, contact:
Amphenol Corporation
Bendix Connector Operations
40-60 Delaware Street
Sidney, NY 13838-1395
607-563-5302
an \(\mathscr{L} \mathscr{P} \mathscr{L}\) company

\section*{A chip carrier socket that wont play "pop goes the circuit."}


The contacts are High Pressure Tin, an AMP proprietary design which creates very high normal forces-a minimum of 200 grams per contact-for maximum retention and reliable interconnection. Short-signal-path contacts float in the housing to accommodate thermal expansion.

Two basic styles of sockets are available: square or 32-position rectangular EPROM and SO-J. Both come in solder

tail or surface mount versions and feature all the important details. Tin-over-nickel plating is applied after the contacts are formed, to assure full plating. We've built in visual indicators for locating pin 1 , and polarizing to aid correct insertion.

Orientation holes in the \(94 \mathrm{~V}-0\) housing floor make registration to the
pc board both fast and simple, ideal for hand or tube-loaded robotic insertion. And the high pin counts make very effective use of real estate.

Call the AMP Information Center at 1-800-522-6752 for literature on HPT PLCC Sockets. AMP Incorporated, Harrisburg, PA 17105-3608.

\section*{AM| Interconnecting ideas}


\section*{Schroff}

\section*{OUR CABINETS WITHSTAND THE TOUGHEST TEST OFALL THE TEST OF TIME}

As a company dedicated to technical excellence, Schroff helps you prepare for the future. So you can be the first to take advantage of it. For example, our Eurorack and Minirack cabinets are designed to help you meet changing market conditions. They're available in more than 45 different sizes, with a wide range of accessories and options, including RFI/EMI shielding, full glass doors (a Schroff exclusive), and much more. So chances are, you'll always be able to offer the features your customers are looking to buy.
Schroff Eurorack and Minirack cabinets are manufactured in the United States. They meet EIA, IEC, VDE and DIN standards. And they're made to deliver maximum rigidity and strength. As a result, you, and your customers, can be sure these cabinets will last as long on the job as they will in the market.
If your present cabinets can't pass the test of time the way Eurorack and Minirack can, it's time you looked at Schroff.

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\section*{Hardware}

into DIP-spaced breadboard socket connections. The boards come in four models that offer from 810 to 2940 tie-points. Each model includes an area containing both distribution and terminal strips that accommodate all DIP sizes, lead components, and wire gauges of 20 to 29 AWG. The phosphor bronze, nickel-plated contacts provide a \(3-\mathrm{m} \Omega\) initial contact resistance and a 10,000 in-outcycle min lifetime rating. Each breadboard features four multipurpose binding posts. \(\$ 16.95\) to \(\$ 59.95\).

Chenesko Products Inc, 21 Maple St, Centereach, NY 11720. Phone (516) 736-7977.

Circle No 684


OPTICAL MODULES
HFBR-X400 fiber-optic transmitter and receiver modules mate directly with both AT\&T's ST connector and with bayonet-style connectors from various manufacturers. You can use them in computer-communication channels, industrial systems, and data-communication links of 3 km or less. The modules interface with \(62.5 / 125-\) - \(50 / 125-\), and \(100 / 140-\mu \mathrm{m}\) multimode fibers.

The module line includes the HFBR-1412, a standard transmitter; the HFBR-1414, a high-power transmitter; the HFBR-2412, a 5M-


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\section*{19" Card Racks allow board size flexibility.}

Design printed circuit boards to the size you need! Amlan's adjustable 19" card racks combine circuit board size flexibility with high quality and a handsome appearance.

Four standard card rack sizes accommodate PCB heights from \(13 / 8^{\prime \prime}\) to 9 13/16". Each card rack accepts board depths from \(31 / 2^{\prime \prime}\) to \(79 / 32\) ".

Fabricated with high strength, clear anodized extruded aluminum and zinc plated steel parts, these rugged card racks offer vibration resistance and ease of assembly. Accessories include aluminum front panels.
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\section*{Hardware and Interconnect Devices}
baud receiver; and the HFBR-2414, a \(25-\mathrm{MHz}\) analog receiver.

The transmitters and receivers come in autoinsertable and wavesolderable DIPs. When operating at \(40^{\circ} \mathrm{C}\), the units feature an MTBF in excess of 5 million hours. To ease the evaluation process, you can order a kit containing a transmitter, receiver, and 3 m of connectorized cable.

Modules, \(\$ 12.50\) to \(\$ 23\) (1000); kit (HFBR-0410), \$49.95.

Hewlett-Packard Co, 1820 Embarcadero Rd, Palo Alto, CA 94303. Phone local office.

Circle No 685

PROTOTYPE PANEL
The 8136-VME940-03D is a

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\(9 \mathrm{U} \times 400-\mathrm{mm}\) wire-wrappable panel for use in VME Bus applications. It features 4-layer construction-two ground planes and two \(\mathrm{V}_{\mathrm{CC}}\) planesthat maximize power distribution. Integral surface-mounted decoupling capacitors increase distributed capacitance and decrease overall impedance, making the panel compatible with high-speed logic devices. The panel accommodates as many as 595 16-pin DIPs when you also use the PGA (pin grid array) areas for the ICs. The PGA areas withstand the high power dissipation that is typical of devices in PGA packages. \(\$ 1583\).

Augat Inc, Box 1037, Attleboro, MA 02703. Phone (617) 222-2202. TWX 710-391-0644.

Circle No 686


\section*{SOCKET}

The PLCC-068 socket accepts 68-pin PLCC (plastic leaded chip carrier) devices. The socket is JEDEC-qualified for type-A carriers and features an \(8 \times 8\)-position contact array that places contacts on \(0.050-\mathrm{in}\). centers. Its glass-filled polyphenylene sulfide insulator carries a \(94 \mathrm{~V}-0\) UL rating, and its

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As technology advances in outer space and undersea, you will need resources to supply parts and components that fulfill the promise of high technology in design, materials, quality and cost efficiency. Utitec knows this. That's why it applies all the tools of Manufacturing Resources Planning (MRP) and Statistical Process Control (SPC) to the development and production of miniature, deep-drawn tubular parts.

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\section*{Hardware and Interconnect Devices}
preloaded, stamped-and-formed, copper-alloy contacts have a tin-lead plating. The socket features a compact low-profile design and perforated construction to ease board cleaning. The socket's design lets you probe the chip-carrier pins in the circuit. The socket operates over -65 to \(+125^{\circ} \mathrm{C}\). \(\$ 3\) (100). Delivery, stock to eight weeks AR0.

Precicontact Inc, Box 798, Langhorne, PA 19047. Phone (215) 757-1202.

Circle No 687

\section*{SOCKET SYSTEM}

The CHG wire-mount socket system is designed for signal-transmission and logic-power applications re-

quiring current ratings between 1 and 3A. Two contact options accommodate \(22 / 24\) or \(26 / 28\) AWG dis-crete-wire or prenotched ribbon cable on a \(0.1-\mathrm{in}\). grid spacing.

You can select 1- and 2-row sockets containing from 2 to 40 or 4 to 80 contacts without polarization or with center-bump polarization. The 2 -row military and DIN versions contain from 10 to 80 contacts and are available with or without the center bump. The contact plating includes a choice of \(0.1-\mu \mathrm{in}\). tin-lead (entire contact) or 10 or \(30 \mu \mathrm{in}\). of gold over \(50 \mu \mathrm{in}\). of nickel. \(\$ 0.96\) (1000) for a 1-row, 10 -position socket with \(30 \mu \mathrm{in}\). of gold plating and no center bump.

3M, Dept EP7-99, Box 2963, Austin, TX 78769. Phone (512) 8341800.

Circle No 688


\section*{CARD FRAMES}

HD167 card frames meet MIL-STD167 and the IEEE mechanical specification draft P1101. The units are constructed of \(1 / 8-\mathrm{in}\). aluminum plate and extrusions, which are conductively finished. Their features include positive locking bars for card guides, which prevent rollover; conductive and nonconductive spacers for use in mounting backplanes; and

\section*{Hardware and Interconnect Devices}
beryllium-copper ground clips to selectively control EMI/RFI emissions on daughter boards. The card frames are available in five standard sizes to suit buses such as the VME Bus, Multibus II, G-64, and Futurebus. The card guides can accept either \(1 / 16\) - or \(3 / 32\)-in.-thick pc boards with \(151 / 4-\) in. depths. From \(\$ 75\).

Bicc-Vero Electronics, 1000 Sherman Ave, Hamden, CT 06514. Phone (203) 288-8001.

Circle No 689


\section*{COAXIAL CABLE}

This subminiature ribbon coaxial cable is highly flexible. You can fold the cable upon itself, bundle it in rectangular or round sections, or group it with other cable for routing.

To terminate each signal set, you secure the signal conductor and companion drain wire to the appropriate connector terminals. The vendor can provide custom-designed multilayer paddle cards or pc boards to eliminate impedance mismatch. You can order the cable in varying lengths, with six to 64 signal conductors, single or dual drain wires, and impedances of 50 to \(130 \Omega\). \(\$ 1.25 / \mathrm{ft}\) ( 100 ft ).

Woven Electronics, Box 667850, Charlotte, NC 28266. Phone (803) 963-5131.

Circle No 690

\section*{BURN-IN SOCKETS}

These low-insertion-force, burn-in sockets come in 20 - to 68 -position versions. They accept square, plastic chip carriers with J leads on \(0.05-\mathrm{in}\). centers.


The units feature liquid crystal polymer housings rated for continuous operation at \(200^{\circ} \mathrm{C}\). You can obtain them with three types of contacts: beryllium copper rated to \(200^{\circ} \mathrm{C}\), beryllium copper rated to \(150^{\circ} \mathrm{C}\), and phosphor bronze rated to \(125^{\circ} \mathrm{C}\).

The contacts feature nickel-boron platings. The sockets have metallocking frames that reduce insertion forces, and positive ejection systems that facilitate manual or automatic loading and unloading and that also
improve heat exchange. Their insertion life specs at 5000 cycles min. From \$8, depending on model and quantity.
Mark Eyelet Inc, 63 Wakelee Rd, Wolcott, CT 06716. Phone (203) 756-8847. TWX 510-600-7291.

Circle No 691


\section*{CONNECTORS}

Mini-Edge pc-board connectors have \(0.35-\mathrm{in}\). profiles and \(0.050-\mathrm{in}\). center-to-center pin spacings. Each of the connectors features 26 -contacts arranged in two 13-position rows and gold-over-nickel plating on

Text continued on pg 249


\section*{We believed so strongly in} Plessey Microsystems' commercial and military computer products,

we bought the company:

\title{
Introducing Radstone Technology, a new name with more computer experience than any company in the business.
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If you've been following the commercial or military microcomputer industries in recent years, you know that the hot company has been Plessey Microsystems. In VME, for example, we soared from entry level to Number 3 worldwide in just a few short years.

As the management team of Plessey Microsystems, we saw an opportunity to go even higher. Our product development and technical support programs were clearly the class of the industry. But still we weren't satisfied. So, we decided to position the company even better to meet the needs of the commercial and military markets we serve. To do this, we bought the company from our corporate parent, Plessey.

\section*{A brand new,}

30-year-old company
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Today, Radstone Technology means business as usual, only better. We've intensified our product development efforts to further accelerate our growing lead in the introduction of advanced products that match market needs precisely.

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Radstone Technology is built on a philosophy that addresses board level hardware, software and systems solutions. We combine the solid technical and marketing bedrock of a 30 -year industry veteran with the flexibility, innovation and responsiveness of a brand new company. A perfect blend: the wisdom and experience of age, the enthusiasm and agility of youth.

For those of you used to dealing with us as Plessey Microsystems, look for an increase in the excellence you've become accustomed to. For those of you yet to take advantage of our proven problem solving strengths, prepare to be impressed.

As for our relationship with our former parent, Radstone Technology will continue to be a valued supplier of board and system level products to Plessey's vast telecommunications and defense efforts. We'll also participate in a continuing technology transfer with Plessey through a cross-licensing arrangement. We couldn't ask for a better partner or customer...nor they for a better supplier. The best is getting even better
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\section*{Hardware and Interconnect Devices}
the card-edge contact areas; the individually replaceable berylliumcopper contacts apply continuous normal force to \(0.022-\mathrm{in}\). platedthrough holes and are arranged in pairs to accommodate double-sided pc boards.
The connectors' 94 V -0-rated polyester housings resists the high temperatures of flow soldering. Their molded-in standoffs facilitate cleaning processes. \(\$ 5.83\) (1000). Delivery, four to six weeks ARO.
Amp Inc, Box 3608, Harrisburg, PA 17105. Phone (717) 564-0100.

Circle No 693

\section*{FLAT CABLE}

In addition to providing high impedance and low capacitance, the 3751 PVC flat cable offers full shielding. It's designed for applications involving certain DEC Q Bus- and Unibuscompatible peripherals.
The cable helps you comply with


FCC regulations for EMI/ESD protection; it features a full \(360^{\circ}\) wrap of extended copper shield. Dielectric PVC material, located above and below the cable inside the shield, maintains precise shield-cable spacing, thus providing the high-impedance and low-capacitance values.
The cable is available with 26,40 , and 5028 AWG conductors on a \(0.05-\mathrm{in}\). center pitch. The impedance measures \(90 \Omega\) with connectors in ground-signal-ground configuration and \(115 \Omega\) in the signal-toground shield configuration. The capacitance specs at \(17.4 \mathrm{pF} / \mathrm{ft}\).
\(\$ 3.20 / \mathrm{ft}\) ( 1000 ft ) for 40 -conductor cable.
3M, Box 2963, Austin, TX 78769. Phone (512) 834-1800.

Circle No 692


\section*{CONNECTORS}

These Amplimite 15-position subminiature D connectors feature preassembled slide latches and locking

\section*{"Erector Set"}


We'll put this versatile TRU-GLIDE Series 375QD steel chassis slide together for you but then we offer a wide choice of adjustable mounting brackets and insertable bar nuts to make it easy for you to finish your chassis installation in any standard 19 " cabinet. Thin design, low profile, as well as smooth balt bearing action and quick disconnect features are pleasant alternatives to the cumbersome slip-stick friction slides now in use. Get complete information on this versatile \(75-\mathrm{lb}\). capacity slide series for your own cabinet... and they're off the shelf!


Jonathan's extensive line of aluminum slides are industry standards in military and aerospace application - with load ratings to 800 lbs .

\section*{Hardware and Interconnect Devices}
posts that save valuable installation time and effort. Compatible with the IEEE-802 standard for LAN interfaces, these tin-plated, metalshell plug and receptacle connectors offer a lifetime spec of 500 mating/ unmating cycles. The plug shells feature grounding indents for EMI/ RFI compatibility. Slide-latch version, \(\$ 2.30\); locking-post model,
\(\$ 1.44\) (1000). Delivery, stock to six weeks ARO.
AMP Inc, Box 3608, Harrisburg, PA 17105. Phone (717) 564-0100.

Circle No 694

\section*{BACKPLANE}

This 12-slot, 9U backplane meets all specifications for any Sunbus mod-

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POMONA is the original line of quality test accessories. The standard for comparison throughout the industry since 1951. So it's little wonder that we are proud of our leadership position. We intend to stay in front with new test products as well as the old reliable ones, and earn your confidence for years to come.

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III Pomona Electronics
}

els that require such a configuration. Based on transmission-line technology, the backplane is mechanically compatible with both VME and Sun card cages. It duplicates the Sun backplane and includes the signal busing necessary to implement the Sun proprietary P2 bus. The 5-layer combination microstrip/stripline design provides high-performance operation to 40 MHz . Bus bars facilitate the power hookup and disconnection. \(\$ 1275\).

Dawn VME Products, 47073 Warm Springs Blvd, Fremont, CA 94539. Phone (415) 657-4444.

Circle No 697

\section*{PROTOTYPING BOARD}

The MD12BRD 4-layer universal prototyping board provides an easy way to evaluate high-speed GaAs logic devices. The board, which has nine sites for 16-pin flatpack ICs and three sites for 20 -pin flatpack ICs, offers easy access to power-supply and RF interconnections. It is suitable for applications reaching impedance levels as high as \(50 \Omega\) and clock speeds as high as 5 GHz . The unit comes with application and assembly information, as well as a list of recommended capacitors, termination resistors, heat sinks, and RF interconnections. \(\$ 100\).
California Eastern Laboratories Inc, 3260 Jay St, Santa Clara, CA 95054. Phone (408) 988-3500.

Circle No 696

\section*{ELECTRIFYINGSOLUTIONS:}


130 to 4000 Watts • SELV Magnetics • 2 to 48 Volts • Made in USA • Rugged Packaging • Fully Regulated • Up to 400 Amps Thermal Protection • Short Circuit Proof • 50 Degree C Power Rating • 3750 VRMS I/O Isolation • Optimum Price/Performance 3601 VETERANS HIGHWAY
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\title{
Integrated tool sets simplify software cross-development
}

The prevalence of various types of workstations has not only automated some software-development tasks-it's also created a need for sophisticated crossdevelopment tools. Using these packages, you can now write software more easily and efficiently for embedded systems or for structurally dissimilar \(\mu\) Ps.

Chris Terry, Associate Editor

Three trends in the computer industry have made good cross-development software tools a necessity rather than a luxury: First, software developers are now using powerful yet relatively inexpensive workstations, with graphics and computational capabilities undreamed of only a short time ago, in large numbers. Second, disparate architectures for workstations and other processors are increasing. Third, the use of embedded microprocessor systems is growing rapidly within application areas such as laboratory instruments, domestic appliances, and process-control systems of all kinds. From these concurrent trends has arisen a strong demand for tightly integrated tool sets that provide all of the facilities necessary for generating, testing, debugging, and integrating software that will run on some specific target system, which is quite different from your development machine.

\section*{Early tools were makeshift}

The idea of using the resources and speed of a large computer to develop software that can run on a machine that lacks development tools is by no means new-some cross-development tools appeared as early as 1970. But at that time, cross-assemblers were scarce, and symbol-


A graphics editor for simplifying microcode development (Trimeter Technologies Corp)
ic debuggers were nonexistent. As for cross-compilers, programmers often defined them as "compilers that are guaranteed to make their users cross."
In those early days, the need for cross-development tools was felt only at a relatively few large installations. Dun \& Bradstreet, for example, had IBM mainframes at their central database installation and Quotron 16-bit minicomputers at their 75 field offices. Because these minicomputers had only a rather primitive assembler and lacked any high-level-language (HLL) tools, Dun \& Bradstreet was forced to commission a Fortran crossassembler, and later a Fortran cross-compiler, so that it could use mainframe resources to develop new application programs for its minicomputers within a reasonable time. Those tools did work, but they were slow and awkward to use.
Modern cross-development tools differ from their precursors in several ways. Intensive, long-term studies of translator principles have yielded algorithms for very sophisticated parsing and efficient code generation. In addition, you can now choose to optimize the code either for speed or for the most effective use of limited memory (there's usually a tradeoff between those two goals). Furthermore, if you're designing
software for embedded systems, you can find compilers that can generate ROMable code for embedded systems. Finally, tightly integrated, retargetable tool sets are replacing single-purpose compilers for cross-development.
Single-purpose compilers and cross-compilers are limited to one language and one target machine. They generate either native machine code for the target or assembly-language source code. If you obtain the latter, you use a cross-assembler to assemble it into relocatable object code. You can then link that object code with routines from the library that the vendor generally supplies with a compiler. In either case, the compiler produces code for only one kind of target processor. Moreover, each compiler vendor tends to use a proprietary set of directives and switches for optimizing code, handling errors, and listing features. So if you change compilers, you have to learn a new set of rules for using them.

\section*{Cross-compilers translate HLLs}

Any of the new tool sets can provide all the softwaredevelopment facilities you need for a number of different target processors. The compiler for a given lan-

Finally, tightly integrated, retargetable tool sets are replacing single-purpose compilers for cross-development.


A typical cross-debugger with multiple windows (Applied Microsystems Corp)
guage usually consists of a front end that translates HLL source code into a processor-independent intermediate language, optimizes the code, and generates a symbol table. A symbolic debugger can then use this table to display HLL source code while executing the corresponding machine code.

\section*{Modular structure makes compilers retargetable}

The addition of a common intermediate language has several important advantages. For any given HLL front end, the vendor can provide a number of hardwarespecific, back-end modules that translate the intermediate language into the native machine languages of different target machines. Conversely, for any given target, the vendor can match the back-end module to compatible front-end compiler modules for different high-level languages. Thus, a tool set can support multiple languages and multiple target types, merely by matching the appropriate front-end and back-end modules. The process is simple, economical, and expandable.

What's more, you can easily call routines written in assembly language from any point in the HLL code.

Although modern compilers generate faster code than their predecessors, you may still need to write the most time-sensitive routines in assembly language to obtain maximum execution speed - especially for real-time embedded systems.

From the user's point of view, the biggest advantage is that the human interface, including all directives and switches, is constant no matter how many different HLLs and target machines you work with. In this way, integrated tool sets can streamline the schedule for any large software project significantly and thus cut costs.

\section*{C and Ada prevail}

You can find cross-development tools for a variety of programming languages, including C, Pascal, Ada, Forth, Modula-2, PL/M, and others, but the predominant languages are C, for commercial products, and Ada, which is mandatory for all projects related to the US government. C has become popular because it provides good low-level access to the hardware. Unlike Basic and Pascal, it hasn't fragmented into numerous dialects. Also many programmers learn to work in C on a Unix machine in college. Whether C is the ideal language for embedded systems is questionable (you could start a riot by even suggesting such a heresy to Forth, Pascal, or Modula-2 proponents), but it is widely accepted, relatively standard, and conditionally portable. Consequently, commercial tool-set vendors who don't wish to support multiple HLLs generally choose C.

The Intertools series from Intermetrics exemplifies highly integrated cross-development tool sets. It contains a retargetable cross-compiler for C, or Pascal; a run-time library and a librarian for managing program libraries; a linker; a locater that closely controls how and where data and code are placed in the target machine's memory; a ROM processor that automates the downloading of initial data values from ROM to the target's RAM when the program starts executing in the target system; and a source-level debugger that can operate in conjunction with a variety of in-circuit emulators. Ancillary tools include a cross-assembler for the target machine, and a formatter for downloading code to the emulator or target. The formats provided include Intel hex, Motorola S-records, and other industry standards.

Intertools is available for VAX/VMS and VAX/ Ultrixc computers; Hewlett-Packard, Sun, and Apollo workstations; and IBM PCs and compatibles. Its target machines include the Intel 8086/186/286 family; Motoro-


The major steps in the cross-development process (Integrated Systems Inc)
la's 6800, 6809, and 68000 families; Zilog's Z80; NEC's V60 and V70; and a variety of 8-bit microcontrollers. The company recently added AMD's 29000 RISC processor to its target list. Compiler prices range from \(\$ 1000\) for a PC host to \(\$ 7000\) for a VAX 8800 host; the corresponding debuggers range from \(\$ 1500\) to \(\$ 8500\).

Intermetrics also sells Ada tools that include a realtime compiler; it runs on IBM mainframes under MVS or VM/CMS. An Ada cross-compiler is also available and runs on VAX machines and generates code for MIL-STD-1750A processors.

\section*{Some tool sets generate microcode}

If you want to develop systems based on bit-slice or other processors for which you have to write the microcode, you have a choice of integrated tool sets too. Step Engineering sells the Metastep Microprogram Language System for \(\$ 3000\). This tool set lets you define the microcode you'll use, and develop microprograms on an IBM PC/AT or compatible. To test the software, you also need MicroStep (\$3695), which includes a plug-in interface card, one or more ROM/ PROM/RAM pods, and a debug and control package. The system supports 25 -nsec Writable Control Store memories as large as 4 k words deep and 128 bits wide.

Another set, the Microcode Assistant from Trimeter Technologies Corp, runs on Apollo workstations and lets you create microword formats graphically, construct complete microprograms using these formats, and then interactively validate their operation. The package has four parts: Using the format editor, you can customize the whole tool set for a specific hardware architecture. With its microprogram editor, you can create and manipulate microcode at the word, block, or
macro level. Its mapping format editor maps, with bit-mapped graphics, whatever is the most convenient logical format into the physical format required by the hardware. You can use its multiwindowed simulator to interactively test, debug, and simulate your microprograms. The complete package costs \(\$ 16,100\).

\section*{Debugging approaches vary}

For embedded systems, you face the compounded problem of trying to write software for still unfinished hardware, whose characteristics are not fully tested. Therefore regardless of what precautions you take to prevent errors in those situations, eventually you must marry the software to the hardware and then thoroughly test and debug it. An in-circuit emulator is essential for running the software and monitoring its performance. Usually a high-speed RS-232C serial line (sometimes a parallel data line) links the emulator to the host development system. This link allows the debugger, running on the host, to pass commands to the emulator and to receive from it reports on all aspects of the performance.
There are two schools of thought on how best to accomplish the integration of the hardware and software. The traditional approach is to supply the design requirements both to the hardware designers and the software designers and to allow them to work more or less independently until the integration phase of the project. At that stage the cross-debugger and in-circuit emulator provide the first indications of how successfully the two groups have met the design requirements.

Given two good teams, whose members communicate with each other throughout the development stages, this approach can work well. If communication in the early stages is inadequate, however, the integration phase may reveal bugs and incompatibilities that can be difficult and expensive to correct. The increasing use of CAE and CASE tools is helping both to bring hardware and software groups together at an earlier stage, and to reduce the time needed to correct any operational problems that arise, but these tools are only just beginning to generate trustworthy code automatically.

\section*{Simulation helps software design, too.}

As hardware costs drop and the computing power and speed of workstations increase, simulators are becoming more and more valuable. Hardware designers, particularly IC designers, have been using simulators for several years. These products can simulate very precisely on a computer the operation of a proposed

\section*{Any of the new tool sets provides all software-development facilities you need for a number of different target processors.}
logic design and so detect timing errors and unforeseen interactions between logic elements, long before the designer is committed to hardware (or silicon).

Recently, several vendors of software development tools have introduced simulators that can execute the instruction set for a target machine entirely in software on the host computer. Thus you can exercise and debug your software at speeds close to real time, and discover potential problems long before the target hardware is operational. In addition, because you can test a software module as soon as it's complete, you can alert the hardware designers to any problems you find. The simulator won't eliminate the need for the final integra-
tion phase in which you use an emulator and the actual hardware, but it can help coordinate the hardware and software aspects of designing into a more smoothly running process.

Applied Microsystems offers Validate/Xel, a sourcelevel cross-debugger and simulator package that supports target systems based on Motorola's 68000 microprocessor family. The cross-debugger works with the vendor's ES 1800 series of 16 - and 32 -bit emulators. The simulator runs compiled code and provides the same debugging facilities as the cross-debugger. Thus you have the opportunity to use the newer methods while development is underway but to revert to traditional

\section*{For more information . . .}

For more information on the cross-development tools discussed in this article, contact the following manufacturers directly, or circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.
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\hline Applied Microsystems Corp & Avocet Systems Inc & Micro Focus & Sierra Systems \\
\hline Box 97002 & Box 490 & 2465 E Bayshore Rd & 6728 Evergreen Ave \\
\hline Redmond, WA 98073 & Rockport, ME 04856 & Palo Alto, CA 94303 & Oakland, CA 94611 \\
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\hline Integrated Systems Inc & First Systems Corp & Microtec Research & SofTech Inc \\
\hline 2500 Mission College Blvd & 865. Manhattan Beach Blvd & Box 60337 & 460 Totten Pond Rd \\
\hline Santa Clara, CA 95054 & Manhattan Beach, CA 90266 & Sunnyvale, CA 94086 & Waltham, MA 02254 \\
\hline (408) \(980-1500\) & (213) 546-5581 & (408) 733-2919 & (617) 890-6900 \\
\hline TLX 5596631 & Circle No 372 & Circle No 379 & Circle No 386 \\
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\hline & Harris Corp & National Semiconductor Corp & Symbolics Inc \\
\hline Intermetrics Inc & Semiconductor Sector & Box 58090 & 11 Cambridge Center \\
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\hline & & Oasys & Telesoft \\
\hline Step Engineering & Hewlett-Packard & 230 Second Ave & 10639 Roselle St \\
\hline Box 3166 & 1501 Page Mill Rd & Waltham, MA 02154 & San Diego, CA 92121 \\
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\hline 200 Hightower Blvi, Suite 100 & 444 Castro St & Portland, OR 97219 & Berkeley, CA 94710 \\
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\hline & & Quantitative Technology Corp 8700 SW Creekside Pl & Whitesmiths Ltd 59 Power Rd \\
\hline identified the following & 1255 Post St, Suite 625 & Beaverton, OR 97005 & Westford, MA 01886 \\
\hline companies as suppliers of & San Francisco, CA 94109 & (503) 626-3081 & (617) 692-7800 \\
\hline cross-development tools: & (415) 771-7001 & Circle No 383 & Circle No 390 \\
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\hline Archimedes Software & Intel Scientific Computers & Box 61029 & \\
\hline 2159 Union St & 15201 Greenbriar Pkwy & Palo Alto, CA 94306 & \\
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methods for the final check. The package is intended for use with Microtec Research's MCC68K C cross-compiler. Prices range from \(\$ 5000\) for an IBM PC or compatible host, to \(\$ 18,000\) for a VAX/VMS host.

\section*{Simulator acts as real-time embedded system}

Integrated Systems Inc offers a simulation tool that is available separately but is a part of a complete CASE system called Autocode. The system is a complete set of design tools for real-time embedded systems, one of which is a module called the Graphical Programming Environment, and the others are a simulator and an automatic code generator. The simulator imitates not only the proposed hardware/software target system, but also the peripheral signals the system responds to and controls. The automatic code generator accepts input from the database of the graphical programming environment, and generates optimized, real-time source code in C, Ada, or Fortran. The source code is ready for compilation or cross-compilation with standard tools. Prices start at \(\$ 10,000\) for each code generator, \(\$ 19,000\) for the graphical programming and simulation tools, and \(\$ 29,000\) for the complete package.

Richard C. Jensen, vice president of Product Development at Applied Microsystems Corp, notes that CAE tools have given design engineers, manufacturing and test engineers, and personnel in auxiliary services, a common language that improves communication between departments considerably. He looks forward to a time when CAE and CASE tools will similarly lead to closer collaboration between the hardware and software designers. Simulators can certainly play a significant role in achieving that goal.

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2. Terry, Chris, "Cross-development tools for PCs and minis let you devlop software for 8 -bit \(\mu \mathrm{Ps}, " E D N\), April 15, 1987, pg 89.

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\section*{Quantum}

\title{
Test-program generator with AI diagnoses populated circuit boards
}

The DES (diagnostic expert system) employs artificial-intelligence techniques to quickly generate debugging routines for the firm's line of analog/digital test systems costing under \(\$ 10,000\). Test programs developed with DES prompt the operator to check nodes, tweak adjustments, and replace components in populated pe boards under test.

Instead of using traditional step-by-step programming, the test-program generator learns to recognize fault conditions in your boards. You create a fault dictionary by either simulating failures in software or adding fault conditions to the dictionary as they occur during normal testing. You expand the software's database of error patterns and their associated causes by simply typing

in the cause of new errors as you encounter them.

Using a graphics editor, you can manually enter analog and digital test vectors, reference data, and tolerance guardbands. Alternatively, the software can learn these parameters from the system under test by
analyzing a statistically valid sample of known-good boards. With proper format-conversion routines, the software can also accept test vectors from software circuit simulators.

The software can also capture a sequence of manually executed test steps, store the sequence as a macro, and then later execute the macro automatically. DES costs \(\$ 1985\), and test systems start at \$6669. A DES demo disk is free.

Array Analysis Inc, 200 Langmuir Lab, Brown Rd, Ithaca, NY 14850. Phone (607) 257-6800; in NY, (800) 451-8514.

Circle No 444

\section*{Unix utility converts 8086 code into executable 68020 code}

XDOS converts binary-code programs written for the IBM PC to binary images for Unix computers. This Unix utility program for 68020based computers includes a binary compiler that performs the code conversion and an emulator for MS-DOS. It permits the end user to simultaneously execute multiple, converted, IBM PC programs in Unix windows.

In a two-stage conversion process, XDOS converts MS-DOS programs without modifying them. First, the binary compiler analyzes instruction flow and generates a proprietary, intermediate data format. Then, the optimizing compiler generates executable code for the target system.

After the compiler has performed the conversion, the end user can

directly execute the program on the Unix system, because the XDOS utility maps MS-DOS, MS-DOS BIOS, and hardware system calls to Unix calls, and also manages calls that invoke MS-DOS data structures.

Programs converted with XDOS are not affected by the MS-DOS limit of 32 M -byte disk volumes and can therefore use the full Unix disk capacity. The programs can read
and write Unix files because the package maps the MS-DOS files onto Unix files. XDOS also provides a Unix utility that reads MS-DOS files. End-user pricing ranges from \(\$ 425\) to \(\$ 2000\), depending on the number of users the Unix system supports.

Hunter Systems, 444 Castro St, Mountain View, CA 94041. Phone (415) 965-2400.

Circle No 445

\title{
PLD design software accepts designs as schematics, waveforms, or tables
}

The 74150A PLD Design System software accepts PLD design specifications as Boolean equations, schematics, waveforms, and state-transition or truth tables. The software runs on the company's Model 9000 Series 300 workstations. It takes your design, minimizes its equations, simulates its operation, checks for glitches, selects target PLDs, partitions the design to fit in one or more PLDs, and generates a PLD fuse map and test vectors.

The software's waveform editor allows you to define a PLD using timing diagrams. For synchronous circuit design, you can create statetransition diagrams (STDs) with a special STD editor that shows indi-

vidual states as boxes linked with state-transition arrows and pop-up state-transition tables.

The software accepts designs without regard to a target PLD. You can also ask the PLD Design System to select the most appropriate device automatically from its

PLD library, and you can specify vendor-type, inventory, or powerconsumption parameters. The PLD library contains devices from Altera, AMD, Intel, Lattice, Monolithic Memories, NEC, Ricoh/ Panatech, Signetics, and VTI.

The PLD Design System can exchange circuit and simulation data with the other software tools in the company's Electronic Design System for complete board-level design entry and simulation. The PLD Design System costs \(\$ 8000\) to \(\$ 14,500\). Delivery is 12 weeks ARO.

Hewlett-Packard Co, 1820 Embarcadero Rd, Palo Alto, CA 94303. Phone local office.

Circle No 446

\section*{Real-time operating system runs at twice the speed of other systems}

According to its maker, the BSO / Realtime Craft real-time operating system (RTOS) runs at least twice as fast as any other 16- or 32-bit RTOS. The system is written in assembly language for Intel 8086/ 286 (286 real mode only), Motorola 68000/20, and National Semiconductor \(32000 \mu\) Ps. Compiler interfaces are available for Intel C and Pascal, Microsoft C, Alsys Ada, and all of the manufacturer's compilers.

Running on an \(8-\mathrm{MHz} 8086\) or \(68000 \mu \mathrm{P}\), the system performs a complete context switch in \(110 \mu\) sec; each of the system executive's 28 primitives requires between 25 and \(60 \mu \mathrm{sec}\) for execution. The application program makes all operat-
ing-system calls via software interrupts.

The system comprises four modules: the real-time executive, an in-put-output supervisor, a file-management system, and a debugger. BSO/Realtime Craft comes burnt into a PROM. The executive, for example, occupies less than 3 k bytes of PROM space.

The input-output supervisor provides hooks for user-written device drivers and allows for dynamic allocation of RTOS objects such as tasks, programs, semaphores, and mailboxes. It also manages the dynamic allocation of memory. Drivers for some common peripherals are available. Each real-time task can
dynamically specify the size of its memory buffers.
The file-management system works with the I/O supervisor to provide cached disk I/O, and it can read Unix files. The maker claims that you can always read your files even in the event of a system crash or after a power failure. BSO/ Realtime Craft costs \(\$ 6300\) to \(\$ 16,000\), depending on the number of modules licensed.

Boston Systems Office Inc, 128 Technology Ctr, Waltham, MA 02254. Phone (617) 894-7800. TWX 710-324-0760. FAX (617) 642-5762.

Circle No 447

\section*{Software}

\section*{Software package merges Unix and real-time executives}

Based around the Unix System V operating system and Ready Systems' VRTX32 real-time executive, VXCEL provides an integrated environment for the development and supervision of real-time multiprocessor VME Bus applications. The package is compatible with the company's range of VME Bus boards, and allows both a processor running Unix and processors running realtime applications to coexist on the same VME Bus backplane without compromising system performance. VXCHIP, a ROM-resident multiprocessor executive with built-in communications channels to the Unix host processor, accompanies each real-time processor.
The Unix host environment allows you to use a wide range of commercially available software tools for program generation. VXCEL's command interpreter, VXSHELL, then allows you to interact with multiple target boards to download, debug, and execute your real-time program code. The debugger displays the VXCHIP status, as well as data structures such as queues, event flags, semaphores, and mailboxes that exist on each of the system's real-time processors.

In addition to providing a multitasking executive, VXCHIP includes facilities for I/O management, file management, distributed processing, Unix host communications, and integrated debugging. Local disk facilites include an MS-DOS-compatible file system.
A message-passing scheme, which uses a shared VME Bus memory architecture, supports distributed processing. This architecture allows task-to-task or task-to-group communications between processors. If system facilities are not di-
rectly available on one processor, you can access them by executing a remote procedure call to another processor. In addition, a name server allows you to make local or global objects visible to other processors in the system. License charges for VXCEL range from \(\$ 6000\) to \(\$ 12,000\), depending on the system configuration and the software modules.

\section*{Plessey Microsystems Ltd,} Water Lane, Towcester, Northants NN12 7JN, UK. Phone (0327) 50312. TLX 31628.

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Plessey Microsystems, 1 Blue Hill Plaza, Pearl River, NY 10955. Phone (914) 735-4661. TWX 710-541-1512.

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\section*{FlexOS \\ Systems Software Solutions for Top to Bottom Integration}


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Flex0S Customer Service
Digital Research Inc.
Box DRI
Monterey, CA 93942

\section*{SIMULATOR}

The Libra harmonic-balance simulation program combines linear analysis in the frequency domain with time-domain analysis of nonlinear elements. The package includes the vendor's Touchstone frequency-domain simulator for linear simulation. By extending the file to include appropriate models, you can add non-linear-analysis capabilities. The simulator will show you time-dependent voltage waveforms at any node; time-dependent current into any nonlinear device; power density and total power; frequency-selective power; and frequency-selective voltage and current, including phase response.

You can transfer all of the data from a simulation to a disk file. If you need steady-state response to sinusoidal waveforms, the package lets you obtain solutions based on the vendor's library of microwave components. A new large-signal and small-signal model for GaAs FET transistors improves the precision of nonlinear simulations of microwave active networks. The program currently runs on VAX/VMS workstations, Apollo Domain-IX workstations, and HP 9000 , Series 300 computers under the HP-UX operating system. From \(\$ 20,000\).

EEsof Inc, 5795 Lindero Canyon Rd, Westlake Village, CA 91362. Phone (818) 991-7530. TLX 384809. Circle No 525

\section*{NEURAL-NET TUTORIAL}

Netwurkz introduces you to associa-tive-memory concepts and their implementation by means of a neural model. This model consists of processing elements, analogous to neurons, that use rules such as the sum of products to produce an output from multiple inputs. The output of one neuron can form part of the input to other neurons to produce an aggregate (a neural net), which can learn complex patterns and recall patterns correctly, even when the input is not an exact match.

The associative-memory demo comes with a PL/D compiler that lets you add new Data statements in the demo or modify existing statements. The example uses approximately 50 neurons, and you can expand the net to a maximum of 1000 neurons. To run the program, you'll need an IBM PC, PC/XT, or PC/AT that's equipped with at least 192 k bytes of RAM and MS-DOS version 2.0 or later. Netwurkz, \(\$ 79.95\); PL/D compiler, \(\$ 124.95\); both programs, \(\$ 154.95\).
DAIR Computer Systems, 3440 Kenneth Dr, Palo Alto, CA 94303. Phone (415) 494-7081.

Circle No 526

\section*{SOFTWARE TOOL}

This version of the chipForth soft-ware-development system allows you to write and debug software for Intel's \(8051 / 8031\) family of microcontrollers without using an in-circuit emulator. Instead, chipForth provides you interactive program development, using only the on-chip RAM of the \(\mu \mathrm{C}\) and a ROM emulator. You can write programs that use only the on-chip RAM and ROM of the \(\mu \mathrm{C}\) or programs that use the 8051's 64 k bytes of external data and program space. You can also implement systems with overlapping data and program space.

The development system uses the Forth programming language, which combines an editor, an assembler, and a compiler. This development software runs on an IBM PC, \(\mathrm{PC} / \mathrm{XT}\), or PC/AT linked to the target system via a serial port. The Forth multitasking kernel that's supplied uses as few as 40 bytes/ task. It imposes no overhead on the \(\mu \mathrm{C}\) 's interrupt handling and does not affect its bit-handling capabilities. £1800.

> Computer Solutions Ltd, Canada Rd, Byfleet, Surrey KT14 7HQ, UK. Phone (09323) 52744. TLX 946240 (Request ref 19012265).

Circle No 529

\section*{IEEE-488 DEVICE DRIVER}

The NI-488 MS-DOS device-driver software package helps you develop instrument-control software on IBM PC and PS/2 computers. The enhancements to the package support Version 4 of Microsoft's QuickBasic. They also include an applications monitor that gives you program-tracing facilities and a special interrupt service request, using the Basic On Pen statement. This feature eliminates the need for continuous polling to capture instrument service requests.

The package includes a QuickBasic language-interface library and a BasicA library. You can instruct the monitor to install breakpoints that show the details of the most recently executed IEEE-488 call on a pop-up screen. The display can also show you a listing of as many as 255 of the preceding IEEE488 calls. This feature obviates inserting debugging statements in the instrument-control source code. The package is included with the vendor's GPIB-PCII, \(\$ 395\); GPIB-PCIIA-2, \$495; and MC-GPIB,\$495 interface boards. Current users of these products may upgrade to the new package at nominal cost.

National Instruments, 12109 Technology Blvd, Austin, TX 78727. Phone (800) 531-4742; in TX, (800) 433-3488. TLX 756737.

Circle No 527

\section*{DATABASE ANALYZER}

IXL: The Machine Learning System combines statistical methods, symbolic data analysis, induction, and deduction to explore and reveal previously unknown data interdependencies and relationships in very large databases. You can specify the level of error you are willing to accept, the kind of rules that the program should use, and the concepts (both exact and inexact) that the program should use in constructing the rules.
The program presents its results in the form of the percentages of

data elements that conform to each rule, with a confidence factor based on the variations in the database contents. The user interface is menu driven; it prompts you for all the information it needs, so you don't need any programming experience to use the program. To run the program, you'll need an IBM PC/XT or PC/AT that's equipped with at least 512 k bytes of RAM and a hard disk. \(\$ 490\).
IntelligenceWare Inc, 9800 S Sepulveda Blvd, Suite 730, Los Angeles, CA 90045. Phone (213) 4178896.

\author{
Circle No 528
}

\section*{MONTE CARLO ANALYSIS}

Monte Carlo Plus runs on the IBM PC and performs risk and sensitivity analyses of data contained in a Lotus 1-2-3 spreadsheet; it provides both risk and sensitivity results in the form of tables and graphs. The program prompts you to supply high- and low-accuracy limits for key variables, and then estimates the statistical probability that a particular result computed by the spreadsheet will exceed or fall short of the predicted value. You can also use the program to determine the effect of independent variables on a dependent result; the program identifies the variables that have the least and the greatest effect on the result. \(\$ 89\).
Suntex National Corp, Box 772868, Houston, TX 77215. Phone (713) 783-9059.

Circle No 530

\section*{DSP SIMULATORS}

The AVSIM321 and AVSIM322 are software simulators/debuggers for the Texas Instruments 32010 and 32020 families of digital signal-processing chips. They run on an IBM PC or a compatible computer, and interactively execute object code under the control of a full-screen symbolic debugger. The screen display shows you the current instruc-

tion stream and the contents of registers, flags, and areas of data memory. You can examine and modify these at any time; by using an Undo key, you can back up, one instruction at a time, through recently executed instructions to determine where an error occurred. You can issue commands either from a menu structure or from a command line. \(\$ 379\) each.
Avocet Systems Inc, Box 490, Rockport, ME 04856. Phone (207) 236-9055.

Circle No 531

\section*{EUREKA FOR MAC}

Eureka: The Solver is a mathematical tool that lets you use your Macintosh to solve mathematical problems, including simultaneous linear equations in multiple variables. You enter an equation in the text-editor window, and the program searches for the variables and finds a solution. You can then verify the solution, plot it, or send a report to the printer or to a disk file. The program can use the Macintosh II's 60881 math coprocessor and color capabilities. \(\$ 195\).

Borland International Inc, 4585 Scotts Valley Drive, Scotts Valley, CA 95066. Phone (408) 438-8400. TLX 172373.

Circle No 532

\section*{ON-LINE MANUALS}

The Norton On-Line Programmer's Guides provide reference material for 8088 assembly language as well as for the Basic, Pascal, and C languages. You load a RAM-resident
access program, which occupies 65 k bytes, and a language database; while you're running an application program, pressing Shift and F1 puts the language-database menu on the screen. You can call up the detailed reference entry or short definitions; or you can search for a key word or look for related cross references.

In the resident mode, you load the access program and guide before running any other program, and they remain available until you uninstall them. In the pass-through mode, you load the guide on the same command line as your application; when your application terminates, the access program is automatically uninstalled, freeing the memory for other programs to use. Access program and one language database, \(\$ 100\); additional language databases, \(\$ 50\) each.

Peter Norton Computing Inc, 2210 Wilshire Blvd, Suite 186, Santa Monica, CA 90403. Phone (213) 453-2361. TWX 650-226-1869.

Circle No 533

\section*{REAL-TIME OS}

You can use the UniFlex/RT multiuser, multitasking operating system for VME Bus systems based on the 68020 processor with the Macintosh II. The real-time features include fast message exchanges, a named enqueue/dequeue mechanism, and shared data pages and text segments. The OS runs on a minimal Macintosh II computer that has at least 1 M byte of main memory, 20 M bytes of disk storage space free, and the 68851 memory-management chip. If you have a larger hard disk, you can partition the disk so that you can use the Mac OS as well as UniFlex/RT.

The system comes with the utilities for real-time, assembly-language software development, including a relocating macroassembler, a linking loader, a library generator, and a symbolic debugger, as well as file- and system-

\section*{WHODOYOUCAILII:AN
 \\ }

The problem with most real-time operating systems is simple, they're not an integrated solution. You end up dealing with a multitude of suppliers for languages, compilers, debuggers and other important development tools. And when something does go wrong, it can be a frustrating experience trying to straighten out the mess.

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maintenance programs. You can obtain such options as a System V-compatible C compiler, X-Windows, a screen editor, or a text processor. \(\$ 750\).

Technical Systems Consultants Inc, 111 Providence Rd, Chapel Hill, NC 27514. Phone (919) 4931451. TWX 510-920-0540.

Circle No 537

\section*{SCHEMATIC CAPTURE}

Spice_Net is a schematic-entry program that runs on an IBM PC and generates an ASCII file for input to a Spice simulator program. You draw your schematic with the aid of a mouse; a single keystroke can place symbols representing parts that are included in the vendor's subcircuit and model libraries. The

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\section*{Software}

Brainpower Inc, 24009 Ventura Blvd, Suite 250, Calabasas, CA 91302. Phone (818) 884-6911.

Circle No 536

\section*{RF SIMULATION}

You can use the enhanced simulation program Touchstone version 1.6 for the design, analysis, and optimization of microwave/RF circuits. A new sparse-admittance matrix-reduction feature permits this version of the simulation program to use computer memory more efficiently than did earlier versions and speeds up the analysis of large, complex circuits. You can include as many as 250 variables and equations in a circuit file. Other added features and capabilities include a sweep progress indicator, print- and plotinterrupt facilities, and the ability to make VSWR measurements and to simulate stripline-cross and stripline-curve elements.

The network-analyzer interface works with the Wiltron 360 vector network analyzer, as well as with HP network analyzers. The program runs on IBM PCs, VAXs, HP 9000 Series 300 machines, and Apollo and Sun workstations. From \(\$ 9900\).

EEsof Inc, 5795 Lindero Canyon Rd, Westlake Village, CA 91362. Phone (818) 881-7530.

Circle No 535

\section*{CROSS-DEVELOPER}

The ST Universal Cross-Development Kit runs on Atari 520, 1040, and Mega-ST computers. The package includes a text editor with which you can write assembly-language programs for a wide variety of \(4-, 8\)-, and 16 -bit \(\mu \mathrm{Ps}\) and microcontrollers. The table-driven crossassembler translates the source code into the target machine's native code. The cross-assembler contains tables for \(20 \mu \mathrm{Ps}\), including the HD64180, Z80, 6502, 68000, 8048, 8051, 8085, 8086/88, and 8096.

When you've assembled your program, you can download the object code to the EPROM emulator, which plugs into the target machine's EPROM socket. The emulator is compatible with EPROMs in the 2716 through 27256 families. The access time for the emulator is 150 -nsec. For downloading purposes, both the cross-assembler and
the EPROM emulator can handle Intel Hex, Motorola S-record, and simple binary formats. Most serial EPROM programmers can operate with at least one of these formats. \(\$ 575\).

Memocom, 1920 Arbor Creek Dr, Carrollton, TX 95010. Phone (214) 446-9906.

Circle No 538

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CIRCLE NO 178

\section*{CAE FOR MAC}

The EDS-I electronic-design software system runs on the Apple Macintosh II. The schematic-capture module lets you generate schematics, using both standard and user-defined component libraries, and automatically produces a net list and a parts list. The layout module can handle pc-boards as large as \(32 \times 32 \mathrm{in}\). and provides zoom and pan features, auto-tooling extraction, multilayer editing, and Gerber photoplotting files. The autorouting module accepts the net list and parts list as input, lets you define routing parameters and options as well as keep-out areas, and lists unfinished nets.

The Gerber translator module lets you convert Gerber files created by other systems into the electronic design software's internal format, or Gerber files produced by the software into PostScript files for use by high-resolution graphics printers in the Linotronics \(300 / 500\) series so you can check the photoplot drawings before committing them to film. You can order modules separately. Full system, \(\$ 1500\).

Vamp, 6753 Selma Ave, Los Angeles, CA 90028. Phone (213) 4665533. TWX 650-262-3069.

Circle No 539

\section*{CASE TOOLS}

The C Documenter and C Scan utilities for the IBM PC aid in documenting and examining programs written in the C programming language. Where such programs are subdivided into separate subsystems, C Documenter generates a set of four reports containing cross-reference information, such as which modules or functions are called by other modules or functions. You can use the reports to define the interface between two program subsys-tems-even those written by different programmmers.

When you examine your programs, C Scan lets you use symbolic
names to locate and display portions of code, regardless of the sourcecode file in which they're contained. By keeping a record of your program's files, C Scan lists the type of items in the program, such as functions, global variables, structures, and union members; upon your selection of an item, C Scan displays all items of that type in your program and lets you locate them. You can obtain the utilities separately or as part of the C Dev cross-development tool set, which includes a C cross compiler, a mimic simulator, and a cross-assembler. C Scan, £195; C Documenter, £195; C Dev, approximately \(£ 2000\).
Real Time Systems Ltd, Viking House, Nelson St, Douglas, Isle of Man, UK. Phone (0624) 26021. TLX 94011289.

Circle No 540

\section*{FAX PACKAGE}

The PC FAX package allows you to use an IBM PC/XT, PC/AT, or compatible computer to send or receive FAX messages to or from any International Group III FAX machine. You can also use it to send telex or electronic mail. The package can transmit or receive any word-processed document, desktop publishing image, or paint-box system image.

You can generate input from the PC's memory or keyboard, from an optional digitizer tablet, or from a hard-copy scanner or an existing FAX machine. In addition, the FAX software can capture and transmit drawings generated by CAD packages. In normal mode, the package provides \(202 \times 98\)-pixel FAX, but you can select a fine mode that increases resolution to \(204 \times 196\) pixels.
Although the standard software operates to International Group III CCITT FAX standards, you can upgrade it to Group IV. The transmission software includes a directory of FAX numbers, and automatic dialing and redialing facilities. You can also program the software to trans-
mit messages during cheap-rate periods, and to poll other FAX or communications systems to check for any FAX messages programmed for transmission to your PC's number.

The PC can receive FAX messages while you're using it for other tasks. When incoming FAX messages are automatically saved to disk, you are alerted by audible and on-screen prompts. You can then recall messages to the PC's screen, zoom in on them to examine fine detail, or output them to a printer or plotter. An optical character recognition (OCR) package that can learn new character fonts is optionally available. The PC-FAX package, including the telephone-line interface hardware, costs \(£ 895\). The OCR package costs \(£ 395\).

Softech Professional Systems Ltd, 9 Tonbridge Chambers, Pembury Rd, Tonbridge, Kent TN9 2HZ, UK. Phone (0732) 362688. FAX (0732) 770263.

Circle No 553

\section*{IEEE-488 SOFTWARE}

This IEEE-488 management software package allows a VME Bus computer system to control an IEEE-488 instrumentation bus via the company's CC-91 IEEE-488 interface card. It allows you to implement IEEE-488 talker, listener, and system controller functions. The software comprises an IEEE488 manager and an IEEE-488 device driver, running under the OS-9/ 68 k operating system, and allowing you to write application programs in C or Pascal.

Running as a background task, the manager configures and initializes the IEEE-488 bus, and arbitrates all communication between the IEEE-488 device-driver software and the application program. To execute an IEEE-488 bus operation, the application program passes a command-parameter block, specifying the device name and address, the action required, and the source

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\text { Industrial }\end{array} \\
\hline \text { X2864H } & 70 \mathrm{nsec}, 90 \mathrm{nsec} & 90 \mathrm{nsec}\end{array}\right]\)\begin{tabular}{c}
\(120 \mathrm{nsec}, 150 \mathrm{nsec}\), \\
180 nsec
\end{tabular} \begin{tabular}{c}
\(120 \mathrm{nsec}, 150 \mathrm{nsec}\), \\
180 nsec
\end{tabular}
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and destination of transferred data via the manager to the device driver. A device descriptor provides the driver with information about the addressed device's IEEE-488 capabilities. The manager also includes a time-out monitor to detect excessive delay in an instrument's response. Sample device descriptors, a menudriven system test program, and sample application programs are included in the software package. Approximately \(\$ 750\).

Compcontrol bv, Stratumsedijk 31, 5600 AD Eindhoven, The Netherlands. Phone (040) 124955. TLX 51603.

Circle No 551
Compcontrol Inc, 15466 Los Gatos Blvd, Suite 109-365, Los Gatos, CA 95032. Phone (408) 3563817. TWX 510-601-2895.

Circle No 552

\section*{DIAGNOSIS TOOL}

The TestBench software package uses artificial intelligence techniques to build expert systems for the diagnosis and repair of malfunctions in complex machines and processes. It consists of three modules: TestBuilder, TestBridge, and TestView. The TestBuilder expert-system development system, which runs on a TI Explorer, helps you acquire an expert technician's knowledge of the diagnostic and repair procedures for a complex machine or process, and builds a knowledge base that combines this information with documentation and rules of thumb. When you have completed the knowledge base, a less expert technician can run the expert system and receive guidance in troubleshooting and repairing the target equipment.

TestBridge translates the information captured by the development system into a form that Test View can use, a similar expert system that runs on an IBM PC/AT or compatible. Thus, you can develop a diagnostic expert system on the Explorer and distribute it to several
repair stations that can run it on the much less expensive PC/AT. Complete TestBench package, including one week of training and one year of software maintenance, \(\$ 42,000\).

Texas Instruments Data Systems Group, Box 2909, Austin, TX 78769. Phone (512) 250-6314.

Circle No 548
Carnegie Group Inc, 5 PPG Place, Pittsburgh, PA 15222. Phone (412) 642-6900. TLX 4970240.

Circle No 549

\section*{VME BUS FORTH}

The PolyForth V4000 software-development system suits VME Inc's V4000 VME Bus CPU board. The V4000 board uses NCR's NC4016 \(\mu \mathrm{P}\), which executes high-level Forth code as its native instruction set. The NC4016 chip runs at 8 MHz ; at this speed, according to the vendor, Forth programs execute 19 times faster than do compiled C programs running on a \(10-\mathrm{MHz} 80286\).

The software package includes complete source code, an optimizing compiler, clock/calendar management facilities, and utilities. You can create ROM-resident Forth programs as large as 64 k bytes for embedded applications. At additional cost, you can obtain libraries of mathematical and database-management routines. Software only, \(\$ 2950\); software and CPU board, \(\$ 5745\).

Forth Inc, 111 N Sepulveda Blvd, Manhattan Beach, CA 90266. Phone (213) 372-8493. TLX 275182.

Circle No 545

\section*{LANGUAGE}

PC-Simula is a version of the Simula object-oriented programming language that you can run on IBM PCs. The language can run under MS-DOS, Xenix, or OS/2 operating systems. An 80386-based version is under development.

The company recommends that
the MS-DOS version for standard IBM PC/XT and PC/AT computers have at least 640 k bytes of RAM. It has tested the Xenix version under Santa Cruz Operation's Xenix Sys-tem-V and under Microsoft Xenix. It has also tested the 0S/2 version on an IBM PC/AT running Microsoft OS/2. When you run the language under Xenix or OS/2, the company recommends that your computer have at least 1.5 M bytes of RAM, a math coprocessor, and a hard disk.

Cross-compilers to transfer programs between MS-DOS and Xenix, and between MS-DOS and OS/2 are available. Single-user license for the MS-DOS version, 10,000 Norwegian Kroner; for the Xenix and OS/2 versions, 15,000 Norwegian Kroner; and for the cross-compilers, 20,000 Norwegian Kroner.

Simula as, Box 4403, Torshov, 0402 Oslo 4, Norway. Phone (2) 156710. FAX (2) 156051.

Circle No 550

\section*{NEURAL NET}

The Awareness software package runs on IBM PCs and consists of programs that demonstrate four neural-network paradigms. The generalization paradigm uses a generalized learning rule and demonstrates that a layered neural network can solve the exclusive-OR function. The associative paradigm exhibits many of the computational capabilities of neural networks, such as preferential learning, fault tolerance, differentiation, and association.

The optimization paradigm is an example of a neural network that can produce solutions to combinatorial optimization problems. The selforganization paradigm is an example of a nearest-neighbor classifier that behaves as an optimal signal processor in the presence of noise. The documentation contains introductory material on neural networks, together with the equations

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- technical data

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\begin{tabular}{|c|c|c|c|}
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\] & Number of output channels OUTPUT modes & unipolar or bipolar. user-selectable \\
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0.1 betweem ranges (max.) \\
any range adjustable to 0 \\
\(\pm\) ppm/degrees \(C\) of \(\operatorname{FSR}\) (max)
\end{tabular}} \\
\hline Rejection ratio & \multirow[t]{2}{*}{72 db} & & \\
\hline integral Inearity & & \multirow[t]{2}{*}{Stability} & \\
\hline error & \(\pm 1\) LSB maximum & & \\
\hline DIFFERENTIAL Linearity & & \multicolumn{2}{|l|}{OFFSET} \\
\hline Error & \multirow[t]{2}{*}{\begin{tabular}{l}
\(\pm\) '? LSB maximum \\
\(\pm 5 \mathrm{ppm} /\) degrees C of FSR (max)
\end{tabular}} & \multirow[t]{2}{*}{Erroc unipolar
Error bipolar} & \multirow[t]{2}{*}{\(\pm 325\) milivor max adjustable to 0} \\
\hline Stability & & & \\
\hline \multicolumn{2}{|l|}{GAIN:} & Unipolar stability & \(\pm 8 \mathrm{ppm} /\) degrees C of FSR (max) \\
\hline Error & \multirow[t]{2}{*}{\(\pm 0.10\) between ranges (max) any range adjustable to 0} & \multirow[t]{2}{*}{Bipolar stability MONOTONICITY} & \multirow[t]{2}{*}{\(\pm 24 \mathrm{ppm} /\) degrees C or FSR (max) 0 to 50 degrees \(C\)} \\
\hline & & & \\
\hline Stability & \multirow[t]{2}{*}{\(\pm 32 \mathrm{ppm} /\) degrees C of FSR (max)} & \multirow[t]{2}{*}{SEttung time} & \multirow[t]{2}{*}{10 microseconds max to wthen 0 1\% FSR tor a 10 volt step with 1000} \\
\hline OFFSET: & & & \\
\hline Error & adjustable to 0 & & \({ }^{\text {ofl load }}\) Protected for sharto common \\
\hline Unipolar stability & \(\pm 24 \mathrm{ppm} /\) degrees C of FSR (max) & PROTECTION & Protected for short to common \\
\hline Bipolar stability & \(\pm 24 \mathrm{ppm} /\) degrees C of FSR (max) & OVERSHOOT & \(\pm 1 \%\) of FSR max \\
\hline MONOTONICIITY & 0 to 50 degrees C & THROUGHPUT & \\
\hline THROUGHPUT to memory & 15.000 conversions/second min. & trom memory & 25.000 conversion/second \\
\hline
\end{tabular}

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\title{
Latched bus provides foolproof I/O
}

\author{
Samuel C Creason \\ Beckman Industrial Corp, La Habra, CA
}

You can reduce the time necessary to interface a \(\mu\) P-controlled system to parallel I/O devices (ADCs, DACs, etc) if you adopt a standard I/O bus comprising latched-I/O lines (Fig 1). Such a bus provides several benefits: It eliminates I/O-device timing problems; it partitions your design task into smaller, independent tasks; and it produces an easily modifiable design.
The only disadvantage of this latched-I/O bus is that accesses are slower than they would be if you tied the devices directly to the \(\mu\) P's system bus. For example, instead of requiring just a single instruction, writing to a device requires four instructions: one each to write bytes to the address and data ports, and two to toggle the data-transfer strobe on and off. Reading from an external device also requires four instructions. Fortunately, this extra overhead is not significant for any but the most I/O-intensive systems.

In Fig 1, the circuit elements to the left of the dashed line are conventional I/O ports. In fact, you could replace them, during development, with any personal computer that has parallel I/O ports. \(\mathrm{IC}_{1}\) is an octal, 3 -state, positive-edge-triggered D-type flip-flop, and
\(\mathrm{IC}_{2}\) is an octal, 3 -state buffer/line-driver/receiver; together, they form an 8-bit, bidirectional I/O data bus. \(\mathrm{IC}_{3}\) and \(\mathrm{IC}_{4}\) are both octal, D-type flip-flops. \(\mathrm{IC}_{3}\) drives the I/O address bus; \(\mathrm{IC}_{4}\) simply acts as a data-transfer strobe. Note that \(\mathrm{IC}_{5}\), a dual 2 - to 4 -line decoder/ demultiplexer, gates \(\mathrm{IC}_{1}, \mathrm{IC}_{2}, \mathrm{IC}_{3}\), and \(\mathrm{IC}_{4}\) during read and write operations.

This scheme has an unusual aspect: I/O address lines \(\mathrm{A}_{3}\) through \(\mathrm{A}_{7}\) from \(\mathrm{IC}_{3}\) drive \(\mathrm{IC}_{6}\) and \(\mathrm{IC}_{7}\) (3- to 8-line decoder/demultiplexers). \(\mathrm{IC}_{6}\) and \(\mathrm{IC}_{7}\) provide eight input and eight output strobes, respectively. (I/O address lines \(\mathrm{A}_{0}, \mathrm{~A}_{1}\), and \(\mathrm{A}_{2}\) are available for other uses.) Note that address line \(\mathrm{A}_{7}\) also controls \(\mathrm{IC}_{1}\) 's 3 -state output. Because this circuit requires a high signal to disable \(\mathrm{IC}_{1}\) 's 3 -state output, input devices must have an I/O address of \(80_{\text {Hex }}\) or higher.

Using \(\mathrm{IC}_{3}\) and \(\mathrm{IC}_{4}\) to generate the input and output strobes directly might seem to be a simple way to reduce chip count and \(\mu \mathrm{P}\) overhead. However, if you employ such a scheme, a glitch in your software could easily enable more than one I/O device at a time. The scheme in Fig 1 trades \(\mu \mathrm{P}\) overhead for foolproof I/O.

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Fig 1-By adopting this foolproof, latched-I/O scheme as a standard for \(\mu P\)-based systems that must handle parallel I/O, you can quickly adapt it to different tasks.

\section*{DESIGN IDEAS}

\section*{State machine controls PLD programming}

\author{
V Lakshminarayanan \\ Sneha Corp, Bangalore, India
}

The 5P8 PAL and the 74273 octal flip-flop in Fig 1 form a finite-state machine that functions as a programmingwaveform timer for numerous Monolithic Memories' PALs (5P8, 5P8A, 8P4, 8P8, 9P4, 9P8, 10P4, 10P8, 11P4, 11P8, 12P4, 12P8, 9R8, 10R8, 11RA8, and 11RS8) and generates the timing diagram of Fig 2. The Boolean statements in Fig 3 reduce to the Boolean


Fig 1-A state machine consisting of a 5P8 PAL and an octal flip-flop generates programming-timing control signals for programming common PALs.
equations in Listing 1. A Boolean-equation compiler such as Palasm can produce a fuse-map program for Fig 1's 5P8 PAL.
The 5P8 develops the three control signals necessary to program PLDs: TVCC controls the application of supervoltages to the target PLD's supply pin; TE is the timing signal for enabling the target PLD; and CO

\section*{LISTING 1-MINIMIZED BOOLEAN EQUATIONS}
\[
\begin{aligned}
& \text { ADD A0 A1 A2 A3 A4 } \\
& \text { DAT BO B1 B2 B3 B4 TVCC TE CO } \\
& \text {;NEXT STATE GENERATOR } \\
& B O=I A 4 * I A O+A 4 * I A 3 * I A^{2} * I A O \quad ; \text { INCREMENT LSB } \\
& B 1=I A 4^{*}\left|A 1^{*} A 0+I A 4^{*} A 1^{*}\right| A O+\left|A 3^{*}\right| A 2^{*} \mid A 1^{*} A 0 \\
& +\left|A 3^{*}\right| A 2^{*} A 1 \text { * } \mid A 0 \quad \text {;INCREMENT BIT1 } \\
& B 2=I A 4^{*} A 2^{*}\left|A 1+I A 4^{*} A 2^{*}\right| A O+I A 4^{*}\left|A 3^{*}\right| A 2^{*} A 1^{*} O \\
& +\mid A 4 \text { * } A 3 \text { * } \mid A 22^{* A 1 * A O} \\
& \text { INCREMENT BIT2 } \\
& B 3=I A 4^{*} A 3^{*}\left|A 2+I A 4^{*} A 3^{*}\right| A 1+I A 4^{*} A 3^{*} \mid A 0 \\
& +I A 4^{*} A 3^{*} \mid A O+I A 4^{*} I A 3^{*} A 2^{*} A 1^{*} A 0 \text {;INCREMENT BIT3 } \\
& B 4=I A 4^{*} A 3^{*} A 22^{*} A 1 \text { * } A 0+A 4^{*}\left|A 3^{*}\right| A 2^{*} \mid A 1 \\
& +A 4^{*} I A 3 \text { * } \mid A 2 \text { * } I A O \\
& \text {;INCREMENT BIT4 }
\end{aligned}
\]
;TIMING WAVEFORMS
\(T V C C=I A 4^{*}\left|A 1^{*} A O+I A 4^{*} A 1^{*}\right| A O+I A 4^{*} \mid A 3^{*} A 1\)
\(+I A 4\) *A2 *IA1 + IA4 * A3 * IA2 ;TIMING FOR VCC
\(T E=I A 4{ }^{*}\left|A 3^{*}\right| A 2^{*} \mid A 1+I A 4^{*} A 3^{*} A 2 * A 0\)
\(+I A 4^{*} A 3\) * \(A 2\) * \(A 1+I A 4^{*} \mid A 3 * I A 2 * I A 0 ; T I M I N G\)
;FOR E
\(C O=I A 4 * \mid A 3\) * \(A 1+I A 4{ }^{*} A 2 * / A 1\)
\(+I A 4^{*} A 3 * / A 2\)
:TIMING FOR OUTPUTS


Fig 2-The state machine in Fig 1 generated these waveforms.

\section*{DESIGN IDEAS}
controls the application of supervoltages to the target PLD's outputs.

If you apply a \(1-\mathrm{MHz}\) clock signal to the circuit, it will generate a \(t_{D}\) (delay time between programming steps) of \(1 \mu \mathrm{sec}\); a \(\mathrm{t}_{\mathrm{PW}}\) (programming pulse width) of \(10 \mu \mathrm{sec}\);
and a \(\mathrm{t}_{\mathrm{vCC}}\) ( \(\mathrm{V}_{\mathrm{CC}}\) supervoltage-application time during programming) of \(14 \mu \mathrm{sec}\).

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Fig 3-Developing the program for the 5P8 PAL in Fig 1 begins with Boolean equations derived from the waveforms in Fig 2.

\section*{Watchdog timer avoids \(\mu \mathrm{P}\) overhead}

\author{
Shen-Feng Hwang \\ Norfin International Inc, Seattle, WA
}

A simple pair of retriggerable one-shots can function as a watchdog timer while eliminating the programming
overhead and dedicated output line associated with conventional watchdog timers. In a typical system, the \(\mu \mathrm{P}\) periodically pulses an internal watchdog timer. If the \(\mu \mathrm{P}\) 's software malfunctions, the \(\mu \mathrm{P}\) will fail to reset the timer. When the timer times out, it triggers the

\section*{DESIGN IDEAS}
\(\mu\) P's nonmaskable interrupt (NMI), thereby restarting the \(\mu \mathrm{P}\).
The circuit in Fig 1 accomplishes the same task but doesn't require that the \(\mu \mathrm{P}\) reset the watchdog timer. \(\mathrm{IC}_{1}\) gates all of the \(\mu \mathrm{P}\)-generated input-enable signals to \(\mathrm{IC}_{2}\), a retriggerable one-shot. In this circuit, the enable signals are active low. Any active-low signal to \(\mathrm{IC}_{1}\) will retrigger the one-shot. The one-shot's time-out period equals \(1.1 \mathrm{R}_{2} \mathrm{C}_{2}\). You should set the time-out
period longer than the longest expected time between two consecutive input-enable signals from the \(\mu \mathrm{P}\). Thus, if the \(\mu \mathrm{P}\) fails to enable the inputs at the proper rate (indicating that the \(\mu \mathrm{P}\) will not respond to inputs and is, in effect, going deaf), the one-shot will trigger the \(\mu\) P's NMI.

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Fig 1—This watchdog timer saves \(\boldsymbol{\mu} \boldsymbol{P}\) overhead and eliminates a dedicated output line. It uses the \(\mu\) P's input-enable signals to retrigger its one-shot rather than employing a specific watchdog-timer command.

\section*{Flip-flop multiplies input frequency}

\section*{Paul D Gracie}

The Microdoctors Inc, Palo Alto, CA
The output frequency of the classical flip-flop circuit of Fig 1a is twice the clock input frequency. Although the circuit generates a narrow, positive pulse at each edge of its Clock input, you cannot cascade the circuit to further multiply the input frequency without some modifications.
Assume that the clock's signal and the flip-flop's Q output are low to begin with. The rising edge of the Clock input passes through the XOR gate unchanged and triggers the flip-flop, causing its Q output to go high. The flip-flop's Q output feeds back to the XOR gate, transforming it into an inverter. The XOR gate then truncates the Clock signal, resulting in a narrow positive pulse at the flip-flop's clock (CK) input. When the Clock signal goes low, the XOR gate inverts it, triggering the flip-flop again. This retriggering generates a second pulse at the flip-flop's CK input, causing


Fig 1-A simple circuit can double the clock input frequency (a). However, you cannot cascade such doublers unless you insert appropriate delays in the feedback paths as in \(\boldsymbol{b}\).
its Q output to go low.
The width of the pulses at the flip-flop's CK input depends on the propagation delay of the XOR gate and the flip-flop. For 74LS devices, this combined delay is about 30 nsec. Because the input Clock signal must stay high until the pulse to the flip-flop's CK input is completed, you can not cascade this basic circuit unless
you lengthen the pulse. Fig 1b shows inverters placed in the feedback path of the first and second stages of a 3 -stage clock-frequency multiplier. This circuit achieves a \(16-\mathrm{MHz}\) output from a \(2-\mathrm{MHz}\) input signal.

To Vote For This Design, Circle No 748

\section*{Flip-flops arbitrate shared resource}

\section*{Aditya Dua \\ Kanazia Digital Systems, Bombay, India}

The simple arbitration circuit of Fig 1 allows two masters to share a common resource. The circuit, consisting of a pair of back-to-back flip-flops, provides one pair of request lines and one pair of grant lines.

In operation, either master can request access to the shared resource by first pulsing the appropriate request line (either \(\mathrm{REQ}_{1}\) or \(\mathrm{REQ}_{2}\) ). If the resource is available, the requester will receive an active-low authorization signal on its respective grant line ( \(\overline{\mathrm{GNT}}_{1}\) or \(\mathrm{GNT}_{2}\) ). If the resource isn't available, the requester must repeatedly pulse its request line and poll its grant line until the grant line goes low. When a master relinquishes control of the shared resource, it must pulse its request line one last time to clear the way for subsequent requests from either itself or the other master.

For example, when the second master requests control of the shared resource (pulsing \(\mathrm{REQ}_{2}\), the clock input of \(\mathrm{IC}_{2}\) ), \(\mathrm{IC}_{2}\) 's D-connected output, \(\overline{2 Q}\), toggles from high to low provided its preset line, \(\overline{2 \mathrm{PR}}\), is inactive. This low output serves as both the grant signal to the second master, \(\overline{\mathrm{GNT}}_{2}\), and serves to force the output of \(\mathrm{IC}_{1}\) high. If the first master tries to toggle its flip-flop, \(\mathrm{IC}_{1}\), it will fail until the second master toggles its own flip-flop one more time.

Note that if both masters try to gain control of the shared resource simultaneously, only one of them will succeed. Also note that, upon power up, the circuit is in an indeterminate state. You must either initialize the circuit with a power-up initialization routine or add an AND gate (Fig 1b).

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Fig 1-A pair of back-to-back flip-flops can arbitrate access to a shared resource. A master requesting access will be blocked until the second master relinquishes control because each resource-grant flip-flop, when active, engages the other flip-flop's preset input. The AND gate in \(\boldsymbol{b}\) forces the circuit to a predetermined state upon power up.

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Number 12 in a series from Linear Technology Corporation

\section*{An LT1013 and LT1014 Op Amp SPICE Macromodel}

\section*{Walter G. Jung}

With the advent of low cost and powerful desktop computers, present day op amp circuit designs can mature more quickly with good simulation tools. One such tool since its inception has been SPICE, the standard analog circuit simulator. However, while PCS and workstations may now be present on more and more desks, a potential bottleneck towards effective simulation has been SPICE models for the more popular parts.

The macromodel approach to simulation of an op amp is viable for many designs, with the great asset of simulation speeds far faster than that of a full device-level circuit. With this design note, Linear Technology Corporation introduces
op amp macromodels to its applications library. It is hoped that eventually most op amps in the product line will be developed as macromodels and made available to customers.

The LT1013 and LT1014 devices are popular single supply LTC op amps, and are thus logical candidates for macromodels. While existing macromodels for the generic 358 and 324 types might suffice for some applications, circuit designs which take advantage of the unique precision and functional features of the LT1013 warrant a model which reflects those features. The schematic diagram of the LT1013 and LT1014 macromodel is shown in Figure 1, and is applicable to one channel of either device.


Figure 1. LT1013 Op Amp Macromodel

Key op-amp specifications for the commercial device are:
offset voltage \(=50_{\mu} \mathrm{V}\) (offset is not simulated)
bias current \(=8 \mathrm{nA}\)
gain \(=1000000\)
slew rate \(=0.4 \mathrm{~V} / \mu \mathrm{S}\)
bandwidth \(=0.8 \mathrm{MHz}\)
Also, the model simulates the input common mode range (which includes ground) and the output characteristics of swinging to ground while sinking current.

This macromodel acts very much like the real LT1013 or LT1014 device which incorporates input common mode clamping, to prevent the sign-reversal errors common to the \(358 / 324\) types. For example, comparison responses for the LT1013 and the 358 are shown in Figure 2. The common conditions of this test are for an overdriven, +5 V singlesupply follower. In both instances, the input signal is \(\mathrm{V}_{\mathbb{N}}\), a -20 V to 20 V sweep fed through 10 k , while the output is \(V(5)\).

Note that with the 358 , the output reverses sign when the input is overdriven below ground. In contrast, the LT1013 model is well behaved, simply clamping the overdrive at ground level....just like the real LT1013 device does!

The model itself is listed on this page, and can be entered by typing it in (carefully!). Registered users of MicroSim's PSPICE simulator will automatically receive this macromodel as part of the model library update with version 3.07 . Interested readers may contact MicroSim at the address or phone number listed at the end of this note for further information.

This LT1013/LT1014 op amp macromodel is being supplied to users as an aid to circuit designs. While it reflects reasonably close similarity to the actual device in terms of performance, it is not suggested as a replacement for breadboarding. Simulation should be used as a forerunner or a supplement to traditional lab testing.


Figure 2. LT1013 Test Circuit: Single-Supply (+5V), Overdriven Follower

This more complete macromodel has been adapted from the Parts program generated LT1013/LT1014 model. This version features closer fidelity to the real part, with input common mode clamping and compensated output clamping. It can be used for large signal and/or single supply applications, where the inputs can potentially be overdriven.

\section*{SPICE List for LT1013 Macromodel}


For LT1013/14 literature call \(800 \cdot 637 \cdot 5545\). For help with an application call (408) 432-1900, Ext. 361.


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Circle No 770

\section*{How to connect plotters to personal computers}

The 4-color brochure, Versatec puts power in personal-computer plotting, tells you how to receive hardcopy information on your IBM PC or compatible or about your Macintosh personal computer. Its diagrams show the configurations, including software, interface or rasterizer, and plotter. Besides discussing the advantages of using electrostatic and thermal-transfer plotters, the folder contains a product table to help you select the right plotter for speed, size, and high-quality output.

Versatec, 2710 Walsh Ave, Santa Clara, CA 95051.

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\section*{Monograph series reports on computer memory}

Memory Pointers, the vendor's newsletter, analyzes current technical developments in the add-in com-puter-memory market and analyzes the various computer memory offerings of Apollo, Sun, Hewlett-Packard, and other computer manufacturers. It provides regular features such as a question-and-answer column and a new-product section.

Clearpoint Inc, 99 South St, Hopkinton, MA 01748.

\section*{Circle No 771}


Listing of computer/electronic products
The vendor's 1988 illustrated cata\(\log\) lists more than 5000 items. The product lines feature a variety of items from computer kits and peripherals to integrated circuits. An 8 -pg insert highlights IBM, Apple, Commodore, and Tandy computer peripherals. Also included is a \(6-\mathrm{pg}\) insert of pin-out data.

Jameco Electronics, 1355 Shoreway Rd, Belmont, CA 94002.

Circle No 774

\section*{Multibus I and II boards cataloged}

This \(35-\mathrm{pg}\) catalog covers a complete line of Multibus I boards and accessories, and four new Multibus II products. It contains photographs,
descriptions, and specifications for more than 30 Multibus I, Multibus II, and VME Bus board-level products. Among the devices it describes are a Multibus II communications controller, a SCSI controller, a flop-py-disk controller, an SBX singleboard computer, and an intelligent prototype board.

Central Data Corp, 1602 Newton Dr, Champaign, IL 61821.

Circle No 773

\section*{Computer printers displayed}

This fold-out brochure illustrates the vendor's full line of computer printers suitable for business, scientific, or personal applications. Besides listing the printing speeds, performance capabilities, and special features of each printer type, it presents the options, accessories, and supplies that are available.

Seikosha America Inc, 1111 MacArthur Blvd, Mahwah, NJ 07430.

Circle No 775


Guide to memory requirements
The 1988 Designer's Guide to Add-In Memory describes a broad range of memory products from the very technical to the managementoriented. The 80-pg catalog contains information about buses currently in use; DEC's latest offerings;
where to find the best price and performance for memory products; and a survey of performance and memory options available from IBM. Also included are the features for the HP-9000 and the MIPS and number of megabytes that are available for the Sun 4/2XX and Apollo DN 4000.

Clearpoint Inc, 99 South St, Hopkinton, MA 01748.

Circle No 777

\section*{STD \(\mu \mathrm{Cs}\) and IEEE-488 interfaces categorized}

The 200-pg Technical Data Book details the vendor's complete line of STD-8088 industrial computer systems and IEEE-488 interfaces of microcomputers. The new-products section focuses on industrial-networking products, IBM-compatible STD DOS systems, interfaces for the IBM PS/2, single-board computers, STD and IEEE-488 drive pack-
ages, and bubble-memory systems. The catalog provides application examples, a system designer's guide to 8088 -based STD Bus systems, and the complete STD-8088 Bus specification. It also includes specifications, configuration guidelines, and ordering information.

Ziatech Corp, 3433 Roberto Ct, San Luis Obispo, CA 93401.

Circle No 778

\section*{Publications feature VME Bus and VME/Plus}

The 560-pg VME Data Book 1988 is the vendor's second edition of its data book on VME Bus products. Published simultaneously, the technical brochure on VME/Plus examines several growth-oriented architectural features. The data book contains nine chapters with numerous photos and illustrations. A prod-uct-selection matrix arranged in columns introduces each chapter.

\section*{Tight Schedule? Complex Project? Tough Problems?}

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}

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-We are proud of our proven record of performance on schedule, within budget
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For a corporate brochure contact Bruce R. Knox, President

\author{
MicroDimensions, Inc. \\ 7345 Production Drive \\ Mentor (Cleveland), Ohio 44060 \\ Phone: (216) 974-8070 • FAX: (216) 974-1648
}


Further, a product guide helps you locate the products you are looking for easily and quickly. The brochure features the most recent VME/Plus 32 -bit devices. It also contains previews of products in the design stage scheduled for release this year.

Force Computers Inc, 3165 Winchester Blvd, Campbell, CA 95008.

Circle No 779

\section*{System components and reference data categorized}

This \(80-\mathrm{pg}\), pocket-sized catalog lists versions of the PC Bus, Multibus, VME Bus, and Q Bus, as well as computers that are \(100 \%\) compatible with IBM PCs. It provides descriptions and specifications for each product. Further, the applica-tion-information and reference-data sections contain useful features, such as application maps.

Diversified Technology, Box 748, Ridgeland, MS 39158.

Circle No 776

\section*{Brochure sums up PC-to-mainframe transfers}

The 6-pg publication Mag Tape Power for Your PC presents three different approaches to processing bulk data transfers between PCs and mainframe computers. It includes software solutions for DOS and Xenix, based on PC systems. It also describes Filetran for selective file-based backup, as well as hardware solutions, using three different tape systems.
Telebyte Technology Inc, 270 E Pulaski Rd, Greenlawn, NY 11740.

Circle No 781

\section*{LITERATURE: COMPONENTS}


\section*{Booklet discusses specialty lubricants}

Lubricants for Electric Contacts and Connectors is the vendor's cata\(\log\) of specialty oils and greases. It covers four application areas: greases for sliding electric contacts, as in electric switches; oils for sliding contacts and potentiometers; potentiometer greases; and lubricants for stationary separable electric connectors. The catalog describes lubricant properties and operatingtemperature ranges, and discusses lubricants you can use to help suppress arcing conditions.

William F Nye Inc, Box G-927, New Bedford, MA 02742.

Circle No 560

\section*{Optoelectronics \\ guide and data book}

The two publications, Optoelectronics Selector Guide (SG87/D) and Optoelectronics Data Book (DL118/D), are divided into products sections: emitters/detectors, isolators, slotted switches, and fiber-optic components. The data book contains 65 new products and their applications, as well as a new section on optoelectronic chips or die. Both books contain an industry cross-reference and a reliability section.

Motorola Inc, Technical Information Center, Box 52073, Phoenix, AZ 85072.

Circle No 561

\section*{Application note defines common voltage terms}

The DPM-10 application note helps clarify your questions about several common terms and problem areas, such as isolated and nonisolated sig-nal-source inputs, common-mode voltages, and safety precautions for wiring primary line power to any device. The note provides wiring diagrams and standard formulas for component selection in voltage dividers, attenuators, and shunt circuits. The publication also helps you measure high- and low-level voltages and signals from resistance bridges, solid-state temperature sensors, and similar devices.

Acculex, 440 Myles Standish Blvd, Taunton, MA 02780.

Circle No 563


\section*{Antennas and accessories}

The vendor's 1988 catalog features antennas and accessories for EMI/ RFI testing, and a product- and antenna-selection guide in the form of a \(20 \times 30-\mathrm{in}\). calendar. In addition to product information, the catalog features FCC and VDE regulations, tables that help you make a selection, and a list of formulas.
EMCO, Box 1546, Austin, TX 78767.

Circle No 564

\section*{Folder introduces absorption material}

This introductory packet includes a folder and a sample of Poron products, the vendor's high-density cellular urethanes. It describes five
simple tests that you can conduct at your desk. The tests indicate energy absorption, resistance to impression, self-healing after puncture, resiliency, and resistance to long-term compression set.

Rogers Corp, 1 Technology Dr, Rogers, CT 06263.

Circle No 565


\section*{Comprehensive catalog heralds upcoming products}

This \(520-\mathrm{pg}\) catalog summarizes products from the vendor's Signal Processing Technologies division. It also features preliminary reports about devices in production, as well as application notes for switched capacitor filters, flash converters, and video D/A converters. The catalog's component data sheets and technical descriptions help you select the signal-processing products you need for designs.

Honeywell Inc, Signal Processing Technologies, 1150 E Cheyenne Mountain Blvd, Colorado Springs, CO 80906.

Circle No 567

\section*{App notes address interfacing problems}

The publication Real World Interfacing Application Notes contains 12 application notes that shed light on common interfacing problems in the laboratory and in industry. It provides tips and expounds on useful circuits for interfacing thermocouples, thermistors, solid-state temperature sensors, pH probes, and piezoresistive pressure trans-
ducers to personal computers. The publication also deals with solidstate relays to activate 110 V ac lines; sensing incident light; and selecting an appropriate method of A/D conversion. These notes do not provide solutions for a particular interfacing problem, but rather guide you in the right direction. They provide sample circuits, and list references and companies for further information.

Real Time Devices Inc, Box 906, State College, PA 16804.

Circle No 566


\section*{Brochure deals with thermostats}

This 4-pg folder provides you with information about the vendor's line of surface and immersion-sensing snap-action thermostats. It includes low-profile, grounded, isolated-contact, and sealed and hermetically sealed units.

Airpax Corp, Husky Park, Frederick, MD 21701.

Circle No 569

\section*{Buyer's guide \\ to French components}

The French Electronic Passive Component Industry Catalog 1988/ 89 lists over 10,000 products from 84 French component manufacturers in both English and French. Product categories include, capacitors, resistors, printed circuits, connectors,
keyboards, switches, ferrites, coils and transformers, quartz filters, antennas, fans, hybrid circuits, cable accessories, enclosures, fiber-optic components, protection devices, and microwave components.

Syndicat des Industries de Composants Electroniques Passifs, 11 rue Hamelin, 75783 Paris Cedex 16, France.

Circle No 574

\section*{Publication presents range of components}

The technical data and drawings that fill the vendor's \(80-\mathrm{pg}\) catalog help you select fuse and fuse-holder variations, as well as power-entrymodule combinations over a range from ac connector and line switch to ac connector. The publication also features rotary voltage selector switches and a state-of-the-art circuit breaker that combines the functions of a time-lag overload and short-circuit release device with those of an on/off line switch.

Schurter Inc, Box 750158, Petaluma, CA 94975.

Circle No 568

\section*{Preferred product catalog}

This 700-pg Preferred Type Range Catalog 1988 spotlights the components that appeal to the majority of the vendor's customers. The catalog provides type numbers, catalog numbers, selection guides, and brief technical information about this preferred product range. Components that meet CECC (Cenelec Electronic Components Committee) requirements are listed at the end of those product sections in which they are described. The catalog has six major sections that cover integrated circuits, semiconductors, electron tubes, capacitors, resistors, materials, and other products.

Philips, Components Div, Box 523, 5600 AM Eindhoven, Netherlands.

Circle No 575


\section*{Publication details based LED lamps}

This \(46-\mathrm{pg}\) catalog describes the company's submidget flange, screw, groove, wedge, and bipin LEDs. It also details miniature bayonet LEDs, and candelabra cluster and telephone slide lamps that come in a selection of sizes. The catalog also features application notes and a cross reference to incandescent bulbs.
Ledtronics Inc, 4009 Pacific Coast Hwy, Torrance, CA 90505.

Circle No 572

\section*{Chart your way through "magnetic phantoms"}

This wall chart of inductor applications helps design engineers rid their circuits of the "phantoms of magnetics," and it helps specifiers understand how to use inductive components to solve circuit-design problems. The chart covers component descriptions, diagrams of typical circuits, and brief design application notes for 11 inductor styles: molded inductors, variable inductors, and tunable coils; chip and surface-mount magnetics; wideband chokes; noise filters; toroids; pot cores; air coils; delay lines; highcurrent chokes; and balun chokes.

Inductor Supply, 15204 Transistor Lane, Huntington Beach, CA 92649.

Circle No 573


\section*{ATE described}

The vendor's 136 -pg 1988/1989 catalog presents its line of automatic test equipment. The volume covers the vendor's inventory and services in these areas: configured ATE systems, test frames, accessories, test fixtures, computers, software, ATE networks and services, and digital-, analog-, and custom-function modules. The publication features a subject index, model-number index, and an ATE configuration guide.

Summation, 11335 NE 122nd Way, Kirkland, WA 98034.

Circle No 755

\section*{Pamphlet describes simulator/analyzer}

This 6 -pg, 4 -color document details the features and options of the TE820A DS1/T1 frame simulator/ analyzer. It describes applications including testing and evaluating, troubleshooting, simulation, analyzing, and field testing. A listing of additional options is included.

Tekelec, 26540 Agoura Rd, Calabasas, CA 91302.

Circle No 758

\section*{Applications for dual-channel analyzers}

The \(32-\mathrm{pg}\) booklet A World of Applications deals with 12 applications where you can use a dual-channel analyzer to identify and help you
solve engineering problems. In the area of acoustics, it focuses on sound-intensity measurements and architectural acoustics. The section on electroacoustics reviews transducer measurements and sound-reinforcement systems. It also provides an analysis of servo systems and materials. Finally, it outlines how dual-channel analyzers can assist you in college courses.

Bruel \& Kjaer Instruments Inc, 185 Forest St, Marlborough, MA 01752.

Circle No 756

\section*{How to use a scope to find video digital signals}

The application note See Digital Controlled Video Signals and Make Precision Timing Measurements Using the 2467 Portable Scope ( 38 W 6797) explores a wide range of measurement techniques from basic video to digital/frame-rate timing. It reports on the measurement of pulses as narrow as 50 nsec and explains how you can see them in full detail on an MCP (microchannel plate) CRT scope. The note provides several illustrations of this wave-form-viewing feature.

Tektronix Inc, Box 1700, Beaverton, OR 97077.

Circle No 765

\section*{Reference for \\ microwave and RF engineers}

The 61-pg brochure Microwave Datamate deals with topics such as microwave applications by frequency, waveguide parameters, connections, transmission lines, power measurement, and scalar network analyzers. Following the first section on general information is an explanation of satellite and terrestrial telecommunications systems. Another section examines waveguide parameters and has a waveguide data chart. Among the subjects examined in the final section are IEEE-488 programming, a status and message-exchange over-
view, S-parameters and transformations, and scattering parameter relationships.

Marconi Instruments, 3 Pearl Ct, Allendale, NJ 07401.

Circle No 761

\section*{Comprehensive guide categorizes test equipment}

The Test Equipment Reference Guide 1987/1988 is a 375-pg catalog that contains technical specifications and prices for more than 4000 reconditioned test instruments, as well as new instruments, power supplies, coaxial components, waveguides and waveguide components, and a line of technical books. Many items are available for short-term rental or lease. The equipment categories include amplifiers, analyzers, avionics and telecommunications test equipment, frequency measuring instruments, generators, bridges, calibration and standards, meters, oscilloscopes, power supplies, RFI/EMI, and microwave components.

Tucker Electronics Co, Box 461966, Garland, TX 75046.

Circle No 759


\section*{Guide to instrumentation}

The vendor's 1988/1989 product guide covers more than 5000 different models from major manufacturers of electronic test and measurement instruments, data-processing equipment, and telecommunications test devices that you can rent, lease, or buy. The \(400-\mathrm{pg}\) reference book contains specifications, descrip-
tions, photos, and other technical data. Included in its listings are analyzers, CAE/CAD equipment, generators, meters, recorders, oscilloscopes, signal modifiers, microcomputers, and general telecommunications test equipment.

US Instrument Rentals, 2988 Campus Dr, San Mateo, CA 94403.

Circle No 764

\section*{A catalog of microwave products}

The vendor's 1988 92-pg, 4-color catalog highlights microwave measurement components, instruments, and systems in the dc to \(60-\mathrm{GHz}\) range. General information before each major product group helps you to make the best choice for your particular needs. The book also features

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complete specifications for precision measurement components; the K Connector coaxial product line, which operates to 46 GHz ; and a new line of \(40-\mathrm{GHz}\) fixed attenuators. Other new products include vector network analyzers, scalar network analyzers, swept-frequency synthesizers, sweep generators, and RF analyzers.
Wiltron Co, 490 Jarvis Dr, Morgan Hill, CA 95037.

Circle No 757

\section*{App note helps you calibrate DP transmitters}

Application Bulletin P-80 tells you how to calibrate differential-pressure (DP) transmitters on site, thus eliminating the task of having to remove the DP cells and take them back to the shop for calibration. It also describes three calibration devices and provides illustrations.

Rochester Instrument Systems Inc, Test \& Calibration Products, 255 N Union St, Rochester, NY 14605.

Circle No 768

\section*{Scope details}

The vendor's \(10-\mathrm{pg}\) brochure describes its SAS-812A digitizing oscilloscope with which you can observe and analyze ultrahigh-speed waveforms. It discusses features such as jitter reduction, an autotrigger function, and remote control with an external controller via RS232 C and IEEE-488 STD interfaces. Detailed descriptions, including illustrations, provide more information about applications and the evaluation of wide-bandwidth differential amplifiers. The publication also includes an overview of the display screen, as well as specifications and options for the scope.

Iwatsu Instruments, 430 Commerce Blvd, Carlstadt, NJ 07072.

Circle No 767

\section*{LITERATURE: COMPUTER-AIDED ENGINEERING}


Journal for Macintosh users
The CAD/CAM Journal for the Macintosh Professional surveys Macintosh advancements in comput-er-aided design and computer-aided manufacturing. The publication covers the expanding field of CAD/ CAM/CAE applications and features stories about evaluations of 2-D and 3-D drafting and design software, communications and CAD, hardware evaluations, and the most recent CAD/CAM/CAE products. \(\$ 20\) (six issues).

The CAD/CAM Journal, 16 Beaver St, New York, NY 1004.

INQUIRE DIRECT

\section*{Publication spreads CAD news}

The CAD Educator publicizes educational opportunities, new products, activities, and applications in the world of AutoCAD. For example, in the March 1988 issue, Joe Oakey's column notes the transformation of the former Manuals and Tutorials List (renamed In Print) into an updated list of text and support materials for the vendor's products; the lead story reports on advanced AutoLisp training offered by a joint venture of Gold Hill and AutoCAD training centers; and a
feature story focuses on the application of AutoCAD for numerical control. The quarterly also lists upcoming trade shows.

AutoDesk Inc, 2320 Marinship Way, Sausalito, CA 94965.

Circle No 753

\section*{Computer-integrated manufacturing ideas}

The \(12-\mathrm{pg}\) publication The Role of Automated Information in Com-puter-Integrated Manufacturing presents the company's philosophy toward computer-integrated manu-facturing-that is, to unify all production, administrative, and engineering functions in one computer system. The vendor's goals include improved productivity and higher quality. The brochure also presents major concepts involved in integrating management information and plant-floor information.

Allen-Bradley Response Ctr,

Dept 5234, Box 92846, Rochester, NY 14692.

Circle No 752

\section*{Scientific- and engineeringsoftware aids discussed}

Lifeboat, a scientific- and engineer-ing-software guide, describes 100 packages designed for use in solving equations, analyzing data, breaking down numbers, and designing 3-D CAD/CAM. The products are listed side by side to make it easier for you to compare them and make a selection. The product categories include circuit design, embedded systems, data acquisition/signal analysis, languages/utilities, Basic, C, cross-assemblers, and Fortran.

Lifeboat Associates Inc, 55 S Broadway, Tarrytown, NY 10591.

Circle No 751


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High character height and easy-to-read display.
Easy-to-use design thanks to high contrast TN.
\begin{tabular}{|c|c|c|}
\hline Model name & \begin{tabular}{c} 
Number of \\
characters
\end{tabular} & \begin{tabular}{c} 
Outline dimensions \\
\((\mathbf{m m})\)
\end{tabular} \\
\hline TLC-671 & \(16 \times 1\) & \(80.0 \times 36.0 \times 12.0\) \\
\hline TLC-241 & \(16 \times 1\) & \(80.0 \times 36.0 \times 12.0\) \\
\hline TLC-491 & \(16 \times 2\) & \(80.0 \times 36.0 \times 12.0\) \\
\hline TLC-731 & \(16 \times 4\) & \(87.0 \times 60.0 \times 12.0\) \\
\hline TLC-501 & \(20 \times 2\) & \(116.0 \times 37.0 \times 12.5\) \\
\hline TLC-721 & \(20 \times 4\) & \(98.0 \times 60.0 \times 12.0\) \\
\hline TLC-691 & \(24 \times 1\) & \(126.0 \times 36.0 \times 12.0\) \\
\hline TLC-771 & \(24 \times 2\) & \(118.0 \times 36.0 \times 12.0\) \\
\hline TLC-601 & \(40 \times 1\) & \(182.0 \times 33.5 \times 13.0\) \\
\hline TLC-591 & \(40 \times 2\) & \(182.0 \times 33.5 \times 13.0\) \\
\hline TLC-1001 & \(40 \times 4\) & \(221.0 \times 76.0 \times 12.5\) \\
\hline
\end{tabular}

\section*{Graphic Display Type with Built-in Controller}

Clear display thanks to high contrast TN
Easy-to-use C/G, RAM and ROM built-in types.
\begin{tabular}{|c|c|c|c|}
\hline Model name & \begin{tabular}{c} 
Number of \\
dots
\end{tabular} & \begin{tabular}{c} 
Outline dimensions \\
\((\mathbf{m m})\)
\end{tabular} & Controller \\
\hline TLC-1021 & \(120 \times 64\) & \(85.0 \times 70.0 \times 20.0\) & T6963C \\
\hline TLC-682 & \(160 \times 64\) & \(125.0 \times 50.0 \times 18.0\) & T6963C \\
\hline TLC-711A & \(240 \times 64\) & \(180.0 \times 65.0 \times 12.0\) & T6963C \\
\hline TLC-1013 & \(160 \times 128\) & \(129.0 \times 104.5 \times 14.0\) & T6963C \\
\hline TLC-1091 & \(240 \times 128\) & \(241.0 \times 125.3 \times 12.0\) & T6963C \\
\hline TLC-1101 & \(160 \times 32\) & \(140.0 \times 40.0 \times 12.0\) & T6963C \\
\hline \multicolumn{4}{|c|}{\begin{tabular}{c} 
"We provide many options. \\
"We can meet the need for customized products.
\end{tabular}} \\
\hline
\end{tabular}

\section*{Graphic Display Type}

Clear display through employment of high contrast TN.
Capable of displaying large-size letters and special letters.
\begin{tabular}{|c|c|c|c|}
\hline Model name & \begin{tabular}{c} 
Number of \\
dots
\end{tabular} & \begin{tabular}{c} 
Outline dimensions \\
\((\mathbf{m m})\)
\end{tabular} & Controller \\
\hline TLC-612 & \(640 \times 32\) & \(320.0 \times 48.0 \times 14.0\) & \((\) T6963C) \\
\hline TLC-761 & \(640 \times 64\) & \(320.0 \times 47.0 \times 14.0\) & \((\) T6963C) \\
\hline TLC-341AK & \(128 \times 128\) & \(93.2 \times 86.6 \times 12.0\) & \((\) T6963C) \\
\hline TLC-531A & \(128 \times 128\) & \(132.2 \times 111.0 \times 13.0\) & \((T 7755)\) \\
\hline TLC-972 & \(480 \times 128\) & \(274.0 \times 114.4 \times 20.0\) & \((T 7755)\) \\
\hline
\end{tabular}

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\section*{How two engineers built -and nearly losttheir business}

\begin{abstract}
Mike Wells and Jack Hancock started their company with plenty of technical know-how but no business experience. Together, they made every mistake in the book. After a series of near-disasters, which gave them an intensive on-the-job business education, they expect \(\$ 8\) million in sales this year.
\end{abstract}

They make the oddest of couples. Mike Wells is 32 years old. After college, he spent only two years in the engineering work place before he became disillusioned with corporate attitudes and fled. Jack Hancock is a 67 -year-old mechanical engineer who worked for the same employer for 31 years. Somehow, the two men strike a perfect balance as business partners.

In 1985, however, two years after they founded Arnet Systems, the company's equilibrium was anything but balanced. The prospects for success looked grim. The Internal Rev-

\author{
Deborah Asbrand, Associate Editor
}
enue Service was demanding payment of \(\$ 30,000\) in witholding taxes. Local banks refused them credit, and representatives of one bank recommended liquidation. A year later, the situation was even worse. The loss of a major contract forced them to lay off nearly half of their employees, and the corporate infrastruc-ture-what there was of it-was about to cave in.

Luckily, Arnet Systems was able to regain its equilibrium. Part of the
ballast is no doubt a function of the differences between Mike Wells and Jack Hancock. "The brashness and risk-taking innate to Mike are foreign to Jack," observes one of their colleagues, Bob McKeown. "And Jack's solidness and conservatism are foreign to Mike." Wells serves as Arnet's president and supervises its long-range growth and market plans. Hancock is the company's chairman and oversees investments, stock transactions, and export activities.

The two men first met in 1972. When Wells was 16 years old, Han-


Mike Wells: "The PC market was easy to enter and attracted lots of young people with no credibility -like me."

> Jack Hancock: "I'd always toyed with the idea [of starting a business], but I never got serious about it until Mike came along."
cock, a 51 -year-old divorced father of two, married Wells' mother, Betsy. The men's early relationship was tenuous at best. "We just kind of avoided each other for a while," Hancock recalls. Eventually, they became friends, and when Wells went off to Vanderbilt University, he chose to study electrical engineering, a decision both men now attribute to Hancock's influence.

In 1980, Wells graduated from Vanderbilt with a master's degree in electrical engineering. He went to work for Merrick Corp, a Nashville company specializing in robotic are welding. Shortly afterwards, the owners sold the business to a large, Houston-based company. Immediately it became apparent that the new managers had little understanding of engineering and little interest in preserving Merrick's small-company atmosphere.

Morale in the engineering department, where Wells and about a dozen others worked, quickly deteriorated. "We were behind, and the managers came in and told us that the products had to ship on the scheduled day," Wells remembers. "We told them that the products weren't ready, that they weren't debugged. They said 'ship them anyway.' Twenty units were shipped, and every one of them came back. It was the beginning of their demise. It ruined their reputation."
Feeling disenfranchised by the
business's bottom-line philosophy, Wells left after two years. He returned to Vanderbilt to earn a doctoral degree in electrical engineering. After one year back on campus, however, he grew restless.

Around the same time, Hancock was considering an early retirement from E I du Pont de Nemours, where he worked as a senior research engineer in the company's textile-fiber division. Hancock saw his early leavetaking as an opportunity not to sit in a rocking chair but to try something new. "I didn't really want to retire," he recalls. "I'd always toyed with the idea [of starting a business], but I never got serious about it."

\section*{Hanging out a shingle}

In April 1983, Wells and Hancock agreed to give entrepreneurship a try. The plan, albeit a bit fuzzy, was to set up a consulting business specializing in industrial-control equipment. Hancock would draw on his expertise in machine design, and Wells would contribute his knowledge of microprocessors.

The chances were slim that the two business novices would make it in the rough-and-tumble electronics industry. For Wells and Hancock, the past few years have been anything but easy. Like many small businesses, Arnet's growth has been a wild roller-coaster ride of highs and lows, but what saved the
company from going out of control was its founders' quickly learned lessons and a rare candor in admitting their mistakes.
The two men capitalized the business themselves, retaining full ownership, and, as they would later come to realize, full responsibility. Hancock took out a second, \(\$ 15,000\) mortgage on his home, and Wells kicked in his \(\$ 2000\) in savings. They rented a small office in downtown Nashville. They named the business Arnet Systems because "the theory was to have a business that started with the letter ' \(A\) ' so we'd be first in the consultants directory. We also wanted something that sounded high tech." After a series of friendly disputes, Wells and Hancock settled on Arnet, which is the latter's middle name.

Neither man enjoyed consulting. They both disliked making cold sales calls and designing products that clients might or might not market. Even worse, they weren't making any money. "We always underestimated how much time we needed on a project-probably by about \(50 \%\)," Wells says. "Usually we got negotiated down and wound up working for \(\$ 10\) /hour rather than the \(\$ 45 /\) hour that we'd planned on." The handful of small contracts that they managed to scare up brought in little money. The first year Arnet lost \(\$ 6000\).

The consulting effort, though,
was part of a larger plan that Wells and Hancock had hatched. They envisioned consulting as a stepping stone to the business they wanted to build. Wells was one of the legions of people attracted to the potential gains of the burgeoning personalcomputer industry. "For a while I thought we would make a standard product in the industrial-control area," he says. "But the area didn't seem to be growing. And, because it was stable, it was hard to enter. The PC market was easy to enter and attracted lots of young people with no credibility-like me."

\section*{Choosing a standard product}

To transform their struggling consulting practice into a fullfledged business, Wells and Hancock needed a standard product. In December 1983, they got their opportunity. A local company was planning to enter the personal-computer market with a software program that monitored telephone switchboards. The company was dissatisfied with the design of the 4-port RS-232C card that a Minneapolis vendor had developed for it, and agreed to let Arnet Systems
redesign the card and implement new features on it.
"Because we were so small and unknown, they had absolutely no faith in us," Wells says. For this reason, the deal they struck required Arnet to absorb the design costs and committed the client to purchase a maximum of only 50 boards. In return, Arnet retained full ownership of the product.

Within four months, Wells and Hancock demonstrated a printed-circuit-board prototype for their client. Not only did the company purchase the required 50 boards, but it continued to buy them at a rate of 50 each month. With that contract earning Arnet \(\$ 25,000\) a month in revenue, Wells and Hancock could begin plowing some of their earnings into the development of their own products. Fortuitously, they got the idea of enhancing the pc board and making it the longawaited product they had been looking to build a company upon from another vendor's product brochure.
The two men also decided they could now afford to hire another full-time engineer. Wells turned to Bob McKeown, a colleague from his
days at Merrick. McKeown, like Wells, had chafed under Merrick's new owners' attempts to impose strict rules on the engineering department. The new manager "felt we lacked commitment because we lacked discipline," McKeown remembers. "His way of making us disciplined was to institute a dress code and to rigidly enforce working hours."

One experience, in particular, cemented McKeown's belief that the new management's policies were largely misguided. The day after a memo instructed all male employees to wear ties at work, McKeown arrived wearing blue jeans-and no tie. His action, he says, carried no message of defiance. To McKeown, it was simply keeping in line with Nashville's southern casualness. "I wore my best dressy jeans and didn't think of it as rebellious."

\section*{An enlightening ride}

His new manager did. He asked McKeown to accompany him for a drive to his ranch 20 miles outside of Nashville. There, the manager pointed to the livestock and explained that there were two kinds of


Bob McKeown: "I still live vicariously through the engineering department."

\title{
Just days before the IRS planned to padlock Arnet's doors, the company found a bank willing to lend them money.
}
cattle-those that accepted nurturing and were allowed to go on producing, and those that wound up on dinner tables. What kind of cattle, the supervisor wanted to know, was McKeown?
McKeown's memory of the incident is vivid. "I thought he was a lunatic," he says. "I started looking for a job and took the first one I got." A year later, when Wells contacted him about joining Arnet, he jumped at the chance.

McKeown began working on some of the company's existing contracts, and Wells began looking around for other product niches. The issue of copy-protection schemes for personal computers was attracting a lot of attention at that time, and Wells got the idea to design protective devices. "I'd read in a databook about using PAL chips to protect software. We started thinking maybe this could be a standard product for us." It was a market different from the add-in one that the company was hoping to penetrate with its expansion board, but it seemed to hold promise. Wells and Hancock had no idea that venturing into two disparate markets would nearly cause the company's downfall.

\section*{Mounting a marketing effort}

In the spring of 1985, Arnet Systems was ready to release both the copy-protection device, which it called Gardware, and the upgradable IBM PC expansion board, which it called Multiport. There was no extra capital to fund the salary of a sales and marketing professional, so Wells, already working hard as head of new-product development, assumed the responsibility for assembling a marketing campaign. He
checked a dozen books out of the library, pored over all of them, and became a self-taught product promoter. He wrote press releases, arranged for photo shoots, concocted direct-mail packages, and worked on company and product brochures.

\section*{Times were improving}

By August, the company's efforts began to show results. Multiport and Gardware were selling well. Gardware, in particular, was generating a great deal of revenue for the young company, thanks to a lucrative contract with Computervision, a Bedford, MA, CAE company that was buying thousands of units each month. Arnet had eight employees and was growing rapidly, and the 7-day workweeks were beginning to pay off.
Any sense of accomplishment, however, was fleeting. Despite steady product sales, the business needed a cash infusion to provide working capital and, more important, to pay the IRS \(\$ 30,000\) for back payroll taxes. Wells and Hancock briefly considered venture-capital sources, but instead opted for more homegrown financing methods. The trouble was they had nearly reached their personal-financing limits. Hancock's home was mortgaged to the hilt, and he had run up large credit-card debts by taking out cash advances.
That fall, Wells and Hancock began offering their employees ownership stakes in the company. McKeown remortgaged his house for \(\$ 11,000\). In return, he received \(10 \%\) ownership in the company. Lisa Ernst, Arnet's controller, contributed her savings and became a \(6 \%\) owner. The company pitch was also heard by miscellaneous aunts, un-
cles, and other relatives with untapped bank accounts.
At the same time, Hancock, Wells, Ernst, and Elaine Floyd, an Arnet sales representative, were also giving the team pitch to area banks in the hopes of securing a loan or a line of credit. Complicating matters was pressure from a local bank through which Arnet had a Small Business Administration loan. Nervous about Arnet's precarious financial condition, the bank sent two representatives to assess the company's condition. The men (both electronics-industry veterans) toured Arnet's facilities and then spent several hours with Wells. Their verdict was liquidation. "They didn't even suggest Chapter 11," Wells says, referring to the bank-ruptcy-law provision that offers a company protection from its creditors while it reorganizes its finances. "They advised us not to compete in the add-in market and to go back to consulting."

Wells refused to comply. "They had the 'you can't compete with the big guys' mentality," he says. Wells's tenacity was a critical motivating force for the others. "Mike just wasn't going to give up," says Ernst. The group continued meeting with other bank officers, and just days before the IRS planned to padlock Arnet's doors, it found a bank willing to advance the necessary funds.
Although Wells, Hancock, et al thought they had orchestrated a reprieve, they had already committed a near-fatal mistake that was soon to become apparent. Without realizing the danger of it, Arnet's managers had allowed the company to become dependent on the revenue from the Computervision contract.

\title{
Arnet nicknamed its 6-week effort "Rambo marketing" and assigned every available staff member to follow up on sales leads.
}
"We had built up the company based on revenues from one customer, and we failed to realize how vulnerable we were," recalls Ernst.

\section*{A bird in the hand . . .}

In early 1986, a Gardware EEPROM began failing in the field. Arnet lacked the resources to trace the problem and rectify it, and Computervision, which had purchased more than \(\$ 600,000\) worth of the EEPROM, grew concerned about quality control. "What was really going on was that between Multiport and Gardware, we were going crazy with customer support," explains Wells. Arnet decided to forgo further development of Gardware. It did plan to continue providing support for its existing customers, however.

Abandoning further development irked Arnet's biggest customer. Computervision cancelled its contract and gave its business to Rainbow Technologies, Arnet's closest competitor. "Our income dropped \(50 \%\) overnight," Wells says. To make up the badly needed revenue, the company mounted an all-out campaign to increase sales of Multiport. Arnet nicknamed the effort "Rambo marketing" and assigned every available staff member to follow up on sales leads that had come in over the past few months.

The intensive 6 -week effort brought in some extra revenue, but not enough. Without drastic measures, the company would be forced to close its doors. "It got scary," remembers Ernst. "There was a lot of pressure, and we knew our survival depended on laying people off." An emergency meeting in June determined that 9 of the company's 22 employees had to be let go. Most of
these had been hired within the previous six months, and they came from a variety of departments within the company. Wells and Hancock met with the soon-to-be-laid-off employees, explained the company's status, and offered them an admittedly meager severance packagetwo weeks' pay and the preparation of their resumés.

\section*{A favorable outlook-again}

While Arnet limped along, the company's sales efforts gradually began to show results. Sales of Multiport started to climb. The company was inching back from the precipice-or so it seemed. To operate efficiently and avoid repeating past mistakes, however, Arnet had to synchronize its inner workings. As president, director of R\&D, and company marketing representative, Wells had assumed responsibility for many of the company's key functions. As Arnet grew, he was stretched ever thinner and was slowly grinding Arnet's progress to a halt. "The company was moving along but there was no way for people to sit down and plan together and get a feeling for what was going on," says Ernst. "Mike was so involved with the marketing that it was as if we had no leader."

Finally, Ernst took Wells aside and minced no words in describing the business's status. Looking around the company, Wells says, he knew she was right. The problems were genuine-and acute. Among other things, the company was constantly running out of inventory and was slow to fill orders. Members of the engineering department were unsure of their direction. "I was making lots of decisions in a vaccuum," Wells concedes. A short
time later, he hired a sales and marketing manager and turned his attention to working full-time as the company president and new-product director.

\section*{Rounding the corner}

For a while, the new organization worked. But even after passing on the sales and marketing responsibilities, Wells remained overextended. By early 1987, the pressure had become too much. At a staff meeting one January morning, he lost his cool. "Everything was behind schedule, and there seemed to be 25 problems to solve. I seemed to be the only one concerned. No one seemed to be taking any initiative or getting fired up.
"I screamed and called [the managers] wimps." Fuming, Wells stormed out of the meeting and left the office for the day. "I wanted to quit." His ego got the best of him, he admits. He almost wanted to leave for a month and see the company crash and burn. "On the other hand," Wells says, "I was begging for help." Several days of introspection ensued. "My ego had definitely become a problem," he continues. "I didn't understand how I could be president and not be in charge of everything."

Among the first remedial steps Wells took was to distribute some of his responsibilities. To relieve himself of day-to-day management duties, he created the position of chief executive officer. "I needed to find someone who liked managing," Wells says. "I hated it all-the meetings, the performance reviews, all the details."

Ernst became chief executive officer, and Wells promoted McKeown to executive vice president.

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McKeown realized that it was a position of prestige, but it also meant leaving behind his engineering tasks. "I had to do a lot of soul searching to decide if it was the right thing for me," McKeown says. "My specialty was never microprocessor technology, so the company had really evolved out from under me. But the excitement of being part of this organization had always been enough compensation for me."

He takes pride in the fact that one of his board designs recently received a top rating from Infoworld magazine, and he adds, "I still live vicariously through the engineering department." Now ensconced in administrative work, McKeown commemorates his days at Merrick with a painting of a cow, which hangs on his office wall.

As another measure to further diffuse the decision-making process, Wells and Hancock created an executive committee to make decisions by consensus. Besides the company's founders, Ernst and McKeown became board members.

Arnet now has 35 employees, 15 of whom own stock in the company. In 1988, Wells and Hancock expect to double their fiscal 1987 profits. They've continued to expand their product line and now produce four communications boards. In April, the Nashville Business Journal bestowed its Small Business of the Year award on Arnet. Buzz Heidtke, a member of the judging panel, visited the company and came away impressed both by Arnet's comeback and by the atmosphere it has created. As Heidtke says, "They got some good contracts and were a bit too successful too early. Now, they seem to really work on quality."

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\section*{POSITIONDESIRED}


Reason for Change:

\section*{PREVIOUS POSITION:}

Job Titie:
Employer: \(\qquad\) From: \(\qquad\) To: \(\qquad\) City: \(\qquad\) Salary: State:
Division: Type of Industry: Salary:
COMPENSATION/PERSONAL INFORMATION


WILL RELOCATE
\(\square\) WILL NOT RELOCATE
\(\square\) OTHER


BS/MS in Computer Science or Electrical Engineering and previous experience with " C ", RMX operating systems, Intel microprocessors, data communications protocols (specifically HDLC) and real-time applications.

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BS/MS in Electrical Engineering and previous experience with digital circuitry/firmware design, analog circuit design, PCM-channel bank and line/trunk interface design or circuit design simulation. ASIC background is a definite plus as well as Intel microprocessor experience.

\section*{SYSTEMS ENGINEERS}

BS/MS in Computer Science or Electrical Engineering and previous experience in digital loop carrier systems, system configuration, system level problem resolution and automated system testing.

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\(\mathrm{BS} / \mathrm{MS}\) in Electrical Engineering or Mechanical Engineering and previous experience in telecommunications systems applications design, digital loop carrier and lightwave transmission. Responsibilities include creating standardized/specialized systems for customer

\section*{—} -

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\title{
LOOKING AHEAD
}

\section*{Electronic still imaging to gross \(\$ 540 \mathrm{M}\) in 1992}

The market for equipment that produces and processes electronic still images, a relatively new commercial technology, should flourish during the next few years. Estimated at \(\$ 68.5 \mathrm{M}\) for this year, the international market for electronic still-image equipment will reach \(\$ 542.6 \mathrm{M}\) by 1992, according to Electro-Imaging Advisors Inc (La Jolla, CA). Improvements in picture quality, together with the usual reductions in price that accompany a rise in sales volumes, will cause the market to grow dramatically by 1990 .
A somewhat new approach to photography, electronic still imaging solves the problems involved in accumulating photos and sending them economically to a distant location in a short period of time. The technology includes the capture of images by a sensor-typically of heat or light-and the subsequent processing, transmission, storage, and display of such images in individual frames. A charged-couple-device sensor, a video camera, or an electronic still-image camera captures the images for processing.
In spite of its youth, the industry has some established standards. All

recording devices use a 2 -in. floppy disk, and the format for recording allows the recording of 25 frames or 50 fields on each disk. The images can be transmitted via a still-frame communication unit, or via telephones or satellite. The person who receives the images can view them on a monitor or on hard copy and then store them on a hard disk for later reference.

EIA has identified seven key commercial markets. Very large companies, for example, will use the technology for presentations, archives, artwork and advertising files, communication systems, and security. Within five years, corporations will have electronic still-imaging systems in 20 to \(25 \%\) of their facilities.

\section*{Demand for T3 test devices is expected to boom}

The digitization of America, fueled by the Bell operating companies and other, independent telecommunications companies, should quadruple the market for T3 transmission test equipment over the next five years, according to Able Telecommunnications Inc, a consulting and market research company based in Milpitas, CA. Currently worth \(\$ 24\) million, that market should attain a \(\$ 100\)
million value by 1992 .
Able defines T3 transmission test equipment as systems that perform separate testing, monitoring, and error-testing functions on T3 lines, which feature digital bit streams to 44M bps. Interchange carriers, large corporations, and other organizations that use fiber-based networks are expected to follow the telecommunications companies' lead
over the next few years.
In terms of unit sales, the market will grow from 3000 this year to 16,700 in 1992. At the same time, the price of a single test system should drop. Right now, T3 test systems average \(\$ 8000\) per unit; by 1992, they should cost about \(\$ 6000\) each.


\section*{dc to 2000 MHz amplifier series}

SPECIFICATIONS
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[t]{2}{*}{MODEL} & \multirow[t]{2}{*}{\begin{tabular}{l}
FREQ. \\
MHz
\end{tabular}} & \multicolumn{3}{|c|}{GAIN, dB} & & \multirow[t]{2}{*}{- MAX PWR. dBm} & \multirow[t]{2}{*}{\[
\begin{aligned}
& \mathrm{NF} \\
& \mathrm{~dB}
\end{aligned}
\]} & \multirow[t]{2}{*}{PRICE
Ea.} & \multirow[t]{2}{*}{\$ Qty.} \\
\hline & & \[
\begin{gathered}
100 \\
\mathrm{MHz}
\end{gathered}
\] & \[
\begin{aligned}
& 1000 \\
& \mathrm{MHz}
\end{aligned}
\] & \[
\begin{aligned}
& 2000 \\
& \mathrm{MHz}
\end{aligned}
\] & Min. (note) & & & & \\
\hline MAR-1 & DC-1000 & 18.5 & 15.5 & - & 13.0 & 0 & 5.0 & 0.99 & (100) \\
\hline MAR-2 & DC-2000 & 13 & 12.5 & 11 & 8.5 & +3 & 6.5 & 1.50 & (25) \\
\hline MAR-3 & DC-2000 & 13 & 12.5 & 10.5 & 8.0 & +80 & 6.0 & 1.70 & (25) \\
\hline MAR-4 & DC-1000 & 8.2 & 8.0 & - & 7.0 & +11 & 7.0 & 1.90 & (25) \\
\hline MAR-6 & DC-2000 & 20 & 16 & 11 & 9 & 0 & 2.8 & 1.29 & (25) \\
\hline MAR-7 & DC-2000 & 13.5 & 12.5 & 10.5 & 8.5 & +3 & 5.0 & 1.90 & (25) \\
\hline MAR-8 & DC-1000 & 33 & 23 & - & 19 & +10 & 3.5 & 2.20 & (25) \\
\hline
\end{tabular}

NOTE: Minimum gain at highest frequency point and over full temperature range.
- 1dB Gain Compression
- +4 dBm 1 to 2 GHz

\section*{designers amplifier kit, DAK-2}

5 of each model, total 35 amplifiers only \$59.95

Unbelievable, until now ...tiny monolithic wideband amplifiers for as low as 99 cents. These rugged 0.085 in.diam.,plastic-packaged units are 50ohm* input/output impedance, unconditionally stable regardless of load*, and easily cascadable. Models in the MAR-series offer up to 33 dB gain, 0 to +11 dBm output, noise figure as low as 2.8 dB , and up to DC-2000MHz bandwidth.
*MAR-8, Input / Output Impedance is not 50 ohms, see data sheet
Stable for source/load impedance VSWR less than \(3: 1\)
Also, for your design convenience, Mini-Circuits offers chip coupling capacitors at 12 cents each. \(\dagger\)
\(\left.\begin{array}{cccl}\begin{array}{c}\text { Size }\end{array} & \text { Tolerance } & \begin{array}{c}\text { Temperature } \\ \text { (mils) }\end{array} & \\ 80 \times 50 & 5 \% & \text { Characteristic }\end{array}\right)\)

\title{
Nowt HEXSense makes today's current-sensing obsolete.
}


Old Method
This circuit uses a fractional value resistor (A) to measure current, causing a voltage drop which increases power losses. Its parasitic inductance also slows down switching speed. To offset these losses, a lower \(\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}\) power MOSFET may be used, increasing circuit cost.

HEXSense offers 5-pin lead forms for vertical and horizontal pcb mounting.


 es
 5



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\title{

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\title{
Design Ideas Special Issue VOLUME I
}

\section*{ELECTRONIC TECHNOLOGY FOR ENGINEERS AND ENGINEERING MANAGERS}


DESIGN 1
Linear load dissipates con
Horace T.Jones
Penrit Datacomm, Gaithersburg. MD
The Fig 1 cireuit presents a constunt-power load to the input voltage \(V_{1 s}\). For the component values nhown, the circuit provides a \(4 W\) load at a nominal \(V_{1 N}\) of \(13 V\) and
maintaina this mower level within \(\pm 0.2 \%\) for input maintains this power level within \(\pm 0.2 \%\) for input
voltages voltages of 9 Lo 17 V . Pot
the constant-power level.
Select \(R_{\text {, a }}\) and \(R_{0}\), po the inverting inpat of \(1 \mathrm{C}_{1}\). Op introa
approximately 6 V at
ing aetion approximately
amps IC.A and IC, form u pulse-width moduhator whose
ing netron
Lo about output duty eycle ( X ) is a linear function of 13 is XH
berating frequency (f)
To Vote For This Denikn, Clrcle No 74

\title{
When World Class Precision and Performance Are Required
}


\section*{Highest precision available with \(10 \mu V\) input offset voltage from Raytheon.}

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mance. Additionally Raytheon offers an 8 -lead SOIC specified at \(\pm 25 \mu \mathrm{~V}\).
\(\square\) Well balanced specs:
\(\mathrm{I}_{\mathrm{B}}: 2 \mathrm{nA}\) maximum Gain: 5 million minimum Power dissipation: 50 mW maximum CMRR: 120 dB minimum PSRR: 110 dB minimumCompanion product: Raytheon's
LT1001 high-precision, high-performance op amp follows RC4077's lead with a very low \(15 \mu \mathrm{~V}\) offset voltage. The LT1001 offers 2 nA offset current and gain of .45 million minimum.
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FFT: Detailed distribution of noise and harmonic distortion vs. input signal. Measurement of sample and hold and ZAD2716 A/D converter only. Computer analysis receives digital information from \(\mathrm{A} / \mathrm{D}\) converter (to avoid DAC reconstruction distortion).


Using an Audio Precision system, a sine wave is digitized by the ZAD-2716, then reconstructed by the ZDA-1801, and filtered by Apogee's 944 S linear phase filter. Graph shows combined total harmonic distortion and noise vs. frequency.

\section*{Introducing the world's finest professional 16-bit stereo digitizing subsystem and 18-bit deglitched reconstruction DAC.}

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If seeing is believing, you'll be impressed by the harmonic distortion and noise plots of our ZAD2716/ZSH202 digitizing subsystem (left) and ZDA1801 deglitched reconstruction DAC (right). Nobody's ever seen plots like those before.

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The ZAD2716-2 coupled with the ZSH202 Dual Sample-Hold can provide stereo digitizing at a 50 kHz sampling rate per channel. The ZSH202 allows simultaneous sampling of both channels within \(\pm 2\) nsec with guaranteed \(-96 \mathrm{db} /\) channel separation. And there's no sacrifice of phase linearity or \(\sin x / x\). In fact, we guarantee \(\sin x / x\) performance of 0.01 dB and phase linearity of 0.1 (degree) over 0 to 20 kHz . (Performance that brings a smile to any Golden Ear.) Cost savings are significant on a per channel basis, because you only need one converter for two channels.

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Your needs extend over the entire audio spectrum. And that's how our products are tested. We test the way they're used. So we can guarantee signal-to-noise plus harmonic distortion of 92 dB over the entire \(0-20 \mathrm{kHz}\) audio range.

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\title{
Today, this memory card will travel 15,000 miles, perform massive data collection, transfer data to a central data file and...
}

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\section*{The Applications Are Virtually Endless.}

From personal computers to printers, portable equipment to telecommunications, the applications are as creative and endless as the imagination. Anywhere transportable memory and massive data storage are required, Mitsubishi's transportable cards offer the highest memory densities available on a device this small.

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\section*{\(\square\) Chip-On-Board}
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\section*{INTRODUCTION}

Every batch of Reader Service cards we receive contains many comments about our Design Ideas section. Most often, readers tell us they've found a particularly useful circuit that helped them solve a tricky design problem. But readers also want to know how they can get a complete set of past circuits and software. Alas, such a compendium doesn't exist. But this Design Ideas Special Issue represents the next best thing-a collection of 50 of the best Design Ideas published in EDN from 1985 through June 1988.
This collection of designs reflects your needs and not the preferences of our technical editors. In fact, you and your colleagues voted for these winning ideas by circling the "bingo card" numbers that appear at the end of each Design Idea we publish. Of the circuit-design tips in this Design Issue, 29 garnered Best-ofIssue honors based on reader preference. The remaining Design Ideas received the second-highest number of reader votes over the 1985 to 1987 time frame.

In planning these Design Ideas Special Issues, EDN hoped to be able to present a good mixture from the standpoint of design disciplines. Happily, our readers' choices made it easy to achieve this goal. This issue contains 21 analog circuit designs and 24 digital designs, and the remaining five ideas highlight programming tips. To achieve some semblance of order, we've divided the ideas into some generic categories and provided an index that lists design and circuit categories. Although the index is a handy place to look for a special circuit, we'll bet that most readers will want to read the ideas one by one, just to see what circuits are available.

As you read these Design Ideas, remember that each one came from a reader who thought someone else might benefit from his or her work. So, if you find this issue helpful, thank your fellow engineers who submitted the ideas we've published over the years. And the next time you have an interesting and useful circuit, consider sharing it with other engineers by submitting it for publication in EDN's regular Design Ideas section.

Besides seeing your name in print and gaining an extra \(\$ 100\), you might find your idea has been selected as the issue winner. Each issue winner collects an extra \(\$ 100\) also. Keep in mind, too, that EDN's editors also choose a grand-prize winner each year. The grand prize includes a check for \(\$ 1500\). Your idea may also be selected to appear in a future Design Ideas Special Issue. You'll find an entry form in the Design Ideas section in most issues of EDN.

Our commitment to short, comprehensive design solutions includes two more Design Ideas Special Issues in 1988. You'll receive another with your next July showcase issue, and we'll publish a third later in 1988. If you enjoy this Design Ideas Special Issue and find it useful, we'd like to hear from you. If you have suggestions for improvements and changes, we'd like to hear those, too. Just send us a note or give us a call. We'd also like to thank all the readers who have submitted Design Idea entries over the years.

Enough introduction. Here are the ideas you've been clamoring for. Good reading.


Tom Ormond
Senior Editor


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\section*{DESIGN IDEAS}

\section*{Dual op amp forms RS-232C driver/receiver}

\author{
Jean-Paul Lambrechts \\ National Semiconductor, Santa Clara, CA
}

To provide a single interface between an RS-232C line and a MOS LSI chip while saving board space and power, you can use a dual op amp (Fig 1) in place of the traditional 1488-type line driver and 1489-type line receiver. Further, the LF353N op amp handles transmission rates as high as 19,000 baud without exceeding the EIA's specification for maximum slew rate.

When node B is high, clamping action of the zener diode \(\mathrm{D}_{2}\) keeps op amp \(\mathrm{IC}_{1 \mathrm{~A}}\) in its linear region of operation. The 3.3 V zener voltage is impressed across resistor \(\mathrm{R}_{3}\), which injects a constant current ( \(84 \mu \mathrm{~A}\) ) into node A . The resulting voltage at node A , by a Thevenin-equivalent analysis, is 1.82 V ; thus, node B is \(1.82+3.3=5.12 \mathrm{~V}\) ( 7 V is the maximum allowed by \(\mathrm{IC}_{2}\) ). This 5.12 V output at \(\mathrm{IC}_{1 \mathrm{~A}}\) is stable for all bipolar-low levels normally encountered at the RS-232C line input, ie, -2 V and below for most terminals.

As the line input rises to the high level, the node A voltage decreases, causing the current through \(\mathrm{D}_{2}\) to reverse direction for some input level below 1.82 V . The resulting forward bias across \(\mathrm{D}_{2}(0.7 \mathrm{~V})\) causes node A to drop to 0.85 V , yielding \(0.85-0.7=0.15 \mathrm{~V}\) at node B . The circuit is actually a Schmitt trigger that behaves as a voltage regulator in both stable states. That is, node B assumes 0.15 V for inputs above 0.91 V and assumes 5.12 V for inputs below approximately 0 V .

In the line-driver section, op amp \(\mathrm{IC}_{1 \mathrm{~B}}\) acts as a comparator, with the node A voltage ( 0.85 or 1.83 V ) on its noninverting input. The corresponding serial output (SO) of \(\mathrm{IC}_{2}\) is 2.4 or 0.4 V , so \(\mathrm{IC}_{1 \mathrm{~B}}\) inverts the SO signal as desired. Diodes \(\mathrm{D}_{1}, \mathrm{D}_{3}\), and \(\mathrm{D}_{4}\) provide overvoltage protection, and \(R_{6}\) limits the current exchanged with the line.

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Fig 1-A dual op amp provides a single-channel interface between an \(R S\)-232C line and a MOS terminal-management IC.

\section*{DESIGN IDEAS}

\section*{Circuit provides controllable resistance}

\section*{H H Eck}

Pulse Electronics Inc, Rockville, MD
The bipolar transistor \(Q_{1}\) in Fig 1 presents a linear, dependent resistance between nodes A and B. By extending the connections from \(\mathrm{IC}_{1}, \mathrm{IC}_{2}\), and \(\mathrm{Q}_{1}\) 's base, you can remotely control a resistance value.
Nodes A and B may assume any voltage value within the output range of op amps \(\mathrm{IC}_{1}\) through \(\mathrm{IC}_{5}\), provided the voltage of node A is more positive than B. You set \(R_{A B}\) (the resistance between \(A\) and \(B\) ) to its maximum value by adjusting \(R_{B}\) to its maximum. \(R_{A B}\) can't go to zero, but the divider action of \(\mathrm{R}_{\mathrm{K}}\) provides a lower (and adjustable) limit for the minimum value for \(R_{A B}\). If \(R_{B}\) alone gives enough range, you may eliminate \(R_{K}\) and \(\mathrm{IC}_{4}\).
\(R_{A B}\) remains linear over a range exceeding 20:1. In the test configuration in Fig 2, \(\mathrm{R}_{\mathrm{AB}}=0.061 \mathrm{R}_{\mathrm{B}} \pm 2 \%\) as \(\mathrm{R}_{\mathrm{B}}\) varies from 34 to \(825 \mathrm{k} \Omega\).
For simplicity's sake, you can assign any convenient value to resistors \(R\) and \(R_{K}(\mathrm{eg}, 5 \mathrm{k} \Omega)\) (Fig 1). To derive an expression for \(R_{A B}\), note that \(V_{1}=\left(V_{A}-V_{B}\right)\), \(\mathrm{V}_{3}=\left(\mathrm{V}_{\mathrm{A}}-0.6 \mathrm{~V}-\mathrm{V}_{2}\right)\), and \(\mathrm{V}_{2}=\mathrm{V}_{1} / \mathrm{K}\), where K is the ratio of \(R_{K}\) 's wiper voltage to \(V_{1}\). Then,


Fig 2-In this test configuration, \(R_{A B}=0.061 R_{B} \pm 2 \%\) as \(R_{B}\) varies from 34 to \(825 k \Omega\).
\[
\mathrm{I}_{\mathrm{C}}=\beta \mathrm{I}_{\mathrm{B}}=\beta\left(\frac{\mathrm{V}_{\mathrm{A}}-0.6 \mathrm{~V}-\mathrm{V}_{3}}{\mathrm{R}_{\mathrm{B}}}\right)=\beta\left(\frac{\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}}{\mathrm{KR}_{\mathrm{B}}}\right),
\]
where \(\beta\) is the transistor's short-circuit current gain. Assuming the transistor's emitter and collector currents are equal,
\[
\mathrm{R}_{\mathrm{AB}}=\frac{\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}}{\mathrm{I}_{\mathrm{C}}}=\frac{\mathrm{KR}_{\mathrm{B}}}{\beta} .
\]


Fig 1-Resistance ( \(\boldsymbol{R}_{\boldsymbol{A} B}\) ) between nodes \(\boldsymbol{A}\) and \(\boldsymbol{B}\) is a linear function of the wiper positions for potentiometers \(R_{B}\) and \(R_{K}\).

\section*{DESIGN IDEAS}

\section*{Linear load dissipates constant power}

\section*{Horace T Jones}

Penril Datacomm, Gaithersburg, MD
The Fig 1 circuit presents a constant-power load to the input voltage \(\mathrm{V}_{\text {IN }}\). For the component values shown, the circuit provides a 4 W load at a nominal \(\mathrm{V}_{\text {IN }}\) of 13 V and maintains this power level within \(\pm 0.2 \%\) for input voltages of 9 to 17 V . Potentiometer \(\mathrm{R}_{10}\) lets you adjust the constant-power level.

Select \(R_{4}\) and \(R_{5}\) to produce a voltage \(V_{\text {Control }}\) of approximately 6 V at the inverting input of \(\mathrm{IC}_{1 \mathrm{~A}}\). Op \(\mathrm{amps} \mathrm{IC}_{1 \mathrm{~A}}\) and \(\mathrm{IC}_{2}\) form a pulse-width modulator whose output duty cycle ( X ) is a linear function of \(\mathrm{V}_{\text {IN }}(\mathrm{X}\) equals \(\mathrm{V}_{\text {control }} / \mathrm{V}_{\mathrm{CC}}\). The modulator's nominal \(33-\mathrm{kHz}\) operating frequency ( \(f\) ) decreases as \(\mathrm{V}_{\text {IN }}\) varies above or below its nominal value:

The modulator's output signal (pin 6 of \(\mathrm{IC}_{2}\) ) drives the analog switch \(\mathrm{IC}_{3}\), which in turn drives the linear current source formed by op amp \(\mathrm{IC}_{1 \mathrm{~B}}\) and MOSFET \(\mathrm{Q}_{1}\). Consequently, \(\mathrm{I}_{\mathrm{IN}}\) is inversely proportional to \(\mathrm{V}_{\mathrm{IN}}\), so the load power \(\mathrm{P}_{\mathrm{IN}}\) is constant:
\[
\mathrm{P}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{IN}} \mathrm{I}_{\mathrm{IN}}=\frac{\mathrm{V}_{\mathrm{CC}} \mathrm{~V}_{\mathrm{REF}}}{\mathrm{R}_{\mathrm{II}}}\left[1+\frac{\mathrm{R}_{4}}{\mathrm{R}_{5}}\right] .
\]

Filter components \(\mathrm{R}_{8}\) and \(\mathrm{C}_{3}\) suppress voltage ripple introduced by the switching action of \(\mathrm{IC}_{3}\). This switching action also limits the control loop's \(3-\mathrm{dB}\) bandwidth to about 480 Hz .

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\[
\mathrm{f}=\frac{\mathrm{X} \cdot \mathrm{R}_{3}(1-\mathrm{X})}{\mathrm{R}_{2} \mathrm{R}_{1} \mathrm{C}_{1}}
\]


Fig 1—Developed for testing the rectifier/filter section of switching power supplies, this circuit provides a constant \(4 W\) load ( \(\pm 0.2 \%\) ) for \(V_{I N}\) in the 9 to 17V range.

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\section*{DESIGN IDEAS}

\title{
Open-loop servo adjusts shaft position
}

\author{
James C Smith \\ NASA, Greenbelt, MD
}

By using digital techniques to control a stepper motor, Fig 1a's circuit lets you manually adjust the position of a remote shaft. (Fig 1b shows one possible application for the system. Others include the remote positioning of flow valves and leveling devices.)

The ICs and the stepper motor require a 12 V supply, which also drives a regulator chip \(\left(\mathrm{IC}_{2}\right)\) that supplies 5 V to a digital potentiometer (not shown). This potentiometer generates 256 pulses for each revolution of its adjustment knob, producing the channel A and B quadrature square waves. These two signals enable flip-flop \(\mathrm{IC}_{3}\) to decode the potentiometer's direction of rotation.
\(\mathrm{IC}_{4}\) generates control signals for the stepper motor, and transistors \(Q_{1}-Q_{4}\) supply the necessary drive current to the motor's windings. ( \(\mathrm{IC}_{4}\) alone can supply 350 \(\mathrm{mA} /\) phase to the motor. If your motor requires more
current, use the IC's data sheet to select an \(\mathrm{R}_{3}\) value that provides base drive appropriate to the external transistors you're using.)

Components \(\mathrm{R}_{4}\) and \(\mathrm{C}_{1}\) filter the supply voltage, \(\mathrm{R}_{5}\) limits the motor current, and the 5 V zener diodes ( \(\mathrm{D}_{1}-\mathrm{D}_{4}\) ) reduce voltage transients by providing a flyback path for motor current when a transistor turns off. Because the motor steps at the pulse rate of channel B, the motor rotation is proportional to the rotation of the potentiometer knob-for knob rotation below about \(1 \mathrm{rev} / \mathrm{sec}\).

Unlike analog servo circuits, this system's positioning capability isn't affected by temperature or long-term component drift. Precision and resolution are limited primarily by the mechanical linkage between the motor and its load.

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Fig 1-This shaft positioner converts the output of a digital potentiometer to signals suitable for driving a stepper motor.

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\section*{Circuit ensures proper RS-232C mating}

\author{
Ralph L Adcock \\ SKP Electronics, Santa Ana, CA
}

Connectors for RS-232C data links aren't keyed, which increases the chance that you might plug the connector in backwards, reversing the transmit and receive pins and disabling a piece of equipment. The circuit of Fig 1 masks this mistake by automatically swapping the transmit and receive signals, allowing you to plug the connector in either way.

The RS-232C transmit and receive signals connect, via \(1-\mathrm{k} \Omega\) resistors, to optocouplers \(\mathrm{IC}_{2}\) and \(\mathrm{IC}_{3}\). These devices produce a logic zero at their outputs (pin 6) in response to an input of either polarity having a magnitude exceeding 3.8 V . (RS-232C signals are a minimum of \(\pm 5 \mathrm{~V}\).) In turn, the outputs cause the J-K flip-flop's output ( \(\mathrm{IC}_{4}\), pin 5) to set the position of analog switch \(\mathrm{IC}_{1}\) correctly.

At power-on, active drivers at each end of the RS232C link maintain a voltage on the transmit and receive lines-regardless of data transmission. If the plug is mated correctly, the voltages cause the optocouplers' outputs to each assert a logic zero. Consequently, the flip-flop's output remains unchanged.

If you insert the plug backwards, you remove voltage from the input of one of the optocouplers, producing a logic one output. The flip-flop will then toggle in response to the first negative clock transition, toggling the switch and restoring the proper RS-232C connections. (The clock input can be either a pushbutton contact closure or a signal of approximately 1 kHz .) The optocouplers' \(250-\mu \mathrm{sec}\) response time prevents the circuit from directing a change of connections during the zero-crossings of a data transmission.

The LED \(\left(D_{1}\right)\) indicates the direction of the plug connection-that is, whether the TXD signal is on pin 2 or 3 of the RS-232C interface. Also, when you apply a continuous clock signal, the LED indicates either that there isn't any RS-232C connection or that the power is off at the other end; the system will hunt under these conditions. A given clock frequency sets a lower limit on the baud rate, though. With a \(1-\mathrm{kHz}\) clock, for example, the system can handle 300 to 19.2 k baud, but it may exhibit false switching at 50 baud.

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Fig 1-If you inadvertently plug in an RS-232C connector backwards (crossing the transmit and receive signals), this circuit will detect the fault and automatically swap the signals back to the proper lines.

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\section*{Circuit increases frequency to \(\times \mathbf{1 0 0 0}\)}

\author{
Frank Michele \\ Brno, Czechoslovakia
}

Using 10 packages of 4000 Series CMOS logic ICs and a few passive components, you can translate low-frequency digital signals to a higher frequency. The circuit connections in Fig 1 yield a frequency translation of \(\mathrm{f}_{\text {oUT }}=600 \times \mathrm{f}_{\text {IN }}\), in which the 600 factor is set by three 4018 divide-by-n counters between nodes A and B. The free-running oscillator \(\mathrm{IC}_{1 \mathrm{C}}\) feeds this counter chain with a signal of approximately 250 kHz . Meanwhile, the 4040 binary counter is reset once per cycle of the input frequency \(\mathrm{f}_{\mathrm{IN}}\), allowing the 4040 to measure input period in terms of output counts from node \(B\). The input period (a 12 -bit binary word) is latched by the two 40174s and fed to the three 4029s, which constitute a synchronous modulo-n down counter. Each output from this counter generates one \(f_{\text {OUT }}\) pulse, which feeds back to reset the 4029 s for a countdown of the next period.

Consider \(\mathrm{f}_{\mathrm{IN}}=1.0 \mathrm{~Hz}\), for example. The 4018 s divide 250 kHz by 600 to produce 416.6667 Hz ; the 4040 counts to 416 before it resets. \(\mathrm{IC}_{1 \mathrm{~A}}\) and \(\mathrm{IC}_{1 \mathrm{~B}}\) each generate a positive pulse for each cycle of \(f_{\text {IN }}\) to strobe the 40174 storage registers and then reset the 4040 counter. The 4029s count down with a modulus of 416 , clocked at 250 \(\mathrm{kHz}: 4 \mu \mathrm{sec} \times 416=1664 \mu \mathrm{sec}\), thus yielding 600 Hz .

You can multiply \(f_{\text {IN }}\) by any available integer \(n\) in the range 8 to 1000 , where \(n\) is the division factor between nodes A and B. (You can alter the connections to each 4018 to provide a division between 2 and 10 ; odd integers require an additional inverter and NAND gate.) The minimum frequency for \(\mathrm{f}_{\mathrm{IN}}\) is about 0.06 Hz for the circuit shown, based on a multiplication of 1000 and the 4040 's capacity of 4096 counts. System power dissipation with a 9 V supply is 100 mW .

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Fig 1-You can multiply frequency \(f_{I N}\) from eight to 1000 times by using 10 CMOS ICs and several passive components.

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\section*{DESIGN IDEAS}

\section*{Divider produces symmetrical output}

\section*{Irwin Cohen}

Hewlett-Packard Co, Rockaway, NJ

In synchronous systems, you must often divide a symmetrical clock waveform by an odd integer and obtain a symmetrical output (ie, a \(50 \%\) duty cycle). Unfortunately, J-K flip-flop dividers usually produce a waveform (labeled A in Fig 1b) in which the high and low intervals differ by one period of the input clock. The Fig 1a circuit corrects this situation by splitting the difference-it lengthens the short interval and shortens
the long one by half a clock period. The circuit works for any odd-integer division producing an asymmetry of one clock period; division by 3 is used as an illustration.

Flip-flop \(\mathrm{IC}_{3}\) 's Q output is set high when waveform A is low; Q goes low with the first positive transition of \(f_{\text {IN }}\) after A returns high (Fig 1b). If \(\mathrm{T}_{1}>\mathrm{T}_{2}\), connect \(\mathrm{IC}_{1 B}\) 's Q output instead of the \(\overline{\mathbf{Q}}\) output to \(\mathrm{IC}_{3}\) 's preset input.

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\(\qquad\)


Fig 1-Odd-integer division performed using a divide-by-3 counter (a) produces the asymmetrical waveform (such as the one labeled \(A\) in part b), but you can add additional circuitry to generate a symmetrical fout signal.

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CIRCLE NO 12

\section*{Low-cost regulator converts \(\mathrm{V}^{+}\)to -5 V}

\author{
Ron Lashley \\ ITT Courier, Phoenix, AZ
}

The power-supply circuit in Fig 1a converts a 12 to 15 V positive supply voltage to -5.2 V and delivers 0 to 50 mA . The parts only cost about \(\$ 1.25\). You should add a preregulator circuit (Fig 1b) when using an unregulated 15 V supply.

During operation, the 555 timer's frequency varies with load current. The resulting positive output pulses (pin 3) have a varying repetition rate but a relatively constant duration. Each pulse turns on \(\mathrm{Q}_{1}\) and rapidly charges capacitor \(\mathrm{C}_{4}\). When pin 3 pulls the plus side of \(\mathrm{C}_{4}\) low, the charge on \(\mathrm{C}_{4}\) transfers to \(\mathrm{C}_{3}\).

Load and line regulation are about \(1.5 \% . \mathrm{C}_{1}\) charges through \(R_{1}\) while pin 3 is high; pin 3 goes low when pin 6 reaches its \(2 / 3 \mathrm{~V}_{\text {CC }}\) threshold. \(\mathrm{C}_{1}\) then discharges through \(R_{3}, Q_{2}\), and the IC's open-collector discharge transistor at pin 7. Transistor \(\mathrm{Q}_{2}\) stays off as long as the negative
output voltage is large enough to turn on the 6.0 V zener diode \(\mathrm{D}_{5}\). When load current discharges \(\mathrm{C}_{3}\) enough to turn off \(\mathrm{D}_{5}, \mathrm{Q}_{2}\) turns on and begins another discharge of \(\mathrm{C}_{1}\). The resulting output ripple is about 100 mV . Unless load current is light, ripple frequency is above the audio range; nonetheless, even if load current varies, ripple is inaudible because no chokes or transformers are present.

The output voltage increases slightly with temperature. You can generate a more negative output voltage by choosing a higher-voltage zener diode, but diode losses limit the output voltage maximum to about -8 V at low output current. \(\mathrm{IC}_{1}\) should not operate above 15 V , especially at full load. If necessary, you can limit the supply voltage by using the preregulator shown in Fig 1b.

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Fig 1-This converter produces -5.2 V from a 12 to 15 V supply (a). To ensure that the IC doesn't operate above 15 V , you might have to use a preregulator (b).

\section*{KEITHLEY ON SEMICONDUCTOR MEASUREMENTS:}

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\section*{DESIGN IDEAS}

\section*{Nonlinear load extends PLL frequency range}

\author{
Basel F Azzam and Christopher R Paul \\ Coherent Communications, Hauppauge, NY
}

A PLL chip such as the 74 HC 4046 in Fig 1 uses an external capacitor and resistor to set the frequency range for an internal voltage-controlled oscillator (VCO). By replacing the fixed resistor \(\mathrm{R}_{4}\) with a nonlinear one, you can extend the VCO's frequency range by a factor of 50 or more. For the component values shown, when pin 11 connects to \(R_{4}\), the range is 17 to 300 kHz ; in contrast, when the pin connects to the nonlinear load, the range is 2 kHz to 2 MHz .
Capacitor \(\mathrm{C}_{1}\) and the current through pin 11 control the PLL's output frequency. Higher current produces a higher frequency. When \(\mathrm{V}_{11}\) equals 0.5 V , for example,
the high- \(\beta\) transistor \(\mathrm{Q}_{1}\) is off and the resistance from pin 11 to ground is \(R_{2}+R_{3}\). As \(V_{11}\) increases, \(Q_{1}\) turns on and draws more current from pin 11. Thus, the effective impedance, Z , is
\[
\mathrm{Z}=\frac{\frac{\mathrm{R}_{2} \mathrm{R}_{3}}{\beta\left(\mathrm{R}_{2}+\mathrm{R}_{3}\right)}+\mathrm{R}_{\mathrm{c}}}{\frac{\mathrm{R}_{3}}{\mathrm{R}_{2}+\mathrm{R}_{3}}-\frac{\mathrm{V}_{\mathrm{BE}}}{\mathrm{~V}_{11}}},
\]
where \(\beta\) is the transistor's beta and \(\mathrm{V}_{\mathrm{BE}}\) equals 0.75 V . EDN
\(\qquad\)


Fig I-By connecting the nonlinear load Z to pin 11 of the PLL chip IC, you can extend the PLL's frequency range by a factor of 50 , as compared with that possible by using a fix.eed resistor ( \(R_{i}\) ).

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\title{
IF chip forms audio decibel-level detector
}

\section*{Robert J Zavrel}

Signetics Inc, Sunnyvale, CA
The NE604 is a low-power IF chip that includes a logarithmic signal-strength output. Fig 1's circuit draws less than 5 mA from a 6 V supply and offers a signal sensitivity of \(10.5 \mu \mathrm{~V}\). Although the chip is intended for cellular-radio and other RF applications, the log output provides an \(80-\mathrm{dB}\) range of response and \(\pm 1.5-\mathrm{dB}\) accuracy in the \(100-\mathrm{Hz}\) to \(10-\mathrm{kHz}\) audio range (Fig 2).

You capacitively couple the audio signal to pin 16 . The log circuit generates approximately \(10 \mu \mathrm{~A}\) per 20 dB of input signal at pin 5 ; you convert this current to voltage by connecting \(100 \mathrm{k} \Omega\left(\mathrm{R}_{2}\right)\) from pin 5 to ground. You can then measure this voltage directly with a voltmeter, or buffer and filter the voltage as shown using op amps \(\mathrm{IC}_{2 \mathrm{~A}}\) and \(\mathrm{IC}_{2 \mathrm{~B}}\). A standard 0 to 5 V meter with a linear decibel scale serves to display 80 dB of signal level. To measure higher audio levels, add a resistive attenuator at the chip's audio input.
\(R_{1}\) and \(C_{1}\) form a lowpass filter. Specifying \(2 \mathrm{k} \Omega\) for \(R_{1}\) provides maximum linearity; you should adjust \(C_{1}\) to change the filter's cutoff frequency. A higher value for \(\mathrm{C}_{1}\) lowers the circuit's output to about 0.6 V when no audio signal is present (Fig 2). Lowering \(\mathrm{C}_{1}\) increases the frequency response, but raises the circuit's output


Fig 2-The dotted line indicates the response of Fig 1's circuit for the \(10(1-\mathrm{Hz}\) to \(10-\mathrm{kHz}\) audio range; the solid line indicates an ideal response. Full scale ( 0 dB ) equals 300 mV p-p.
when no audio signal is present. The filter \(\mathrm{R}_{3} / \mathrm{C}_{3}\) provides a tradeoff between meter damping and ripple attenuation. If both a quick response and low ripple are required, you must substitute a more complex, active lowpass filter.

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Fig 1-You can measure audio signal levels to \(\pm 1.5-\) dB accuracy using a dual op amp and an FM IF chip (normally used for cellular-radio (and other RF applications). Sensitivity is \(10.5 \mu \mathrm{~V}\); power consumption is 30 mW .

\section*{KEITHLEY ON SOURCES: \\ QUALIFIED SOURCE}

If you think there isn't much difference between a source and a power supply, then it's time you tried a Keithley source The advantages are many

\section*{SENSITIVITY \& DYNAMIC RANGE}

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As further evidence, consider the \(\pm 50 \mu \mathrm{~V}\) resolution of Keithley's Model 230 voltage source, necessary to achieve fine levels of control when extracting families of curves on semiconductors Higher voltages to 100 V are available for biasing devices and supplying power to circuits
For calibration applications that need accurate low current, the Model 263 sources as low as \(\pm 50 \mathrm{aA}\). Lastly, for calibration of charge ranges on electrometers and nuclear instruments, the Model 263 will source to \(20 \mu \mathrm{C}\) Voltages to 20 V are also available.

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\hline Current & \(\pm 0.5 \mathrm{pA}\) to 100 mA & 220 \\
\hline Current & \(\pm 0.5 \mathrm{nA}\) to 100 mA & 224 \\
\hline Voltage \(/\) & \(\pm 1 \mathrm{mV}\) to 100 V & 228 \\
\hline Current & \(\pm 100 \mu \mathrm{~A}\) to 10 A & \\
\hline Calibrator & \(\pm 50 \mathrm{aA}\) to 20 mA & 263 \\
& \(\pm 0.5 \mathrm{fC}\) to \(20 \mu \mathrm{C}\) & \\
& \(\pm 5 \mu \mathrm{~V}\) to 20 V & \\
& \(1 \mathrm{k} \Omega\) to \(100 \mathrm{G} \Omega\) & \\
\hline
\end{tabular}

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Unlike most power supplies, Keithley sources can both source and sink when connected to the DUT \(\triangle\) This four-quadrant operation can either supply power to the device or act as a programmable load up to the full rated power of the output Sink capability can also help reduce transients and dissipate energy stored in circuit reactances

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\section*{DESIGN IDEAS}

\title{
Dual optoisolator detects zero crossings
}

\section*{Paul Galluzzi}

Dynamics Research Corp, Wilmington, MA
By using the dual optoisolator shown in Fig 1a, you can build a zero-crossing detector that gives you TTL-level output pulses. You can then use those pulses to drive a counter or as an input to a control system. Fig 1b compares the circuit's input waveform and its output pulses.

This circuit accepts the \(\pm 15 \mathrm{~V}\) output swing of an Inductosyn-driven PLL demodulator. Current through \(\mathrm{R}_{1}\) into \(\mathrm{IC}_{1}\) 's \(\mathrm{LED}_{1}\) (an HP HCPL-2630 dual optoisolator) holds \(\mathrm{V}_{2}\) at nearly 0 V when \(\mathrm{V}_{1}\) is at its maximum
positive value. Then, when \(V_{1}\) crosses through zero as it heads for its -15 V negative maximum, \(\mathrm{LED}_{1}\) ceases emitting; this shuts off \(Q_{1}\). Pullup resistor \(R_{2}\) causes \(V_{2}\) to reach 5 V . When \(\mathrm{V}_{1}\) goes negative, diode \(\mathrm{D}_{1}\) conducts to protect \(\mathrm{LED}_{1}\).

The zero-crossing detector's input circuit is symmetrical. Resistor \(\mathrm{R}_{3}\) provides a wired-OR output connection. The Schmitt triggers \(\left(\mathrm{IC}_{2 \mathrm{~A}}\right.\) and \(\left.\mathrm{IC}_{2 \mathrm{~B}}\right), \mathrm{C}_{1}\), and \(\mathrm{C}_{2}\) ensure that a clean pulse results from each zero crossing, resulting in an effective circuit for medium-speed inputs.

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Fig 1-A dual optoisolator and a handful of components make up a circuit that provides a glitch-free pulse for each detected zero crossing.

\section*{PCM filter offers lowpass, bandpass options}

\author{
David J Donovan \\ Harris Corp, Melbourne, FL
}

The Harris HC-5512, 5512A, and 5512C pulse-codemodulated (PCM) filters are switched-capacitor monolithic circuits that were originally designed for PCM codec filtering in systems that sample at 8 kHz . However , by taking advantage of the devices' receive lowpass filter and transmit highpass filter, you can build a wide range of lowpass and highpass filter circuits with cutoff frequencies ranging from 200 Hz to 10 kHz (Fig 1). To program the filter's cutoff frequency, you simply vary the device's input clock frequency.

The chip's transmit side uses a \(200-\) to \(3400-\mathrm{Hz}\) bandpass filter, which comprises a fifth-order elliptical lowpass filter cascaded with a fourth-order highpass filter. The receive side uses a \(3400-\mathrm{Hz}\) lowpass filter with \(\sin \mathrm{X} \div \mathrm{X}\) correction. Second-order antialiasing filters precede both the transmit and receive filters.

To specify a \(3-\mathrm{dB}\) cutoff frequency, you first connect the chip's CLK \({ }_{0}\) input (pin 14) to either \(\mathrm{V}_{\mathrm{CC}}\), GND, or \(\mathrm{V}_{\mathrm{BB}}\). (These connections result in internal clock frequencies of \(2.048,1.544\), and 1.536 MHz , respectively.) \(\mathrm{V}_{\mathrm{CC}}\) 's and \(\mathrm{V}_{\mathrm{BB}}\) 's nominal values are +5 V and -5 V , respectively. You then set the filter's cutoff frequency by applying an external clock signal to pin 12, whose frequency lies between 300 kHz and 5 MHz (the device won't work at frequencies out of this range).

For a \(300-\mathrm{kHz}\) external clock frequency, the lowpass filter will have a cutoff frequency of 1 kHz , and the bandpass filter will have a lowpass cutoff frequency of 200 Hz and a highpass cutoff frequency of 1 kHz . At 5 MHz , the lowpass filter will have a cutoff frequency of 10 kHz , and the bandpass filter will have a lowpass cutoff frequency of 5800 Hz and a highpass cutoff frequency of 9 kHz .

Between 300 kHz and 5 MHz , the cutoff frequencies for both the lowpass and bandpass filters will vary nearly linearly with the programming frequency, with some compression occurring at either end of the programming frequency's spectrum. The bandpass filter's bandwidth will remain constant at 3200 Hz , except for some compression at the lower end of the programming spectrum. The least amount of compression will occur for the highest internal clock frequencies (ie, when pin 14 is connected to \(\mathrm{V}_{\mathrm{CC}}\) ).

At 3300 Hz , the chip's receive filter provides an inherent gain. You can compensate for this gain by providing pole-zero compensation at the receive filter's


Fig 1-You can program this PCM filter to act as either a lowpass or highpass filter. By varying the chip's external clock frequency between 300 kHz and 5 MHz , you can select filter cut-off frequencies between 200 Hz and 10 kHz .
output. For the values shown in Fig 1, the chip uses a pole at 1 kHz and a zero at 3300 Hz to provide the compensation. The chip's transmit channel provides an inherent gain of 3 dB ; you can increase this gain to as much as 20 dB by adjusting resistors \(\mathrm{R}_{1}\) and \(\mathrm{R}_{2}\).

Both the transmit and receive outputs can drive a 3.2 V p-p signal into a \(10-\mathrm{k} \Omega, 25-\mathrm{pF}\) load. The poweramp outputs, \(\mathrm{PWR}_{0}{ }^{+}\)and \(\mathrm{PWR}_{0}{ }^{-}\), can drive a 5 V p-p signal into a \(300 \Omega\) single-ended load or a 10 V p-p signal into a \(600 \Omega\) balanced load.

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\section*{Current-folding circuit breaker resets itself}

\author{
Shalom Bukimer \\ Tadiran, Syosset, NY
}

The circuit breaker shown in Fig 1 limits load current to 0.5 A ; if the load is heavy enough (below \(1.6 \Omega\) ), it further limits load current to about 50 mA . You don't have to reset the breaker; it resets itself shortly after the overload is removed. A delay introduced by capacitor \(\mathrm{C}_{1}\) lets the circuit breaker ignore overloads of brief duration.
An increasing load current creates a voltage across \(R_{1}\), which lowers the voltage at comparator \(\mathrm{IC}_{1 A}\) 's noninverting input. When the \(\mathrm{R}_{1}\) drop equals approximately 0.2 V , the comparator output goes low, reducing the load current via \(Q_{3}\) and \(Q_{1}\). Comparator \(\mathrm{IC}_{1 \mathrm{~B}}\) monitors the falling load voltage \(\mathrm{V}_{\mathrm{L}}\) as \(\mathrm{Q}_{1}\) throttles the load current, and it turns off \(\mathrm{Q}_{1}\) when \(\mathrm{V}_{\mathrm{L}}\) reaches approximately 0.8 V .

Equally important, \(\mathrm{R}_{2}\) provides 50 mA through the load to enable comparator \(\mathrm{IC}_{1 \mathrm{~B}}\) to sense a rise in the load impedance. When \(R_{L}\) rises a few ohms above its threshold, \(\mathrm{IC}_{18}\) 's output goes low, which turns on \(\mathrm{Q}_{1}\) and reconnects the source and load.

The circuit has the following characteristics:
- The trip point is well defined and only slightly affected by changes in temperature.
- The circuit breaker makes a fast transition from off to on, and it consumes little power in either state.
- \(\mathrm{Q}_{2}\) and the LED may be omitted unless you require a visual overload indicator.
- You can manually open the circuit breaker by shorting \(\mathrm{IC}_{1 \mathrm{~A}}\) 's noninverting input to ground.

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Fig 1-Load currents higher than 0.5 A trip this electronic circuit breaker. The circuit resets automatically when the overload is removed.

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\section*{DESIGN IDEAS}

EDITED BY TARLTON FLEMING

\section*{Current monitor uses Hall sensor}

\section*{Paul Galluzzi}

Dynamics Research Corp, Wilmington, MA
The Fig 1 circuit uses a Hall-effect sensor, consisting of an IC that resides in a small gap in a flux-collector toroid, to measure dc current in the range of 0 to 40 A . You wrap the current-carrying wire through the toroid; the Hall voltage \(\mathrm{V}_{\mathrm{H}}\) is then linearly proportional to the current (I). The current drain from \(V_{B}\) is less than 30 mA .
To monitor an automobile alternator's output current, for example, connect the car's battery between
the circuit's \(V_{B}\) terminal and ground, and wrap one turn of wire through the toroid. (Or, you could wrap 10 turns-if they'd fit-to measure 1A full scale.) When \(\mathrm{I}=0 \mathrm{~V}\), the current sensor's (CS's) \(\mathrm{V}_{\mathrm{H}}\) output equals one-half of its 10 V bias voltage. Because regulators \(\mathrm{IC}_{1}\) and \(\mathrm{IC}_{2}\) provide a bipolar bias voltage, \(\mathrm{V}_{\mathrm{H}}\) and \(\mathrm{V}_{\text {OUT }}\) are zero when I is zero; you can then adjust the output gain and offset to scale \(V_{\text {out }}\) at 1 V per 10A.

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Fig 1-This circuit's Hall-effect current sensor CS, enables Vort to register de current at 1V per 10A.

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\section*{Build a l-chip, \(300-\mathrm{MHz}\) gateable oscillator}

\section*{Donald Trimble \\ Harris GASD, Melbourne, FL}

Using a Plessey SP16F60DC, you can build the 300MHz ring oscillator shown in Fig 1. The oscillator uses two ECL OR gates and an optional delay line. Without a delay line, the oscillator runs at approximately 300 MHz . Using a delay line reduces the oscillator's operating frequency. For example, a \(9-\mathrm{ft}\) RG-316/U delay line produces \(30-\mathrm{MHz}\) oscillations.
In Fig 1, the two ECL OR gates and a delay line are connected as a ring oscillator. Fig 2's timing diagram shows the relationship between the inputs and outputs of the OR gates. When the oscillator's enable input (pin


Fig 1-Two ECL gates and an optional delay line provide a \(300-\mathrm{MHz}\) gateable oscillator.

\(\mathrm{T}_{1}\) IS THE PROPAGATION DELAY OF IC \({ }_{1 \mathrm{~A}}\).
\(T_{2}\) IS THE PROPAGATION DELAY OF IC \(\mathbf{1 B}_{18}+J D\) (ASSUMED TO BE 0 ).
Fig 2-The Fig 1 oscillator always starts in the same state. Either \(I C_{1 B}\) 's internal propagation delay by itself or the delay and an optional feedback loop determine the frequency of oscillation.

7 of \(\mathrm{IC}_{1 \mathrm{~A}}\) ) goes low, the output at \(\mathrm{IC}_{1 \mathrm{~A}}\) 's pin 2 goes low and the output at its pin 3 goes high. The signal from pin 3 drives a feedback path comprising \(\mathrm{IC}_{1 \mathrm{~B}}\) and the optional delay line.
\(\mathrm{IC}_{1 \mathrm{~B}}\) guarantees oscillation by acting as a nonsaturating differential amplifier. One input of the amplifier is tied to an internally generated bias point called \(\mathrm{V}_{\mathrm{BB}}\). The amplifier's other input comes from external pin 13. The amplifier compares this input to \(\mathrm{V}_{\mathrm{BB}}\) (Ref 1).
For \(\mathrm{IC}_{18}\) 's noninverting (pin 15) output to change state, pin 13's input must go sufficiently above or below \(\mathrm{V}_{\mathrm{BB}}\) (Ref 2). This noninverting output drives either the optional delay line or \(\mathrm{IC}_{1 \mathrm{~A}}\), completing the ring. \(\mathrm{IC}_{1 \mathrm{~B}}\) 's 550 -psec typical propagation time allows \(\mathrm{IC}_{1 \mathrm{~A}}\) to settle to an ECL logical low level.

You can use any of \(\mathrm{IC}_{1 \mathrm{~A}}\) 's unused inputs ( pin 7 is used in Fig 1) to gate the oscillator. Note that although the oscillator always starts in the same state, it can stop in either a high or low state, unless you use a counting or control scheme to prevent it. Letting all unused inputs float low gives you a free-running oscillator. Fig 3 shows the oscillator's performance. You can obtain similar results using 100 K Series ECL parts.

\section*{References}
1. Digital IC Handbook, Plessey Semiconductor, pg 115.
2. F 100 K ECL User's Handbook, Fairchild Advanced Bipolar Div, pgs 2-5.


Fig 3-The Fig 1 free-running oscillator furnishes a clean output that varies between -0.5 and -2.25 V . Here, the vertical scale is 500 \(\mathrm{mV} / \mathrm{div}\); the horizontal scale, \(500 \mathrm{psec} / \mathrm{div}\). The horizontal line at the top of the display represents the zero level.


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\hline & \(2280^{\text {TM }}\) & 80186 & 68070 \\
\hline Package & \[
\begin{aligned}
& 68 \text {-pin } \\
& \text { PLCC/CMOS } \\
& \hline
\end{aligned}
\] & 68 -pin LCC/NMOS & \[
\begin{aligned}
& \text { 84-pin } \\
& \text { PLCC/CHMOS }
\end{aligned}
\] \\
\hline Typical Power & 375 mW & 2W & 800 mW (est) \\
\hline Speed & \(10-25 \mathrm{MHz}\) & \(8-12.5 \mathrm{MHz}\) & 10 MHz \\
\hline Memory Support & \[
\begin{aligned}
& \text { 16 MbPhysical } \\
& \text { Paged }
\end{aligned}
\] & 1 Mb Physical Segmented & 16 Mb Physical
8 or 128 Segments \\
\hline 16-bit Registers & 12 General & 8 General & 15 Dedicated \\
\hline Instruction Pre-fetch & 256-Byte Assoc. Cache; Burst Mode & 6-Byte Queue & None \\
\hline Multiprocessor
Support & Local or Global & Local only & Local only \\
\hline Wait Logic & Programmable & Programmable & Hardwire \\
\hline DMA & 4 Channels, 6.6 \(\mathrm{Mb} / \mathrm{s}\) @ 10 MHz & 2 Channels 2Mb/s@8MHz & 2 Channels, 3.2 \(\mathrm{Mb} / \mathrm{s}\) @ 10 MHz \\
\hline Counter/Timers & 316-bit & 316-bit & 216-bit \\
\hline Serial//0 & 1 Full-Duplex UART & None & 1Full-Duplex UART \\
\hline DRAM Controller & 10-bit Refresh & None & None \\
\hline Price (100) & \$33 & \$43 & \$50 \\
\hline
\end{tabular}

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\section*{DESIGN IDEAS}

\section*{Crystal oscillator sets pulse width}

\section*{T G Barnett \\ London Hospital Medical College, London, England}

You can use the circuit in Fig 1a to set precise pulse widths that are independent of component aging, pow-er-supply variations, and temperature. The monostable multivibrator \(\left(\mathrm{IC}_{1}\right)\) is configured in the nonretriggerable mode, and the RC time constant is set wider than the desired pulse width. In Fig 1b, the time constant is set at 15 msec ; this value produces an output pulse that's 10 msec wide. The pulse width equals RC. The Q output of \(\mathrm{IC}_{1}\) is connected to the reset pin of the programmable oscillator \(\left(\mathrm{IC}_{2}\right)\), and the output of the oscillator is inverted and connected to the reset pin of \(\mathrm{IC}_{1}\).

The programming pins of \(\mathrm{IC}_{2}\) are set to give a \(50-\mathrm{Hz}\) output frequency in the form of a square-wave output with equal high and low periods of \(10 \mathrm{msec} . \mathrm{IC}_{1}\) is triggered by the rising edge of an input pulse (A in Fig 1b). The Q output goes high, resetting the oscillator. The oscillator's output remains low for 10 msec ; when the oscillator's output goes high, the inverted output \((\bar{Q})\) resets \(\mathrm{IC}_{1}\), and the oscillator's reset pin goes low until the next trigger signal on \(\mathrm{IC}_{1}\). The PXO \(600\left(\mathrm{IC}_{2}\right)\) can be programmed for 57 different frequencies over 0.005 Hz to 1 MHz , so you can choose a wide range of pulse widths.

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Fig 1-This circuit (a) lets you set precise pulse widths (b) that are independent of component aging, power-supply variations, and temperature. The monostable multivibrator \(\left(I C_{1}\right)\) is configured in the nonretriggerable mode, and the \(R C\) time constant is greater than the desired pulse width.


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\section*{80m radio transmitter uses power MOSFETs}

\section*{Robert G Culter}

Tektronix Inc, Beaverton, OR
By using power MOSFETs as active elements, you can build the 80 m amateur-radio transmitter of Fig 1 with parts that cost less than \(\$ 20\). The circuit consists of a keyed crystal oscillator/driver and a high-efficiency final amplifier.
In the oscillator section, an inexpensive color-burst TV crystal determines output frequency. In addition, the \(700-\) to \(1200-\mathrm{pF}\) input capacitance \(\mathrm{C}_{\text {iss }}\) of MOSFET \(Q_{2}\) constitutes an essential part of the oscillator's feed-back-the oscillator won't operate without \(Q_{2} . Q_{1}\) retains enough gain for oscillation while driving amplifier \(\mathrm{Q}_{2}\) in a \(50 \%\)-duty-cycle (approximate) switching mode.
The output stage achieves \(84 \%\) efficiency rather than the \(50 \%\) you'd expect with a class-C amplifier. When \(Q_{2}\)
turns off, current through inductor \(\mathrm{L}_{3}\) causes the drain voltage to rise well above the 24 V supply (the 100 V zener diode \(\mathrm{D}_{1}\) limits this voltage excursion) and remain high for part of the conduction cycle as well. The high drain voltage allows the FET to deliver a given amount of power with less internal dissipation and hence with greater efficiency than if the drain voltage remained constant.
The output impedance-matching network is based on Q2's drain impedance \(R_{0}\), which is twice the dc value as a result of the \(50 \%\) duty cycle:
\[
\mathrm{R}_{0}=\mathrm{V}_{\mathrm{cc}}{ }^{2} / 2 \mathrm{P}_{0}=24^{2} /(2 \times 45)=6.4 \Omega
\]

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Fig 1-This 80m amateur-radio transmitter includes one MOSFET for the oscillator and a second MOSFET for the final amplifier.

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\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[t]{2}{*}{SERIES} & \multirow[t]{2}{*}{CHARACTER FORMAT} & \multirow[t]{2}{*}{MODEL NO.} & \multirow[t]{2}{*}{\[
\begin{aligned}
& \text { NO. OF } \\
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& \text { X CHAR. }
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\]} & \multirow[t]{2}{*}{CHAR HEIGHT (MM)} & \multirow[t]{2}{*}{} & \multirow[t]{2}{*}{\[
\begin{array}{|c}
\text { BI- } \\
\text { DIRECT } \\
\text { BUS }
\end{array}
\]} & \multirow[t]{2}{*}{CURSOR POSITION} & \multicolumn{2}{|l|}{USER PROG. FONTS} \\
\hline & & & & & & & & ROM & RAM \\
\hline \multirow{5}{*}{SCPB} & \multirow[t]{2}{*}{\[
\begin{gathered}
\text { 14-SEGMENT } \\
\text { + COMMA } \\
\text { +D.P. }
\end{gathered}
\]} & FU169SCPB-S1A & \(1 \times 16\) & 9.4 & S/P & \(\chi\) & X & & \\
\hline & & FU209SCPB-S1A & \(1 \times 20\) & 9.0 & S/P & X & X & & \\
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\] & \(2 \times 40\) & 5.0 & S/P & X & X & \[
\underset{(- \text { S31A })}{\text { X }}
\] & 4 \\
\hline
\end{tabular}

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\section*{DESIGN IDEAS}

\section*{Use simple circuit to program EPROMs}

\section*{Richard Poindexter}

Telex Computer Products, Tulsa, OK
You can program 21V, 28 -pin EPROMs (eg, models 2764 or 27128 ) using a circuit that occupies little space on a pe board (Fig 1). The signal interface consists of an 8 -bit control word and an 8 -bit data bus.
The switching regulator \(\left(\mathrm{IC}_{3}\right)\) provides the EPROM's programming voltage by stepping the 5 V supply to 21 V . The 12 -bit binary counter ( \(\mathrm{IC}_{1}\) ) simplifies the generation of EPROM addresses. After resetting the counter to 0 by applying a pulse to pin 11 , you can step
the EPROM through its addresses by applying pulses to the counter's clock input, pin 10. Refer to the EPROM data sheet for details on controlling the other inputs ( \(O_{1}, O_{2}\), and \(\mathrm{O}_{5}\) to \(\mathrm{O}_{7}\) ).

You can verify the contents of the EPROM against your original object code by the same process of strobing the counter for each address and reading the data bus for that address. Again, refer to the data sheet for details on reading data from the EPROM.

EDN


Fig 1-Program EPROMs using a binary counter for address generation and a step-up switching regulator to supply the 21V programming voltage.

\section*{DESIGN IDEAS}

\section*{Simpson's rule solves double integrals}

\author{
A Cameron \\ Defence Research Centre, Adelaide, SA, Australia
}

Although one generally uses Simpson's rule to approximate single integrals, you can extend the technique for use in solving double integrals. The C language routine of Listing 1 contains the required algorithm plus an example, demonstrating that you can apply Simpson's rule to certain complex double integrals that would normally require the application of numerical techniques. In addition, the technique applies equally well to integrals of a higher order.
You can obtain the algorithm by substituting the standard Simpson integration formula,
\[
\begin{aligned}
S_{x}\left(y_{j}\right)= & f\left(x_{0}, y_{j}\right)+f\left(x_{n}, y_{j}\right) \\
& +4 \sum_{i=1}^{\frac{n-2}{2}} f\left(x_{2 i-1}, y_{j}\right)+2 \sum_{i=1}^{\frac{n-2}{2}} f\left(x_{2 i}, y_{j}\right),
\end{aligned}
\]
for the double integral's inner integral. Reapplying Simpson's rule to the outer integral allows you to express the original integral equation in algebraic terms. After further simplification, it should be apparent that this technique consists of an integration along
one axis for each interval on the orthogonal axis, followed by an application of Simpson's rule on the accumulated results along the orthogonal axis.

If \(\mathrm{S}_{\mathrm{x}}\left(\mathrm{y}_{\mathrm{n}}\right)\) represents Simpson's rule applied along the \(x\)-axis as a function of the position \(y_{n}\) on the \(y\)-axis, then
\[
\begin{aligned}
\text { SIMPSON }= & \frac{h_{x} * h_{Y}}{9}\left(\mathrm{~S}_{\mathrm{x}}\left(\mathrm{y}_{0}\right)+\mathrm{S}_{\mathrm{x}}\left(\mathrm{y}_{\mathrm{n}}\right)+4 * \mathrm{~S}_{\mathrm{x}}\left(\mathrm{y}_{1}\right)\right. \\
& \left.+2 * \mathrm{~S}_{\mathrm{x}}\left(\mathrm{y}_{2}\right) \ldots \text { etc }\right)
\end{aligned}
\]
where \(h_{X}\) and \(h_{Y}\) represent the increments between steps:
\[
\mathrm{h}=(\text { upper limit-lower limit)/(number of steps). }
\]

The example in Listing 1 is a partial solution for the total radiated power through a hemispherical surface. Two quarter-wavelength monopole antennas, separated by a quarter wavelength and fed by signals that are out of phase by a quarter wavelength, are the source of the radiated power. (The result should be 3.829042.)

EDN

LISTING 1-C LANGUAGE ROUTINE
```

Simpson integration technique for
evaluating double integrals.
\#include "math.h"
float fxy[100][100],fy[100];
float pi;
main() {
float f();
float llx,lly,ulx,uly,x,y;
float hx,hy,ef,of,simpson;
int nosx,nosy,i,j;
/*---------------------------------------------
Simpson integration constants.
nos -> number of strips
ul -> upper limit of integration
ll -> lower limit of integration
h -> incremental value per strip

```


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}

\section*{DESIGN IDEAS}

\section*{Simple circuit tests twisted-pair cables}

\author{
Mark D Braunstein Contel Information Systems, Fairfax, Va
}

Using the system shown in Fig 1, you can quickly test a cable containing twisted-wire pairs and detect open or reversed pairs, shorted pairs, and shorts between unrelated pairs. The tester consists of an active test set that plugs into one end of the cable, and a passive terminator that plugs into the other end. (An RS-449 cable is used as an example.)

A battery or a dc supply delivers 15 to 24 V to the test set. The voltage regulator \(\left(\mathrm{IC}_{1}\right)\) is connected as a current regulator to supply a nominal 25 mA to the LED strings at each end of the cable. The cable in this example contains eight twisted pairs, and for a good cable, all eight LEDs in the test set \(\left(\mathrm{D}_{\mathrm{A}}\right.\) through \(\mathrm{D}_{\mathrm{H}}\), which are series-connected segments of a bar-graph display) and all eight LEDs in the terminator ( \(\mathrm{D}_{1}\)
through \(\mathrm{D}_{8}\) ) will light. If a twisted pair is open or reversed, the corresponding LED on the terminator will be extinguished; if a pair is shorted, corresponding LEDs at both ends will be extinguished; and if any two unrelated wires of different pairs are shorted, all intervening LEDs in the strings at both ends will be extinguished. For example, if pins 4 and 6 are shorted, LEDs \(\mathrm{D}_{\mathrm{A}}, \mathrm{D}_{\mathrm{B}}, \mathrm{D}_{1}\), and \(\mathrm{D}_{2}\) will not light.
You can add a heat sink to the \(\mathrm{IC}_{1}\) regulator as a safety precaution, but normal tester operation is well within the regulator's power-dissipation limits. Even with many shorted pairs, a dissipation of 700 mW would cause no more than a \(60^{\circ} \mathrm{C}\) junction temperature, and the IC is guaranteed to turn itself off at \(160^{\circ} \mathrm{C}\). The complete tester costs less than \(\$ 50\) to build.

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\(J_{1}\) AND \(J_{2}\) ARE GC ELECTRONIC 41-2411s OR CINCH DC-375S.

Fig 1-By driving two LED strings from a common current source, you can quickly check a cable of twisted-pair wires for short circuits, open circuits, and pair-to-pair shorts.


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\begin{tabular}{|l|c|c|}
\hline \multicolumn{3}{|c|}{ Reprogramming Failure Rate } \\
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\hline Flash Memory & \(<.01 \%\) & \(.1 \%\) \\
\hline EPROM & \(2 \%\) & - \\
\hline EEPROM & - & \(5 \%\) \\
\hline
\end{tabular} code changes almost instantaneously. Over phone lines,

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\section*{DESIGN IDEAS}

\section*{Isolation amp uses balanced modulators}

Moshe Gerstenhaber, Steve Miller, and Chuck Kitchen
Analog Devices Semiconductor, Wilmington, MA
The isolation amplifier of Fig 1 can faithfully transfer low-frequency information across a potential of 3 kV or more. Over the \(\pm 10 \mathrm{~V} \mathrm{~V}_{\text {IN }}\) range, \(\mathrm{V}_{\text {out }}\) deviates no more
than \(350 \mu \mathrm{~V}\)-that is, less than \(0.0018 \%\), which corresponds to a linearity of better than 14 bits. Offsetvoltage drift and gain drift are less than 2 ppm and 50 \(\mathrm{ppm} /{ }^{\circ} \mathrm{C}\), respectively. Typical bandwidth and output noise are 35 Hz and 1 mV rms. If necessary, you can improve these two specs by optimizing the output-filter characteristics.


Fig 1-This isolation amplifier employs PWM and optocoupling to transfer low-frequency signals across a barrier of 3 kV or more.

\section*{DESIGN IDEAS}

Each balanced modulator ( \(\mathrm{IC}_{2}\) and \(\mathrm{IC}_{7}\) ) is connected as a \(\pm\) unity-gain block (Fig 2). \(\mathrm{IC}_{2}\) and op amp \(\mathrm{IC}_{3}\) constitute a triangle-wave generator that produces a \(2.2-\mathrm{kHz}, 11 \mathrm{~V} \mathrm{pk}\) waveform at the inverting input of comparator \(\mathrm{IC}_{4}\). With \(\mathrm{V}_{\text {IN }}\) applied to its other input, the comparator produces a PWM output in which \(\mathrm{V}_{\text {IN }}\) is encoded as a duty cycle.

This modulated signal passes through the optocoupler and, together with 10 V from reference \(\mathrm{IC}_{6}\), directly drives \(\mathrm{IC}_{7}\), the second modulator. \(\mathrm{IC}_{7}\) 's \(\pm 10 \mathrm{~V}\) output (Fig 3) has a duty cycle that corresponds precisely to \(\mathrm{V}_{\text {IN }}\). Finally, the 3-pole, lowpass output filter reconstructs \(\mathrm{V}_{\text {IN }}\) by extracting the duty-cycle waveform's average value.

Because the isolation amplifier's linearity depends on duty-cycle ratios, variations of waveform frequency and amplitude have negligible effect on the circuit's accuracy. What's more, the precision voltage references ( \(\mathrm{IC}_{1}\) and \(\mathrm{IC}_{6}\) ) provide a temperature-stable output. The isolation amplifier's transfer function is


Fig 3-Looking at these waveforms, you can see how the Fig 1 circuit uses PWM to encode \(V_{\text {IN }}\) as a duty cycle.
resistors \(R_{2}\) and \(R_{3}\) set the triangle-wave amplitude ( \(\mathrm{IC}_{3}\), pin 6) to 11 V peak, which sets a minimum duty cycle for \(\pm 10 \mathrm{~V}\) inputs. To restore the proper output level, you set the filter gain to 1.1 by adjusting the \(20-\mathrm{k} \Omega\) potentiometer.

EDN
\[
\mathrm{V}_{\text {OUT }}=\frac{\mathrm{V}_{\mathrm{IN}}\left(\mathrm{~V}_{\mathrm{REFF}}\right)}{\mathrm{V}_{\mathrm{REF}}\left(\mathrm{R}_{氵} / \mathrm{R}_{2}\right)} \text { (1.1). }
\]

Excessively low duty cycles degrade linearity. So,


Fig 2-This balanced modulator/demodulator is configured as a \(\pm\) unity-gain device.

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\section*{Standard logic provides interrupts for Z80}

\section*{Don Johnson}

Network Sciences Corp, Newport Beach, CA
The circuit of Fig 1a implements vectored interrupts in a Z80-based system without requiring Z80-peripheral devices-a counter/timer circuit (CTC), for example, or a serial I/O (SIO) chip. First, configure the Z80 for mode 2 interrupt operation. In this mode, the \(\mu \mathrm{P}\)
responds to a low level on its \(\overline{\mathrm{INT}}\) input by using the contents of its 8 -bit I register to access a memory-based table of interrupt-handler addresses. (You must store this table on a modulo-256 address boundary.) During the interval when \(\overline{M 1}\) and \(\overline{\mathrm{O} R Q}\) are low (Fig 1b), the interrupting device supplies an 8 -bit vector that enables the \(\mu \mathrm{P}\) to select an address from the table.

As shown, Fig 1a supports eight interrupts. You can


Fig 1—This circuit handles as many as eight interrupts in a Z80-based system without using specialized peripheral devices (a). The timing diagram ( \(b\) ) shows that the interrupting device must identify itself by supplying an s-bit vector while \(\overline{M 1}\) and \(\overline{I O R Q}\) are low.

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\section*{DESIGN IDEAS}
increase this number to 16 by adding an LS00 gate and another 74LS148; the second 74LS148 uses the first 74 LS 148 's \(\mathrm{E}_{1}\) input (pin 5, connected to ground in Fig 1a) as a chip-enable input.

The interrupt signal from each interrupting device must go low and remain low until reset by the inter-rupt-handler subroutine. If the interrupting device generates a pulse instead of a level change, you must supply an edge-triggered flip-flop to capture and retain
the interrupt signal. Listing 1 contains assembly-code fragments for initializing the Z80's I register, for entering and exiting from the interrupt-handler subroutines, and for setting up the interrupt-handler address table.

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\section*{LISTING 1-CODE FRAGMENTS FOR INTERRUPT SYSTEM}


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\section*{DESIGN IDEAS}

\section*{Subroutine plots data from Basic programs}

\author{
Brown Porter Jr \\ Unisys Corp, Bristol, TN
}

Listing 1, a graphics subroutine for the IBM PC, is capable of plotting a graph of the data generated by your Basic program, using any or all quadrants of the

Cartesian plane. It gives you scale options of linear, semilog ( X or Y axis), or \(\log\)-log (one to five cycles); it provides automatic ranging and scaling; and it lets you set the Y scale, typically from \(1 / 4\) to a full page ( \(2^{1 / 2}\) to 10 in.). It also prints the title, subtitle, and scale labels in text mode, using upper- and lower-case characters.

\section*{LISTING 1-GRAPHICS SUBROUTINE}
```

1000, *~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~
1010* **** PLOT1 S/R ****
1020.
1030 HT%=15 Set height of graph
1040 CLS: KEY OFF
1050 DIM A\$ (62), B$(62)
1 0 6 0 \text { GOSUB 2120' *Define plot char S/R}
1070 IF (XSCALE$="G") OR (XSCALE$="g") THEN 1080 ELSE 1120
1080 GOSUB 1790 'Scale X S/R
1090 IF (YSCALE$="G") OR (YSCALE$="g") THEN 1100 ELSE 1120
1100 GOSUB 1860 'Scale Y S/R
1110, --- Find Min and Max ---
1120 FOR N=1 TO AF
1130 IF N }<>1\mathrm{ THEN 1160
1140 XMAX=X(1): XMIN=X(1): YMAX=Y(1): YMIN=Y(1) 'Intl regs
1150 GOTO 1240
1160 IF XMAX > =X (N) THEN 1180 FFind Xmax
1170 XMAX=X(N)
1180 IF XMIN<X(N) THEN 1200 'Find Xmin
1190 XMIN=X (N)
1200 IF YMAX >=Y(N) THEN 1220 'Find Ymax
1210 YMAX=Y(N)
1220 IF YMIN<Y(N) THEN 1240 FFind Ymin
1230 YMIN=Y(N)
1240 NEXT N
1250 FOR I=1 TO N
                Y(I)=Y(I)-YMIN "Translate all X & Y values
                X(I)=X(I)-XMIN 'to avoid negative numbers
    NEXT I
1290 XDIV=60/(XMAX-XMIN): YDIV=HT%/(YMAX-YMIN) ,Calculate scale factors
1300 XAXIS%=HT%-YMAX*YDIV: YAXIS%=INT (.5-XMIN*XDIV), Locate akes
1310 PRINT YUNIT$
1 3 2 0 GOSUB 2050 "Print top border
1330, --- Build Image Line ---
1340 FOR I=1 TO 60
1350 IF I<>YAXIS% THEN 1380 *Y axis ?
1360 A$(I)="+" 'Yes
1370 GOTO 1390
1380 LET A$(I)=" " 'no
1390 NEXT I
1400 IF HT%<>XAXIS% THEN 1420
1410 FOR I=1 TO 60: A$(I)="-": NEXT I 'Yes
1420 FOR I=1 TO N
1430 IF Y(I)<=(HT%-.5)/YDIV THEN 1470 value Y(I) = line value?
1440 K%=X(I)*XDIV 
1450 A$ (K%)=CURVE1 \$
1460 Y(I)=-Y(I)
1470 NEXT I
1480 A$(61)=YBORDER$ 'Right boundry
1490 V=YMIN+HT%/YDIV
1500 IF (YSCALE$="G") OR (YSCALE$="g") THEN 1510 ELSE 1520
1510 GOSUB 1930
1520 IF HT%/5<>INT (HT%/5) THEN 1560 Frint every fifth line
1530 GOSUB 1970
1540 PRINT TAB(10); YBORDER$;"-"
1550 GOTO 1570
1560 FRINT TAB(10); YBORDER$;" "; "Frint non label lines
1570 FOR I=1 TO 61: PRINT A\$(I);: NEXT I Print image line
1580 PRINT
1590 HT%=HT%-1 Next line
1600 IF HT%>=0 THEN 1340 'Finished plotting?

```


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\section*{LISTING 1-GRAPHICS SUBROUTINE (Continued)}
```

1610 FOR I=1 TO 7
PRINT TAB(I*10+2); YBORDER$; Print }X\mathrm{ axis divisions
1620 PRINT
1640 GOSUB 2050
1650 FOR I=0 TO 6
        V=XMIN+I*10/XDIV
        IF (XSCALE }$="G") OR (XSCALE$="g") THEN 1680 ELSE 1690
        GOSUB 1930
        FRINT TAB(I*10+9); "Frint x labels
        GOSUB 1970
    NEXT I
1720 PRINT: PRINT: PRINT TAB(27) XUNIT$
1730 PRINT: PRINT: PRINT TAB(27) TITLE$: PRINT
1740 FRINT TAB(27) SUBTITLE$
1750 RETURN
1760 END
1770 ,
1780 --- LOG SCALE x S/R ---
1790 FOR I=1 TO AP: X=X (I)
1800 ON SGN (X)+2 GOTO 2170, 2190,1810
1810 X=LOG(10*X)/LOG(10): IF X<0 THEN 2210
1820 X(I)=X: NEXT I
1830 RETURN
1840,
1850 * --- Log Scale Y S/R ---
1860 FOR I=1 TO AP: Y=Y(I)
1870 ON SGN(Y)+2 GOTO 2170, 2190,1880
1880 Y=LOG(10*Y)/LOG(10): IF Y<0 THEN 2210
1890 Y(I)=Y: NEXT I
1900 RETURN
1910.
1920 , --- Log Scale Label S/R ---
1930 V=10^V/10
1940 RETURN
1950.
1960, --- Axes Labeling 5/R ---
1970 IF V<>0 THEN 1980: FRINT V; GOTO }202
1980 IF V>99 THEN 2010
1990 IF V<0!-99 THEN 2010
2000 PRINT USING "\#\#\#.\#\#";V;: GOTO 2020

```

```

2020 RETURN
2030 *
2040 = --- Border S/R ---
2050 FOR I=1 TO 61: B$(I)=XBORDER& : NEXT I Fill top/bot border
2060 PRINT TAB(11);
2070 FOR I=1 TO 61: FRINT B$(I);: NEXT I Print top/bot border
2080 PRINT
2090 RETURN
2100.
2100, --- Flot Char. Table ---
2120 XBORDER =CHR*(196) IBM SM Star Epson
2130 YBORDER\&=CHR(179) (179)
2140 CURVE1\$ =CHR\$ (254) ,
2150 RETURN
2160.
2170 PRINT"Negative numbers are illegal on log scales."
2180 GOTO 995
2190 FRINT"Zero is illegal on log scales."
2 2 0 0 GOTO 995
2210 FRINT"This program will plot log scales down to 0.1 with printing resolutio
n. Units smaller then 10^-1 should be converted to milli, micro etc. to preser
ve plotting accuracy."
2220 GOTO 995

```

The subroutine has minimal impact on your main program, and it precludes the necessity of writing a special plotting routine, or of trying to visualize a curve by looking at a list of output data. Listing 2 is an example of how the subroutine operates in a short program. The program includes a For/Next loop that generates discrete points on a continuous, 2-dimensional curve called a strophoid.
You establish an X-Y array (lines 30 and 900 to 915 ), select linear or log scales (lines 920 and 930), and then
provide the title and scale labels (lines 940 to 970 ). To center the titles, add or subtract spaces between the quotation mark and the first character. The graph's vertical dimension depends on line 1030 (HT\%=nn) in Listing 1. The variable "nn" lets you size charts for inclusion in technical reports; 15 is a good size for screen viewing, and 30 is a good starting value for printing.
As you can see, the subroutine displays the graph on a monitor. Using the echo print method, you can make a

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LAMB ELECTRIC DIVISION

\section*{DESIGN IDEAS}

hard copy by pressing Ctrl and then pressing PrtSc (Fn, Echo on the IBM Jr) and then running the program. Don't forget to clear the echo mode by repeating this key sequence after the printing is finished. If you don't want or need to preview the graph on the screen, change all PRINT statements to LPRINT ones, which route the output directly to the printer.

To use different printers or change the appearance of your graph, look up the special characters for plotting the curve and borders, which you can find at line 2110 in Listing 1. If you're using a printer that doesn't have an IBM mode, and therefore inserts unwanted spaces in
the vertical borders, try adding lines 1052, 1054, and 1745:

1052 PRINT CHR\$(27);" (a"
1054 PRINT CHR\$(27);"A";CHR\$(6)
1745 PRINT CHR\$(27);"(a"
'Clear printer presets
'Print W/O spaces
'Return printer to text mode

If this doesn't work, you'll have to consult your printer manual for a command that allows printing without spaces.

EDN


\section*{DESIGN IDEAS}

\title{
Pocket calculator eases filter design
}

\author{
Arthur Delagrange and Reynold Douyon \\ Naval Surface Weapons Center, Dahlgran, VA
}

By using this filter-design method and Fig 1's circuit, you can pick the poles and zeros for a bandpass filter and then calculate the corresponding component values on a pocket calculator.

A previous article (Ref 1) described a circuit for a floating synthetic inductor that you can use to convert any prototype lowpass filter to a bandpass or bandstop filter. David C Bidwell (Ref 2) revised the bandpassfilter design using a more efficient circuit. His conversion, however, entailed "complex polynomial-manipulation procedures requiring extensive computer manipulation."

The method presented here achieves an efficient design. It involves tedious calculations, but you can perform them on a pocket calculator having square root and trigonometric functions.
The method consists of calculating new poles and zeros for the bandpass filter and determining component values directly from the poles and zeros. This example uses a 3-pole, 2-zero prototype filter that specs


Fig 1-Use this building-block circuit to generate the pole-zero pairs used in bandpass-filter design.
\(1.25-\mathrm{dB}\) ripple and \(39-\mathrm{dB}\) stopband (Ref 3 ). The prototype filter has zeros at
\[
\mathrm{S}= \pm \mathrm{j} 2.6003
\]
and poles at
\[
S=-0.48307
\]
and
\[
S=-0.20690 \pm j 0.96264 .
\]

The bandpass transformation makes the substitution
\[
S \rightarrow \frac{S^{2}+\omega_{0}^{2}}{S}
\]

Substituting for the pole-zero pair at
\[
S=-A+j B
\]
and solving by means of the standard quadratic equation and Euler's formula gives the new poles or zeros:
\[
S=\frac{-A \pm j B}{2} \pm(\cos \theta \mp j \sin \theta) \sqrt[4]{\left(\frac{A^{2}-B^{2}-4 \omega_{0}^{2}}{4}\right)^{2}+\left(\frac{A B}{2}\right)^{2}},
\]
where
\[
\theta=\operatorname{ARCTAN} \frac{\mathrm{AB} / 2}{\left(\mathrm{~A}^{2}-\mathrm{B}^{2}-4 \omega_{0}^{2}\right) / 4+\sqrt{\left(\frac{A^{2}-B^{2}-4 \omega_{0}^{2}}{4}\right)^{2}+\left(\frac{A B}{2}\right)^{2}}}
\]

In this example, \(\omega=1\).


Fig 2-You can achieve a 39-dB stopband and 1.25-dB ripple using this bandpass filter. Scale the capacitor values to change its center frequency.


Everybody promises, but nobody delivers a realtime, emulator-based C-debug environment like Arium's ECHO. 16-bit, true multitasking and UNIX \({ }^{\text {® }}\)-based, ECHO gives you more power, speed and menu-driven features to handle the 68000 and other \(\mu\) Ps better than the HP 64000 , or anything else.

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Stack-Relative Trigger lets you trigger on the addresses and values of stack-relative variables-a "must" for effective C-debug where the address of an automatic variable is different each time the function is called and is determined at execution. Here, a read of the local variable "nrecur" is included in the trigger sequence.

TimeStamp \({ }^{\text {TM }}\) and variable display are two further features that are a must for real-time C-debug. Note the display of two instances of a structure in array "starray." The contents of these structures, as for any C variable, can be changed right on the screen.


\section*{DESIGN IDEAS}

This equation gives zeros at
\[
S= \pm j 0.3398
\]
and
\[
\mathrm{S}= \pm \mathrm{j} 2.9406 .
\]

It has a real pole at
\[
S=-0.2415 \pm \mathrm{j} 0.9704 .
\]

Finally, the equation has complex poles at
\[
S=-0.05843 \pm j 0.6246
\]
and
\[
\mathrm{S}=-0.1485 \pm \mathrm{j} 1.5872 .
\]

To build this filter, you need three sections: bandpass, highpass, and lowpass. The bandpass section can be an ordinary Sallen-Key resonator (Ref 4), but the highpass and lowpass sections require zeros. Fig 1's
circuit can provide these zeros. Its transfer function (omitting the gain factor) is
\[
\begin{gathered}
\frac{\mathrm{S}^{2}+\frac{\left[(\mathrm{G} \beta+\beta-\mathrm{G}) \mathrm{R}_{1} \mathrm{C}_{1}+(\mathrm{G} \alpha+\alpha-\mathrm{G}) \mathrm{R}_{2} \mathrm{C}_{2}-\mathrm{GR}_{1} \mathrm{C}_{2}\right] \mathrm{S}}{(\mathrm{G} \beta+\beta-\mathrm{G}) \mathrm{R}_{1} \mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2}}}{\mathrm{~S}^{2}+\frac{\left(\mathrm{R}_{1} \mathrm{C}_{1}+\mathrm{R}_{2} \mathrm{C}_{2}-\mathrm{GR}_{1} \mathrm{C}_{2}\right) \mathrm{S}}{\mathrm{R}_{1} \mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2}}+\frac{1}{\mathrm{R}_{1} \mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2}}} \\
+\frac{+\frac{(\mathrm{G} \alpha+\alpha-\mathrm{G})}{(\mathrm{G} \beta+\beta-\mathrm{G}) \mathrm{R}_{1} \mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2} .}}{\frac{1}{\mathrm{~S}^{2}+\frac{\left(\mathrm{R}_{1} \mathrm{C}_{1}+\mathrm{R}_{2} \mathrm{C}_{2}-G \mathrm{R}_{1} \mathrm{C}_{2}\right) \mathrm{S}}{\mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2}}}+\frac{1}{\mathrm{R}_{1} \mathrm{R}_{2} \mathrm{C}_{1} \mathrm{C}_{2}}} .
\end{gathered}
\]

The lowest frequency zero pair and the lowest frequency pole pair combine to give the highpass-filter section. Similarly, the highest frequency pole and zero pairs give the lowpass-filter section. For each section, you can divide the product of the poles by the product of the zeros and write the transfer function in the form
\[
\frac{S^{2}+D}{S^{2}+A S+B}
\]

Identifying these transfer-function terms with the


Fig 3-The bandpass filter's performance changes only slightly when you change its center frequency from 1.6 kHz (solid curve) to 16 kHz (broken curve) by scaling capacitor values.

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\hline \begin{tabular}{l} 
Multiplexer channels
\end{tabular} & - & 4 & 4 \\
"On-the-fly" operation & NO & YES & YES \\
\begin{tabular}{l} 
Data Transfer Rate (Mbytes/Second) \\
Automatic chaining of command
\end{tabular} & 8 & 10 & 20 \\
\begin{tabular}{l} 
and data blocks
\end{tabular} & NO & YES & YES \\
Package & PLCC & PLCC & PLCC \\
& & LCC & LCC \\
& & PGA & PGA \\
\hline
\end{tabular}
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\section*{DESIGN IDEAS}

\section*{COMPONENT VALUES FOR SECTIONS WITH ZEROS}
\begin{tabular}{|c|c|}
\hline HIGHPASS & LOWPASS \\
\hline \[
\begin{aligned}
\text { LET } \beta & =1 ; C_{1}=C_{2}=1 \\
R_{1} & =\frac{\left(1-\frac{D}{B}\right)}{A} \\
R_{2} & =\frac{1}{B R_{1}} \\
G & =\frac{\left(R_{1}+R_{2}-\frac{A}{B}\right)}{R_{1}} \\
\alpha & =\frac{G+\frac{D}{B}}{G+1} \\
R_{\text {SERIES }} & =\frac{R_{1}}{\alpha} \\
R_{\text {SHUNT }} & =\frac{R_{1}}{(1-\alpha)}
\end{aligned}
\] & \[
\begin{aligned}
\text { LET } \alpha & =1 ; R_{1}=R_{2}=1 \\
C_{2} & =\frac{\left(1-\frac{B}{D}\right)}{A} \\
C_{1} & =\frac{1}{B C_{2}} \\
G & =\frac{\left(C_{1}+C_{2}-\frac{A}{B}\right)}{C_{2}} \\
\beta & =\frac{G+\frac{B}{D}}{G+1} \\
C_{\text {SERIES }} & =\quad \beta C_{1} \\
C_{\text {SHUNT }} & =(1-\beta) C_{1}
\end{aligned}
\] \\
\hline
\end{tabular}
terms in the complete transfer function and solving the equation yields the parameters listed in the table. Included at the bottom are the values for splitting the input resistor or capacitor to give a Thevenin-equivalent voltage-divider ratio \(\alpha\) or \(\beta\).
Fig 2 shows the complete filter circuit. You need just seven precision resistors and three op amps. The filter's output comes from an op amp and doesn't need buffering. The values calculated for this prototype have been scaled up by a factor of \(10^{4}\) in frequency ( 1.6 kHz ) and
\(10^{4}\) in impedance ( \(10 \mathrm{k} \Omega\) ). Components were selected to \(1 \%\) tolerance, and the first stage's gain and Q were then adjusted to equalize the amplitude in the peaks of the filter's response (Fig 3). Note that the vertical scale is expanded in the passband. The solid line shows the response of Fig 2's circuit with a \(1.6-\mathrm{kHz}\) center frequency. The actual performance is quite close to that predicted by theory. The broken line gives the response for the same circuit with capacitor values reduced by a factor of 10 to give a new center frequency of 16 kHz .

EDN

\section*{References}
1. Delagrange, Arthur, "Make passive filters active with a floating synthetic inductor," EDN, June 23, 1983, pg 277.
2. Bidwell, David C, "GICs yield efficient bandpass filter," \(E D N\), January 12, 1984, pg 289.
3. Zverev, A I, Handbook of Filter Synthesis, John Wiley and Sons, New York, 1967.
4. Huelsman, L P, and Allen, P E, Introduction to the Theory and Design of Active Filters, McGraw-Hill, New York, 1980.

\section*{Components enhance amplifier's bandwidth}

Eric Filseth
National Semiconductor Corp, Santa Clara, CA
Two modifications to the standard hookup for an

LH0002 current amplifier will boost its \(30-\mathrm{MHz}\) frequency response from that shown in Fig 1a (standard performance using \(\pm 5 \mathrm{~V}\) power supplies) to that of Fig 1b.


Fig 1-These input (top) and output (bottom) waveforms show the frequency-response improvement that results when you modify the LH0002 current amplifier (whose response is shown in a) by adding the external components in Fig 2. The improvement illustrated in \(\boldsymbol{b}\) results when \(R_{S}=1 \mathrm{k} \Omega, R_{L}=50 \Omega\), and frequency equals 30 MHz .

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\section*{DESIGN IDEAS}

First, you increase the input-stage bias currents tenfold by adding \(500 \Omega\) resistors \(\mathrm{R}_{\mathrm{A}}\) and \(\mathrm{R}_{\mathrm{B}}\) to pins 10 and 6 (Fig 2). The higher bias current increases the input transistors' \(\mathrm{f}_{\mathrm{t}}\) and provides more base drive for the output transistors. Note that the output devices' beta is probably less than 10 at 30 MHz , so each will need at least a \(3-\mathrm{mA}\) base current to provide the required \(30-\mathrm{mA}\) output current. Also, higher base current provides more drive to the input capacitance of these transistors; the increased drive increases the amplifier's slew rate.
Adding \(\mathrm{R}_{\mathrm{A}}\) and \(\mathrm{R}_{\mathrm{B}}\) increases the quiescent supply current, of course, and lowers the amplifier's input impedance to about \(15 \mathrm{k} \Omega\). The variation of input impedance over the operating bandwidth is less, however, than for an unmodified LH0002. At high frequencies, the modified circuit's input impedance (particularly the reactive part) is actually greater. The main
advantage, though, is the speed and bandwidth obtained for \(\pm 5 \mathrm{~V}\) applications. (Incidentally, you must use the 10 -pin DIP version of the LH0002; pins 6 and 10 offer connections that are not present on the 8 -lead TO-5 version.)
The second modification is useful in applications with high source resistance, where speed is limited by the product of source resistance and LH0002 input capacitance. The input capacitance is mostly collector-base capacitance from \(Q_{3}\) and \(Q_{4}\); you can bootstrap (minimize the effects of) this capacitance by adding resistors \(\mathrm{R}_{\mathrm{X}}\) and \(\mathrm{R}_{\mathrm{Y}}\), and then adding capacitors \(\mathrm{C}_{1}\) and \(\mathrm{C}_{2}\) to bypass the output stage (Fig 2). The amplifier's output voltage now drives its input capacitance via the capacitors.

EDN


Fig 2-You can add components to extend the bandwidth of an LH0002 amplifier. This circuit increases the input-stage bias currents and bootstraps the amplifier's input capacitance.

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\hline CXB11020 & Quad 2-in EXOR/NOR & 490 ps & 1.5 GHz & 680 mW & 24 FLAT \\
\hline CXB11030 & Quint Line Receiver & 410 ps & 1.5 GHz & 650 mW & 24 FLAT \\
\hline CXB11040 & Dual D Flip Flop & 620 ps & 3.2 GHz & 520 mW & 24 FLAT \\
\hline CXB11050 & Triple Fan-out Buffer & 590 ps & 1.5 GHz & 720 mW & 24 FLAT \\
\hline CXB11060 & 4-Stage Ripple Counter & & 3.4 GHz & 720 mW & 24 FLAT \\
\hline CXB11070 & Decision Circuit & & 3.2 GHz & 430 mW & 24 FLAT \\
\hline CXB11080 & Laser Driver & & 2.0 GHz & 740 mW & 16 FLAT \\
\hline CXB11090 & Quad D-FF with Master Reset & 620 ps & 3.4 GHz & 790 mW & 24 FLAT \\
\hline CXB11100 & 16 to 1 Multiplexer & 610 ps & 1.5 GHz & 680 mW & 24 FLAT \\
\hline CXB11110 & Look Ahead Carry Block & 580 ps & 1.5 GHz & 610 mW & 24 FLAT \\
\hline CXB11120 & Phase Frequency Detector & 720 ps & 0.8 GHz & 500 mW & 24 FLAT \\
\hline CXB11130 & 4 to 1 Multiplexer & & 2.0 GHz & 950 mW & 24 FLAT \\
\hline CXB11140 & 1 to 4 Demultiplexer & & 2.5 GHz & 1100 mW & 24 FLAT \\
\hline CXB11300 & \begin{tabular}{l}
9,8, 4-bit \\
Multiplexer
\end{tabular} & & 1.6 GHz & 730 mW & 32 FLAT \\
\hline CXB11310 & \[
\begin{aligned}
& \text { 9, 8, 4-bit } \\
& \text { Demultiplexer }
\end{aligned}
\] & & 1.6 GHz & 1000 mW & 32 FLAT \\
\hline CXB11320 & 9, 8, 4-bit Universal Shift Register & & 1.3 GHz & 910 mW & 32 FLAT \\
\hline CXB11330 & 22, 15, 7-Stage Scrambler & & 1.6 GHz & 600 mW & 24 FLAT \\
\hline CXB11340 & 22, 15, 7-Stage Descrambler & & 1.6 GHz & 610 mW & 24 FLAT \\
\hline CXB11350 & 8-16 bit Comparator & & 1.3 GHz & 630 mW & 32 FLAT \\
\hline CXB11360 & 8 -bit Universal Counter & & 1.2 GHz & 730 mW & 32 FLAT \\
\hline CXB11370 & 8 -bit Shift Matrix & 1250 ps & & 700 mW & 24 FLAT \\
\hline CXB11380 & 4-bit Arithmetic Logic Unit & 1460 ps & & 680 mW & 24 FLAT \\
\hline
\end{tabular}
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\section*{DESIGN IDEAS}

\section*{Audio AGC circuit has \(40-\mathrm{dB}\) dynamic range}

\section*{Norman M Hill}

Zetron Inc, Bellevue, WA
The automatic-gain-control (AGC) circuit of Fig 1a operates on \(\pm 5 \mathrm{~V}\) supplies and provides good fidelity for audio signals over an input range of 40 dB . The circuit maintains an output of \(0 \mathrm{dBm}( \pm 1.1 \mathrm{~V})\) by varying the input attenuation from 0.1 to 10 ( -20 to 20 dBm ).

JFET \(Q_{1}\) operates as a variable resistor. \(Q_{1}\) and resistor \(R_{1}\) divide the input signal in response to the control voltage on the gate of \(Q_{1}\). Op amp \(\mathrm{IC}_{1 \mathrm{~A}}\) (with \(\mathrm{R}_{4}\) and \(\mathrm{R}_{5}\) ) then provides enough gain to accommodate input signals of \(\pm 20 \mathrm{dBm}\) (Fig 1b).

To ensure low distortion in a JFET, you should add approximately one-half the drain-source voltage to the gate voltage. The voltage divider \(\left(\mathrm{R}_{2} / \mathrm{R}_{3}\right)\) and the buffer
amplifier ( \(\mathrm{IC}_{1 \mathrm{~B}}\) ) provide this gate bias by charging and discharging capacitor \(\mathrm{C}_{2} . \mathrm{R}_{8}\) isolates the gate-source diode in \(Q_{1}\).
The peak-detector amplifiers ( \(\mathrm{IC}_{\mathrm{IC}}\) and \(\mathrm{IC}_{1 \mathrm{D}}\) ) have a gain of 20 . Their outputs reside at negative-saturation levels when \(\mathrm{V}_{\text {out }}\) is between 1.1 and -1.1 V ; the resulting \(Q_{1}\) gate voltage ( -4 V ) allows maximum circuit gain. When \(V_{\text {out }}\) makes an excursion outside this range, one of the peak-detector outputs rises in the positive direction, discharging \(\mathrm{C}_{2}\) and driving \(\mathrm{Q}_{1}\) 's gate in a more positive direction. Thus, \(Q_{1}\) regulates the output by increasing the input-signal attenuation. \(\mathrm{C}_{2}\) discharges through the attack resistor, \(\mathrm{R}_{6}\), until the output level returns to 0 dBm . When the input level subsides, \(\mathrm{C}_{2}\) slowly recharges through the decay resistor, \(\mathrm{R}_{7}\). For a \(20-\mathrm{dBm}\) change, the attack time is 5 msec and the


Fig 1-Using a JFET to control input attenuation, this AGC circuit (a) provides a dynamic range of \(\pm 20 \mathrm{dBm}\) (b).

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Memory systems are a prime area for significant improvements in overall system throughput. Read how TI's memorymanagement ICs can get you in and out of memory faster no matter which processor you choose.

You can now solve a problem whose solution has eluded design engineers for years: How to catch memory speeds up to CPU speeds. The solution lies with TI's advanced memorymanagement circuits, and you can use them with whichever processor best suits your application.

\title{
Texas Instruments can help processor speeds.
}


A universal architecture enables these TI devices to work with - and enhance - virtually any high-speed microprocessor or bus structure, even custom engines.

In addition, your component count is cut because these are single-chip VLSI circuits. Your design time and effort are shorter and easier because of

TI's comprehensive Memory Management Design Kit (see page 4).

\section*{TI addresses your major memory-design concerns}

To immediately improve memory-access time, use both main and cache memories, as shown in the block diagram. This approach can produce up to a 3 X increase in system performance.

Frequently accessed data and instructions are stored in a few high-speed static random-access memories and "tagged" by a TI industry-standard cache controller (SN74ACT2151/4). These \(2 \mathrm{~K} \times 8\) CMOS controllers are the fastest available and can support deep cache architectures of 16 K or even 32 K .

\section*{TI's MegaChip Technologies}

Our emphasis on volume manufacturing of high-density circuits is the catalyst for ongoing advances in how we design, process, and manufacture semiconductors and in how we serve our customers. These are our MegaChip \({ }^{\text {TM }}\) Technologies. They are the means by which we can help you and your company get to market faster with better, more competitive products.
tions on chip to improve flexibility and speed and to allow for custom timing routines. This controller supports nibble- and page-mode access and scrubbing-mode refresh to increase memory output.


High-speed memories can be designed with less effort and implemented more cost-effectively by using Tl's family of universal memory-management ICs. These devices, all of which are contained in TI's Memory Management Design Kit, will work with and enhance almost any high-speed processor.

This scheme is cost-effective because slower, less expensive dynamic randomaccess memories (DRAMs) can be used for main memory.

When you must assure system integrity, use of an error-detection-and-correction (EDAC) circuit can improve system reliability 500 -fold. Since this approach is necessary with memory arrays larger than half a million bits, TI offers its leadership 32-bit EDAC.

The SN74AS632 detects dual-bit errors and detects and corrects single-bit errors while avoiding processor wait states. And at 25 ns for error detection, it meets your high-performance needs.

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Soon to come: An ASIC (applicationspecific integrated circuit) solution.

Reducing over/undershoot is accomplished by TI's 2000 Series buffers and drivers - 25 -ohm series-damping resistors on the output prevent false reads at DRAM input. For example, the SN74BCT2828 driver can reduce undershoot by \(40 \%\) compared to traditional approaches. TI's 2000 Series has a high-drive current suitable for VME and MULTIBUS \({ }^{\circledR}\) II bus structures.

You can use any or all of TI's memory-management ICs to obtain the superior performance that marks a market winner. And there's no design rule that says your memory-management chips and your CPU have to come from the same supplier.

Turn page for more information.

\title{
The tools you need to design a high-performance memorymanagement system are between these
}

\section*{covers:}

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-SN74ALS6301 16K to 1 Mbit DRAM Controller
-SN74BCT2828 10-bit Buffer/ Driver with series-damping resistor
-TIBPAL16R8-10 and TIB82S105B High-speed Programmable-logic Devices for user-defined timing control
-TMS4464 256K DRAM
- Memory Management Applications Handbook containing applications reports and briefs that supply valuable insights into memory-management system design.
- Data sheets on TI circuits designed for efficient memory management.
- Memory-management-product software graphic-symbol libraries and supporting documentation for use with Futurenet \({ }^{\text {TM }}\) or Mentor Graphics \({ }^{\text {TM }}\) CAE systems.
For more information on TI's
Memory Management Design Kit, call 1-800-232-3200, ext. 3203, or contact your nearest TI field sales office or authorized distributor.

\section*{Texas Instruments Incorporated}

SDVø63ED800C

\section*{P.O. Box 809066}

Dallas, Texas 75380-9066
YES, please send me more details on TI's universal memorymanagement ICs.
\begin{tabular}{llll}
\hline NAME & & \\
\hline TITLE & & \\
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\section*{DESIGN IDEAS}
decay time is 1 sec .
For a small sacrifice in audio quality, you can reduce the parts count by substituting a simplified peak detector (Fig 2). The simpler circuit causes a faint buzzing, but this noise is audible only with some stereo equipment.

You can also shift the \(40-\mathrm{dBm}\) gain range. Changing \(R_{4}\) to \(30 \mathrm{k} \Omega\), for example, sets the range from -30 to 10 dBm . (With the higher gain, you need a capacitor in series with \(R_{5}\) to prevent excessive voltage offset.) Because the noninverting input of op amp \(\mathrm{IC}_{1 \mathrm{~A}}\) is noise-sensitive, it should be physically small and remote from the noisy peak-detector outputs.

EDN


Fig 2-You can save parts by using this peak detector in the Fig 1 circuit.

\section*{Three-rail power supply uses four diodes}

\author{
Luis de Sa \\ Universidade de Coimbra, Coimbra, Portugal
}

The circuit shown in Fig 1 generates three supply voltages using a minimum of components. Diodes \(\mathrm{D}_{2}\) and \(D_{3}\) perform full-wave rectification, alternately charging capacitor \(\mathrm{C}_{2}\) on both halves of the ac cycle. On the other hand, diode \(\mathrm{D}_{1}\) with capacitor \(\mathrm{C}_{1}\) and diode \(\mathrm{D}_{4}\) with capacitor \(\mathrm{C}_{3}\) each perform half-wave rectification. The full- and half-wave rectification arrangement is satisfactory for modest supply currents drawn from the
-5 and +12 V regulators ( \(\mathrm{IC}_{3}\) and \(\mathrm{IC}_{2}\) ).
You can use this circuit as an auxiliary supply in a \(\mu \mathrm{P}\)-based instrument, for example, and avoid the less attractive alternatives of buying a custom-wound transformer, building a more complex supply, or using a secondary winding (say 18 V ac ) and wasting power in the 5 V regulators.

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Fig 1-This simple power supply generates three regulated voltages using a minimum of components.

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}

\section*{DESIGN IDEAS}

\section*{Circuit generates frequency difference}

\author{
Steve Momii \\ University of Washington, Seattle, WA
}

Fig 1's output pulse rate equals the difference in pulse rates of the inputs \(\mathrm{F}_{0}\) and \(\mathrm{F}_{1}\), where \(\mathrm{F}_{1}<\mathrm{F}_{0}\). (You must generate \(2 \mathrm{~F}_{0}\), but the duty cycle need not be \(50 \%\).)

The easiest way to understand the circuit's operation is in terms of sampling theory. Consider the exclusiveOR gate \(\mathrm{IC}_{2}\) as a modulator, in which \(\mathrm{F}_{1}\) modulates the carrier frequency \(\mathrm{F}_{0}\). The resulting waveform contains the sum and difference of \(\mathrm{F}_{0}\) and \(\mathrm{F}_{1}\). By using \(\mathrm{IC}_{1 \mathrm{~B}}\) to sample this waveform at \(2 \mathrm{~F}_{0}\), you generate an aliasing frequency that folds around the Nyquist frequency \(\mathrm{F}_{0}\); ie, the output difference frequency becomes
\[
\mathrm{F}_{\mathrm{D}}=2 \mathrm{~F}_{0}-\left(\mathrm{F}_{0}+\mathrm{F}_{1}\right)=\mathrm{F}_{0}-\mathrm{F}_{1} .
\]

Of course, variations are posssible. If \(\mathrm{F}_{1}=\mathrm{F}_{0} / \mathrm{n}\) (using a divide-by-n counter), then \(\mathrm{F}_{\mathrm{D}}=\mathrm{F}_{0}(1-1 / \mathrm{n})\). And, if \(\mathrm{F}_{1}=\mathrm{F}_{0} \cdot \mathrm{~m} / \mathrm{n}\) (using a rate multiplier), then \(\mathrm{F}_{\mathrm{D}}=\mathrm{F}_{0}(1-\mathrm{m} / \mathrm{n})\).

EDN


Fig 1-Two ICs generate the frequency difference between \(F_{0}\) and \(F_{1}\).

\section*{Multivibrator achieves near-50\% duty cycle}

Edward W Rummel
Robin Baker Associates Inc, Short Hills, NJ
The 3 -component astable multivibrator of Fig 1 provides an output with a near-50\% duty cycle. Using this circuit, you don't need the additional flip-flop usually required to obtain such a duty cycle. What's more, the CMOS timer shown costs a penny less than often-used equivalent bipolars ( 100 -piece price).

Three factors contribute to the output symmetry. The capacitor charges and discharges through the same external resistor. An internal resistive divider sets accurate switching thresholds within the chip (the bipolar types use dividers as well). And most important, \(\mathrm{IC}_{1}\) 's CMOS output stage switches fully between ground and \(\mathrm{V}_{\mathrm{CC}}\), avoiding the errors due to asymmetry that are often found in a TTL timer's output. The IC's internal switching-threshold tolerances can cause a deviation of several percent from the desired \(50 \%\) duty


Fig 1-Based on a CMOS version of a 555 timer, this astable multivibrator's output symmetry depends on the tolerance of the internal switching thresholds.
cycle. To meet a tighter specification, you might have to select from a group of ICs.

EDN

\section*{DESIGN IDEAS}

\section*{Digital inputs program audio compressor}

Jeffery L Sharp
Philips Subsystems and Peripherals Inc, Knoxville, TN

With the circuit in Fig 1, you can select one of four
different gain-compression slopes ( \(1: 1,2: 1,3: 1\), or \(4: 1\) ) without changing the circuit's unity-gain point. A compressor is characterized by two attributes of its amplitude transfer function-the slope and the \(0-\mathrm{dB}\) gain point (Fig 2).


Fig 1-With this circuit, you can select gain-compression ratios of 1:1, 2:1, s:1, and 4:1 without changing the unity-gain point. By changing inputs \(G_{0}\) and \(G_{1}\) of the analog multiplexer, you control the detector current in each half of the NE572 (IC 2 ), and, therefore, the gain of each section.

\title{
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}


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\(\qquad\)

\author{
\footnotetext{
\footnotetext{

}
} \\ \(\square\)
}

\section*{Marconi}

Instruments 182

\section*{DESIGN IDEAS}

The gain of amplifier \(\mathrm{IC}_{1 \mathrm{~A}}\) is determined by the NE572 variable-gain cells ( \(\mathrm{IC}_{2 \mathrm{~A}}\) and \(\mathrm{IC}_{2 \mathrm{~B}}\) ). Each half of the NE572 multiplies the value of the current passing through the \(\Delta \mathrm{G}\) section by the value of the current flowing into the IC's detector section. When gate inputs \(\mathrm{G}_{0}\) and \(\mathrm{G}_{1}\) equal 0 , the detector currents are constant. The gain of \(\mathrm{IC}_{1 \mathrm{~A}}\) is the same for all input amplitudes, resulting in a \(1: 1\) transfer slope. For the other gateinput combinations ( 01,10 , and 11 ), the detector currents are selected by the 4052 analog multiplexer \(\left(\mathrm{IC}_{3}\right)\); the currents provide gains that are inversely proportional to the second, third, and fourth powers of the input-signal amplitude. In all cases, 1V rms produces unity gain.

EDN


Fig 2-The input/output transfer-function curves of the circuit in Fig 1 show the four possible signal-compression slopes and the unity-gain point for each slope.

\section*{Transducer improves ultrasonic ranging}

\author{
Mitchell Lee \\ National Semiconductor, Santa Clara, CA
}

In an ultrasonic ranging system, a single transducer


Fig 1-The ringing that follows excitation of a piezoelectric transducer can last for 20 msec , masking the echoes from targets within 10 ft of the transmitter. In this photo, scale is \(5 \mathrm{~V} / \mathrm{div}\) in the vertical axis and \(200 \mu \mathrm{sec} / \mathrm{div}\) in the horizontal axis.
often does double duty: First, it behaves like a loudspeaker while generating the transmit pulse; then-ideally-it immediately stops ringing and behaves like a microphone while listening for the return echo. Piezo-


Fig 2-Ring duration for an electrostatic transducer is less than \(1 / 20\) th that of a piezoelectric type, allowing easy recognition of an echo that occurs \(800 \mu\) sec after the transmit pulse begins. Here, scale is 5 V/div in the vertical axis and \(200 \mu \mathrm{sec} / \mathrm{div}\) in the horizontal axis.

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\section*{National Semiconductor}

\section*{DESIGN IDEAS}
electric transducers, however, have a high Q and can continue to ring for 20 msec or more after excitation has ceased, thus masking return echoes representing more than 10 feet of range (Fig 1).

Broadband electrostatic transducers, however, lower the minimum range by reducing the ring time (Fig 2). You can combine an electrostatic transducer with an ultrasonic transceiver IC (Fig 3) to build a ranging system that senses objects at distances from 4 in . to more than 30 feet.

Transducer \(\mathrm{Y}_{1}\) 's broadband characteristic simplifies tuning. The secondary of \(\mathrm{T}_{1}\) resonates with the \(500-\mathrm{pF}\) capacitor \(\mathrm{C}_{1}\) at a frequency between 50 and 60 kHz . You tune \(\mathrm{L}_{1}\) to this frequency by using an oscilloscope to note the maximum echo sensitivity at pin 1 . Step-up transformer \(\mathrm{T}_{1}\) provides 150 V bias for the transducer.

EDN
\(\qquad\)


Fig 3-This ultrasonic ranging system features an ultrasonic transceiver \(\left(I C_{1}\right)\) and an electrostatic transducer \(\left(Y_{1}\right)\). Its sensing range is 4 in. to more than 30 ft .

\section*{Cut your losses}


\section*{High current, low dropout regulators reduce power loss}

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LT1083 DROPOUT VOLTAGE VS. OUTPUT CURRENT

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Cut your losses with our triplethreat power regulator family. LT1083/84 devices are packaged in TO-3 metal cans and TO-247 plastic. The LT1085 is in TO-3 and TO-220 plastic, and all are offered in MIL-STD-883 versions. Pricing begins at \(\$ 3.70\) each, in quantities of 100. For literature contact: LINEAR TECHNOLOGY CORPORATION, 1630 McCarthy Blvd., Milpitas, CA 95035. 800-637-5545.

\section*{DESIGN IDEAS}

\section*{Software provides rapid parity check}

Jerold R Thompson
Sperry Corp, Salt Lake City, UT
When the need arises to check or generate the parity of character codes in software, the programmer must
\begin{tabular}{|lll|}
\hline & LISTING 1 & \\
MOVE.B & & D0,D1 \\
LSR.B & & \#4,D1 \\
EOR.B & & D1,D0 \\
MOVE.B & & D0,D1 \\
LSR.B & \#2,D1 \\
EOR.B & & D1,D0 \\
MOVE.B & D0,D1 \\
LSR.B & \#1,D1 \\
EOR.B & D1,D0 \\
\hline
\end{tabular}
usually develop an algorithm because very few microprocessors have instructions for this purpose. Such algorithms involve counting the number of 1 -valued bits and then testing whether the count is odd or even.

The algorithm will run faster if its code simulates a hardware parity checker such as that of Fig 1. The 68008 code of Listing 1 emulates the three logic levels of Fig 1. After executing this routine, you can test bit 0 of register D0. The bit will be cleared (0) if the character code had even parity, or it will be set (1) if the parity was odd.

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Fig 1-You can quickly test 8-bit character codes for parity by emulating this hardware in a software routine.

\section*{VCO generates frequencies above 40 MHz}

\section*{Doug Farrar}

Apple Computer Inc, Cupertino, CA
The VCO of Fig 1 is inexpensive, operates from 5 V , and generates square, TTL-compatible complementary outputs. In addition, the circuit lends itself to integration because it uses only npn transistors and standard digital functions.
\(\mathrm{V}_{\mathrm{IN}}\) and \(\mathrm{R}_{\mathrm{T}}\), along with transistor \(\mathrm{Q}_{\text {IE }}\) and its emitter resistor, set the input current \(\mathrm{I}_{\mathrm{T}}\). Current-mirror transistors \(Q_{1 B}\) and \(Q_{1 D}\) also each sink \(I_{T}\). Then, the timing capacitor \(\mathrm{C}_{\mathrm{T}}\) works with current-source transistors \(\mathrm{Q}_{1 \mathrm{~A}}\) and \(Q_{1 C}\) to set the output half-cycle interval. Flip-flop \(\mathrm{IC}_{2 \mathrm{~B}}\) buffers the heavily loaded Q and \(\overline{\mathrm{Q}}\) outputs of \(\mathrm{IC}_{2 \mathrm{~A}}\).

A high level at \(\mathrm{IC}_{2 \mathrm{~A}}\) 's Q output, for example, forces \(Q_{1 A}\) 's base voltage near \(\mathrm{V}_{\mathrm{CC}}\), which clamps node A about

\section*{Technology Update}

\title{
NEW TI/CEPT PRIMARY RATE INTERFACE IC'S
}

\section*{A Transition to ISDN}

Silicon Systems offers two new line interface transceivers that can provide an optimum solution to users for easing the transition to ISDN. The new SSI 78P233 conforms to the North American Tl standards, and the SSI 78P234 conforms to the world-wide CEPT standards.

The new devices can be utilized, with the telephone wiring which is already in place, as line interface transceivers in the verification of voice and data transmission into an all-digital, high speed ISDN communications network.

\section*{Converting DS-1 and CEPT-levels} to TTL
The demand for high rates of transmission has intensified the demand for what are commonly termed T1 interface chips-those IC's which provide the interface functions necessary to convert DS-1 or CEPTlevel signals to TTL-level and conversely for such applications as private automatic branch exchanges, channel banks, multiplexors, communications processors, digital electronics switching systems, and Tl network process systems.

Historically, interface devices have had a spotty track record. Older designs require manual tuning for clock recovery. New designs trade off added features for increased cost. Some newer devices are only available in the form of two-chip sets. Other more hopeful entries into the marketplace have been either untested or lagging, and facsimile transmission is accelerating, both in North America and overseas.

\section*{High Performance, Low-noise} Bipolar Transceivers
The SSI 78P233, conforming to the Tl's 1.544 Mbps standard transmission rate and the SSI 78P234, conforming to CEPT's 2.048 Mbps
transmission rate, are complete line interface chips. As single-chip analog front-end devices, they are actually transceiver devices consisting of receiver, transmitter, and loopback sections.

Both single chip devices are designed in Silicon Systems' lownoice, high performance bipolar technology - best suited to combine high speed functions and low cost for dependable equipment line
interfacing. And both devices are available for plug compatibility in 24 -pin DIP and SO packages and require only a single 5 -volt power supply. The SSI 78P234 is additionally available in the smaller 20 pin DIP and SO versions.
For more information, send for the SSI 78P233 data sheet: Silicon Systems, 14351 Myford Road, Tustin, CA 92680. Phone: (714) 731-7110.

78 P233 Block Diagram


The receiver performs clock and data recovery on the incoming signal, and then converts them to TTL-level. This is accomplished through the use of a trimmed analog phase locked loop and an external precision resistor. No other tunable or precision components are required. Easy interfacing to deframer circuitry is achieved through the use of standard TTL-level receive outputs.

The transmitter accepts TTL data and clock signals to be transmitted, combines and

\begin{abstract}
converts them to alternate mark inversion (AMI)-level and then equalizes and shapes outgoing pulses according to line length, as selected by the user. Featured are six different line equalization settings for pulse shaping at the DSX-1 level, which eliminates the need for off-chip adjustable networks.

The loopback section is used for test and diagnostic operations on the line side or local side of the chip.
\end{abstract}

\section*{DESIGN IDEAS}


Fig I-Two ICs form a voltage-controlled oscillator that generates TTL-compatible, complementary outputs with maximum frequency above 40 MHz .
one \(V_{\text {BE }}\) below \(V_{C C}\). IC \(_{2 A}\) 's \(\bar{Q}\) output is low, so resistors \(R_{3}\) and \(R_{4}\) force \(Q_{1 c}\) 's base to about 1.7 V . This action allows node B's voltage to decrease according to a slope determined by \(\mathrm{I}_{\mathrm{T}}\) and \(\mathrm{C}_{\mathrm{T}}\). When B reaches the switching threshold \(\mathrm{V}_{\mathrm{TH}}\) of \(\mathrm{IC}_{2 \mathrm{~A}}\) 's Reset input (about 1.4 V ), the flip-flop changes state.
Next, \(Q_{1 C}\) clamps node \(B\) near \(V_{C C}\) minus \(V_{B E}\), but the capacitor (charged to \(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{BE}}-\mathrm{V}_{\text {TH }}\) ) forces node A to \(2\left(\mathrm{~V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{BE}}\right)-\mathrm{V}_{\text {TH }}\). Because \(\mathrm{Q}_{1 \mathrm{~A}}\) is back-biased, node A ramps down to the switching threshold of \(\mathrm{IC}_{2 \mathrm{~A}}\) 's Set input, which toggles the flip-flop again and renews the cycle.
This description of circuit operation applies to timing capacitors greater than 1000 pF . The circuit operates properly for smaller values, but the waveforms are dominated by parasitics; you must use experimental data in place of theoretical predictions of performance. The output waveform remains square, however, for frequencies to 40 MHz and above. Fig 2 shows measured values of input current \(I_{T}\) vs output frequency for different values of timing capacitance.
Small timing capacitors also allow lower excursions for the waveforms at nodes A and B. Clamps on these nodes ensure that the excursions go low enough to set and reset flip-flop \(\mathrm{IC}_{2 \mathrm{~A}}\) without saturating the current sources \(Q_{1 B}\) and \(Q_{1 D}\) : Resistors \(R_{1}-R_{4}\) bias the transistor bases \(Q_{1 A}\) and \(Q_{1 C}\) (at 1.7 V low) to clamp nodes A and B at about 0.9 V min .

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Fig 2-This graph illustrates how different values of Fig 1's timing capacitor \(C_{T}\) provide different output frequencies for a given value of input current.

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\section*{DESIGN IDEAS}

\section*{Serial interface buffers parallel data}

\section*{Steve Walker}

ITT Federal Electric Corp, Vandenberg AFB, CA
The data interface of Fig 1 accepts byte-parallel data and transmits it in a bit-serial format. The FIFO memory \(\left(\mathrm{IC}_{1}\right)\) performs the parallel-to-serial conversion and also accumulates 16 data bytes before initiating the serial transmission. The circuit is useful in applications that require a slow accumulation of data for subsequent transmission in one high-speed burst, or in applications in which data transmission on a byte-by-byte basis requires too-frequent service from the data source.

The FIFO memory ( 328 -bit bytes) can serialize output data or deserialize input data. Each rising edge of the Data Strobe signal (pin 18) clocks a byte of data
into the memory (Fig 2). Following the sixteenth byte, the half-full signal (Flag output, pin 19) goes high, which sets the flip-flop ( \(\mathrm{IC}_{2}\), ) which in turn activates the \(\mathrm{VCO} \mathrm{IC}_{3}\).
The VCO generates the system's serial clock signal. You can vary this signal from 200 to 1000 kHz by adjusting the \(10-\mathrm{k} \Omega\) potentiometer, or you can generate lower frequencies by connecting a larger capacitor to the VCO. (For operation above 500 kHz , however, you would have to use the Am2812A FIFO memory instead of the Am2812.)

The FIFO generates an output-ready pulse (OR, pin 3) for each bit of serial data transmitted. Counter \(\mathrm{IC}_{5}\) counts these pulses and signals the transmission of 128 bits (16 bytes) by asserting a logic-high signal on its


Fig 1-IC \(C_{1}\) provides a 16-byte buffer memory for parallel-input data and formats the data for serial transmission to \(I C_{7}\). \(I C_{;}\)then reorganizes the data in byte-parallel format.

\section*{DESIGN IDEAS}

QD2 output (pin 8). This signal triggers the one-shot \(\mathrm{IC}_{6}\), causing it to produce \(1-\mu \mathrm{sec}\) pulses at outputs Q and \(\overline{\mathrm{Q}}\). These pulses reset the flip-flop (which turns off the VCO), the counter, and the FIFO memory. When the data source sees the Clock Enable output go low, it can load another 16 bytes into the FIFO memory, provided the receiver signals that it is ready.

A second FIFO memory, \(\mathrm{IC}_{7}\), can receive and buffer the serial data. Its Flag output indicates reception of a complete block of data; you must then clear the device by asserting a CLEAR FIFO signal at pin 4. The IC
has 3-state outputs controlled by the OE signal at pin 11, so you can connect the chip directly to a data bus.

If desired, you can use the one-shot's \(\overline{\text { RESET }}\) output to interrupt the data source. You can add extra circuitry to increase the block size to 32 bytes, and you can cascade additional FIFO memories to increase the buffer memory (in 32-byte increments).

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Fig 2-These waveforms for the Fig 1 circuit illustrate the serial transmission of one 16-byte block of data.

\section*{Enhanced op amp delivers 100 V p-p}

\author{
Barry Kline \\ Technicare Corp, Solon, OH
}

If you need an amplifier that provides more than \(\pm 50 \mathrm{~V}\) output swing along with the high gain and low offset of a high-performance op amp, consider the Fig 1 circuit. It employs a gain stage \(\left(Q_{1}-Q_{4}, R_{1}-R_{4}\right)\) to multiply the op amp's \(\pm 10 \mathrm{~V}\) output swing to the desired level. The combined op amp and gain stage may be regarded as a high-voltage amplifier.

The gain stage is based on a design by Jerald Graeme (Ref 1), but adds current feedback via \(R_{4}\) to achieve three performance improvements: reduction of openloop output impedance (reduced sensitivity to changes
in load current); reduction of output current \(I_{A}\) from the op amp; and increased dynamic range due to a reduction in signal voltage across the current-sensing resistors \(R_{1 A}\) and \(R_{1 B}\).
\(Q_{1}\) and \(Q_{3}\) act as cascode stages for the op-amp supply currents \(\mathrm{I}_{\mathrm{A} 1}\) and \(\mathrm{I}_{\mathrm{A} 2} . \mathrm{Q}_{2}\) and \(\mathrm{Q}_{4}\) sense these currents and provide amplified output currents \(I_{1}\) and \(I_{2}\) :
\[
\begin{align*}
& \mathrm{I}_{01}=\mathrm{I}_{\mathrm{A} 1} \times \frac{\mathrm{R}_{1}}{\mathrm{R}_{2}} \\
& \mathrm{I}_{02}=\mathrm{I}_{\mathrm{A} 2} \times \frac{\mathrm{R}_{1}}{\mathrm{R}_{2}} \tag{1}
\end{align*}
\]

Also, because the difference in supply currents is equal to the op-amp output current \(\left(I_{A}=I_{A 1}-I_{A 2}\right)\),

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\section*{Component Solutions For Your Power System}

\section*{DESIGN IDEAS}
\[
\begin{equation*}
\mathrm{I}_{01}-\mathrm{I}_{\mathrm{O} 2}=\left(\mathrm{I}_{\mathrm{A} 1}-\mathrm{I}_{\mathrm{A} 2}\right) \times \frac{\mathrm{R}_{1}}{\mathrm{R}_{2}}=\mathrm{I}_{\mathrm{A}} \times \frac{\mathrm{R}_{1}}{\mathrm{R}_{2}} . \tag{2}
\end{equation*}
\]

Feeding this current imbalance back to the op-amp output terminal provides negative feedback for the booster stage. Recall that:
\[
\begin{equation*}
\mathrm{I}_{01}-\mathrm{I}_{02}=\mathrm{I}_{\mathrm{A}} \times \frac{\mathrm{R}_{1}}{\mathrm{R}_{2}} \tag{3}
\end{equation*}
\]

In addition,
\[
I_{4}=\left(I_{01}-I_{02}\right)-I_{L}=\left(I_{A} \times \frac{R_{1}}{R_{2}}\right)-I_{L}
\]

But, \(I_{A}=I_{3}-I_{4}\).
\[
\begin{array}{ll}
\text { So, } & I_{4}=\left(I_{3}-I_{4}\right) \times\left(\frac{R_{1}}{R_{2}}\right)-I_{L}  \tag{4}\\
\text { Or, } & I_{4}\left(1+\frac{R_{1}}{R_{2}}\right)=\left(I_{3} \times \frac{R_{1}}{R_{2}}\right)-I_{L}
\end{array}
\]

Substitute \(V_{A}\) for the op amp's output voltage:
\[
\begin{aligned}
& \mathrm{I}_{4}=\frac{\mathrm{V}_{\text {OUT }}-\mathrm{V}_{\mathrm{A}}}{\mathrm{R}_{4}} \\
& \mathrm{I}_{3}=\frac{\mathrm{V}_{\mathrm{A}}}{\mathrm{R}_{3}} \\
& \mathrm{I}_{\mathrm{L}}=\frac{\mathrm{V}_{\text {OUT }}}{\mathrm{R}_{\mathrm{L}}} .
\end{aligned}
\]

Therefore:
\[
\begin{equation*}
\frac{V_{\text {OUT }}-V_{A}}{R_{4}}\left(1+\frac{R_{1}}{R_{2}}\right)=\frac{V_{A} R_{1}}{R_{3} R_{2}}-\frac{V_{\text {OUT }}}{R_{L}} . \tag{6}
\end{equation*}
\]

The equation can be solved for voltage gain of the booster stage:
\[
\begin{align*}
\frac{V_{\text {OUT }}}{V_{A}} & =\frac{1+\frac{R_{4}}{R_{3}}\left(\frac{R_{1}}{R_{1}+R_{2}}\right)}{1+\frac{R_{4}}{R_{L}}\left(\frac{R_{2}}{R_{1}+R_{2}}\right)}  \tag{7}\\
& =\left[1+\frac{R_{4}}{R_{3}}\left(\frac{R_{1}}{R_{1}+R_{2}}\right)\right]\left(\frac{R_{L}}{R_{0}+R_{L}}\right) .
\end{align*}
\]

Output impedance \(\left(R_{0}\right)\) is:
\[
\begin{equation*}
\mathrm{R}_{0}=\frac{\mathrm{R}_{4} \times \mathrm{R}_{2}}{\mathrm{R}_{1}+\mathrm{R}_{2}} . \tag{8}
\end{equation*}
\]

To calculate the op amp's output current \(\left(\mathrm{I}_{\mathrm{A}}\right)\) :
\[
\begin{align*}
I_{A} & =I_{3}-I_{4}=\frac{V_{A}}{R_{3}}-\frac{V_{\text {OUT }}-V_{A}}{R_{4}} \\
& =V_{A}\left(\frac{1}{R_{3}}+\frac{1}{R_{4}}-\frac{V_{\text {OUT }}}{V_{A} R_{4}}\right) . \tag{9}
\end{align*}
\]

The no-load booster gain is obtained by setting \(\mathrm{R}_{\mathrm{L}}\) to infinity in Eq 7:


Fig 1-Gain stage with current feedback boosts a conventional op amp's output to more than \(100 \mathrm{~V} p\)-p.

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Fig 2-The Fig 1 circuit produces 100V p-p sine- and square-wave outputs-a and b, respectively. Slew response to the square wave is a product of the OP-27 slew rate (2.8V/usec typ) and gain of the booster stage (approximately 5.5).
\[
\begin{equation*}
\frac{\mathrm{V}_{\text {OUT }}}{\mathrm{V}_{\mathrm{A}}}=1+\frac{\mathrm{R}_{4}}{\mathrm{R}_{3}}\left(\frac{\mathrm{R}_{1}}{\mathrm{R}_{1}+\mathrm{R}_{2}}\right) . \tag{10}
\end{equation*}
\]

Substituting the Eq 10 no-load booster gain into Eq 9 yields:
\[
\begin{equation*}
I_{A}=\frac{V_{A}}{R_{3}}\left(\frac{R_{2}}{R_{1}+R_{2}}\right) \tag{11}
\end{equation*}
\]

The use of current feedback reduces the output impedance and the required op-amp output current by a factor of \(R_{2} /\left(R_{1}+R_{2}\right)\). Also, reducing \(I_{A}\) reduces signal current variations in the supply lines.

Be careful not to exceed the power-dissipation capacity of components \(Q_{2}, Q_{4}, R_{2}, R_{4}\), and \(R_{L}\). In addition, quiescent current in \(Q_{2}\) and \(Q_{4}\) is proportional to the current gain \(R_{1} / R_{2}\), so choose this ratio carefully. EDN

\section*{Reference}
1. Graeme, Jerald G, Designing with Operational Ampli-fiers-Applications Alternatives, McGraw-Hill, 1977, pg 14.

\title{
Generate bipolar logic levels from a 5 V supply
}

\author{
Edward W Rummel \\ Robin Baker Associates Inc, Short Hills, NJ
}

The circuit in Fig 1 converts serial TTL data to the bipolar levels compatible with RS-232C transmission. It employs a charge-pump technique to generate the necessary negative supply voltage by using an inexpensive 3P2T analog switch (CD4053) and two \(10-\mu \mathrm{F}\) capacitors. The circuit saves money in some cases (the CD4053 costs \(\$ 0.87\); the ICL7660 de/dc converter costs \(\$ 2.40(100)\), on the other hand).

Note that the CD4053 has a built-in level shifter but requires a square-wave drive (the ICL 7660 includes the square-wave oscillator)-you can use a clock signal if available, or build a simple RC oscillator (eg, the CD40106 in Fig 1). Thus, two sections of the 3P2T switch produce the negative supply voltage, and the third section shifts the voltage level of the TTL signal to produce an RS-232C bipolar format. (You can invert this data by swapping the connections to pins 3 and 5 of the CD4053.)

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Fig I-Two inexpensive ICs convert serial TTL data to a bipolar format suitable for transmission on \(R S\)-232C lines.

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\hline 27 CX 322 & 24 & 300 mil & \(4 \mathrm{~K} \times 8\) & \(35,40,45 \mathrm{~ns}\) & \(40 \mathrm{~mA}^{*}\) \\
\hline 27 CX 641 & 24 & 600 mil & \(8 \mathrm{~K} \times 8\) & \(40,45,55 \mathrm{~ns}\) & 60 mA \\
\hline 27 CX 642 & 24 & 300 mil & \(8 \mathrm{~K} \times 8\) & \(40,45,55 \mathrm{~ns}\) & 60 mA
\end{tabular}

\footnotetext{
* User-programmable \(500 \mu \mathrm{~A}\) low-power standby mode
}

\section*{DESIGN IDEAS}

\section*{Basic program linearizes thermistors}

\section*{Mick Murray}

\author{
Lab-Line Instruments Inc, Melrose Park, IL
}

You can reduce the nonlinearity of a negative-tempera-ture-coefficient (NTC) thermistor (TH-1) by adding a series resistor (RS) and a parallel resistor (R). Fig 1 contains the resulting network, designated \(\mathrm{R}_{\mathrm{EQ}}\). To define optimum values for the resistors, the Basic program of Listing 1 uses an iterative process, selecting from standard values in data statements 480 through 560 .

Using a 3 -point straight-line approach, the program can linearize any NTC thermistor. First, the program prompts you for the lowest and highest temperatures of interest and the corresponding thermistor resistance
(which you can find in the manufacturer's data sheet). Next, it calculates the exact mid-range temperature and prompts you for the resistance at that point as well. You divide each resistance value by 1 k before entering it in the program.

In the example of Fig 1, thermistor TH-1 varies 127 to \(2.64 \mathrm{k} \Omega\) over the temperature range of 20 to \(130^{\circ} \mathrm{C}\). Resistance change in the lower half of this range ( 20 to \(75^{\circ} \mathrm{C}\) ) is 9.3 times greater than that in the upper half. By comparison, the computer-selected network \(\mathrm{R}_{\mathrm{EQ}}\) varies only 1062 to \(977 \Omega\) over the same temperature range, but the change in each half of the range is nearly equal. (Note, however, that the network's linearity error increases as you widen the specified temperature range.)


Fig 1-The Basic program in Listing 1 selects the standard values for resistors \(R\) and \(R S\) that provide the most nearly linear relationship between \(R_{E Q}\) and temperature.

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\section*{DESIGN IDEAS}

\section*{LISTING 1-LINEARIZATION PROGRAM}
```

10 REM
20 REM
3 0 ~ R E M ~ T H E R M I S T O R ~ L I N E A R I Z A T I O N ~ P R O G R A M ~
4 0 ~ R E M ~ C O P Y R I G H T ~ 1 9 8 6 ~ M I C K ~ M U R R A Y ~
50 REM LAB-LINE INSTRUMENTS, INC.
6 0 ~ R E M ~ M E L R O S E ~ P A R K , ~ I L . ~ 6 0 1 6 0 ~
70 REM
80 REM ***********************************
90 A$=STRING$(70,"*")
100 FOR T=1 TO 24:PRINT:NEXT
110 INPUT"INPUT LOW TEMPERATURE:";L
120 INPUT"INPUT LOW TEMP RESISTANCE:";RL
130 INPUT"INPUT HIGH TEMPERATURE:";H
140 INPUT"INPUT HIGH TEMP RESISTANCE:";RH
150 M = ((H-L)/2)+L
160 PRINT"THE CALCULATED MID-POINT TEMPERATURE = ";M;"DEGREES."
170 PRINT"INPUT THERMISTOR RESISTANCE AT ";M;" DEGREES:";
1 8 0 ~ I N P U T ~ R M : S T = 1 0 0 0 ~
190 REM DE=DELTA, ST=STORE
200 PRINT A$:PRINT:PRINT
210 DIM A(96):DIM B(96)
220 FOR X=1 T0 96:READ A(X):B(X)=A(X):NEXT X
230 FOR I=1 TO 96
240 FOR J=1 T0 96
250 LO=((RL+A(I))*B(J))/(RL+A(I)+B(J))
260 MI =((RM+A(I))*B(J))/(RM+A(I)+B(J))
270 HI=((RH+A(I))*B(J))/(RH+A(I)+B(J))
280 X=ABS(HI-MI)
290 XI=ABS(MI-LO)
300 DE=ABS (X1-X)
310 IF DE<=ST THEN ST = DE:RS=A(I):R=B(J):GOS|B 380
320 NEXT J
3 3 0 ~ N E X T ~ I ~
340 IF SD=0 THEN GOTO 580
350 PRINT" END OF CALCULATIONS."
360 PRINT C.HR$(7);:FORP=1T01000:NEXT:G0T0360
370 END
380 PRINT"BEST SO FAR:"
390 PRINT"RS =";A(I);"OHMS, AND R = ";B(J);"OHMS."
4 0 0 ~ P R I N T " ~ * * * ~ ( A T " ; L ; " D E G R E E S , ~ R t h = " ; L O ; " O H M S . ) " ~ "
410 PRINT" *** (AT";M;"DEGREES, Rt.h=";MI;"OHMS.)"
420 PRINT" *** (AT";H;"DEGREES, Rth=";HI;"OHMS.)"
430 PRINT"DELTA1 =";X;"OHMS, AND DELTA2=";X1;"OHMS."
440 HL=ABS (HI-LO)
450 PRINT"THE TOTAL CHANGE IN RESISTANCE FROM";L;"TO";H;"OEGREES IS";HL;"OHMS."
450 PRINT:PRINT A\$:PRINT"Workina....":PRINT
470 RETURN
4 8 0 DATA 1.00,1.02,1.05,1.07,1.10,1.13,1.15,1.18,1.21,1.24
4 9 0 ~ D A T A ~ 1 . 2 7 , 1 . 3 0 , 1 . 3 3 , 1 . 3 7 , 1 . 4 0 , 1 . 4 3 , 1 . 4 7 , 1 . 5 0 , 1 . 5 4 , 1 . 5 8
5 0 0 ~ D A T A ~ 1 . 6 2 , 1 . 6 5 , 1 . 6 9 , 1 . 7 4 , 1 . 7 8 , 1 . 8 2 , 1 . 8 7 , 1 . 9 1 , 1 . 9 6 , 2 . 0 0 ~
510 DATA 2.05,2.10,2.15,2.21,?.26,2.32,2.37,2.43,2.49,2.55
5 2 0 ~ D A T A ~ 2 . 6 1 , 2 . 6 7 , 2 . 7 4 , 2 . 8 0 , 2 . 8 7 , 2 . 9 4 , 3 . 0 1 , 3 . 0 9 , 3 . 1 6 , 3 . 2 4 , 3 . 3 2 ~
5 3 0 DATA 3.40,3.48,3.57,3.65,3.74,3.83,3.92,4.02,4.12,4.22,4.32,4.42
5 4 0 DATA 4.53,4.64,4.75,4.87,4.99,5.11,5.23,5.36,5.49,5.6?,5.76,5.9
5 5 0 DATA 6.04,6.19,6.34,6.49,6.65,6.81,6.98,7.15,7.32,7.5,7.68,7.87
5 6 0 ~ D A T A ~ 8 . 0 6 , 8 . 2 5 , 8 . 4 5 , 8 . 6 6 , 8 . 8 7 , 9 . 0 9 , 9 . 3 1 , 9 . 5 3 , 9 . 7 6 ~
5 7 0 REM THE FOLLOWING IS FOR THE SECOND DECADE OF RS VALUES:
580 SD=1:RESTORE
590 FOR N=1 TO 96
600 READ A(N):B(N)=A(N):A(N)=10*A(N)
6 1 0 ~ N E X T ~ N : G O T O 2 3 0 ~

```

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\section*{DESIGN IDEAS}

With each iteration, the program checks the network resistance \(\mathrm{R}_{\mathrm{EQ}}\) at the low, middle, and high temperatures. It seeks to achieve an equal resistance change in the upper and lower halves of the temperature range by trying all combinations of R and RS by substituting values from the data statements. It prints a result, searches for a better result, prints that, and continues
(Fig 2). Remember to denormalize the final result by multiplying the resistance values by 1 k . Typical execution time is about 10 minutes.

EDN
INPUT LOW TEMPERATURE: 20
INPUT LOW TEMP RESISTANCE: 127
INPUT HIGH TEMPERATURE: 130
INPUT HIGH TEMPERATURE RESISTANCE: 2.64
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INPUT THERMISTOR RESISTANCE AT 75 DEGREES: 12.96
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\(R S=1\) OHMS, AND R \(=1\) OHMS.
    \(\star \star \star\) (AT 20 DEGREES, \(\mathrm{R}_{\mathrm{EQ}}=.9922482\) OHMS.)
    \(\star \star \star\) (AT 75 DEGREES, REQ \(=.9331552\) OHMS.)
    \(\star \star \star\) (AT 130 DEGREES, \(\mathrm{R}_{\text {EO }}=.7844828\) OHMS.)
DELTA1 \(=.1486723\) OHMS, AND DELTA \(2=.059093\) OHMS
THE TOTAL CHANGE IN RESISTANCE FROM 20 TO 130 DEGREES IS . 2077653 OHMS.

Working
BEST SO FAR:
RS \(=8.66\) OHMS, AND R \(=1.07\) OHMS.
    \(\star \star \star\) (AT 20 DEGREES, \(R_{E O}=1.061627\) OHMS.)
    \(\star \star \star\) (AT 75 DEGREES, \(R_{\text {EQ }}=1.019542\) OHMS.)
    \(\star \star \star\) (AT 130 DEGREES, \(R_{E Q}=.9774454\) OHMS.)
DELTA1 \(=.0420962\) OHMS, AND DELTA2 \(=.04208481\) OHMS.
THE TOTAL CHANGE IN RESISTANCE FROM 20 TO 130 DEGREES IS .08418102 OHMS.
Working
END OF CALCULATIONS.

Fig 2—This example of Listing I's output shows two of the 14 iterations that the program prints in determining the \(R_{E Q}\) network in \(\operatorname{Fig} 1\). \(R=1.07 \mathrm{k} \Omega\) and \(R S=8.66 \mathrm{k} \Omega\) are the final values.

\section*{Test probe measures 4 to 220 V}

Christos S Koukourlis
Demokritos University of Thrace, Xanthi, Greece

Using inexpensive components, you can fit a simple probe circuit (Fig 1) into a pencil-sized enclosure.

When both LEDs are on, the probe indicates the presence of an ac voltage; either LED alone indicates the presence and polarity of a dc voltage.

The diode-bridge arrangement allows a 1 -way current source ( \(R_{1}, R_{2}, Q_{1}\), and \(Q_{2}\) ) to light either LED (or both) when the probe is activated by a test voltage.

\section*{DESIGN IDEAS}

Diodes in series with the LEDs provide the necessary peak-inverse voltage rating; the four diodes should each withstand \(300 \mathrm{~V} \min . \mathrm{R}_{3}\) and \(\mathrm{C}_{1}\) provide a spike-suppression network to protect the current-source transistors when the probe is connected to a high voltage. (Ed Note: The author recommends BF459 transistors. We substituted \(2 N 3439\) transistors (which are rated 350 V min collector-emitter sustaining voltage) only for want of data on the BF459.)

EDN


Fig 1-This simple test probe indicates ac or dc voltage over a 4 to 220 V range. The current source ( \(R_{1}, R_{2}, Q_{1}\), and \(Q_{2}\) ) ensures a constant current in the LEDs for any acceptable test voltage.

\section*{Wien-bridge filters enhance tone control}

\section*{Frédéric Boes}

Brugge, Belgium
Most audio tone controls affect midband gain, and they often create booming or hissing sounds when activated. You can avoid these problems by using a dual Wienbridge filter to provide independent control of the treble and bass frequencies.


Fig 1-This audio filter's two Wien-bridge networks provide \(\pm 9 \mathrm{~dB}\) of tone control for treble and bass frequencies.


Fig 2—The frequency response of the dual-filter tone control (Fig 1) shows little effect at 1 kHz but \(\pm 9-d B\) variation in the treble and bass regions. For comparison, the dashed lines show the response of a Boxandal tone control.

Experiments with equalizers indicate that the optimum center frequencies are about 100 Hz and 8 kHz . Using the relation \(\mathrm{f}=(2 \pi \mathrm{RC})^{-1}\), set the Fig 1 values accordingly:
\[
\begin{aligned}
& 100 \mathrm{~Hz}: \mathrm{R}_{1}=15 \mathrm{k} \Omega ; \mathrm{C}_{1}=0.1 \mu \mathrm{~F} \\
& 8 \mathrm{kHz}: \mathrm{R}_{2}=16 \mathrm{k} \Omega ; \mathrm{C}_{2}=1.3 \mathrm{nF} .
\end{aligned}
\]
\(\mathrm{R}_{3}\) and \(\mathrm{C}_{3}\) provide stability. You obtain a \(\pm 9-\mathrm{dB}\) variation of treble and bass by adjusting the potentiometers \(R_{4}\) and \(R_{5}\), respectively. The filter's frequency response is shown in Fig 2.
(Ed Note: The LF356 BiFET op amp is a good choice for this application: It provides low \(I_{B}\), low noise, and a good slew rate.)

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\section*{DESIGN IDEAS}

\section*{Variable reference extends converter's resolution}

\author{
Leonard Sherman \\ National Semiconductor, Santa Clara, CA
}

You can increase an A/D converter's apparent resolution by deriving part of its reference voltage from the input signal. The resulting input/output relationship, however, is nonlinear. In Fig 1, the 8-bit converter's sensitivity is 5 mV (10-bit resolution) for small input signals, yet the input range is 5 V .

The FET-input op amp \(\mathrm{IC}_{1 \mathrm{~A}}\) buffers the input signal;
\(\mathrm{IC}_{1 \mathrm{~B}}\) sums the input with a fixed voltage (in this case, the 5 V power supply). The result is a variable reference voltage for the converter, yielding the transfer function \(\mathrm{D}=\left(256 \times 4 \times \mathrm{V}_{\text {IN }}\right) \div\left(3 \times \mathrm{V}_{\text {IN }}+5\right)\) for the voltage and resistor values shown. D is the digital-output code. Note that the ADC0820 will exhibit increased noise and nonlinearity for \(\mathrm{V}_{\text {REF }}\) values of less than 1.25 V . EDN


Fig 1-You can increase an \(A / D\) converter's apparent resolution by combining the input signal with a fixed reference voltage and feeding this sum to the converter's reference input.

\section*{UART forms RS-232C/Centronics interface}

\section*{K S Perianayagam and U K Kalyanaramudu Bharat Electronics Ltd, Jalahalli, Bangalore, India}

You can use the circuit shown in Fig 1 to create an interface between an RS-232C output and a Centronics
input. Many personal computers contain only RS-232C output ports that cannot send information to line printers with Centronics ports. The Fig 1 interface, which has parts that cost about \(\$ 10\), is simpler and cheaper than adding a Centronics port to a personal computer.
The CDP-1854 UART ( \(\mathrm{IC}_{1}\) ) converts the serial-data

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\section*{DESIGN IDEAS}
stream from the RS-232C port into a byte-wide parallel transfer. You configure the UART by connecting its control pins to 5V and GROUND as shown in Fig 1. By selecting one of the outputs from the MC14411 bit-rate generator \(\left(\mathrm{IC}_{2}\right)\), you match the clock rate of the UART to the baud rate of the RS-232C line. The 74LS244 3 -state buffer ( \(\mathrm{IC}_{3}\) ) drives the signals from the UART's parallel outputs to the Centronics port.

The 1489A receiver IC converts the serial data from the RS-232C connection into TTL format and applies it to the UART. When the UART has received eight bits, it assembles them into a data byte, drives the byte to the 74LS244, and asserts the Data Available (DA) pin. The assertion triggers the monostable multivibrator \(\left(\mathrm{IC}_{4}\right)\), which delivers a voltage pulse to the strobe signal \(\overline{\mathrm{STB}}\), indicating to the Centronics port that data is
ready at the outputs of \(\mathrm{IC}_{3}\). The assertion of DA also asserts the CTS through the 1488A line-driver IC. CTS inhibits the RS-232C port from transmitting.
(Ed Note: The values of the components \(R\) and \(C\) determine the length of the strobe \(\overline{S T B}\). Given a minimum specification of 500 nsec for a Centronics port, \(R\) and C should be \(20 \mathrm{k} \Omega\) and 40 pF , respectively.)

When the Centronics port has latched the data byte, it asserts its acknowledge signal \(\overline{\mathrm{ACK}}\), which is connected to the UART's data-acknowledge input, \(\overline{\mathrm{DAR}}\). Asserting \(\overline{\mathrm{DAR}}\) causes the UART to deassert DA, allowing the RS-232C port to send another eight bits of data.

EDN


Fig 1-The CDP-1854 UART assembles serial RS-232C data into bytes for a Centronics port. \(I_{2}\) provides a baud rate that matches the \(R S-232 C\) transmission rate, and \(I C_{s}\) buffers the data byte to the Centronics port. Signals DA and \(\overline{D A R}\) control the timing of the two ports.

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\section*{DESIGN IDEAS}

\section*{Build step-up/step-down regulator}

\section*{Jade Alberkrack}

Motorola Semiconductor Products, Phoenix, AZ
Using a single inductor, a switching-regulator IC, and a few external components, you can build the step-up/ step-down battery-output regulator circuit in Fig 1. The circuit's novelty stems from its use of a single inductor to provide the voltage transformation. The circuit combines two basic switching-regulator configurations: a step-up circuit and a step-down circuit. Table

1 lists the regulator's specifications.
Fig 2's collection of switching circuits shows the evolution of Fig 1's circuit. The voltage step-down circuit in Fig 2a generates a voltage lower than that supplied by the battery. Fig 2b's circuit steps up the voltage provided by a battery. By combining the circuits as shown in Fig 2c and using a switching-regulator IC to provide \(\mathrm{Q}_{1}\) and \(\mathrm{Q}_{2}\), you get the design in Fig 1.

In Fig 2c's circuit, inductor L stores energy while \(Q_{1}\) and \(\mathrm{Q}_{2}\) are on (during \(\mathrm{T}_{\mathrm{ON}}\) ). When the transistors are

TABLE 1-SWITCHING-REGULATOR SPECIFICATIONS
\begin{tabular}{l|l|l}
\multicolumn{1}{c|}{ TEST } & \multicolumn{1}{|c}{ CONDITIONS } & \multicolumn{1}{c}{ RESULTS } \\
\hline LINE REGULATION & \(\mathrm{V}_{\mathrm{IN}}=7.5 \mathrm{TO} 14.5 \mathrm{~V}, \mathrm{I}_{\mathrm{OUT}}=120 \mathrm{~mA}\) & \(\Delta=22 \mathrm{mV} ; \pm 0.11 \%\) \\
\hline LOAD REGULATION & \(\mathrm{V}_{\mathrm{IN}}=12.6 \mathrm{~V}, \mathrm{I}_{\mathrm{OUT}}=10 \mathrm{TO} 120 \mathrm{~mA}\) & \(\Delta=3.0 \mathrm{mV} ; \pm 0.015 \%\) \\
\hline OUTPUT RIPPLE & \(\mathrm{V}_{\mathrm{IN}}=12.6 \mathrm{~V}, \mathrm{I}_{\mathrm{OUT}}=120 \mathrm{~mA}\) & \(95 \mathrm{mVP}-\mathrm{P}\) \\
\hline SHORT-CIRCUIT CURRENT & \(\mathrm{V}_{\mathrm{IN}}=12.6 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=0.1 \Omega\) & 1.54 A \\
\hline EFFICIENCY & \(\mathrm{V}_{\mathrm{IN}}=7.5 \mathrm{TO} 14.5 \mathrm{~V}, \mathrm{I}_{\mathrm{OUT}}=120 \mathrm{~mA}\) & \(74 \%\)
\end{tabular}


Fig 1-Using a single inductor, a switching-regulator IC, and a few external components, you can build a step-up/step-down switching regulator.


Fig 2-These three switching circuits show the evolution of Fig 1's circuit. The voltage step-down circuit in a generates a voltage lower than that surplied by the battery. The circuit in \(\boldsymbol{b}\) steps up the voltage provided by a battery. By combining these circuits as shown in cand using a switching-regulator IC to provide the switching function and \(Q_{2}\), you derive the design in Fig 1.
switched off, the inductor discharges into output capacitor \(\mathrm{C}_{0}\) and into the circuit's load, thus forward-biasing diodes \(\mathrm{D}_{1}\) and \(\mathrm{D}_{2}\). During \(\mathrm{T}_{\text {OFF }}\), Fig 2c's circuit is electrically identical to the basic step-up configuration shown in Fig 2b, but the circuit's output voltage results only from the inductor's discharge and is relative to ground instead of to \(\mathrm{V}_{\text {IN }}\). Thus, the circuit can provide an output voltage greater than, less than, or equal to its input voltage.

Fig 2b's voltage step-up circuit lacks current limiting. If you short or overload the circuit's output, you can destroy inductor L or diode \(\mathrm{D}_{2}\); they form a direct path from \(\mathrm{V}_{\text {IN }}\) to \(\mathrm{V}_{\text {out. }}\) However, in Fig 2c's circuit, transistor \(\mathrm{Q}_{2}\) appears in series between \(\mathrm{V}_{\text {IN }}\) and \(\mathrm{V}_{\text {out }}\); when you add a control circuit to \(\mathrm{Q}_{2}\), the transistor can limit the regulator's output current.

EDN
(Ed Note: The author included the design equations (Ref 1) for determining the component values in Fig 1. These equations were omitted for the sake of brevity.)

\section*{Reference}
1. Application Note AN-920A, Motorola Semiconductor Products Inc, 5005 E McDowell Rd, Phoenix, AZ 85008.

\title{
Circuit measures capacitor's resistance
}

\author{
Fred Brown \\ Consulting Engineer, Lake San Marcos, CA
}

To test large numbers of capacitors in production runs, use the circuit in Fig 1a, which lets you use an oscilloscope to determine a capacitor's series resistance at a glance. You drive the circuit with a square wave. If the capacitor has no series resistance, the oscilloscope will display a perfect sawtooth wave. If the capacitor has series resistance, however, the oscilloscope will display a sawtooth wave with a square wave superimposed on it (Fig 1b). The larger the superimposed square wave, the greater the series resistance.
To obtain a quick approximation of the capacitor's series resistance, look at the ratio of \(e_{r}\) to \(e_{C}\). For \(R C \gg T, e_{C}=E T \div R C\), where \(T\) is the square wave's period and \(\mathrm{e}_{\mathrm{C}}\) is the voltage across the capacitor. The ratio of \(e_{r}\) to \(e_{C}\) equals the ratio of the capacitor's series resistance to \(\mathrm{T} \div \mathrm{C}\). For nonpolarized capacitors, you can omit the dc polarizing voltage.

EDN


Fig 1-You can quickly determine a capacitor's series resistance by using the circuit in a. By looking at the relative magnitudes of \(e_{r}\) and \(e_{C}\) (b), you can decide whether the capacitor's resistance lies within acceptable limits.

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\title{
Use 3-state buffer as PLL phase detector
}

\section*{Tom Lange \\ Rochester NY}

Using a 3 -state buffer as a PLL's phase detector, you can keep the PLL synchronized with a pulse traineven when pulses are missing. As Fig 1 shows, the circuit is simple: To implement it, just connect the pulse source to the buffer's output-enable ( \(\overline{\mathrm{OE}}\) ) input, and connect the PLL's VCO output to the buffer's input.

As the timing diagram in Fig 2 illustrates, when an input pulse occurs, the VCO's output is high, the 3 -state buffer's output is high, and capacitor C is being charged. In the pulse's second portion, the VCO's output goes low, the buffer's output goes low, and the capacitor discharges. Once the pulse ends, it places the


Fig 1-Using a 3-state buffer as a PLL's phase detector, you can keep the PLL synchronized with a pulse train-even when pulses are missing.


Fig 2-The output of Fig 1's VCO is high when an input pulse occurs; simultaneously, the 3-state buffer's output is high, and capacitor C charges. Then, the VCO's output goes low, the buffer's output goes low, and the capacitor discharges.


Fig 3-Using a PLD with 3 -state outputs, you can implement Fig 1's circuit. Just connect the pulse source to a gate's output enable and the VCO's output to the gate's input. Because these PLDs have incerted outputs, you'll have to invert some signals.

3 -state buffer in its high-impedance state, thus causing the capacitor's charge to remain at a constant level.

On the other hand, if the falling edge of the VCO's output occurs after the center of the pulse, as it does in Fig 2, the net charge on the capacitor increases. Thus, the VCO's speed input, and hence VCO operating frequency, increases. Conversely, if the falling edge of the VCO's output waveform occurs earlier than the center of the pulse, the VCO's operating frequency decreases.

If the 3 -state buffer receives no input pulse at all, it remains in its high-impedance state; the capacitor charge remains constant; and the VCO frequency stays the same.

You can also implement this circuit using a programmable logic device (PLD) that has 3 -state outputs (eg, a \(16 \mathrm{~L} 8,16 \mathrm{R} 4\), or 20L10 PAL). To use one, connect the pulse source to a gate's output-enable pin and the VCO's output to the gate's input. Because these devices have inverted outputs, you'll have to invert some signals. Fig 3 shows one possible PAL implementation of this scheme, including the PALASM equations. EDN

\section*{Program a sine-wave oscillator's frequency}

\author{
Shvalbe Gershon \\ Holon, Israel
}

Fig 1 depicts a sine-wave oscillator that produces a constant-amplitude, programmable-frequency audio signal. You derive the basic timing frequency from a \(\mu \mathrm{P}\)
and a programmable-interval timer chip.
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\section*{THE FIRST CHOICE} IN TOUCH CONTROL

\section*{DESIGN IDEAS}
programmable-interval timer (PIT). The PIT includes three fully independent, 16 -bit counters, but in this case it only uses one. By programming the 0 counter's initial state and count, you can obtain a desired frequen\(\mathrm{cy}\left(\mathrm{F}_{\mathrm{CLK}}\right)\) at the \(\mathrm{Q}_{0}\) output ( pin 10 ).

The CMOS divider \(\mathrm{IC}_{5}\) divides \(\mathrm{F}_{\text {CLK }}\) by 100 and feeds the result to the input of a fourth-order Butterworth lowpass filter \(\left(\mathrm{IC}_{6}\right)\), which extracts the fundamental sine-wave frequency. (Filter gain is about \(1 / 3\); you may need an output amplifier, as shown within the dotted lines.) The filter realizes a fourth-order response by cascading two second-order functions: The A side has a
gain of 0.3 and a \(Q\) of 0.504 , and the \(B\) side has a gain of 1 and a \(Q\) of 1.306 .

Because the programmed frequency \(\mathrm{F}_{\mathrm{CLK}}\) also serves as the switched-capacitor filter's clock input, the filter's output and cutoff frequencies maintain a fixed relationship. Thus, the output remains at a constant amplitude for any frequency you program over a 5 -decade range.

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Fig 1-This microcomputer-controlled system produces a constant-amplitude, programmable-frequency audio output. Frequency range spans five decades to 20 kHz .


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\section*{DESIGN IDEAS}

\section*{MOSFET circuits yield higher \(\mathrm{BV}_{\mathrm{Dss}}\)}

\author{
Tosh Mizuno \\ Dalmo Victor Co, Belmont, CA
}

You can create a composite MOSFET whose \(\mathrm{BV}_{\text {DSS }}\) is double that of a single MOSFET by combining two devices with a diode and two resistors (Fig 1). Even though you can buy an MTP1N100 MOSFET from Motorola that has a 1000 V minimum \(\mathrm{BV}_{\text {DSS }}\), you may need a higher value for use in a gated image intensifier, for example, or a TWT grid modulator.

With a voltage \(\mathrm{V}_{\mathrm{DD}}\) between the composite drain and source, \(\mathrm{V}_{\mathrm{G}_{2}}=1 / 2 \mathrm{~V}_{\mathrm{DD}}\). Therefore, the voltage at P equals \(1 / 2 V_{D D}\) minus \(Q_{2}\) 's gate-source threshold voltage \(V_{G S(T H)}\), but \(\mathrm{V}_{\mathrm{P}} \approx 1 / 2 \mathrm{~V}_{\mathrm{DD}}\) for large \(\mathrm{V}_{\mathrm{DD}}\). The 12 V zener diode \(\mathrm{D}_{1}\) ensures that the gate of \(\mathrm{Q}_{2}\) is sufficiently positive for \(\mathrm{Q}_{2}\) to remain in saturation under all conditions. Thus, Q2's on-resistance remains low, and the composite MOSFET's electrical characteristics depend only on the characteristics of \(\mathrm{Q}_{1}\).

Typical applications of this composite MOSFET in-


Fig 1-Double a MOSFET's BV Dss by connecting two units in series This composite device behaves as shown in the equivalent diagram.
clude the common-source connection (Fig 2a) and the source-follower connection (Fig 2b). For the source

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Fig 2-You can use the composite MOSFET in conventional high-V DD \(_{\text {applications such as common-source (a) and source-follower (b) }}^{\text {( }}\) ( connections.
follower, you must add capacitor \(\mathrm{C}_{\mathrm{G}}\) to compensate for the gate capacitance of \(Q_{2}\). Because this gate capacitance doesn't generally equal the sum of \(\mathrm{C}_{\text {rss }}\) and \(\mathrm{C}_{\text {iss }}\), you must experiment to determine the required compensation value. (For example, you might want to select
a capacitor that achieves less than \(-90^{\circ}\) of phase shift at the highest signal frequency of interest.)
\(\qquad\)

\section*{Transistor array squares control current}

\section*{Burkhard Braach}

Wandel \& Goltermann, Eningen, West Germany

A simple 5-transistor array and a resistor (Fig 1) generate a square-law relationship between \(\mathrm{I}_{\mathrm{iN}}\) and \(\mathrm{I}_{\text {out. }}\) The circuit is useful in PLL frequency synthesizers and other closed-loop systems requiring square-law amplification in the feedback path.

Assume that the transistor base currents are negligible and that \(Q_{1}-Q_{2}\) and \(Q_{4}-Q_{5}\) have negligible baseemitter offset voltages. These transistor pairs then form ideal current mirrors, and their collector currents equal the input current:
\[
\begin{equation*}
\mathrm{I}_{1}=\mathrm{I}_{2}=\mathrm{I}_{4}=\mathrm{I}_{5}=\mathrm{I}_{\mathrm{IN}} . \tag{1}
\end{equation*}
\]

The \(Q_{1}\) and \(Q_{3}\) collector currents are
\[
I_{1}=I_{s} e^{\frac{q V_{B E 1}}{k T}} \text { and } I_{3}=I_{s} e^{\frac{q V_{B E}}{K T}} \text {, }
\]
respectively, and their ratio is


Fig 1-This transistor-array circuit performs square-law amplification of \(I_{I N}\).


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\section*{Thomas\&Betts}

\section*{DESIGN IDEAS}
\[
\begin{equation*}
\frac{I_{3}}{I_{1}}=e^{\frac{q\left(V_{B E 3}-V_{B E 1}\right)}{k T}} \tag{2}
\end{equation*}
\]
where \(I_{S}=\) saturation current, \(q=\) electron charge, \(\mathrm{K}=\) Boltzmann's constant, \(\mathrm{T}=\) absolute temperature, and \(V_{\text {BE1 }}\) and \(V_{\text {BE3 }}\) are the base-emitter voltages for transistors \(Q_{1}\) and \(Q_{3}\).

Because \(\mathrm{V}_{\mathrm{BE} 3}=\mathrm{V}_{\mathrm{BE} 1}+\mathrm{R}_{1}\left(\mathrm{I}_{1}+\mathrm{I}_{2}\right)\), you can write \(\mathbf{E q} 2\) as
\[
I_{3}=I_{1} e^{\frac{\left(4 I_{1}+I_{2}\right) \mathrm{R}_{1}}{k T}}
\]

Substituting \(\mathrm{I}_{\text {IN }}\) from Eq 1 yields
\[
\mathrm{I}_{3}=\mathrm{I}_{\mathrm{IN}} \mathrm{e}^{\frac{2 q 1_{1 N} \mathrm{R}_{1}}{\mathrm{KT}}}
\]

Thus, transistor \(Q_{3}\) provides an exponential function that you can expand as a power series in the form of \(\mathrm{e}^{\mathrm{x}}=1+\mathrm{x} / 1!+\mathrm{x}^{2} / 2!+\mathrm{x}^{3} / 3!+\ldots\) to yield
\[
\begin{gather*}
\mathrm{I}_{3:}=\mathrm{I}_{\mathrm{IN}}\left[1+\frac{2 \mathrm{qI}_{\mathrm{IN}} \mathrm{R}_{1}}{\mathrm{KT}}+\left(\frac{2 \mathrm{qI}_{\mathrm{IN}} \mathrm{R}_{1}}{\mathrm{KT}}\right)^{2} / 2+\right. \\
\left.\left(\frac{2 \mathrm{q}_{1 \mathrm{~N}} \mathrm{R}_{1}}{\mathrm{KT}}\right)^{3} / 6+\cdots\right] . \tag{3}
\end{gather*}
\]

The output current is \(\mathrm{I}_{\text {out }}=\mathrm{I}_{3}-\mathrm{I}_{5}=\mathrm{I}_{3}-\mathrm{I}_{\text {IN }}\). Substituting for \(\mathrm{I}_{3}(\mathbf{E q} 3)\) eliminates the linear term, so the series begins with the quadratic term:
\(\mathrm{I}_{\text {OUT }}=\left(\frac{2 \mathrm{qR}}{\mathrm{KT}}\right) \mathrm{I}_{\mathrm{IN}}^{2}+\left(\frac{2 \mathrm{qR}_{1}}{\mathrm{KT}}\right)^{2} \mathrm{I}_{\mathrm{IN}}^{3} / 2+\left(\frac{2 q \mathrm{R}_{1}}{\mathrm{KT}}\right)^{3} \mathrm{I}_{1 \mathrm{~N}}^{4} / 6+\ldots\).
In short, the circuit produces a useful squaring characteristic for low \(2 \mathrm{qR}_{1} / \mathrm{KT}\) ratios and low input currents. Fig 2 shows the measured and calculated results for a CA3096 transistor array and a \(27 \Omega\) resistor. You can extend the \(20: 1\) output-current range by using transistor pairs with tighter \(\mathrm{V}_{\mathrm{BE}}\) matching and


Fig 2-These curves illustrate the performance of Fig I's circuit. The curve labeled A represents the ideal squaring function, the curve labeled B shows the calculated function, and the two curves labeled C form an envelope for the results obtained using five different CA3096 arrays.
higher betas. If desired, you can reverse the outputcurrent polarity by inserting a resistor with a value \(2 \mathrm{R}_{1}\) in the emitter of \(Q_{4}\) (remove the \(Q_{1}-Q_{2}\) emitter resistor in this case).

EDN

\section*{Digital power controller handles 1 kW}

\author{
John A Haase \\ Colorado State University, Fort Collins, CO
}

With this power controller (Fig 1) you can vary the power to a load according to a 2 -decade BCD word, obtained either manually or from a computer. For an
open-loop system you can deliver at least 1 kW to a resistive load, in \(1 \%\) increments ( 0 to \(99 \%\) ). Or, you can combine the controller with a sensor and heater, for example, to regulate temperature. The controller delivers an integral number of line cycles every 1.67 sec ; switching occurs near the zero-voltage crossing so the

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\section*{DESIGN IDEAS}
load current has no dc component.
AC line voltage is regulated to about 7 V by \(\mathrm{IC}_{1}\) for use as a supply voltage \(\left(\mathrm{V}_{\mathrm{S}}\right)\). The zero-crossing detector \(\mathrm{IC}_{2}\) also takes its supply voltage directly from the line
via \(R_{4}\) and \(R_{5}\). In addition, the line drives the 4 N 26 optocoupler, producing a \(60-\mathrm{Hz}\) clock signal to the rate multipliers \(\mathrm{IC}_{3}\) and \(\mathrm{IC}_{4}\).

The cascaded rate multipliers accept a 2-decade BCD


Fig 1-A low-power circuit provides digital control of 1 kW , or more, in \(1 \%\) steps. A zero-crossing detector ensures that only an integral number of full line cycles reach the load during any \(1.67-\mathrm{sec}\) control period.

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word and provide an output ( \(\operatorname{pin} 6, \mathrm{IC}_{3}\) ) of one-tenth the word value times 60 Hz . (In this system, you enter the control word manually using the quad switches \(S_{1}\) and


Fig 2-This circuit is timed to fire the SCRs \(\mathbf{Q}_{2}\) and \(\mathbf{Q}_{4}\) (Fig 1) so that only one full line cycle (highlighted area) is passed to the load for each pulse from \(I C_{5 B}\), pin 4 .
\(S_{2}\).) The shunting effect of \(R_{5}\) ensures that the LED \(\mathrm{D}_{4}\) 's on-period is well within a positive half-cycle of the line voltage (Fig 2). Consequently, the count-advance pulses from \(\mathrm{IC}_{5 A}\) to the rate multipliers occur once per line cycle and in advance of the zero crossing.

Then, for each output pulse via the Schmitt trigger \(\mathrm{IC}_{5 \mathrm{~B}}\), the zero-crossing detector \(\mathrm{IC}_{2}\) emits two pulses at pin 4, which are optically coupled to the gates of SCRs \(\mathrm{Q}_{2}\) and \(\mathrm{Q}_{4}\). As a result, the appropriate SCR turns on during two consecutive half-cycles of the line voltage (highlighted areas in Fig 2).

This circuit can replace an autotransformer in many applications, offering less weight, less volume, and better control resolution. The circuit also has the following characteristics:
- \(\mathrm{S}_{3}\) is an on-off switch-no power to the load in the idle position.
- Brightness of the neon lamp \(\mathrm{NE}_{1}\) gives an approximate indication of load power.
- Power switching is optically isolated from the control circuits.
- No partial- or half-cycles of line voltage are delivered to the load.
- Power cycles are uniformly spaced for power-of-2 control words. Note, however, that other control words may cause a noticeable variation in power over the 1.67 -sec control interval.

EDN

\title{
Circuit checks remaining battery capacity
}

\author{
Rolf Zinniker \\ Swiss Federal Institute of Technology, Zurich, Switzerland
}

The test circuit in Fig 1 gives an indication of the capacity remaining in a battery. By noting the time (in seconds) that the LED remains on after you depress the test switch \(\mathrm{S}_{1}\), you can read the remaining battery life (in hours) from the curve in Fig 2a. The circuit has proven reliable in testing NiCd-, carbon-, and alkalinetype batteries, all of which have a discharge-vs-time characteristic similar to the NiCd curve shown in Fig 2b.

Closing \(\mathrm{S}_{1}\) activates the circuit by applying voltage from the battery under test. Voltage \(V_{1}\) jumps to a value \(\mathrm{V}_{0}=\mathrm{V}_{\mathrm{R}} \mathrm{R}_{2} /\left(\mathrm{R}_{2}+\mathrm{R}_{3}\right.\) ) when the switch closes ( Fig 3) and then increases with a time constant \(T=\mathrm{C}_{1}\left(\mathrm{R}_{2}+\mathrm{R}_{3}\right)\). The divider \(R_{4} / R_{5}\) fixes \(V_{2}\). The reference circuit \(\mathrm{IC}_{1}\) sets \(\mathrm{V}_{\mathrm{R}}\) to approximately 2.5 V .

The op amp's output remains high (LED on) until \(\mathrm{V}_{1}\) rises to the level of \(V_{2}\), when the LED turns off. You calculate the on-time \(t_{O N}\) as follows:
\[
\mathrm{t}_{\mathrm{ON}}=\mathrm{T} \cdot \ln \left(\frac{\mathrm{~V}_{\mathrm{R}}-\mathrm{V}_{0}}{\mathrm{~V}_{\mathrm{R}}-\mathrm{V}_{2}}\right) .
\]

\section*{DESIGN IDEAS}


Fig 1-This test circuit checks the remaining charge in a battery; the charge is proportional to the time the LED remains on after you depress \(S_{I}\).

The on-time is zero when \(V_{2}<V_{0}\) and infinite when \(V_{2}>V_{R}\). By substituting \(t_{o N}\) for \(t\) and \(V_{2}\) for \(V_{1}\) in Fig 3, you obtain a curve for \(t_{\text {on }}\) as a function of \(V_{2}\).

You must choose voltages \(\mathrm{V}_{\mathrm{S}}\) and \(\mathrm{V}_{\mathrm{E}}\) (Fig 2b) that correspond to the desired range for \(\mathrm{t}_{\mathrm{ON}}\). Then, set the resistor values as follows:
\[
\frac{\mathrm{R}_{4}}{R_{5}}=\frac{\mathrm{V}_{\mathrm{R}}}{\mathrm{~V}_{\mathrm{S}}-\mathrm{V}_{\mathrm{R}}} ; \frac{\mathrm{R}_{3}}{\mathrm{R}_{2}}=\frac{\mathrm{V}_{\mathrm{E}}}{\mathrm{~V}_{\mathrm{S}}-\mathrm{V}_{\mathrm{E}}} .
\]

You can make \(R_{5}\) a variable resistor for convenience in calibrating the circuit. Because the curves of Fig 2b and Fig 3 have a similar rate of change over the range of interest, the resulting relation (Fig 2a) between \(\mathrm{t}_{\mathrm{ON}}\) and remaining battery lifetime is quite linear. To use the circuit, simply depress \(S_{1}\) and hold it. If the LED remains on for several seconds, the battery contains more than \(70 \%\) of its rated charge. If the LED remains on for only 0.5 sec , for example, the battery is good for approximately four hours while loaded as in Fig 2b, that is, with \(100 \Omega\).

EDN


Fig 2-The calculated on-time in seconds for the LED in Fig 1 corresponds to the battery's remaining life in hours, based on a \(100 \Omega\) load ( \(\boldsymbol{a}\) ). The curve in \(\boldsymbol{b}\) shows the measured rate of discharge for a NiCd battery (four Varta RS600 cells loaded with 100 \(\Omega\) ).


Fig 3-This curve shows the variation of \(V_{l}\) following closure of \(S_{l}\) and caused by the charging of capacitor \(C_{1}\).

\section*{Simple hardware drives multiple displays}

\section*{R Jayapal \\ Bharat Heavy Electricals Ltd, Trichirappalli, India}

In Fig 1, one 8-bit port drives five displays, each of which can include any reasonable number of digits.

Regardless of the number of digits, each display (Fig 2) requires only two signals, Clock and Clear. This approach includes less hardware and fewer interconnections than are found in the more conventional multiplexing and serial-communications methods used in microprocessor-based systems.

\section*{DESIGN IDEAS}


Fig 1-One 8-bit port drives five displays, each with an arbitrary number of digits. Each display requires only two signals, Clock and Clear.


Fig 2-A typical display in Fig 1 includes a digit display, r-segment decoder, and BCD counter for each digit.

Port bits \(P_{0}, P_{1}\), and \(P_{2}\) select the display to be updated: 000 for Display 1, 001 for Display 2, etc. These bits should be 111 when the displays aren't being updated.

To load Display 1, first clear it by issuing a softwaregenerated pulse on \(\mathrm{P}_{3}\). Next, initialize the 8253 pro-
grammable-interval timer for BCD counting and load it with the decimal number to be displayed. After loading \(\mathrm{P}_{0}-\mathrm{P}_{2}\) with 000 and starting the timer, the CPU can continue with other processing. When the 8253 's count reaches the desired value, the 8253 issues an interrupt to the CPU.

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\section*{DESIGN IDEAS}

The interrupt service routine simply loads \(\mathrm{P}_{0}-\mathrm{P}_{2}\) with \(000, \mathrm{P}_{3}-\mathrm{P}_{7}\) with 00000 , and returns. Thus, Display 1 reads the desired value, and you use a similar procedure to update the other displays. Table 1 gives the bit
patterns required at the 8 -bit port for handling each of the five displays.


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\hline Die Size [K sq mil] & 106 & 160 & 160 \\
\hline Transistors [K] & 93 & 275 & 270 \\
\hline Clock Speed (MHz] & 10 & 16 & 20 \\
\hline Max Power (Watts] & \(3.5-4\) & 2.5 & \(2-2.5\) \\
\hline Package & 68 PLCC & 132 PGA & 124 PGA \\
\hline Multiplexed Add/Data & YES & NO & NO \\
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\hline \begin{tabular}{l} 
Max Memory Access \\
(No Wait States]
\end{tabular} & 200 NS & 67 NS & 45 NS \\
\hline \begin{tabular}{l} 
Prog. Wait States
\end{tabular} & YES & NO & NO \\
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