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ELECTRONIC TECHNOLOGY FOR ENGINEERS AND ENGINEERING MANAGERS


1



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The VME Consortium needed an economical, yet highly functional VME bus interface chip, to minimize design time . . . and to help raise the VME standard to higher levels.
"We looked at the leading suppliers," said Joe Ramunni, consortium chairman (and president of Mizar), "and VTC came out on top. Their CMOS standard-cell ASIC approach gave us the high drive capability we needed, optimized for bus interfacing. And, it proved much more cost-effective, with higher performance, than gate array technology.'
The VME Consortium is made up of such firms as Plessey Microsystems, Omnibyte Corporation, Mizar Inc., Ironics Inc., Heurikon Corporation, Matrix Corporation, and Clearpoint Inc., among others. What did they look for in a supplier?
"We needed a credible business partner," said Ramunni, "with a proven track record, who could provide a turnkey package . . . both design and fab. A supplier that could produce in quantity, and provide technical support to the market at large.
"We also needed a firm with an international marketing structure, because we expect this chip to be the de facto standard worldwide.
"But, we needed people we could work with, too. VTC had the right 'comfort factor'."

Jack Regula, consortium technical director (and VP-R\&D, Ironics) added: "Our requirements for high speed, high gate-count, low power consumption, and VME bus drive capability were all met well with VTC's 1-micron CMOS standard cell library. And we were extremely impressed with VTC's facilities, its people, and its customer list."

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## ELECTRONIC TECHNOLOGY FOR ENGINEERS AND ENGINEERING MANAGERS



On the cover: Pressure sensors and transducers are achieving high marks for performance, thanks in part to device manufacturers using CAD/CAE techniques. See pg 106. (Photo courtesy Nova Sensors)

## SPECIAL REPORT

Pressure sensors and transducers
By employing CAD/CAE and new fabrication technologies, pressure sensor and transducer manufacturers have been able to improve product performance without exacting any price penalties. In the siliconbased sensor area, in fact, prices are actually going down as performance figures improve-a combination that opens up a host of new applications.-Tom Ormond, Senior Editor

## DESIGN FEATURES

## CAE software uses algorithms instead of schematics

Drawing logic schematics isn't the only way to create ASIC designs. You can also use algorithms to automatically generate the net list for your hardware design-and you can do so in about one-tenth of the time that it often takes to complete a schematic.-Jay R Southard, Algorithmic Systems Corp

## Single VLSI chip <br> helps you implement EPA factory networks

The Enhanced Performance Architecture (EPA), a subset of the MAP specification, allows networks to provide fast, noise-immune communication over short distances. By applying a single VLSI chip that implements most of the EPA functions in silicon, you can speed your network's response time without a huge software-development effort.-Rhonda Alexis Divvin and Anne-Marie Larkin, Motorola Inc

## Static system design 159 exploits low-power CMOS features

Static CMOS architectures let you customize a design for optimum performance and minimum power dissipation. Capitalizing on the characteristics of CMOS means addressing clock-frequency control before starting the initial design process.-Walter J Niewierski, Harris Semiconductor
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Design, Electronics, Electronic Products, Computer Design, and ESD.
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Third-party graphics-software packages for IBM PCs and compatible computers can ease your software-writing chores (pg 55).

EDN magazine now offers Express Request, a convenient way to retrieve product information by phone. See the Reader Service Card in the front for details on how to use this free service.

## TECHNOLOGY UPDATE

## Third-party PC graphics packages <br> 55 streamline your programming tasks

Attention-grabbing, colorful graphics can dramatically enhance the software you write, regardless of whether you're developing it for your company's internal use or for customers.-Margery S Conner, Regional Editor

## Limited selection of monolithic PLL ICs 65 suffices for current applications

Monolithic phase-locked-loop (PLL) circuits provide an elegant way to achieve frequency demodulation or digital selection of frequency.
-Tarlton Fleming, Associate Editor

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## EDITORIAL

Establishing a tariff that funds US R\&D efforts might be a suitable alternative to mandating artificially high semiconductor prices.

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## How to crack 386 protected mode.



Unlock selectors and descriptor tables. Break open task state segments (TSS) and call gates. Microtek's In-Circuit-Emulator (MICE) cracks 80386 protected mode with real-time, non-intrusive emulation to 20 MHz .

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Our Software Analysis Workstation (SAW) delivers hardware-based, real-time software analysis in a source code environment. For source code development, our Microtec ${ }^{\circledR}$ Research products provide you
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And MicroCASE ${ }^{\text {TTM }}$ backs all of its products with solid applications support, both at the local and factory level.

Microtek enjoys a long track record of being first to market with quality support for many major microprocessors. We were the first alternative for 80386 emulation. We're first to provide non-proprietary support for 80386 protected mode. And you can count on the same support in the future.

For more information and product literature on our full line of MICE, call us at (408) 253-5933.

Crack 80386 protected mode, and open the door to exciting design possibilities - with the best 80386 emulation system now available. From MicroCASE.

# NEWS BREAKS 

## 8-BIT HALF-FLASH ADC INCLUDES TRUE SAMPLE/HOLD AMPLIFIER

The LTCl099 from Linear Technology Corp (Milpitas, CA, (800) 637-5545) is a halfflash 8-bit A/D converter that converts in $2.5 \mu \mathrm{sec}$ typ. A true sample/hold amplifier is included on the chip; it allows the ADC to convert 5V p-p signals at as much as 167 kHz , or signals with slew rates as high as 20V/usec. The LTCl099 is pin compatible with the AD7820 and the ADC0820; these older devices can complete the A/D conversion in $1.5 \mu \mathrm{sec}$, but have pseudo-S/H amplifiers that can accept signals with a maximum slew rate of $0.1 \mathrm{~V} / \mu \mathrm{sec}$ or a bandwidth limit of 7 kHz . (To acquire signals with higher frequencies, these older devices require an external $\mathrm{S} / \mathrm{H}$ amplifier, which is usually expensive.) The LTC1099 costs \$8 (100).-David Shear

## EVALUATION MODULE PROVIDES INTERFACE TO CONSUMER FLECTRONIC BUS

The SEM-200 Smarthome evaluation module from CyberLynx (Boulder, CO, (303) 444-7733) provides easy access to the Consumer Electronic Bus (CEBus) currently being developed as a standard under the sponsorship of the EIA. The CEBus is a low-cost, multimedia LAN designed to interconnect appliances, audiovisual equipment, and other household devices. The $\$ 450$ SEM-200 acts as a network node and LAN analyzer that monitors network traffic. You can connect a PC or a dumb terminal to the unit's RS-232C interface to observe activity on the CEBus network. The module provides digital inputs and relay outputs that allow you to add CEBus capability to a wide range of product prototypes. CyberLynx and Texas Instruments (Dallas, TX) are jointly developing a line of interface ICs for the CEBus.-Steven H Leibson

## PC/AT GPIB INTERFACE OFFERS 1-MHz DMA TRANSFERS

If you have an IBM PC/AT or compatible computer, you can use the AT-GPIB expansion card from National Instruments (Austin, TX, (515) 250-9119) to speed your IEEE-488 communications. This $\$ 495$ interface lets your 80386- or 80286 -based computer communicate with as many as 13 instruments at data-transfer speeds reaching 1 M byte/sec. Bidirectional FIFO buffers let the device read at 1 M byte/sec, write at 700 k bytes $/ \mathrm{sec}$, and execute GPIB commands at 320 k bytes $/ \mathrm{sec}$. The AT-GPIB comes with the NI-488 software package, which includes BASICA, Quick BASIC, and a binary MS-DOS device driver with more than 30 IEEE-488 functions. The board can use one of 11 selectable lines to interrupt the host PC.-J D Mosley

## ANALOG MEASUREMENT MODULE OFFERS 50-kHz THROUGHPUT

For applications requiring 16 -bit A/D conversions, you can select the AMM2 Master Analog Measurement Module from Keithley Instruments (Cleveland, OH, (216) $248-0400$ ) for $50-\mathrm{kHz}$ data transfer. The $\$ 1155$ module offers 16 single-ended or eight differential analog inputs, a crystal-controlled oscillator for jitter-free FFT analysis, and an external trigger. The AMM2 provides programmable operating parameters that let you control gain and configure inputs. The unit automatically performs internal calibration when used with the manufacturer's Soft500 and Quick500 data-acquisition software.-J D Mosley

## MICROCOM NETWORKING PROTOCOL INCLUDED ON MODEM CHIP SET

Sierra Semiconductor (San Jose, CA, (408) 263-9300) is changing the program of its SCllOll modem processor: The new version, the SCllO13, will implement the Microcom Networking Protocol (MNP). This protocol has been used for many years to provide error-free communications in the banking industry. The MNP is a data-
communications protocol for full-duplex, error-free communication over ordinary voice-grade telephone lines. Software-communication packages use similar techniques, but these tend to be machine dependent. MNP exists in firmware within the modem, so any computer or terminal can be used. The protocol includes a data-compression algorithm that provides an effective data rate as high as 2900 bps with a 2400 -bps modem. The SCllO13 modem processor and the SCllOO6 modem analog peripheral together cost $\$ 33.50(25,000)$. You'll also need some extra components such as a RAM, a USART, and some logic circuits; you can obtain all the necessary parts for \$7.50. In effect, you'll be able to build a modem that implements the MNP for less than \$ll over the cost of producing a standard 2400-bps modem.-David Shear

## GRAPHICS CHIP AND MATH COPROCESSOR INCREASE DRAWING SPEED

The TMS34020-the latest version of the TMS34010 graphics processor from Texas Instruments (Dallas, TX, (800) 232-3200 ext 700)—can make use of the company's new TMS34082 floating-point coprocessor to accelerate graphics functions such as coordinate transformations. In addition, the 34020 has $25 \%$ more microcoded graphics primitives, so it can render the primitives faster than can software-based primitives. All 34010 object code is upwardly compatible with the 34020 . The 34010 has a 32-bit internal and a 16 -bit external data bus; the 34020 has an internal and an external 32 -bit data bus. The 34020 runs at from three to 20 times the speed of the 34010. Samples of both the 34020 graphics processor and the 34082 math coprocessor will be available in the fourth quarter of 1988 at $\$ 500$ each; production quantities will be available in the first quarter of 1989. You can expect to see the price of the 34020 drop by an order of magnitude within the next 18 months.-Margery S Conner

## INTERFACE UNIT CONVERTS SCSI TO ETHERNET

The Nodem allows you to connect your personal computer to Ethernet via your computer's SCSI port. The unit is completely external; it doesn't use your computer's expansion slots, and it has its own wall-mounted power supply. You adapt the unit to your system configuration by using software supplied with the unit. The Nodem works with IBM PCs and compatibles, the IBM PS/2, and the Apple Macintosh. It's transparent to AppleTalk.

The units are made by Adaptek (Milpitas, CA, (408) 945-8600), but will be marketed primarily by LAN-system suppliers and value-added resellers. The standard Ethernet version will cost approximately $\$ 545$; the Cheapernet and twisted-pair versions will sell for about $\$ 595$. The units will be available with software for Macintosh systems in September; software for IBM systems will be available in early 1989.-Richard A Quinnell

## MULTITASKING, RESIDENT BASIC COMPILERS TARGET EMBEDDED SYSTEMS

The MTBasic compilers from Softaid Inc (Columbia, MD, (301) 964-8455) provide multitasking Basic and a kernel operating system for diskless, embedded-processor systems employing the Zilog Z80 or Z280, the Hitachi 64180, or the Intel $8088 \mu$ Ps. The compilers reside in a ROM on the target system and offer a conventional programming environment similar to that provided by compilers running on personal computers. The vendor supplies source code for each compiler, allowing you to tailor features of the language for your specific application. Each compiler costs $\$ 6500$. -Steven $H$ Leibson


## THE COMPETITION IS STILL TALKING ABOUT THEIR 10-BIT FLASH ADC

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## NEWS BREAKS: international

## ADA COMPILERS MAKE TRANSPUTERS SUITABLE FOR MILITARY SYSTEMS

If you're using Inmos Transputers in embedded systems for military contracts, you'll probably be required to write the system software in the Ada programming language. Two compilers being developed by Alsys SA (La Celle Saint-Cloud, France, TLX 697569; in the US: Waltham, MA, (617) 890-0030) can help. The compilers generate code for any of the Inmos 16- and 32-bit integer Transputers or the IMS-T800 floating-point Transputer. One is a cross-compiler that runs on DEC VAX computers under VMS; the other is a host compiler that runs under MS-DOS on IBM PCs and compatibles to which you've added a T800-based add-in card. For a medium-sized VAX, the cross-compiler is expected to sell for around Fr 250,000 . The PC-based host compiler will probably cost around Fr 50,000.

Although these compilers won't be available until the middle of 1989, you can start software development now by using one of the company's currently available VAXbased validated Ada compilers. Because all the company's Ada compilers share a common-root front-end and intermediate code, the software you write with one of the available compilers will recompile to Transputer code when run through the Transputer's Ada compiler. To meet military-hardware requirements, Inmos (Bristol, UK, TLX 444723; in̨ the US: Colorado Springs, C0, (719) 630-4000) plans to introduce a version of the T800 Transputer that's qualified to MIL-STD-883C during the first quarter of 1989. The company will follow that product with military versions of the 16and 32 -bit integer Transputers later the same year.-Peter Harold

## SIFMENS AND AMD AGREE TO DEVELOP PIN-COMPATIBLE ISDN CHIP SET

Siemens Components (Munich, Germany, TLX 521000; in US: Santa Clara, CA, (408) 980-4500) and Advanced Micro Devices (Sunnyvale, CA, (408) 732-2400) have agreed to jointly develop and supply ICs for the telecommunications and data-communications industry, with emphasis on an ISDN chip set. You will be able to obtain pin-compatible devices from both companies, selecting from a set of 15 devices ( 13 Siemens chips and two AMD chips) that are based on a common architecture. PSB2llO ISDN terminal adapter circuit, PEB2085 ISDN subscriber-access controller, and the Am'79C401 integrated data-protocol controller. The companies will begin by offering each other's parts; then they'll exchange manufacturing technology to build a true alternate-source capability for the entire set of devices.-Richard A Quinnell

## TOSHIBA FORMS TUNNEL JUNCTION FROM SUPERCONDUCTOR MATERIAL

According to reports in the Japanese press, Toshiba Corp has created a thin-film technology for high-temperature superconducting material that may make commercial superconducting computer chips possible. The company has successfully combined lead and yttrium-based superconductor material, forming a tunnel junction. The firm used this procedure to make a prototype, and confirmed that the device's voltage could be used in on/off switching functions. These results reportedly represent the first time anyone has confirmed such a function in a superconducting device.-Joanne Clay

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|  | $(\mathrm{MHz})$ |
| :---: | :--- |
| MODEL | $\mathrm{f}_{\mathrm{L}}$ to $\mathrm{f}_{\mathrm{u}}$ |
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| MAN-2 | $0.5-1000$ |
| MAN-1LN | $0.5-500$ |
| -MAN-1HLN | $10-500$ |

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$1,2,3,4,5,6,7,8,9,10,12,15,20,30$, and 40
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| *Freq. <br> $(\mathrm{MHz})$ | Atten. Tol. <br> (Typ.) | Atten. Change, (Typ.) <br> over Freq. Range |  | VSWR (Max.) |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
|  |  | DC-1000 | $1000-1500$ | DC-1000 MHz | $1000-1500 \mathrm{MHz}$ |
| $\mathrm{DC-1500} \mathrm{MHz}$ | $\pm 0.3$ | 0.6 | 0.8 | 1.3 | 1.5 |

*DC-1000 MHz (all 75 ohm or 30 dB models) $\mathrm{DC}-500 \mathrm{MHz}$ (all 40 dB models)
Model Availability
SAT (SMA) CAT (BNC) NAT (N)
Model no. = a series suffix and dash number of attenuation.
Example: CAT-3 is CAT series, 3 dB attenuation.
Precision 50 ohm terminations only $\$ 7.95$ (1-24) DC to $2 \mathrm{GHz}, 0.25 \mathrm{~W}$ power rating, VSWR less than 1.1 BNC (model BTRM-50), TNC (consult factory) SMA (model STRM-50), N (model NTRM-50)

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"building blocks" of system silicon. And you won't comply with customer and industry requirements if you don't do complete "system" functional testing. With conventional test systems it means two of everything. Two testers, two test programs, two insertions, two data bases. And more than twice the time to get to market.

The A500 allows you to do it all with one system. So there's only one system to program. One insertion to make for both component and functional testing. And only one data base to work with. Which means significantly less time to market.

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The heart of the A500 is Teradyne's unique Vector Bus II architecture. It integrates analog and digital VLSI test capability at the system level. Which means you won't have to build special applications hardware for every new device you design. Vector Bus II eliminates that costly custom-work bottleneck


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Differential Inputs
Separate Trigger and Reset Inputs
310 mW Power Dissipation
APPLICATIONS
ATE
Pulse Deskewing
Phase Correction
High Stability Timing Circuits
Waveform Generators

## PRODUCT DESCRIPTION

The AD9500 is a digitally programmable delay generator which includes virtually all the circuits needed for generating time delays for digital pulses. It provides 256 programmed delays in a user-specified full scale range which can be varied from more than $100 \mu$ s to as little as 2.5 ns . On the latter scale, it can resolve increments as small as 10 picoseconds.

Its output is delayed from the input by a time which is directly proportional to the 8 -bit digital input code. Using groups of AD9500 devices is an excellent way to adjust signal timing skews and delays in various system applications

Differential TRIGGER and RESET control inputs are designed primarily for ECI. signal levels but will also function with analog and TTL levels. An on-board ECL reference midpoint allows both to be driven by either single-ended or differential ECL circuits.
Two temperature ranges and three packages are available. For industrial temperatures of $-25^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$, order the AD9500BP or the AD9500BQ. The "P" suffix indicates a 28 -pin PLCC; the "Q" designates a 24 -pin ceramic "skinny" DIP with 0.3 " package width. For extended temperatures of $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$, order the AD9500TE ( 28 -pin LCC) or AD9500TQ (24-pin skinny" DIP).


AD9500 Functional Block Diagram

## PRODUCT HIGHLIGHTS

1. The AD9500 delay generator is an extremely versatile but remarkable, casy-to-use timing device. A few basic configurations can be expanded and extended into multiple applications
2. Accurate control of pulse timing is critical for all digital electronic systems, and many systems require that delays be controlled digitally. Until now, the majority of systems using that technique used discrete LSI devices which may consume up to one watt or more of power. The AD9500 performs the same function with 300 mW of power.
3. Like a high speed counter, the AD9500 can be programmed with a binary digital word. This makes the unit a variable delay device, in effect, a digital-to-time converter (DTC The digital input word scales the time reference of the AD9500 in essentially the same way a digital word scales voltage or current references in a DAC


## SIGNALS \& NOISE

## EE seeks source of basic, general-purpose algorithms

The company I work for designs medical test equipment, and most of our work involves 8-bit microprocessors in embedded-control applications. We use IBM PCs and compatibles as development tools, and there is plenty of software for that environment. However, there is no source that I can find for writing the fundamental algorithms that we take for granted in a high-level language. What I'm very anxious to find are sources for algorithms in general terms that can solve the problems we often encounter.
For instance, floating-point routines, sine, cosine, tangent, square root, graphics, and ASCII-to-binary routines are becoming much more important to our needs. Where can the fundamental algorithms be found to implement these functions?

The ideal solution for me would be
a reference book (or books) that described at a flowchart level how to code these algorithms independently of a specific processor. This method allows the designer to pick the best hardware solution without depending on the availability of some special-purpose software that is written for only one $\mu \mathrm{P}$.
In the accompanying table, I've listed some of the particular needs that I've run across; it's a wish list of mysteries I'd love to see revealed.
If you could suggest some places
to look or publications that answer some of these questions, I'd appreciate it very much. I think this need is shared by a great many engineers. Perhaps someday your magazine could publish the solutions to these algorithms in generic format.
Alan Clark
Dynatech Nevada
Carson City, NV
(Ed Note: We'd like to enlist the readers' help on this one. If you know of any sources for the algo-

| FUNCTION |  |
| :--- | :--- |
| FLOATING POINT | ADESCRIPTION SUBTRACT, MULTIPLY, DIVIDE |
| TRIGONOMETRY | SINE, COSINE, TANGENT |
| SQUARE ROOT | INTEGER AND FLOATING-POINT SOLUTIONS |
| INTEGER TO FLOATING | CONVERSION |
| FLOATING TO INTEGER | CONVERSION |
| GRAPHICS | HOW TO MODEL A BIT-MAPPED PLANE; ROUTINES FOR <br> DRAWING A LINE BETWEEN TWO POINTS; CIRCLES AND <br> OTHER SHAPES |
| ASCII TO INTEGER | CONVERT ASCII STRING TO SIGNED/UNSIGNED INTEGER |
| INTEGER TO ASCII | CONVERT INTEGER TO SIGNED/UNSIGNED ASCII STRING |

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## SIGNALS \& NOISE

rithms Alan is seeking, please send the information to the Signals \& Noise Editor, EDN, 275 Washington St, Newton, MA 02158. We'll publish the information in this column in a future issue.)

## Fault-critical applications need bugless processors

Steven H Leibson's article "Faulttolerant design spans terrestrial and space applications" (EDN, April 28, 1988, pg 180)-while naturally, in the space available, is able to offer only superficial treatment of this important subject-is particularly timely with regard to Intel's announcement of its 80376 microprocessor for embedded applications.

Embedded processors, especially high-end 32 -bit machines, are primarily intended for fault-critical applications such as those outlined in Steve's article. To use a processor in
a fault-critical application, one must have a high level of confidence in that processor's integrity-its freedom from undocumented faults that can lead to abnormal behavior. The 80386, upon which the 80376 is based, however, is becoming notorious for its number of errors, and Intel's Stepping Information (essentially a bug list showing possible ways to work around those bugs) is now at Revision D, and possibly even at Revision E. Is this an architecture one can have confidence in for a critical application?

That question leads me to a crucial point that Steve doesn't mention in his article: the problems of com-mon-mode faults and the question of software in redundant systems.

The inexperienced may be tempted to think that redundancy is a potential panacea, but that's not so. Suppose that you construct a dual-redundant system, using the same processor and the same soft-
ware in each channel. If that processor contains a fault that causes the system to crash for some combination of data and instructions, both channels will fail simultaneously.

What is needed is a processor that has been proven to contain no bugs, and that behaves in a predictable manner for such cases as illegal op codes. Such a processor would not, of course, protect against erroneous (but legal) software operations, but it would remove at least one level of uncertainty from the design of fault-critical systems. Such a processor is already commercially available: It has a RISC architecture, 32 -bit data, predictable behavior, and an architecture that has been mathematically verified to be correct. I refer, of course, to the MAS1908 Viper, which is available from Marconi and supported by Assurance Systems. Unfortunately, in the US this device seems to suffer from the NIH ("not invented here") syn-


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[^2]

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drome, so I and others will have to continue to ensure our safe descent into Logan Airport at night, in a thunderstorm, by the unproven (but not disproven) power of prayer.
Cris Whetton
President
Assurance Systems
Levittown, PA

## Resistor would solve discharge-current problem

The Schmitt-trigger switchdebounce circuit appearing in the Design Idea "Switch debouncer uses few parts" (EDN, February 18, 1988, pg 244) will ultimately be unreliable, because excessive discharge current from the $4.7-\mu \mathrm{F}$ capacitor will damage the switch contacts (S1). This problem would be even more troublesome should the designer choose a switch designed for low-current switching (an appropriate choice for the type of circuit shown). A $220 \Omega$ resistor in series with the switch would solve the problem and leave the circuit's performance essentially unaffected.
Rod Deakin
Director of Engineering Services Arix Corp
San Jose, CA
(Ed Note: In fact, the circuit contains a number of problems. The author is submitting a corrected version of the Design Idea; it'll appear in a future issue of EDN.)

## Omission

Part 5 of the Decade 90 Series (EDN, April 28, 1988, pg 180) neglected to mention the location of Fail-Safe Technology Corp, a consulting firm specializing in reliablesystem design. The company is based in Los Angeles, CA.

## WRITE IN

Send your letters to the Signals and Noise Editor, 275 Washington St, Newton, MA 02158. We welcome all comments, pro or con. All letters must be signed, but we will withhold your name upon request. We reserve the right to edit letters for space and clarity.

# Programmable IC Filters Easy To Design With Improved Performance 



Filters are basic building blocks required in many systems applications. Most are built using discrete op amps, resistors, capacitors and inductors. This approach works for precision applications, but requires manual tuning of the elements. Once designed, these filters cannot be changed easily.

An alternate approach has been available for some time where the analog signal is sampled. This makes it tuneable with a clock. The circuit can be efficiently integrated on-chip using switched capacitor filter technology. This approach has been successful when used for application specific filters, however, for general-purpose filter applications it has its shortcomings: (1) the devices are difficult to use and require lots of external support circuitry; (2) very little software exists to help with the design process; (3) the devices typically exhibit excessive noise characteristics and change in value of programmed center frequency or selectivity (Q); (4) the devices only handle up to 20 kHz center frequency. Users have had to be very knowledgeable about sampled filters to successfully build one.

A new family of general purpose dual 2nd order programmable filters from Maxim, the MAX260-268, provide some solutions that can be used for precision applications over a wide frequency range.

## Microprocessor or PinProgrammable

Under microprocessor control the MAX260/261/262 filters can be programmed for center frequency ( $\mathrm{f}_{0}-$ 6-bits) and selectivity (Q - 7-bits). Filter characteristics can be reprogrammed in milliseconds. If $\mu \mathrm{P}$ control is not desired, pin programmed versions of
the MAX263/264/265/266 provide hard-wired control. The MAX265 and 266 resistor programmed filters provide infinite $Q$ and center frequency resolution for precisely tuned notches and applications which require fine Q control. The MAX267/268 dedicated bandpass filters round out the family.

In addition to the MAX260 programmable filters, Maxim offers the DC-accurate, "zero offset" MAX280 filter in an 8-pin package. And, for continuous non-switched filtering applications, including anti-aliasing, the MAX270 family of digitally programmed RC active filters will be introduced later in 1988.

## Programming the Filter from a PC

Complex nth order filters can be designed effortlessly in minutes with a MAX260-268 and Maxim's powerful software. This software reduces hours of searching through filter textbooks and performing hand calculations to a few simple keystrokes. It is PC-based, menu-driven and easy to use. A designer need only specify the basic parameters: type of filter (highpass, lowpass, bandpass, notch or allpass), and the frequency response. The 6color graphics routine then plots the simulated frequency response of the composite complex filter, so it can be examined on the computer screen to optimize its performance for desired results before you build it. For development work, the filters can even be programmed by connecting them to a PC's printer port without requiring $\mu \mathrm{P}$ programming or adjustments.

## Improved Performance

Center frequencies of the MAX260 family range from .01 Hz to 140 kHz . This extends bandwidth up to six times higher than previous switched capacitor filters. 100 uV Wideband RMS noise represents less than $1 / 2$ LSB in a precision 12-bit system. The MAX260268 4th order and MAX280 5th order filters are three to four times quieter than
(Please see Programmable Filters on back)

## Replace Two DG211's With The MAX333

Aquad Single Pole Double Throw $(4 \times$ SPDT $)$ analog switch from Maxim can now replace two DG303s, two IH5043s, or a DG211 plus DG212-at a lower price per channel than any of the alternatives.

The MAX333 is specifically designed for large volume applications such as disk drives, environmental controls and telecom.

At $\$ 2.37$ (1000-up), the cost per channel for the MAX333CPP-2 is less than \$0.60!

## Four Switches in One

A single pole double throw (SPDT) switch is really two switches that are operated by a single command. When one of the switches is turned on (closed), the other switch is simultaneously turned off (opened). An SPDT analog switch is therefore ideal for switching between two input signals.

The MAX333 quad SPDT switch has four distinct SPDT switches on a single chip. Previously, the best a designer could get was two SPDT switches in a package.


## Guaranteed +12V Operation

In addition to dual supply operation guaranteed over $\mathrm{a} \pm 5 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$, the MAX333 can operate with just one supply where ground serves as the negative supply. Single supply operation is guaranteed for supplies ranging from +10 V to +30 V . Electrical specs for

Programmable Filters (cont'd)

| Maxim's Family of Filter Products |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Part No. | Freque | ncy Range | Prog. Interface | Price* |
| Dual 2nd Order-Universal Filters |  |  |  |  |
| MAX260 | 0.01 Hz | to 7.5 kHz | Microprocessor | \$6.50 |
| MAX261 | 0.40 Hz | to 57.0 kHz | Microprocessor | 6.50 |
| MAX262 | 1.0 Hz | to 140.0 kHz | Microprocessor | 7.50 |
| MAX263 | 0.4 Hz | to 57.0 kHz | Pin Strap | 6.50 |
| MAX264 | 1.0 Hz | to 140.0 kHz | Pin Strap | 7.50 |
| MAX265 | 0.4 Hz | to 57.0 kHz | Resistor and Pin Strap | 6.50 |
| MAX266 | 1.0 Hz | to 140.0 kHz | Resistor and Pin Strap | 7.50 |
| Dual 2nd Order-Bandpass Filters |  |  |  |  |
| MAX267 | 0.4 Hz | to 57.0 kHz | Pin Strap | 6.50 |
| MAX268 | 1.0 Hz | to 140.0 kHz | Pin Strap | 7.00 |
| Dual 2nd Order-Lowpass RC Filter |  |  |  |  |
| MAX270 | 1.0 Hz | to 25.0 kHz | Pin Strap or Microprocessor | 5.00 |
| 5th Order Lowpass Filter-Zero DC Offset |  |  |  |  |
| MAX280 | 0.1 Hz | to 20.0 kHz |  | 3.55 |
| Universal 2nd Order Filter |  |  |  |  |
| MF10 | 0.1 Hz | to 30.0 kHz | Resistor | 1.70 |
| * 1000-up. |  |  |  |  |

filters like the MF-10. Dynamic Range is in excess of 88 dB !
$1 \%$ clock-to-fo ratio accuracy and only 5ppm/ ${ }^{\circ} \mathrm{C}$ variance in center frequency and gain are guaranteed over the full military temperature range.

## No External ComponentsMinimum Size

The MAX260 filters eliminate up to 8 resistors needed to program previous filters. All filters are available in DIPs and $0.3^{\prime \prime}$ wide SOIC surface mount packages at no extra cost.

## Ease of Design

A sample bandpass filter with the following desired parameters:

| Center frequency (fo) | $=15.6 \mathrm{kHz}$ |
| :--- | :--- |
| Pass bandwidth | $=15.6 \mathrm{kHz}$ |
| Stop bandwidth | $=46.8 \mathrm{kHz}$ |
| Max. passband ripple | $=1.0 \mathrm{~dB}$ |
| Min. stopband attenuation | $=20 \mathrm{~dB}$ |



Using the filter software, the order (number of poles), $\mathrm{f}_{\mathrm{o}}, \mathrm{Q}$ and number of sections is determined. The software generates the digital coefficients required by the filter, ' N .' 'The values of ' N ' are listed in Figure 1, as well as the chosen clocks and mode selection.

In this example, two 2nd order filter sections are required. This can be implemented using one MAX262.

Both halves of the filter operate in Mode 1 (Fig. 2) and use the same clock source, which is switched from 1 MHz to 2 MHz and 3 MHz , resulting in center frequencies of $15.6 \mathrm{kHz}, 31.3 \mathrm{kHz}$ and 47.0 kHz , respectively (Fig. 3).


Replace Two DG211's (cont'd)
12 V operation are shown on the data sheet as well as the specs for typical analog supplies of $\pm 15 \mathrm{~V}$. The MAX333 does not require a separate +5 V logic supply for either single or dual supplies and is completely TTL/CMOS logic compatible.

## Only 2 mW Power Dissipation

The MAX333 requires only $130 \mu \mathrm{~A}$ positive supply current and $10 \mu \mathrm{~A}$ negative supply current. Its analog signal range includes both the positive and negative supplies. ON resistance is 140 ohms and ON leakage only 0.2 nA . OFF leakage is 0.02 nA with a turn off time of just 50ns.

The MAX333 makes a useful flying capacitor multiplexer that multiplexes two channels differentially, removing any common mode signal present. Line rejection in industrial applications benefits from such an arrangement because the circuit can reject line interference and present the signal conditioning amplifier with an improved signal to noise ratio.

In telecommunications, the switching between different banks of channels is more cost effective with a MAX333. Not only is the cost per channel lower, but the device requires less board space than the alternativesespecially if the surface mount (SOIC) package is used.

In disk drive circuits the MAX333 enables multiplatter or disk read/write amplifier circuits to be switched into a common data separator/servo control circuit.

The MАХЗз3 complements Maxim's broad line of proprietary analog switches and multiplexers. Maxim supplies both industry standard analog switches (DG200, DG300, IH5040, and IH5140 families) and 'improved' proprietary devices such as the low power, high speed MAX331/332, an upgrade of the popular DG201/202.

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(CIRCLE 11)
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\author{

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International Conference on Handheld Computing, Corvallis, OR. Eric Gakstatter, ICHC, 301 NE Byron Pl, Corvallis, OR 97330. (503) 752-5456. August 4 to 6.

CASE Benchmarks: A Product Comparison Seminar, Atlanta, GA. Digital Consulting Inc, 6 Windsor St, Andover, MA 01810. (617) 470-3870. August 10 to 12.

Microwave Circuit Design I: Linear Circuits (short course), College Park, MD. University of Maryland Center for Professional Development, University Blvd at Adelphi Rd, College Park, MD. (301) 9857195. August 15 to 19.

Microwave Circuit Design II: Nonlinear Circuits (short course), College Park, MD. University of Maryland Center for Professional Development, University Blvd at Adelphi Rd, College Park, MD. (301) 985-7195. August 22 to 26.

Midcon, Dallas, TX. Electronic Conventions Management, 8110 Airport Blvd, Los Angeles, CA 90045. (800) 421-6816; in CA, (213) 772-2965. August 30 to September 1.

Surface Mount '88, Marlborough, MA. MG Expositions Group, 1050 Commonwealth Ave, Boston, MA 02215. (800) 223-7126; in MA, (617) 232-3976. August 30 to September 1.

Modern Electronic Packaging (seminar), Santa Clara, CA. Technology Seminars, Box 487, Lutherville, MD 21093. (301) 269-4102. September 7 to 9.

International Test Conference, Washington, DC. Doris Thomas, ITC, Box 264, Mount Freedom, NJ 07970. (201) 267-7120. September 12 to 14.

Worst-Case Circuit Analysis (seminar), Boston, MA. Design and Evaluation, 1000 White Horse Rd, Suite

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C Programming Workshop (short course), Seattle, WA. SSC, Box 55549, Seattle, WA 98155. (206) 5273385. September 12 to 15.

12th International Fiber Optic Communications and Local Area Networks Exposition, Atlanta, GA. Information Gatekeepers, 214 Harvard Ave, Boston, MA 02134. (800) 323-1088; in MA, (617) 2323111. September 12 to 16.

Connector and Interconnection Technology Symposium, Dallas, TX. Electronic Connector Study Group, 104 Wilmot Rd, Suite 201, Deerfield, IL 60015. (312) 940-8800. October 3 to 5.

IEEE International Conference on Computer Design: VLSI in Computers and Processors, Port Chester, NY. Gail Clanton, IEEE, 1730 Massachusetts Ave NW, Washington, DC 20036. (202) 371-1013. October 3 to 5.

Autotestcon, Minneapolis, MN. Steve Palmer, Unisys, 3333 Pilot Knob Rd, Eagan, MN 55121. (612) 456-2349. October 4 to 6.

Buscon/88 East, New York, NY. Conference Management Corp, 200 Connecticut Ave, Norwalk, CT 06856. (203) 852-0500. October 4 to 6.

Electronic Imaging Conference East, Boston, MA. MG Expositions Group, 1050 Commonwealth Ave, Boston, MA 02215. (800) 223-7126; in MA, (617) 232-3976. October 4 to 6.

Power Electronics East, New York, NY. Conference Management Corp, 200 Connecticut Ave, Norwalk, CT 06856. (203) 852-0500. October 4 to 6.

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## Spend it here



A guest editorial in the June 21st Wall Street Journal describes the causes of today's memory-chip shortages and defends the 1986 semiconductor pact between Japan and the US. According to the author, the Japanese priced dynamic-RAM chips below cost in the early 1980s in order to drive US manufacturers out of the DRAM business. Thus, they aimed to secure the world's semiconductor markets for the Japanese manufacturers. It's true that dumping products into a market can force out domestic suppliers-as many US DRAM manufacturers found out. However, the semiconductor pact took an illogical and shortsighted approach that continues to hurt US semiconductor suppliers and consumers.

In essence, the pact forces Japanese suppliers to charge a "fair" price for their DRAM chips. However, by raising prices to artificially high levels, we give foreign manufacturers extra profit, and we fund their research. The pact gives US industries some time to regroup, but that time is hard to put a price on. As an alternative to the pact, the US could impose a tariff, but according to the Journal's contributor, a tariff would result in immediate shortages-although the author doesn't explain why. Also, multinational companies could circumvent any tariff by purchasing low-cost DRAMs in overseas markets where tariffs don't exist.

In most cases, tariffs are counterproductive-companies use them as a shield against foreign competitors. After all, if the government forces up the price for a commodity, there's no incentive for domestic companies to increase manufacturing productivity or take any other measures to compete against low-cost foreign imports. But a short-term tariff on imported semiconductors just might work. If semiconductor prices must go up, I'd rather have the extra money stay in the US than go to fund Japanese researchers' next DRAM breakthrough.

So, let's eliminate the "fair" prices and let the Japanese and others charge what the market will bear. Then let's have the US government impose a tariff on every foreign DRAM that enters the US, whether it's in a tube of chips or in a telephone manufactured in Hong Kong. We'll set the tariff at $\$ 1$ per chip.

The funds collected from the tariff will go to an industry-research group such as Sematech and to US university labs working on semiconductor research. However, a semiconductor-tariff law must have a few restrictions. First, no one will be permitted to license or transfer any of the tariff-funded research information or process technology to a foreignowned or -dominated company. Second, the tariff will diminish to zero over the next two years, and it won't be reimposed. If we detect dumping again, however, a tariff will go into effect again. To get the ball rolling, I'll contribute $\$ 1$ for every foreign-made DRAM chip in my PC-even though I bought them before the dumping started. Just tell me where to send the check.


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## TECHNOLOGY UPDATE

# Third-party PC graphics packages streamline your programming tasks 

Margery S Conner, Regional Editor

Attention-grabbing, colorful graphics can dramatically enhance the software you write, regardless of whether you're developing it for your company's internal use or for customers. When you're writing software for IBM PCs and compatible computers, however, you'll find that addressing the PC's graphics hardware is a difficult task. To lighten your programming chores, consider using third-party graphicssoftware packages.
Third-party graphics packages for PCs come in varying degrees of complexity. Many are simply libraries of graphics subroutines that you'd otherwise have to write yourself. These packages range from general-purpose libraries (often called graphics toolkits), to libraries of specialized scientific subroutines. Other third-party packages include a graphics editor that helps you develop your display format in a hurry, and a complex graphics platform that offers a complete graph-ics-programming environment. By taking advantage of any these packages, you can develop graphics displays faster and for a wider variety of display adapters than you could on your own.
Until recently, most C and Pascal compilers for the IBM PC provided minimal support for color graphics displays. Third-party packages have filled that graphics void. However, since EGA and VGA adapters -with their relatively high resolu-tion-have become commonplace, the compiler vendors have beefed up the graphics libraries available with their compilers.

Borland International (Scotts Val-


Although the Halo '88 graphics library explicitly supports only 2-D renderings, you can develop 3-D renderings with it. Mega CADD (Seattle, WA) used Halo ' 88 as the graphics library for its Mega CADD 3D, a CAD software package.
ley, CA), for example, which sells two of the most popular versions of C and Pascal, now also offers the Borland Graphics Interface (BGI). BGI is a common graphics interface: Whether you're programming in Borland's Turbo C or Turbo Pascal, the graphics library is the same. A further advantage of using BGI is that if you're developing software for resale, you don't have to pay a royalty fee, as you do when you use most of the third-party packages for that purpose.

Keep in mind, however, that BGI is essentially just a library of primitives, which are the basic graphics commands, such as the commands to draw a point, line, circle, and polygon, and the commands to fill a circle and a polygon. Third-party graphics packages not only offer more-elaborate graphics commands, but also give you the option of programming in coordinates other than Cartesian ones. (Cartesian coordi-
nates are specific to a particular display.)

## Toolkits complement compilers

If your compiler already includes a good amount of graphics support, the complexity of your application's graphics will determine whether you'll need to lay out the extra money for a third-party graphics package. Chuck Batterman, graphics manager for Borland, explains: "If you don't need the high-level windowing and world coordinates supplied in the third-party packages, then BGI will do the job for you. And if you do need the more exotic graphics, then you can get your feet wet with the BGI and know that the third-party system is a reasonable investment." In other words, if you don't know whether you need a third-party package, you can find out by using the graphics supplied with your compiler to create a prototype of your display.

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Halo '88 from Media Cybernetics, for example, gives you the choice of specifying graphics positions in Cartesian, world, or normalized world coordinates. Unlike Cartesian coordinates, world coordinates are not specific to a particular display. Normalized world coordinates are world coordinates expressed as a number between zero and one.

World coordinates are useful when your software may have to run on a variety of graphics boards, because display coordinates vary from one board to another. If you use Cartesian coordinates, you must know the resolution of each supported graphics device, and your software won't be device independent. With Halo '88's implementation of world coordinates, you specify the resolution of the graphics, and Halo ' 88 maps the coordinates in your program to the device coordinates of your display adapter.

Because the compilers' graphics libraries have graphics drivers that access the EGA and VGA boards directly through the hardware, they are not device independent. Device independence allows a graphics program developed for one board-say, a nonintelligent EGA-to run on another board, such as an intelligent graphics-coprocessor-based board. When you use a device-independent program, you can specify your system's board at run time, and the program will then automatically send the correct level of information to that board. For the EGA, for example, the software must calculate the position of each pixel; for a board based on Texas Instruments' (Houston, TX) TMS34010, the software only needs to send the graphics primitives.

Third-party packages support a wide range of EGA and VGA boards that offer resolutions beyond the

IBM standard, as well as boards with graphics coprocessors. Halo '88, for example, supports 144 different display adapters.

Also keep in mind that the graphics libraries supplied with compilers don't support a wide range of hardcopy output devices, but third-party packages commonly support a plethora of plotters and printers.

## Port software from OS/2

A third-party graphics package that provides for portability between MS-DOS and the forthcoming OS/2 is crucial for you if your software will have to run under both operating systems. The Graphics Development Toolkit (GDT) from Graphics Software Systems is available in both OS/2 and MS-DOS versions, and their source code is compatible. GDT also supports IBM's new 8514/A intelligent display adapter. Because IBM hasn't re-

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| GRAPHICS SOFTWARE SYSTEMS INC | GRAPHICS DEVELOPMENT TOOLKIT | - |  | - | - | - | - | - | - | \$495 | SOFTWARE THAT USES MS-DOS VERSION IS PORTABLE TO OS/2 |
| METAGRAPHICS SOFTWARE CORP | METAWINDOW | - | - | - | - | - | - | - | - | \$195 | HAS DOS-RESIDENT DRIVER: BINDINGSPECIFIC VERSION FOR C COMPILERS IS \$95 |
|  | METAWINDOWIPLUS | - |  | - |  |  | - | - | - | \$275 | LINKABLE LIBRARY: GENERATES .EXE FILE |
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## TECHNOLOGY UPDATE

leased much information about this board's hardware, writing to it through an interface such as the GDT's is the best way at present to ensure that your software will run on the board.

Virtually all graphics toolkits must appeal to two different kinds of customers: the programmer who uses the graphics features only in programs for his or her company's internal use, and the programmer who uses the packages to develop a software-based product. Metagraphics lets you choose from three versions of its graphics package. The standard Metawindow package costs $\$ 195$ and supports a wide variety of compilers for C, Pascal, and Fortran. The company also offers a version for $\$ 95$ that supports either Microsoft C, Turbo C, or Turbo Pascal. Both these packages have DOS-resident drivers. Every time you run the program, you must first load the graphics driver into memory. (The driver can also be deleted from memory from the DOS level.) These programs are best for personal use or internal company use.

For $\$ 275$, you can obtain Metawindow/Plus, which includes a linkable graphics library; your application program links in only the graphics functions it actually uses, and it consists of a single .EXE file. Further, the code for Metawindow/ Plus is more compact than that of the DOS-resident driver version. The Plus version also supports display adapters with resolutions higher than those of the standard EGA and VGA boards. The $\$ 495$ Metawindow/Premium is also linkable, and it supports display adapters that have onboard graphics coprocessors, such as the Genesis 1280 from National Design (Austin, TX), which is based on TI's TMS34010.

## Down-and-dirty math library

If you're developing software that will display graphs of specialized math functions rather than using more general-purpose graphic elements, such as windows, consider


Fig 1—This robot comprises elements stored in four segments: the base, the upper arm, the lower arm, and the torch. Because of the hierarchical relationship among the elements, a command to rotate the base will cause the arms and the torch to rotate along with it. The robot was developed at Cornell University with the Hoops graphics package.
using Graphic from Scientific Endeavors. Its $\$ 395$ price includes source code. The software performs all the calculations for a high-resolution Tektronix format of $4096 \times 3120$ pixels and shrinks it down to suit the lower-resolution EGA/VGA format. It stores the higher-resolution display on disk. Because most printers, even inexpensive ones, have higher resolution than the VGA's $640 \times 480$ pixels, you can print out a high-resolution hard copy of your math functions

Because the functions are compu-tation-intensive, it's best to use the Graphic package with a math coprocessor. For example, a $6-\mathrm{MHz}$ PC/AT without a math coprocessor takes about 90 seconds to complete a typical plot with 600 trigonometric function calls; with a math coprocessor, it takes under 10 seconds.

With any of these graphics libraries, you still have to figure out the coordinates at which you want graphical elements such as windows
and icons to appear. This task can quickly become tedious unless you use a package such as Courseware Applications Inc's Drawbridge, a graphics editor that lets you position the elements on the screen with a pointing device such as a mouse. Once you've determined the placement of the elements, Drawbridge determines all of the coordinates and generates the library calls needed to display them. Drawbridge does not contain the subroutines themselves; instead, it acts in conjuction with a graphics library. It currently supports the Microsoft C, Turbo Pascal, and Turbo C libraries ( $\$ 129$ each), and the Metawindow package (\$49).
If your graphics needs are com-plex-that is, if you require lots of transformations and 3-D manipula-tions-you may be able to justify using a graphics platform such as Hoops from Ithaca Software Inc. Hoops supports 2-D and 3-D rendering as well as hidden-line removal

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and shading from multiple light sources.

Most graphics libraries are procedural; they comprise many detailed, explicit drawing instructions, but have no structure that relates the various elements of the display. Hoops stores the display's graphical elements in a hierarchical database. The software lets you manipulate the display by creating the graphical elements and defining the relationships between the elements. Once you've established the relationships between the elements, you can modify the image (change the perspective, say) with just a single command.

The database's basic units are segments, or named places in the database where you can store related primitives. You create a graphics element by grouping the primitives that describe the element into a segment. Then you use declarative commands to define which elements in the database will be displayed, where they'll be displayed, and how they'll be rendered (in other words, what attributes they'll have, such as color or pattern).

The primitives in a segment are drawn according to the attributes set within the segment. Because the


Fig 2-This tree illustrates the hierarchical relationship among the segments of the robot in Fig 1. Any attributes, such as color or pattern, that are set for a segment can be inherited by the segment below it.
segments are related hierarchically, if a segment doesn't specifically set the attribute, it inherits the attribute setting from the segment that precedes it in the hierarchy.

The robot in Fig 1 provides an illustration of this characteristic. The display shows both the robot itself and a selection menu. As you can see from Fig 2, the robot figure breaks down into four components: the base, the upper arm, the lower arm, and the torch. The primitives used to draw the lower arm, for example, are stored segments corresponding to that component. The color attribute is set to yellow. The segment under the lower-arm segment contains the primitives for drawing the torch. You don't have to specify the yellow color again for the torch, because it will inherit that

## For more information

For more information on the graphics packages discussed in this article, contact the following manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.

Courseware Applications Inc
475 Devonshire Dr
Champaign, IL 61820
(217) 359-1878

Circle No 705
Ithaca Software Inc
The Clinton House
Ithaca, NY 14850
(607) 273-3690

Circle No 706
Graphics Software Systems Inc
9590 SW Gemini Dr
Beaverton, OR 97005
(503) 641-2200

TLX 4994839
Circle No 707

Media Cybernetics Inc
8484 Georgia Ave, Suite 200
Silver Spring, MD 20910
(301) 495-3305

Circle No 708
Metagraphics Software Corp
269 Mount Hermon Rd
Scotts Valley, CA 95066
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# Limited selection of monolithic PLL ICs suffices for current applications 

Tarlton Fleming, Associate Editor

Monolithic phase-locked-loop (PLL) circuits provide an elegant way to achieve frequency demodulation or digital selection of frequency. As a result, PLL ICs have become vital components in modern TVs, car radios, frequency-shift-keyed (FSK) systems, and disk-drive electronics. You might expect, then, to find a robust market for PLL circuits.

Currently, though, domestic IC companies offer a sparse selection of new PLL devices. This lack of activity is surprising when you consider the extensive applications support that was once available to an engineer intent on designing a phaselocked loop into his system. It's doubly surprising when you consider the current emphasis on designing discrete-component PLL circuits for frequency-agile, military-RF systems. The truth is that much PLL work has moved offshore, along with the design and manufacture of certain high-volume, PLL-based products-car radios, for instance. Fortunately, enough applications remain in the US to support a continuing demand for domestically produced PLL ICs.

A PLL is a small servo system in which negative feedback forces the frequency of a voltage-controlled oscillator (VCO) equal to that of the applied signal, $\mathrm{f}_{\mathrm{IN}}$ (Fig 1). This action causes the PLL to sense and track the $\mathrm{f}_{\mathrm{IN}}$ frequency, thereby reconstructing $f_{\text {IN }}$ and removing noise in the process (which is a form of filtering).

If you monitor the VCO's input instead of its output, the PLL will function as a frequency-sensing device suitable for FM detection or


Fig 1-The basic phase-locked-loop circuit is a small servo system. The voltage output is proportional to frequency, and the frequency output is a noise-free equivalent of $f_{I N}$.

FSK demodulation. These two configurations (taking the PLL's output from the VCO's input or its output) account for most applications. If you add a divide-by-M counter at the input and a divide-byN counter within the loop, at the VCO output, the PLL can also produce a digitally selectable frequen$\mathrm{cy}, \mathrm{f}_{\mathrm{IN}}(\mathrm{N} / \mathrm{M})$.
Many low-frequency applications (approximately 1 MHz and below) use the industry-standard CMOS PLL, whose generic part number is 4046. Manufacturers of this type include GE/RCA, Motorola, National Semiconductor, and Signetics. The Motorola MC14046BCP is a representative part: It costs $\$ 0.93$ in quantities of 100 .

The 4046 devices include two types of digital phase detectors on chip. (Most other PLLs feature one type, not both.) Your 4046 application determines whether a type I or type II phase detector is most ap-
propriate, and you then make external connections for including that detector in the loop.

## Type I vs II phase

The type I phase detector is simply an XOR gate. The output signal, after passing through an external lowpass filter, produces a linear ramp in which the amplitude is proportional to the phase ( 0 to $180^{\circ}$ ) between $f_{\text {IN }}$ and the onboard VCO signal. The ramp reverses and repeats when the PLL is out of lock, forming a sawtooth waveform. The type I phase detector provides the widest lock range (the $f_{\text {IN }}$ range for which the loop will remain locked) for input signals with a duty cycle of $50 \%$. When the loop is locked at the VCO's free-running frequency, the phase detector generates a dc level at half the ramp amplitude-equivalent to a $90^{\circ}$ phase difference.

If nonzero phase in the locked condition is unacceptable, you can

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| :---: | :---: | :---: | :---: |
|  | OP-97 | OP-07 |  |
| V | 25 | 25 | $\mu \mathrm{V}$ Max |
| TCV ${ }_{\text {OS }}$ | 0.6 | 0.6 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ Max |
| Avo | 300 | 300 | $\mathrm{V} / \mathrm{mV}$ Min |
| $\mathrm{I}_{\mathrm{B}}$ @ $25^{\circ} \mathrm{C}$ | $\pm 100$ | $\pm 2,000$ | pA Max |
| $\mathrm{I}_{\mathrm{B}}$ @ $125^{\circ} \mathrm{C}$ | $\pm 250$ | $\pm 4,000$ | pA Max |
| $\mathrm{I}_{\text {SY }} @ \pm 15 \mathrm{~V}$ | 600 | 4,000 | $\mu \mathrm{A}$ Max |

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use the type II (edge-controlled) phase detector, which ignores waveform duty cycles by operating on the two input signals' positivegoing digital edges. While in the locked condition, the phase detector allows no phase difference between the VCO and $\mathrm{f}_{\mathrm{IN}}$ signals. The type II detector provides a large capture range (the range of $f_{\text {IN }}$ for which an unlocked PLL can achieve the locked condition) that is independent of the external filter's time constant.
When unlocked, the detector produces a high or low de level according to the relative frequency values for $\mathrm{f}_{\mathrm{IN}}$ and the VCO signal (vs the ac output of a type I detector). This circuit, also known as a phase-frequency detector, was an important advance in the development of PLLs at the time of its introduction.

Exar and Signetics are noted for their PLL product lines. Exar's XR-210, for example, suits FSK applications and includes a VCO, an XOR-gate phase detector, and a voltage comparator. The device operates with a 5 V supply over frequencies of 0.5 Hz to 20 MHz . Housed in a 16 -pin ceramic DIP, an XR-210CN sells for $\$ 2.78$ (100).

The more-complex XR-2211, sec-ond-sourced by Raytheon, also fits data-communications tasks such as tone decoding and FSK demodulation. The 2211 has an analog-multi-plier-type phase detector (Gilbert cell) that subtracts the two input signals. It provides outputs for lock detection and FSK demodulation, and it operates from 0.01 Hz to 300 kHz . Exar's version costs $\$ 2.89$ (100).

For applications that require temperature stability, Exar's precision XR-2212 PLL operates from 0.01 Hz to 300 kHz and exhibits a $20-\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ temperature coefficient of the VCO frequency. The XR-2212 comes in a 16 -pin DIP and costs $\$ 2.88$ (100).

Signetics uses a bipolar process in its high-frequency NE568 design in which the $\mathrm{f}_{\mathrm{T}}$ for the npn transistors
exceeds 6 GHz . The result is a $150-$ MHz PLL suitable for incorporation in satellite receivers, fiber-optic video links, and VHF-range FSK demodulators.

Realizing that proper pc-board and component layout is essential at high frequencies, Signetics includes an evaluation circuit with the NE568's data sheet, as well as a pc-board pattern and parts list. According to the manufacturer, with a good layout and surface-mount capacitors, the device demodulates $\pm 10 \%$ deviations from a $70-\mathrm{MHz}$ IF and exhibits a linearity error of less than $4 \%$. An NE568N comes in a 20-pin DIP, draws 60 mA from a 5 V supply, and costs $\$ 1.79$ (100).

## Digital vs analog PLLs

Although overwhelmingly digital, most PLL ICs are called analog PLLs because their VCO input is an analog error voltage. The SN74LS297 from Texas Instruments, however, is all digital. The device has a counter and increment/ decrement circuit in place of the conventional lowpass filter and VCO. To program the counter, you use a 4 -bit digital word, which corresponds to setting the filter time constant in an analog PLL circuit. (Both types of adjustments tradeoff
loop bandwidth and phase jitter.) The SN74LS297's operating-frequency range is dc to 50 MHz . A 16-pin, plastic-DIP version costs $\$ 11.69$ (100).

Unlike the discrete-IC PLLs discussed so far, Sierra Semiconductor's PLL34M exists only as a CMOS cell in the vendor's semicus-tom-ASIC library. This standardcell PLL circuit operates at 30 MHz (or as high as 60 MHz , with tweaking by the manufacturer). It includes a phase detector, I/O drivers, and a bandgap reference for setting the VCO frequency. The chip area is 3224 mils ${ }^{2}$. By combining the PLL334M with other cells in the library-you have a choice of 250 digital, 50 analog, and 20 EEPROM cells-you can configure an ASIC suitable for use as a data separator in disk drives, for example. You can buy a PLL34M, plus 500 logic gates, enclosed in a 44-pin surface-mount package for $\$ 3.50(100,000)$.

## Prescalers offer alternatives

Siemens combines its PLLs with appropriate prescaling circuits to produce ICs capable of handling the high frequencies involved in digital channel selection in TVs and radios. These products are partial PLLs because they operate in conjunction

## For more information

For more information on the PLL products discussed in this article, contact the following manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.

## Exar Corp

2222 Qume Dr
San Jose, CA 95131
(408) 732-7970

Circle No 712
Motorola Semiconductor Products Inc
3501 Ed Bluestein Blvd
Austin, TX 78721
(602) 244-6900

Circle No 713
Siemens Components Inc
2191 Laurelwood Rd
Santa Clara, CA 95054
(408) 980-4500

Circle No 714

Sierra Semiconductor Corp
2075 N Capitol Ave
San Jose, CA 95132
(408) 263-9300

Circle No 715

## Signetics Corp

Box 3409
Sunnyvale, CA 94088
(408) 991-2000

Circle No 716
Texas Instruments
Box 225012
Dallas, TX 75265
(214) 995-6611

Circle No 717

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```
                        2.0
```

                    12.7
                        Div
    

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| Western Region | $408245-2060$ |
| :--- | ---: |
| So. California | $714583-2250$ |
| Central Region | $312640-7785$ |
| Eastern Region | $617745-7400$ |
| Japan | $042333-8111$ |
| England | 0734788666 |
| West Germany | $08992692-666$ |

with an external tuning chip that contains the VCO. Model SDA 3202-2, for instance, includes an RF amplifier, a phase-frequency detector, and a serial-data $\mu \mathrm{P}$ interface. Counter circuits on the chip can scale signal frequencies as high as 1.3 GHz . The phase detector then compares the result with a 7.8125 kHz reference (obtained from an onboard oscillator divider and an external crystal). In an 18-pin DIP, the device draws 55 mA from a 5 V supply. It's priced at $\$ 4.63$ (100).

Siemens uses the same approach in its SDA 2121; a device that helps provide digital tuning in AM and FM radio receivers. The CMOS chip can accept a VCO input as high as 35 MHz (AM) or 150 MHz (FM). It comes in a 20 -pin DIP and costs $\$ 5.14$ (100).
Similarly, the TBB 200 is a CMOS PLL circuit optimized for use in $\mu \mathrm{P}$-controlled mobile telephones and radios. When added to an external VCO tuner and prescaler, the chip forms a compact system suitable for use in circuits operating above 900 MHz . The TBB 200 comes in a 14 -pin DIP and costs $\$ 5.61$ (100).

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algorithms for improved optimization; DeMorganization (helpful when negating complex expressions); and simulation, to help you verify your design.

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- A manual for the CUPL software.
- A Programmable Logic User Guide (PLUG) diskette, which lets you browse interactively through the workings of programmable logic. - "My First PAL Design," a booklet that leads you step by step through programmable logic design.

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| Part | RAMs <br> Supported | Package |
| :--- | :--- | :--- |
| KS84C21-25CL | $256 \mathrm{~K}, 1 \mathrm{MB}$ | 68 -pin PLCC |
| KS84C22-25CL | $256 \mathrm{~K}, 1 \mathrm{MB}, 4 \mathrm{MB}$ | 84 -pin PLCC |

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But don't take our word for it. Request our reliability report, data sheet, and samples today. Then compare it for yourself.

| Product   <br> Specification Samsung 28C64 Leading Competifor's |  |  |
| :--- | :--- | :--- |
| ISB (standby current) | $100 \mu \mathrm{~A}$ | $150 \mu \mathrm{~A}$ |
| ICC (active current) | 30 mA | 50 mA |
| TAA (address access time) | 200 ns | 200 ns |
| TRC (read cycle time) | 200 ns | 200 ns |
| TWC (write cycle time) | $2-5 \mathrm{~ms} /$ byte | $2-5 \mathrm{~ms} /$ byte |
| VCC | $5 \mathrm{~V} \pm 10 \%$ | $5 \mathrm{~V} \pm 10 \%$ |
| Page mode | Yes | Yes |
| Endurance (write cycle) | 10,000 | 10,000 |
| Data retention | 10 years | 10 years |
| Packaging | DIP, PLCC | DIP, PLCC |

## The KM28C64/65 at a glance.

$8 \mathrm{~K} \times 8$ CMOS EEPROM $=28$-pin JEDEC byte-wide memory pinout (DIP, PLCC) $=$ Single 5V $\pm 10 \%$ Vcc supply $\quad$ Performance: 200/250ns $\quad$ Current: standby ( $\max ) 100 \mu \mathrm{~A}$; active $(\max ) 30 \mathrm{~mA}=32$ byte page write: $5 \mathrm{~ms} \pm 4$-cell bridge for enhanced reliability $=$ Write completion indicator: Data polling, Rdy/busy (for KM28C65) ■ Endurance 10,000 cycles $■$ Data retention 10 years



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Samsung CMOS and NMOS EEPROMs.

| Part Type | Organizafion | Speed | Features | Tochmology | Pinout | Avallabillity |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| KM28C64P | $8 \mathrm{~K} \times 8$ | 200,250 | Data Polling, 32-byte Page Mode, Low Power, 5 ms (max) write time | CMOS | 28 pin | From <br> Stock |
| KM28C65P | $8 \mathrm{~K} \times 8$ | 200,250 | Data Polling, 32-byte Page Mode, Low Power, 5 ms (max) write time | CMOS | 28 pin | From Stock |
| KM2816AP | $2 \mathrm{~K} \times 8$ | 250,300,350 | 10 ms (max) write time | NMOS | 24 pin | From Stock |
| KM2817AP | $2 \mathrm{~K} \times 8$ | 250,300,350 | Ready/Busy, 10ms (max) write time | NMOS | 28 pin | From Stock |

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| Part Type | Resolution |  | Linearity |  | Conversion Speed | Industry Part |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | A/D | D/A | A/D | D/A |  |  |
| KSV3110N-10 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 1 / 2$ LSB | 20 MSPS |  |
| KSV3110N-9 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 1$ LSB | 20 MSPS |  |
| KSV3110N-8 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 2 \mathrm{LSB}$ | 20 MSPS |  |
| KSV3110N-7 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 4 \mathrm{LSB}$ | 20 MSPS |  |
| KSV3100AN-8 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 2$ LSB | 20 MSPS | UVC3101 |
| KSV3100AN-7 | 8 bits | 10 bits | $\pm 1 / 2$ LSB | $\pm 4 \mathrm{LSB}$ | 20 MSPS | UVC3101 |

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The Samsung $\mathbf{A} / \mathbf{D}$ and $\mathrm{D} / \mathbf{A}$ Lines.

| Part Type | Resolution |  | Linearify |  | Conversion Speed | Industry Part |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | A/D | D/A | A/D | D/A |  |  |
| KSV3208N | 8 bits |  | $\pm 1 / 2$ |  | 20 MSPS |  |
| KAD0820ACN | 8 bits |  | $\pm 1 / 2$ |  | $1.5 \mu \mathrm{sec}$ | ADC0820BCN |
| KAD0820AIN | 8 bits |  | $\pm 1 / 2$ |  | $1.5 \mu \mathrm{sec}$ | ADC0820BCJ |
| KAD0820BCN | 8 bits |  | $\pm 1 \mathrm{LS}$ |  | $1.5 \mu \mathrm{sec}$ | ADC0820CCN |
| KAD0820BIN | 8 bits |  | $\pm 1$ LS |  | $1.5 \mu \mathrm{sec}$ | ADC0820CCJ |
| KAD0808IN | 8 bits |  | $\pm 1 / 2$ |  | $100 \mu \mathrm{sec}$ | ADC0808CCN |
| KAD0809IN | 8 bits |  | $\pm 1$ LS |  | $100 \mu \mathrm{sec}$ | ADC0809CCN |
| KDA0800CN |  | 8 bits |  | $\pm 1 / 2$ LSB | * 100 nsec | DAC0800LCN |
| KDA0801CN |  | 8 bits |  | $\pm 1$ LSB | * 100 nsec | DAC0801LCN |
| KDA0802CN |  | 8 bits |  | $\pm 1 / 4$ LSB | * 100 nsec | DAC0802LCN |
| KDA0806CN |  | 8 bits |  | $\pm 2$ LSB | * 150 nsec | DAC0806LCN |
| KDA0807CN |  | 8 bits |  | $\pm 1$ LSB | * 150 nsec | DAC0807LCN |
| KDA0808CN |  | 8 bits |  | $\pm 1 / 2$ LSB | ${ }^{*} 150 \mathrm{nsec}$ | DAC0808LCN |
| KS7126CN | 3-1/2 digit |  | $\pm 1 / 2$ |  | 333 msec | TSC7126 |
| KS25C02 | CMOS 8-bit successive approx. register |  |  |  |  | DM2502 |
| KS25C03 | CMOS 8-bit successive approx. register |  |  |  |  | DM2503 |
| KS25C04 | CMOS 12-bit successive approx. register |  |  |  |  | DM2504 |

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To learn about others, please contact us.
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| Part | Function | Application |
| :--- | :--- | :--- |
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| KS5902XX | Voice synthesizer <br> (internal ROM) | Toys; simple sound <br> generation |
| KS5911 | Voice recording and reproducing <br> (talking back type) | Talk-back <br> answering machines |
| KS5912XX | Natural sound <br> generation | Toys; natural sound <br> effect |


| Major Linear ICs |  |  |
| :---: | :---: | :---: |
| Regulators |  |  |
| 3T Positive |  |  |
| KA78TXXCF |  | 3 Amp |
| $\begin{aligned} & \text { KA78TXXCT } \\ & \text { LM317T } \end{aligned}$ |  | 3 Amp |
|  |  | 1.5 Amp |
| MC78XXCT |  | 1 Amp |
| MC78MXXCT |  | 0.5 Amp |
| MC78LXXACZ |  | 0.1 Amp |
| LM723CN |  | 0.1 Amp |
| 3T Negative |  |  |
| KA337T |  | 1.5 Amp |
|  |  | 1 Amp |
|  |  | 0.5 Amp |
| MC79MXXXACZ |  | 0.1 Amp |
| Switching |  |  |
| KA78S40CN |  | KA3524N |
| REF Voltage |  |  |
| KA431CZ (TL431CLP) <br> KA431N (TL431CP) |  | KA336Z-2.5,5 |
|  |  | KA385Z-1.2 |
| Comparafors |  |  |
| KA319N (LM319N) |  | LM2903N |
| KA361N (LM361N) |  | LM311N* |
| KA710CN (LM710N) |  | LM3302N |
| KS374N (TLC374N) |  | LM339N/AN* |
| LM2901N |  | LM393N/AN* |
| Op Amplifiers |  |  |
| KA301N/AN (LM301N)* <br> KA733CN (LM733CN) |  | LM348N* |
|  |  | LM358N/AN* |
| KA9256 (POWER AMP) |  | LM741CN* |
| KF351N (LF351N) LM2902N |  | MC1458N* |
|  |  | MC3303N |
| LM2904N |  | MC3403N* |
| LM324N/AN* |  | MC4558N* |
| Telecommunication ICs |  |  |
| KA2410N Tone Ringer |  |  |
| KA2411N Tone Ringer |  |  |
| KA2418N | Tone Ringe | /bridge diode |
| KA2412FN Speech Network |  |  |
| KA2413N DTMF |  |  |
| KS5808N DTMF (MK5089) |  |  |
| KS5805AN Pulse (MK50992) |  |  |
| KS5805BN Pulse (MK50993) |  |  |
| KS55820N DTMF/Pulse |  |  |
| KT3040J CODEC Filter |  |  |
| KT5116J CODEC |  |  |
| LM567N* Tone Decoder |  |  |
| LM567LN Tone Decoder (Low Power) |  |  |
| KS5812N | Quad UART |  |
| Timers |  |  |
| KS555N (CMOS)* |  | NE555CN* |
| KS555HN (CMOS)** |  | NE556CN* |
|  |  | NE558CN |
| RS-232 Interfaces |  |  |
| MC1488N* |  | MC1489/AN* |
| Audio ICs |  |  |
| LM386N 0.5 Watt Power Amp |  |  |
| KA2201N 1.2 Watt Power Amp |  |  |
| KA2206 2.3 Watt Dual Power Amp |  |  |
| KA22062 4.6 Watt Dual Power Amp |  |  |
|  | 5.8 Watt Du | al Power Amp |
| KA22101 23 Watt Power Amp |  |  |
| KA2243 AM/FM IF \& DET |  |  |
| KA22441 FM IF \& DET |  |  |
| KA22495 FM Front End |  |  |
| KA2263 FM MPX |  |  |
| 45 other audio ICs available |  |  |
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| A2914A Color TV VIF/SIF |  |  |
| KA2153 NTSC Chroma \& Deflection |  |  |
| KA6101 Video R.G.B. Interface |  |  |
|  |  |  |
| 40 other $v$ | ideo ICs ava | ilable |
| Others |  |  |
| KA2580AN (UDN2580AN) |  |  |
| KA2588AN | (UDN2588 |  |
| A2651N (UCN5815AN) |  |  |
| KA2615 LED/Lamp Driver |  |  |
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| KA2288 7 Dot LED Meter Driver |  |  |
| KA2181 Infrared Amp |  |  |
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| :--- | :--- | :--- |
| KSD5010* | KSD5012* | KSD5013* |
| KSD5014 | KSD5016 | KSD5017 |
|  |  |  |
| 3.5 amps |  |  |
| KSD5011* |  |  |
| KSD5015 |  |  |
| *Damper diode built-in transistor |  |  |
| SOT-23 |  |  |

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BCX71G MMBR5179

MMBT4403 MMBTA43


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TIP102 TIP126
TIP105 TIP127

## MJE SERIES

| MJE170 | MJE210 | MJE800 |
| :--- | :--- | :--- |
| MJE171 | MJE340 | MJE801 |
| MJE172 | MJE350 | MJE802 |
| MJE180 | MJE700 | MJE803 |
| MJE181 | MJE701 | MJE2955T |
| MJE182 | MJE702 | MJE3055T |

TO-92

| 2N3904 | 2N5210 | MPSA42 |
| :--- | :--- | :--- |
| 2N3906 | 2N5400 | MPSA43 |
| 2N4123 | 2N5401 | MPSA55 |
| 2N4124 | 2N5550 | MPSA56 |
| 2N4125 | 2N5551 | MPSA70 |
| 2N4126 | 2N6427 | MPSA92 |
| 2N4400 | 2N6428 | MPSA93 |
| 2N4401 | 2N6515 | MPSH10 |
| 2N4402 | 2N6517 | MPSH17 |
| 2N4403 | 2N6520 | MPSH20 |
| 2N5086 | MPSA05 | MPSH24 |
| 2N5087 | MPSA06 | PN2222A |
| 2N5088 | MPSA14 | PN2907A |
| 2N5089 | MPSA20 |  |

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TO-3P Package N-Chamnel Types IRFP120 IRFP422 IRFP121 IRFP423
IRFP122 IRFP430 IRFP122 IRFP430
IRFP123 IRFP431 IRFP123 IRFP431
IRPP130 IRFP432 IRFP131 IRFP433 IRFP132 IRFP440
IRFP133 IRFP441
SSM4N50 SSM20N50
SSM4N45 SSM4N45 SSM20N45
SSM5N40 SSM5N40
SSM25N40 SSM25N40
SSM5N35 SSM5N35 SSM25N35 SSM7N20
SSM8N20 SSM8N20
SSM7N18 SSM7N18
SSM8N18 SSM8N18
SSM7N15 SSM3N70 SSM3N70
SSM4N70 SSM4N70 SSM6N70 SSM7N12 $\begin{array}{ll}\text { SSM10N70 } & \text { SSM8N12 } \\ \text { SSM4N60 } & \text { SSM12N10 }\end{array}$ $\begin{array}{ll}\text { SSM4N60 } & \text { SSM12N10 } \\ \text { SSM6N60 } & \text { SSM10N10 }\end{array}$ $\begin{array}{ll}\text { SSM6N60 } & \text { SSM10N10 } \\ \text { SSM8N60 } & \text { SSM12N08 }\end{array}$ $\begin{array}{ll}\text { SSM8N60 } & \text { SSM12N08 } \\ \text { SSM15N60 } & \text { SSM10N08 }\end{array}$ $\begin{array}{ll}\text { SSM15N60 } & \text { SSM10N08 } \\ \text { SSM4N55 } & \text { SSM12N06 }\end{array}$ SSM6N55 $\quad$ SSM10N06 SSM8N55 SSM12N05 SSM15N55 SSM10N05 IRFP140 IRFP442 IRFP141 IRFP443
IRFP142 IRFP450 IRFP142 IRFP450
IRFP143 IRFP451 IRFP143 IRFP451
IRFP150 IRFP452 IRFP151 IRFP453 IRFP152 SSH3N70 IRFP153 SSH4N70
IRFP220 SSH6N70 IRFP220 SSH6N70
IRFP221 SSH10N70 IRFP221 SSH10N70 IRFP222 SSH4N60
IRFP223 SSH6N60 IRFP223 SSH6N60
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IRFP231 SSH15N60 IRFP231 SSH15N60
IRFP232 SSH4N55 IRFP232 SSH4N55
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IRFP341 SSH8N12 IRFP342 SSH12N10
IRFP343 SSH10N10 IRFP343 SSH10N10
IRFP350 SSH12N08 IRFP350 SSH 12N08
IRFP351 SSH10N08 IRFP351
IRFP352
SSH10N08 IRFP353 SSH10N06
IRFP420
SSH12N05 IRFP420 SSH12N05
IRFP421 SSH10N05

## TO-3P Package

 P-Channel Types IRFP9121 IRFP9221 IRFP9122 IRFP9222 IRFP9123 IRFP9223IRFP9130 IRFP9230 IRFP9130 IRFP9230
IRFP9131 IRFP9231 IRFP9131 IRFP9231
IRFP9132 IRFP9232 IRFP9132 IRFP9232
IRFP9133 IRFP9233 IRFP9140 IRFP9240 IRFP9141 IRFP9241 IRFP9142 IRFP9242 IRFP9143 IRFP9243
TO-3 Package N-Chamnel Types IRF120 IRF242 IRF121 IRF243 IRF122 IRF250 IRF123 IRF251 IRF130 IRF252 IRF131 IRF253 IRF132 IRF320 IRF133 IRF321 $\begin{array}{ll}\text { IRF140 } & \text { IRF322 } \\ \text { IRF141 } & \text { IRF323 }\end{array}$ $\begin{array}{ll}\text { IRF141 } & \text { IRF323 } \\ \text { IRF142 } & \text { IRF330 }\end{array}$ $\begin{array}{ll}\text { IRF143 } & \text { IRF331 } \\ \text { IRF150 } & \text { IRF332 }\end{array}$ $\begin{array}{ll}\text { IRF151 } & \text { IRF333 }\end{array}$ $\begin{array}{lll}\text { IRF151 } & \text { IRF333 } \\ \text { IRF152 } & \text { IRF340 }\end{array}$ IRF153 IRF341 IRF220 IRF342 $\begin{array}{ll}\text { IRF221 } & \text { IRF343 } \\ \text { IRF222 } & \text { IRF350 }\end{array}$ $\begin{array}{ll}\text { IRF222 } & \text { IRF350 } \\ \text { IRF223 } & \text { IRF351 }\end{array}$ IRF230 IRF352 IRF231 IRF353 IRF232 IRF420 IRF233 IRF421 IRF240 IRF422 IRF241 IRF423
P-Ch Package P-Channel Types IRF9120 IRF9220 IRF9121 IRF922 IRF9122 IRF9222 IRF9123 IRF9223 IRF9130 IRF9230 IRF9132 IRF9232 IRF9133 IRF9233 IRF9140 IRF9240 IRF9141 IRF924 IRF9142 IPF9242 RF9143 IRF9243

TO-220 Package -Channel Types IRF510 IRF741 IRF511 IRF742 RF512 IRF743 RF513 IRF820 RF520 IRF821 $\begin{array}{ll}\text { IRF521 } & \text { IRF822 } \\ \text { IRF522 } & \text { IRF823 }\end{array}$ $\begin{array}{ll}\text { IRF522 } & \text { IRF823 } \\ \text { IRF523 } & \text { IRF830 }\end{array}$ RF530 IRF831 IRF531 IRF832 $\begin{array}{ll}\text { IRF532 } & \text { IRF833 } \\ \text { IRF533 } & \text { IRF840 }\end{array}$ $\begin{array}{ll}\text { RF533 } & \text { IRF840 } \\ \text { RF540 } & \text { IRF841 }\end{array}$ | RF540 | IRF841 |
| :--- | :--- |
| RF541 | IRF842 | RF542 IRF843 RF543 SSP3N70 $\begin{array}{ll}\text { RF610 } & \text { SSP4N70 } \\ \text { RF611 } & \text { SSP4N60 }\end{array}$ $\begin{array}{ll}\text { RF611 } & \text { SSP4N60 } \\ \text { RF612 } & \text { SSP6N70 }\end{array}$ $\begin{array}{ll}\text { RF612 } & \text { SSP6N70 } \\ \text { RF613 } & \text { SSP6N60 }\end{array}$ $\begin{array}{ll}\text { IRF613 } & \text { SSP6N60 } \\ \text { IRF620 } & \text { SSP6N55 }\end{array}$ $\begin{array}{ll}\text { RF620 } & \text { SSP6N55 } \\ \text { RF621 } & \text { SSP4N60 }\end{array}$ $\begin{array}{ll}\text { SSF621 } & \text { SSP4N60 } \\ \text { SSP4N55 }\end{array}$ SSP4N55

SSP4N50 $\begin{array}{ll}\text { RF623 } & \text { SSP4N50 } \\ \text { SSP4N45 }\end{array}$ $\begin{array}{ll}\text { RF631 } & \text { SSP5N45 }\end{array}$ $\begin{array}{ll}\text { RF631 } & \text { SSP5N40 } \\ \text { RSP5N35 } & \text { SSP5N }\end{array}$ $\begin{array}{ll}\text { RF633 } & \text { SSP5N35 } \\ \text { SSP7N20 }\end{array}$ $\begin{array}{ll}\text { RF640 } & \text { SSP7N20 } \\ \text { SSP8N20 }\end{array}$ IRF641 SSP7N18 $\begin{array}{ll}\text { IRF642 } & \text { SSP7N18 } \\ \text { SSP8N18 }\end{array}$ $\begin{array}{ll}\text { IRF642 } & \text { SSP8N18 } \\ \text { IRF643 } & \text { SSP7N15 }\end{array}$ $\begin{array}{ll}\text { IRF643 } & \text { SSP7N15 } \\ \text { IRF710 } & \text { SSP8N15 }\end{array}$ $\begin{array}{ll}\text { IRF710 } & \text { SSP8N15 } \\ \text { IRF711 } & \text { SSP7N12 }\end{array}$ $\begin{array}{ll}\text { IRF711 } & \text { SSP7N12 } \\ \text { IRF712 } & \text { SSP8N12 }\end{array}$ $\begin{array}{ll}\text { IRF712 } & \text { SSP8N12 } \\ \text { IRF713 } & \text { SSP12N10 }\end{array}$ $\begin{array}{ll}\text { IRF720 } & \text { SSP12N10 } \\ \text { SSP10N10 }\end{array}$ $\begin{array}{ll}\text { IRF720 } & \text { SSP10N10 } \\ \text { IRF721 } & \text { SSP12N08 }\end{array}$ $\begin{array}{ll}\text { IRF721 } & \text { SSP12N08 } \\ \text { SSP10N08 }\end{array}$ IRF723 SSP10N08 $\begin{array}{ll}\text { IRF723 } & \text { SSP12N06 } \\ \text { IRF730 } & \text { SSP40N06 } \\ \text { IRF731 } & \text { SSP10N06 }\end{array}$ $\begin{array}{ll}\text { IRF731 } & \text { SSP10N06 } \\ \text { IRF732 } & \text { SSP12N05 }\end{array}$ $\begin{array}{ll}\text { IRF732 } & \text { SSP12N05 } \\ \text { IRF733 } & \text { SSP10N05 }\end{array}$ IRF740

## P-Chamel Types

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## PRODUCT UPDATE

## Video and system controller IC promises low-cost $16 / 32$-bit color computers

Used in conjunction with the company's $68070 \mu \mathrm{P}$, the 66470 video and system controller IC allows you to build a complete $16 / 32$-bit processing and color display system, using as few as 10 ICs. You can also utilize the device with other 68000 family $\mu$ Ps. The 66470 incorporates a dynamic RAM controller for video/ system RAM, display control logic, a pixel accelerator, and an interface for an optional graphics coprocessor. Its system-controller functions include reset frequencing, address decoding, interrupt, DTACK generation logic, and a watchdog timer.
The dynamic RAM controller supports dynamic RAMs with capacities as high as 1 M bit, and the address decoding provides chip selects for as much as 1.5 M bytes of video/system RAM, 0.5 M bytes of ROM, and 1 k byte of peripheral I/O space. You can configure as much as 1 M byte of the RAM as video memory, containing 4 - or 8 -bit pixels to provide $768 \times 560$-pixel 16 -color, or $384 \times 280$-pixel 256 -color displays, respectively. Lower screen resolutions are also possible. You can program the video start address anywhere within the 1 M byte of video RAM; each screen pixel occupies either a nibble or a byte in memory, thus simplifying the fetching of pixel information for modification by the system's CPU or the 66470's pixel accelerator. You can configure the video RAM for logical or physical screen modes and generate either full screens or screens with a programmable border color.
The video-display logic provides a digital output for use with an external D/A converter or color look-up table. The display timing is compatible with European, Japanese, and US standards for TV and Teletext.


Using the 68070/66470 chip set, this single-Eurocard board contains all the components required to implement a 16/32-bit color computer, and still has room to spare.

You can synchronize the 66470 with a TV signal or with another 66470, and a RAM-control mechanism allows you to operate the video RAM as a frame grabber.
Associated with the video RAM are two areas of memory that contain instructions to control the display. The first of these, designated the image control area (ICA), is accessed before the start of each frame-that is, during the display's vertical retrace period. The second area, designated the dynamic control area (DCA), contains as much as 64 bytes of information for each horizontal line in the display. The instructions in the dynamic control area for a particular display line are fetched and executed during the horizontal retrace period that precedes the display of that line.
The ICA/DCA instruction list in-
cludes instructions to reload the video start address, generate a processor interrupt, and change the border color. You can also include instructions to control a back-end processor that's connected to the video output port. The ability to change the video start address on a frame-by-frame or line-by-line basis allows you to achieve a variety of special effects, including horizontal and vertical smooth scrolling, or division of the screen into a number of horizontal subscreens.
To manipulate individual pixels in the image, you can use the 66470 's on-chip pixel accelerator rather than the system CPU. This accelerator can perform pixel operations in a single instruction cycle that would take several instruction cycles if you performed them using the 68000's instruction set. Operations that the
pixel accelerator can perform include copying source pixels to destination pixels, exchanging source and destination pixels, transferring source pixels to destination pixels with an associated color change to the current foreground or background color, changing pixels to the current foreground or background color according to a bit-map, and testing pixel color.
You can make most of these operations conditional on whether or not the source pixel's color coincides with the currently defined transparent color. Other pixel accelerator operations, which can execute concurrently with some of the operations described above, include horizontal zooming or shrinking of the display by a factor of 2 , transparency test inversion, and a range of logical operations on the result of the pixel operation.
The pixel accelerator doesn't operate as a coprocessor that needs to
take control of the CPU's address and data bus. Instead, you use the CPU to set up the required pixel operation in the pixel accelerator, then use the CPU to move data to and from registers in the pixel accelerator. The pixel accelerator is automatically triggered to perform the pixel operation when the CPU writes data to the appropriate register.
To further speed image manipulation, you can program the pixel accelerator to operate with implicit addressing. In this mode, the appropriate register in the pixel accelerator is automatically connected to the dynamic RAM's data bus when the CPU reads or writes data from or to memory. As a result, you can execute many pixel operations with a single 68000 move instruction. In addition to controlling the pixel accelerator from the system's main CPU, you can also control it via the 66470's coprocessor interface.

The 66470 is a CMOS part that operates from a single 5 V supply and draws a maximum operating supply current of 65 mA at a clock frequency of 30 MHz . It is packaged in a 124 -pin quad flatpack, surfacemount package. Initial samples of the 66470 have been implemented in a gate array and are available at a cost of around $\$ 50(100)$. The company expects that full custom parts with enhanced functionality will be available in 1989 at a cost of around $\$ 25$ (100), but that a high-volume price for the $68070 / 66470$ chip set will be around $\$ 30$ by 1991.

## -Peter Harold

Philips, Components Division, 5600 AM Eindhoven, The Netherlands. Phone (040) 757189. TLX 51573.

Circle No 726
Signetics Corp, 811 E Arques Ave, Sunnyvale, CA 94088. Phone (408) 991-4571.

Circle No 727


See EEM Vol. C, pgs. 1185-1190



## Big 8-bit performance. Tiny 18 -pin package. Miniscule price.

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since demand for 28 products, and their umportance to the indlustry conilinues to increase, we are developus Super $8^{*}$ is not dearly estabiabled as the high end 28 , and the Mioro8 can be expected to sate as mico un with be other new 78 Fanily memberc you should be Retping an gee on are:

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of on board
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Mighty impressive bit bang for your buck
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You really ought to see for yourself just what the mighty Micro8 can mean to your design application. Why wait? Contact your local Zilog sales office or your authorized distributor today. Zilog, Inc., 210 Hacienda Avenue, Campbell, CA 95008, (408) 370-8000.

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# Data logger acts as system analyzer in both benchtop and field environments 

If your embedded-control application includes the 6800 or 68000 family, consider using the Step- 120 Mo bile Incident Logger (MIL). Although it's geared specifically for the automotive industry, the unit offers a combination of features that also make it suitable for use as a system analyzer in both benchtop and field environments.
The MIL requires an IBM PC or compatible computer to act as a host for data transfer and for initialization. Once initialized, however, the unit operates independently of the host and doesn't need to remain connected to it. Nonvolatile RAM in the unit retains the setup conditions and the acquired data, so you can use the data logger to collect data in the field as well as on the test bench.

The unit records three types of data: processor-bus data, RAM variables, and external events. You can specify match (or comparison) words for each data type, using the match words for triggering and for data filtering. When filtering, the unit stores only those samples that meet the match-word conditions. The data logger can store 32 k samples, which you can configure as a single-sample memory or as four 8 k -sample memories for repeated sample runs.

The bus section of the unit monitors the behavior of the processor. The unit will record a 16 -bit address bus, an 8 -bit data bus, and a 3 -bit status bus. (You can upgrade these bus widths to 24 bits, 16 bits, and 8 bits, respectively.) In addition, you can have the unit monitor a series of RAM addresses that you've defined. The data logger then records any transactions at those addresses.

A special feature of the unit is its code-coverage mapper. This mapper


This multipurpose instrument functions as a data logger, logic-state analyzer, and chart recorder. Together with a host computer, the Step-120 Mobile Incident Logger offers a versatile tool for embedded-control applications.
records all addresses within a 64 k address range that have been accessed by a particular type of bus cycle. You can use this feature to verify that test programs completely exercise your code. You can also use it to identify frequently used sections of code, highlighting areas where improvements in code efficiency would have the greatest impact.

The data logger has four analog channels and eight event channels. The analog inputs have a selectable range of 0 to 5 V or 0 to 20 V ; they offer 8 -bit A/D resolution. The event channels work at logic voltages, but can tolerate as much as 125 V continuously.

When you've collected your data, you must use the host computer to retrieve it. Once the host contains the data, you can access that data in a variety of ways. You can view the complete data record with the data
types interleaved chronologically, or you can examine only one or two data types. You can plot the analog data as a function of time. And you can view the bus activity as dissassembled code or as source code from some C compilers.

The unit operates from dc power with a 9 to 16 V input range, which reflects the instrument's automotive orientation. Its operating-temperature range is -25 to $+65^{\circ} \mathrm{C}$, and it tolerates vibration and mechanical shock. Without a host, the data logger costs $\$ 14,900$ and is available for delivery within 30 days ARO.
-Richard A Quinnell
Step Engineering, 661 E Arques Ave, Sunnyvale, CA 94088. Phone (408) 733-7837.

Circle No 728

## IGBTs. The ${ }_{\wedge}$ comparisons.

## Compare Power



Compare Speed


The choice is now obvious when you require both power and switching speed in your power systems. Insulated Gate Bipolar Transistors (IGBTs) offer both.
Similar to Bipolar, IGBTs have high conductivity with faster switching speeds ( $t f=0.5 \mu \mathrm{~S}$ typ). And like MOSFETs, IGBTs have high input impedance/high speed with lower on-resistance (as low as 8 m ).

So now, for applications like low-noise inverters and servo-motor drives, there really isn't any reason to compromise. The solution you've been looking for is available: IGBT. The performance combination that doesn't make you give up one thing to get another.

Leading the way, again, this time with the broadest line of IGBTs anywhere in the world. Toshiba.
TOSHIBA AMERICA, INC. Semiconductor Products Division

## Compare Selection



[^5]
# Industrial IBM PC/AT-compatible system employs an $80386 \mu \mathrm{P}$ and 23 STD Bus slots 

Targeting industrial-control applications, the STD Busmaster-386 system couples a $16-, 20-$, or $25-\mathrm{MHz}$ $80386 \mu \mathrm{P}$ to the STD Bus. Compatible with the IBM PC/AT and MS-DOS, the system provides 23 STD Bus slots. It also includes a watchdog timer and an individual interrupt acknowledge/disable function.

The system operates in high-vibration and electrically noisy environments, and works over a 0 to $70^{\circ} \mathrm{C}$ temperature range. The company packages the system in an ano-dized-aluminum card cage that you can mount in an EIA-standard 19-in. rack or a NEMA-rated box.

To couple the 80386 to the STD Bus, the designers partitioned the system into two functional and physical sections. The $80386 \mu \mathrm{P}$ resides on a PC/AT card that hosts as much as 16 M bytes of 32 -bit dynamic RAM and an optional 80387 math coprocessor. The board plugs into a passive 3 -slot PC/AT-compatible mother board that is positioned in the upper portion of the card cage.

A second PC/AT-compatible board includes a video controller that is EGA, CGA, and Hercules compatible. The multifunction card includes two serial ports, a parallel port, a dual-floppy-disk-drive controller, and a SCSI host adapter. The company offers an optional ST-506 Winchester controller or LAN-controller board for the third slot. In addition, you can integrate the board of your choice in the system.

The lower portion of the card cage houses the 23 STD Bus slots and a 24th slot that interfaces the PC/AT bus to the STD Bus. A plug-in board acts as the interface between the


An $80386 \mu$ and a 23-slot STD Bus card cage suit the STD Busmaster-386 system for industrial applications that require more computational power than typical STD Bus CPU cards offer.
two buses. The 24th slot employs a DIN connector. The 80386 can address 128 k bytes of 8 -bit STD Bus memory and 504 contiguous 8 -bit I/O locations.
The STD Bus operates as if a 4 -, 6 -, or $8-\mathrm{MHz} \mathrm{Z80}$ were the CPU. The bus-interface card passes wait states, generated by STD Bus cards, directly to the 80386 . The CPU automatically inserts wait states on memory and I/O accesses to the STD Bus to emulate a Z80. You can select the STD Bus's clock frequency to be the CPU clock frequency divided by 2 or 4 , or you can select it to be 7.16 or 3.58 MHz .
Eight 3 -slot groups generate individually vectored interrupts that you can separately enable and disable. You can enable or disable the watchdog timer with software. You
can program timeout selection to 0.25 or 1.0 sec . The system's brownout reset activates a system reset when the $V_{\text {CC }}$ drops below 5 or $10 \%$ (the percentage is software selectable) of the nominal inputs.
The company offers models for MS-DOS version 3.3; Unix-386 System V; Xenix System V; and QNX, a real-time version of Unix. The company will offer 8088- and 80286based systems and a system with only eight STD Bus slots. A $16-\mathrm{MHz}$ model with no DRAM installed, a 23 -slot card cage, and MS-DOS costs $\$ 3995$. Delivery, stock to 90 days.-Maury Wright
Computer Dynamics Sales, 107 S Main St, Greer, SC 29651. Phone (803) $877-8700$.

Circle No 731

## THETHRIL LOF DESIGNING A A SUPPORTCRCUTRY ENDSINAFLASH.

The thrill of designing voltage references, input buffers, timing, and adjustment circuitry is gone. We do it for you. And in hybrid


FFT of HS 1068 Flash Converter, Fin $=1.123 \mathrm{MHz}$, Fsample $=20 \mathrm{MHz}$ packages smaller than you'd expect for flash converters alone.

Consider these three "pop-in" converters. Our 8-bit HS1068 flash A/D samples at 20 MHz minimum, has no spurious or missing codes, aperture jitter of only 60 ps , and true $1 / 2$ LSB 8-bit 47dB signal-to-noise ratio. Our 8-bit SP1070 samples at 25 MHz minimum and requires only $1 W$ of power. And our dual-flash SP1072 provides the performance of two SP1070s while saving you considerable real estate.


FFT of HS1068 Flash Converter, Fin $=9.678 \mathrm{MHz}$, Fsample $=20 \mathrm{MHz}$

Naturally, we test each device to ensure maintenance of SNR as input frequencies climb. They'll be distor-tion-free in your hypersensitive DSP applications.

And, we are the only company you can rely on to deliver full MIL-STD-1772 and MIL-STD-883C, self-contained, hybrid flash converters off the shelf. For data sheets, or a copy of our 382-page 1988 catalog, write SIPEX Corporation, Six Fortune Drive, Billerica, MA 01821, or call 1-800-272-1772. In Massachusetts call (617) 663-7811.


# Digital scope simultaneously captures data from two sources at 500 M samples/sec 

When Tektronix engineers set out to design the Model 2440 digitalstorage oscilloscope, the company was concerned, of course, with performance and price. But equally important was ease of use; using the scope had to feel natural to people accustomed to using high-performance analog scopes. The result is a product that looks and acts much like a familiar analog instrument, yet provides many more features and operating conveniences. According to Tektronix, it offers the industry's highest real-time sampling rate- 500 M samples/sec.

The instrument's predecessor, the Model 2430A, which the company will continue to sell, has achieved a high degree of acceptance. The 2440 has the look and feel of the 2430 A , but its $300-\mathrm{MHz}$ vertical bandwidth improves on the earlier model's bandwidth by a factor of 2 ; its ability to take 500 M samples/sec betters its predecessor's ability by a factor of 5; and its greatest sweep speed ( $2 \mathrm{nsec} /$ div in equivalent-timesampling mode) is faster by a factor of 2.5 . (In real-time sampling mode, the 2440 's 100 -nsec/div max sweep speed is 5 times that of the 2430A.)

Moreover, the 2440 sacrifices none of the 2430A's performance. The simultaneous sampling of two inputs, 8 -bit resolution, $2-\mathrm{mV} / \mathrm{div}$ max sensitivity ( $200 \mu \mathrm{~V} /$ div in average mode), and 1024-point record length are unchanged, as are the $6.3 \times 13 \times 18.9$-in. dimensions and the $24-\mathrm{lb}$ weight. Furthermore, the cost of the 2440 is only $29 \%$ more than that of the 2430A.

Its conveniences include auto setup that automatically configures the sensitivity and sweep speed to let you see the waveform at the input you select in the mode of your


The front panel of the 2440 digital-storage oscilloscope intentionally resembles those of high-performance analog scopes, because market research has indicated that users want scopes whose operation is intuitive to them.
choice. Auto pass/fail testing allows you to define a waveform envelope. You can then operate the scope unattended, and if any waveform falls outside your predefined envelope, the scope will store it and record the time at which it occurred. An automeasure mode can, among other things, provide a numeric display of 20 parameters of a captured waveform.

The scope's dual timebases provide many display modes. For example, you can delay the A or B sweeps by a number of B-trigger events that you select ( $2^{16} \max$ ). You can also supply your own sampling clock with a frequency from 1 to 100 MHz . A glitch-capture feature enables the scope to recognize pulses that are 2 nsec wide at the $50 \%$ amplitude points (at any sweep speed-with $85 \%$ probability). Ex-
cept at the highest sweep speed, such pulses could fall between samples.

Among the unit's interfacing features is an IEEE-488 port that can send waveforms, cursor measurements, and instrument configuration to a company HC 100 color pen plotter, an HP 7400 Series plotter, or an HP Think-Jet printer. A rearpanel output also provides an analog representation of the channel-2 waveform. The price of the 2440 is $\$ 11,500$, and the 2430A costs $\$ 8900$.
-Dan Strassberg
Tektronix Inc, Box 1700, Beaverton, OR 97075. Phone (800) 4262200.

Circle No 725


## When You WantFast 1 Mb EPROMs.

Mitsubishi delivers with a family of $150 \mathrm{~ns}, \mathrm{CMOS}$ 1 megabit UV EPROMs. Available now. And, for applications requiring even higher performance, 120 ns versions are on the way.

Now, you can combine the advantages of fast access times and low power CMOS operation in the organizations and pin-outs you need.

There's a 128 K x 8 that's pin-compatible with 1 Mb mask ROMs for easy conversion when you reach high volume production. Another device offers the JEDEC-standard pinout allowing easy upgrade from lower density $E P R O M$ s. Plus, there's a $64 \mathrm{~K} \times 16$, ideal for applications requiring wider data paths. And, they're all available in 150 ns speeds. From Mitsubishi.


## When You Want Cost-Effective Packaging Options.

If you want all the advantages of Mitsubishi EPROMs, but don't need to reprogram, Mitsubishi's $1 \mathrm{Mb}, 200 \mathrm{~ns}$ CMOS OTP ROMs provide cost-effective alternatives for volume production.

And, for maximum flexibility, Mitsubishi's 1 Mb OTP ROMs are available in PDIP and PLCC packages, with SOP available soon.

For fast EPROMs, or cost-effective packaging options, call or write: Mitsubishi Electronics America, Inc., Semiconductor Division, 1050 East Arques Avenue, Sunnyvale, CA 94086. (408) 730-5900.

| MITSUBISHI EPROMs |  |  |  | Access Time (ns) | Package (CERDIP) | MITSUBISHI OTP ROMs |  |  |  | $\begin{array}{\|l} \hline \text { Access } \\ \text { Time (ns) } \end{array}$ |  | Package Options |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Density | Organization | Part No. | 100120150170200250 |  |  | Density | Organization | Part No. | 200 | 250 |  |
| CMOS | 128K | $16 \mathrm{~K} \times 8$ | M5M27C128K | ■ ■ | 28 pin | CMOS | 256 K | $32 \mathrm{~K} \times 8$ | M5M27C256 | $\square$ |  | 28 pin PDIP and SOP |
|  | 256 K | $32 \mathrm{~K} \times 8$ | M5M27C256K | - $\quad$ - | 28 pin |  | 1 Mb | $128 \mathrm{~K} \times 8$ | M5M27C100 | $\square$ |  | 32 pin PDIP, PLCC and SOP |
|  | 512K | $64 \mathrm{~K} \times 8$ | M5M27C512AK | - ■ | 28 pin |  | 1 Mb | $128 \mathrm{~K} \times 8$ | M5M27C101 | ■ |  | 32 pin PDIP, PLCC and SOP |
|  | 1 Mb | $\begin{aligned} & 128 \mathrm{~K} \times 8 \\ & 64 \mathrm{~K} \times 16 \end{aligned}$ | M5M27C100K/M5M27C101K M5M27C102K |  | $\begin{aligned} & 32 \text { pin } \\ & 40 \text { pin } \end{aligned}$ |  | 1 Mb | $64 \mathrm{~K} \times 16$ | M5M27C102 | ■ |  | 40 pin PDIP and 44 pin PLCC |
| NMOS | 64K | $8 \mathrm{~K} \times 8$ | M5L2764K | - | 28 pin | NMOS | 64 K | $8 \mathrm{~K} \times 8$ | M5M2764 |  | ■ | 28 pin PDIP |
|  | 128K | $16 \mathrm{~K} \times 8$ | M5L27128K | - | 28 pin |  | 128K | $16 \mathrm{~K} \times 8$ | M5M27128 |  | - | 28 pin PDIP |
|  | 256K | $32 \mathrm{~K} \times 8$ | M5L27256K | - | 28 pin |  | 256K | $32 \mathrm{~K} \times 8$ | M5M27256 |  | $\square$ | 28 pin PDIP and SOP |
|  | 512 K | $64 \mathrm{~K} \times 8$ | M5L27512K | - ■ | 28 pin |  | 512 K | $64 \mathrm{~K} \times 8$ | M5M27512 |  | $\square$ | 28 pin PDIP and SOP |

[^6]
## NEW 2400 BPS CCITT V. 22 BIS SINGLE-CHIP MODEM



- One-chip multi-mode modem IC for V. 22 bis / V. 22 / V. 21 and Bell 212A/ 103 applications
- FSK ( 300 BPS), DPSK ( 600,1200 BPS), or QAM (2400 BPS) encoding
- All modem functions included in a single chip
- Integrated DSP for high performance adaptive equalization receive capability
- Fully compatible with SSI K212, K221, and K222 1-chip modems
- Interfaces directly with standard microprocessors (8048, 80C51 typical)
- Single +12 V or +5 V supply
- CMOS technology for low power consumption ( 120 mW @ 5 V )
Silicon Systems now offers the industry's most highly integrated modem IC-the SSI K224. It is a single-chip modem IC that provides all the functions needed to construct a V. 22 bis compatible modem, capable of 2400 BPS full-duplex operation over dial-up lines. The SSI K224 offers excellent performance and a high level of functional integration in a single 28 pin DIP. This device meets world-wide standards and supports all modes of operation, allowing both synchronous and asynchronous communication. The SSI K224 is ideal for use in either free-standing or integral system modem products such as lap-tops, P.C.s and portable terminals, or wherever full-duplex 2400 BPS data communications over the 2 -wire switched telephone network is desired The SSI K224 is pin and software compatible with the SSI K212, K221 and SSI K222 single-chip modem IC's, allowing system upgrades with a single component change.
For more information on the SSI K224 and the complete SSI $K$-Series modem IC family, contact: Silicon Systems, 14351 Myford Road, Tustin, CA 92680 Phone: (714) 731-7110, Ext. 575.


# Modular 1500W power system suits fault-tolerant designs 

By using the 29D Series modular power supplies with the company's Powersystem enclosures, you can implement 2 -, 4 -, or 6 -supply faulttolerant subsystems. The supplies' power ratings range from 225 to 300 W . A 6 -supply fault-tolerant subsystem can provide 1500 W . When used in parallel, the supplies feature automatic load sharing. The family of products targets applications in telecommunications, data communications, and computer systems.

To implement a fault-tolerant subsystem, you simply connect one more supply than you need. For example, six 300 W supplies actually provide 1800 W in parallel, but only provide 1500 W when a supply fails. The company offers isolation diodes for failed supplies, and the enclosures allow for "hot plug-in"-in other words, they let you replace a supply without taking the system off line.

The supplies provide five isolated floating outputs with voltages ranging from 5 to 24 V dc. The first output supplies 5 V and 30 or 40 A . Outputs 2 through 5 offer combinations of $5,12,15$, and 24 V .

The supplies meet all domestic and international safety standards. They require 110 or 220 V ac input power, which is field selectable. You can adjust the first four outputs to $\pm 5 \%$ of the nominal output voltage. The fifth output tracks the first output. The supplies' line regulation is $\pm 0.2 \%$ on all outputs. The first four outputs provide $\pm 0.2 \%$ load regulation. The semiregulated fifth output features a $\pm 5 \%$ load regulation. A step-load change or a $25 \%$ shift in rated maximum load causes an output change of less than $3 \%$. The output recovers to within $1 \%$ in less than $400 \mu \mathrm{sec}$.


Six modular power supplies in parallel implement a 1500W fault-tolerant subsystem when mounted in the Powersystem enclosure. The enclosure offers "hot plug-in" capability; that is, it lets you replace a supply without taking the system off line.

To create fault-tolerant subsystems, you can connect the supplies in parallel. You can parallel the first output of a supply only with the first output of another supply. However, you can parallel outputs 2 through 4 within the same supply or with outputs from other suppliesas long as the outputs have equal voltages and polarities. When connected in parallel, outputs 1 through 4 share current proportionally within $\pm 10 \%$ of each output's $50^{\circ} \mathrm{C}$ current rating (for forced-air cooling). You can't use the fifth output for parallel connections.

The enclosures are 7 in . high and $17 \frac{1}{2}$ in. deep, and the widths vary: The 2 -, 4 -, and 6 -supply units are 5.7, 11.3, and 19 in . wide, respectively. A supply with isolation diodes and outputs of $5 \mathrm{~V} / 30 \mathrm{~A}, 12 \mathrm{~V} /$ $3.75 \mathrm{~A}, 12 \mathrm{~V} / 2.25 \mathrm{~A}, 24 \mathrm{~V} / 3 \mathrm{~A}$, and $12 \mathrm{~V} / 2.25 \mathrm{~A}$ costs $\$ 508$ (25). The enclosures range from $\$ 158$ (25) for a 2 -supply unit to $\$ 484$ (25) for a 6 supply unit. Both the supplies and the enclosures are available from stock.-Maury Wright

Bonar Powertec, 20550 Nordhoff St, Chatsworth, CA 91311. Phone (818) 882-0004. TLX 277483.

Circle No 730

# LET THIS NEW MODEM CHIP make liff east for you 

## The Only 5-Volt, 2400 BPS, l-Chip Modem For Low-Power Applications

 If modem chips have a place in your company's products, here's how a new lowpower single-chip modem from Silicon Systems can make life easier for everyone.Now your product designers can put this new l-chip 2400 BPS modem into their designs and replace three or more oldstyle modem IC's. The K224L draws less than 120 mW from a single 5 -volt supply, eliminating the need for additional power supplies and making integration of the modem function possible in lap-tops, portable terminals, and other batteryoperated applications.

Your marketing department will be happy to know that this single chip fully supports all the world-wide operating modes-V. 22 bis, V.22, V.21, and Bell 212A/ 103. They'Il like the competitive edge that state-of-the-art single-chip modem technology can give to your company's products. And, of course, your management team can only be proud of an engineering and marketing group that has the good sense to make the most of such innovative components that enhance the features of your products, reduce costs, and boost company sales.

Call Now! (714) 731-7110, Ext. 3575

For more information on the new Silicon Systems 73K224L, or any of SSi's fully compatible K-Series 1-chip modems, contact: Silicon Systems, 14351 Myford Road, Tustin, CA 92680.

"Where we design to your applications."



## If you want to increase the value of your PWB real estate, Du Pont can help you do it.

With VACREL ${ }^{*}$ photopolymer film solder mask. The material that enables you to reliably encapsulate dense circuitry and tent via holes, permitting more electrical functions in less PWB real estate; to let you create higher density PWBs than ever before.

Instead of placing vias and components on separate areas of the board, you simply design vias under the components. And because VACREL tents the vias, solder stays only in the places where you want it; you avoid bridging, shorts, and contamination. It's also easier to clean under components.

In addition, you use less solder, minimizing assembly weight, because you're only covering connections. Hole-filling defects are also avoided. The bottom line is VACREL can increase assembly yields and cut manufacturing costs per function allowing you to get the most from your PWB real estate.

If you want more value out of your PWB real estate, ask for VACREL photopolymer film solder mask. It's the right material for PWB designs. Just call Du Pont at 1-800-237-4357. Or write: Du Pont Company, Room G-51176, Wilmington, DE 19898.

## READERS' CHOICE

Of all the new products covered in EDN's May 12, 1988, issue, the ones reprinted here generated the most reader requests for additional information. If you missed them the first time, find out what makes them special: Just circle the appropriate numbers on the Information Retrieval Service card, refer to the indicated pages in our May 12, 1988, issue, or use EDN's Express Request service.


## TURBO C DEBUGGER

The A-COM debugger is a full-screen, source-level debugger for use with Borland International's (Scotts Valley, CA) Turbo C programming tools (pg 276).

## A-Com.

Circle No 605


## $33 / 4$-DIGIT DMM

The DM79 can provide 10 times the resolution supplied by $31 / 2$-digit meters on many measurements (pg 258).

Beckman Industrial Corp. Circle No 604

## RACK-MOUNT PC/AT

The QPC-7000 IBM PC/AT-compatible computer meets EIA stand-
ards for mounting in a standard
19-in. rack (pg 244).
Qualogy Inc.
Circle No 603


- CMOS EPROMs

The 27 F 64 and 28 F 256 flash memories are 64 k -bit and 256 k -bit devices, respectively, that feature in-circuit electrical erasing and reprogramming (pg 88).
Intel Corp.
Circle No 601

## 5-YEAR BATTERY

The Model PS-1242 is a 12 V , sealed lead-acid battery. It's maintenance free, rechargeable, usable in any position, and suitable for both standby and deep-cycle applications (pg 243).
Power-Sonic Corp.
Circle No 602

# NEW 500 MEGASAMPLES/SEC 300 MHz DIGITAL PORTABLE 



# TEK PUSHES THE LIMIT... 

## AT THE PUSH OF A BUTTON.

## AUTO SETUP




AUTO MEASURE

AUTO PASS/FAlL



AUTO SEQUENCE

## ONLY FROM TEK.

The new Tek 2440 Digital Oscilloscope is the only portable offering a $500 \mathrm{MS} / \mathrm{s}$ sampling rate-on two channels simultaneously.

## Capture signals competitors can't.

Complete confidence in results.

You can digitize, store and process single-shot phenomena to 200 MHz and repetitive signals to 300 MHz . At the touch of a button.
This is a scope built for solving tough problems fast.

Exclusive Tek digital features let you capture glitches as short as 2 ns with continuous sampling; automatically catch intermittent failures in babysitting applications; and see waveform changes virtually as they occur thanks to the high update rate.

With the 2440's $500 \mathrm{MS} / \mathrm{s}$ sample rate, 8 -bit vertical resolution, and $0.0015 \%$ crystal-controlled timebase accuracy, you'll have complete confidence in the accuracy of your waveform data. You'll know what you're seeing is what's actually happening in your circuit.

The 300 MHz 2440 and 2432 , along with the 150 MHz 2430 A , drive like the scopes you're used to. They feature a familiar front panel and simple, one-level menus.

A digital family you already know how to use.

Built-in automation simplifies operation even more. The scopes set up at the touch of a button. And you can characterize waveforms instantly: press another button and read out your choices from 21 different measurements.

On the design bench, the 2400
DSOs are effective standalone test systems. Full GPIB programmability, pass/fail testing and time-saving measurement sequencer make them equally powerful system components. With off-the-shelf software you can fully automate measurements and extend waveform analysis.

Don't settle for less. Get the total performance package only Tek delivers, feature for feature, spec for spec.

## Powerful system components.

## WHEN THE MEASUREMENT COUNTS, COUNT ON TEK.

No matter what you measure, Tek measures best.

The standard of excellence in portable scopes.

Analog or digital, standalone or automated, no other portables match Tek 2400 Series Oscilloscopes for sheer breadth of line, range of performance and wide choice of options.

Their proven reliability and built-in versatility are a result of Tek's dedication to superior waveform acquisition tailored to individual needs.

> Answers at the touch of a button.

Six four-channel analog oscilloscopes combine pushbutton simplicity and leading-edge performance. They bring unprecedented efficiency to your design lab, production line or field service site.

The top-of-the-line 2467 with Tek's patented microchannel plate CRT even makes single-shot pulses to 350 MHz visibleat the touch of a button.
A complete range of options further extend each scope's capabilities.
Scope for scope, feature for feature, Tek has the best solution for your measurement needs.

Ask your Tek representative for a demonstration. With the 2440 joining the family, there's an even

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 better selection-in sample rate, performance and price. For more information, please return the reply card or call Tek direct:1-800-426-2200.



# Our networks give more for less... less space, less time, less cost. 

Mepco/Centralab thick-film networks save space and design time in all of your circuits - computer, communications, military, automotive, industrial control and instrumentation.
Choose from our broad range of highly integrated standards - cost-effective conformal networks, EMI/RFI networks, ECL terminators, MIL-R-83401 networks, capacitor banks, and more. We can also custom-design a space-saving unit to meet your requirements, including MIL designs.
With quality-designed Mepco/Centralab networks, you get the extra advantage of ordering just one part instead of many discrete components. So you simplify your purchasing, handling, stocking and assembly operations. And, since you only have to install a single part that's $100 \%$ pre-assembled and tested, you automatically improve your system's electrical and mechanical performance, and its overall reliability.
Look into space-saving, time-saving Mepco/ Centralab networks for your commercial, industrial and military-grade products. They're your assurance of high reliability, plus just-intime delivery and competitive pricing.
Ask for our catalogs. Write to Mepco/ Centralab - the active leader in passive components.


## FDDI. <br> From deskwork to network.



Good news for networks!
The X3T9.5 Task Group, under the procedures of ANSI Accredited

Standards Committee X3, has reaffirmed approval of the Media Interface Connector (MIC) for the proposed FDDI (Fiber Distributed Data Interface) Physical Layer Medium Dependent (PMD) document.

More good news! AMP has the complete fiber optic interconnection system-the AMP OPTIMATE Fixed Shroud Duplex System-that meets all FDDI PMD requirements. And includes all the physical components you need to make your fiber optic network a reality.

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CIRCLE NO 77

By employing CAD/CAE and new fabrication technologies, pressure sensor and transducer manufacturers have been able to improve product performance without exacting any price penalties. In the silicon-based sensor area, in fact, prices are actually going down as performance figures improve-a combination that opens up a host of new applications.
$9{ }^{\circ}$

Pressure sensor and transducer manufacturers are offering less expensive, yet more capable, devices through the use of a largely automatic micro-machining process of producing silicon wafers.
 (Photo courtesy Micro Switch)

# Pressure sensors and transducers 

Tom Ormond, Senior Editor

Improved performance at a lower cost is the ultimate goal of any product designer. When it comes to strain-gauge-type pressure sensors and transducers, manufacturers have achieved this goal and made the devices practicable for numerous applications in the automotive, medical, and military fields. In addition, advances in device-fabrication and -packaging techniques allow today's pressure sensors and transducers -including devices based on older piezoelectric and linear-variable-differential-transformer (LVDT) tech-niques-to easily handle the harshest industrial environments. Manufacturers are also integrating more electronics into the device package, which makes it easier for users to apply the sensors and transducers.

Amplifiers represent one type of integration. Although most of today's transducers generate usable unamplified outputs, you can get products that offer logic-level output signals while operating from low voltage supplies. Transducer manufacturers also include integral electronics that address thermal-related problems. Just about every transducer available today includes temperature-compensation circuitry, which allows the units to provide very accurate output readings in environments involving significant temperature variations. On the other hand, some manufacturers also offer uncompensated products for those users working in a benign environment, who don't want to pay for something they don't need.

Three of the most popular sensor technologies are strain-gauge (piezoresistive), LVDT, and piezoelectric. Nowadays, you'll find that most of the newsworthy
technological developments concern strain-gauge types of sensors. These devices typically use one of three sensing elements: an unbonded metallic-filament strain gauge, a bonded metallic-foil strain gauge, or a bonded piezoresistive (semiconductor) gauge. Regardless of element technology, however, strain-gauge sensors and transducers find their way into diverse applications.

## Solid-state sensors-where the action's at

Gulton Industries' GMS110 Series of pellicloid thinfilm strain-gauge transducers, for instance, fit demanding military as well as aerospace applications. Packaged in a welded, hermetically sealed housing, the solid-state devices are available in versions that measure full-scale absolute or gauge pressures of 15 to 5000 psi .
The sensors consist of a thin-film circuit deposited on a micromachined silicon diaphragm. Pressure across the diaphragm produces strain in the metal-film sensing elements, causing slight changes in their resistances. These sensing elements are configured as a 4 -arm, fully active strain-gauge bridge with integral trimming resistors. The fabrication process places the compensation resistors close to the sensing elements to minimize thermal shift in the span and zero parameters: $0.005 \%$ full-scale output (FSO) $/{ }^{\circ} \mathrm{F}$ max.
The GMS110 Series operates from 10 V dc; the transducer output is $2.5 \mathrm{mV} / \mathrm{V} \mathrm{min}$. Repeatability and hysteresis are $0.05 \%$ FSO max, and nonlinearity equals $\pm 0.25 \%$ max. The overall operating range spans -320 to $+350^{\circ} \mathrm{F}$; the compensated range is -65 to $+250^{\circ} \mathrm{F}$. The transducers cost $\$ 1050$.

## Strain-gauge-type transducers find their way into diverse applications.



Featuring 0.2-mV/kPa sensitivity, Motorola's MPX2000 Series temperature-compensated transducers spec $\pm 0.05 \%$ full-scale linearity over -40 to $+125^{\circ} \mathrm{C}$.

Designed to provide absolute-pressure measurements in low-pressure applications, NPH Series sensors from NovaSensor are available in 0 - to $7-\mathrm{kPa}$ (approximately 0 to 1 psi ) and 0 - to $15-\mathrm{kPa}$ versions (approximately 0 to 2 psi ). The 0 - to $7-\mathrm{kPa}$ unit provides an unamplified output voltage of 75 mV . In OEM quantities, NPH sensors cost $\$ 12$. NovaSensor attributes its ability to achieve such a price/performance ratio to CAD-analysis and advanced-silicon-etching (micromachining) techniques (see box, "Computerbased techniques facilitate sensor design").

The NPH Series specifies respective nonlinearity and hysteresis errors of $\pm 0.2 \%$ and $\pm 0.05 \%$ max. Thermal hysteresis is less than $\pm 0.25 \%$ FSO, and thermal error is $1 \%$ max over the 0 to $70^{\circ} \mathrm{C}$ compensated operatingtemperature range. The sensors are also available in uncompensated versions, as are models that provide differential-pressure readings. All models are compatible with noncorrosive gas and moist-air media. A pressure port with an outside diameter of $3 / 16 \mathrm{in}$. is standard, but similar tubes or vent holes for either TO-5 or TO-8 packages are available to satisfy specific OEM requirements.

Endevco's strain-gauge Model 8515B is a rugged, miniature transducer designed for applications requiring surface-pressure measurements. It utilizes a sculp-


Offering 1- and 2-psi FS measurement ranges, NPH sensors from NovaSensor feature maximum nonlinearity and hysteresis errors of $\pm 0.2$ and $\pm 0.5 \%$, respectively.
tured, piezoresistive pressure-sensing diaphragm and readily mounts on curved surfaces without affecting laminar-air or hot-gas flow. For a flush fit, you can recess the transducer and lead wires into the mounting surface. The transducer includes a protective screen to protect against particle impingement.

The $\$ 495$ transducer is available in full-scale versions of 15 psia (pounds per square in. absolute) (Model 8515B-15) and 50 psia (Model 8515B-50). The $0.03 \times 0.25$-in. case contains hybrid temperature-compensation circuitry that stabilizes operation over the -65 to $+250^{\circ} \mathrm{F}$ temperature range. Full-scale output measures 300 mV . For the 8515B-50 and 8515B-15 versions, sensitivity is $6 \pm 2 \mathrm{mV} / \mathrm{psi}$ and $20 \pm 6.7 \mathrm{psi}$, respectively. Combined nonlinearity (hysteresis, nonrepeatability, and nonlinearity) equals $0.5 \%$ FSO. Thermal zero shift measures $4 \%$ max.

SCC Series strain-gauge transducers from Sensym are designed for cost-sensitive applications where precise accuracy over a wide temperature range is not a critical design consideration. Typical applications include pneumatic controls, automotive diagnostics, medical equipment, and environmental controls. Each member of the series costs $\$ 9.90$ and features an IC sensor element housed in a single-in-line package. You have a choice of three devices with full-scale pressure

## Computer-based techniques facilitate sensor design

Thanks to CAD and
micromachining techniques, the latest generation of silicon sensors are sophisticated devices with performance levels far exceeding those found in oldertechnology chips. Chips with onboard, laser-trimmed calibration and temperature-compensation resistors, for example, are now available in high volume. You can also buy extremely lowpressure sensor chips-less than 1 psi full scale-in large quantities. Today, maximum sensor output nonlinearity is routinely $0.25 \%$.

All these achievements require the most elaborate design, analysis, and manufacturing automation tools and techniques available. Finite element modeling (FEM) programs, for instance, are mechanical-analysis tools that designers can use to predict the operation and performance of complex 3-dimensional shapes such as sensors. Such tools allow designers to calculate and optimize the sensitivity, nonlinearity, resonant frequency, and damping coefficients of new sensor structures before making commitments to expensive tooling.

In addition, FEM programs enable designers to predict the difficult, but important, interactions between the chip and the package. High-performance, high-speed computer workstations, and programs such as ANSYS, make it possible to obtain highly accurate results. By reducing the development time
and experimental tooling and materials costs, modeling significantly reduces sensor cost. In effect, a designer can build, test, and optimize a sensor on the computer before any investment in actual hardware takes place.

Circuit-analysis software performs a function similar to that of FEM. According to the application requirements, the computer will integrate the sensor's electrical performance with that of the signal-conditioning circuitry. Designers can, for example, test active correction coefficients and temperature coefficients before actually building a sensor. And, they can use feedback from such computer testing to modify the required chip and circuit specifications to increase performance or decrease cost.
Test, calibration, and trim procedures also have become more sophisticated as the complexity of sensor chips increases. For advanced sensor systems, designers must use computers and modeling to obtain cost-effective solutions for electrical-design, manufacturing, test, calibration, and final-trim specifications.

As users demand sensors with higher accuracy specifications, the need for precision and longterm stability in semiconductor diffused or implanted resistors becomes crucial. Silicon process modeling can predict process variations. These predictions provide input on device specifications and distributions in these specifications before fabrication
of the first chips begins. Sensor designers can use these same models to evaluate new processing concepts for improvements in product yield, longterm stability, and device reliability.
After defining and modeling device specifications, designers must convert their engineering concepts into hard tooling.
Again, computer-design systems are essential. Computer-graphics terminals can quickly and accurately digitize the various mask levels necessary to fabricate sili-con-based sensors. These systems can also rapidly implement design modifications to speed up development schedules.
Computer-aided test systems complete the computer-aided design/analysis process. During the release of early-productionrun sensors, computer-aided test systems provide designers with immediate feedback on device performance and correlate this information with design goals. Computer-aided test systems also give engineering and manufacturing groups continuous, online, up-to-date, readily available information on yield and on the distribution of crucial specifications for mature sensors already in production.

## Acknowledgment

EDN would like to thank Kurt Peterson and F Pourahmadi of NovaSensor for their invaluable assistance in supplying background information for this box.

Transducers are available for applications where low cost is more critical than is precise accuracy over a wide temperature range.


Housed in convenient, low-cost single-in-line packages, Sensym's SCC Series sensors are designed for applications where precise accuracy is not a prime system consideration.
ranges of 5,15 , and 30 psig (pounds per square in. gauge). All units feature internal temperature compensation for operation over 0 to $50^{\circ} \mathrm{C}$. Although optimized for this temperature range, SCC sensors do operate from -40 to $+125^{\circ} \mathrm{C}$. The guaranteed span-temperature error is less than $0.75 \%$ FSO, and the combined linearity, hysteresis, and repeatability error is less than 0.5\% FSO.

IC Sensors' Models 151 and 154 are solid-state straingauge devices that utilize a 316 stainless-steel housing to provide compatibility with most harsh media. The devices are available in gauge-, sealed-gauge-, and absolute-pressure versions that handle full-scale readings of 5 to 300 psi . Static accuracy is $\pm 0.25 \%$, and the operating range spans -10 to $+80^{\circ} \mathrm{C}$.

Model 154 includes integral laser-trimmed resistors that provide calibration and temperature compensation over the 0 to $50^{\circ} \mathrm{C}$ range. When the sensor is drawing a $1.5-\mathrm{mA}$ supply current, the nominal output span is 100 mV . For the Model 151, you must provide compensation by using three external resistors (the manufacturer supplies the correct resistance values). Again, for a $1.5-\mathrm{mA}$ supply current, the nominal output span equals 100 mV . The zero-pressure output equals $\pm 1 \mathrm{mV}$, and the zero and span temperature coefficients correspond to $\pm 1 \%$. The sensors cost $\$ 60$ (OEM qty); delivery is stock to 60 days ARO.

Motorola's MPX2200 silicon, piezoresistive pressure


Compatible with various liquid and gas media, PT11 sensors from Revere feature a full-scale measurement capability ranging from 100 to $10,000 \mathrm{psi}$.
sensors each contain, on-chip, a monolithic shear-stress strain gauge, diffused thermisters, and a thin-film resistor network. The pressure sensors are laser trimmed and therefore provide span and offset calibration as well as temperature compensation over -40 to $+125^{\circ} \mathrm{C}$.

The MPX2200 Series costs $\$ 10$ in quantities of 50,000 and is available in either basic-element packages for absolute- and differential-pressure measurements or in ported packages for measuring differential, absolute, gauge, gauge-vacuum, gauge-axial, and gauge-axialvacuum pressure. Over the 0 - to $30-\mathrm{psi}(0$ to 200 kPa ) absolute- and differential-pressure ranges, the sensors' 0 - to $40-\mathrm{mV}$ analog output is extremely linear. At $25^{\circ} \mathrm{C}$, the linearity equals $\pm 0.1 \%$ FSO typ and $\pm 0.25 \%$ FSO max. The sensitivity is $0.2 \mathrm{mV} / \mathrm{kPa}$, and the typical zero-pressure offset is $\pm 0.05 \mathrm{mV}$. Temperature effects
on hysteresis, full-scale span, and offset are $\pm 0.5 \% \mathrm{FS}$, $\pm 0.5 \% \mathrm{FS}$, and $\pm 0.5 \mathrm{mV}$, respectively.

Micro Switch's 14PC Series are also piezoresistive, strain-gauge types residing on a silicon chip, but these sensors feature four piezoresistors, an integral sensing diaphragm, and are enclosed in miniature, wave-soldercompatible packages. Internally, an elastomer O-ring surrounds the pressure-sensitive diaphragm, leaving it exposed to pressure media applied through the ports, which are ultrasonically welded in place. You can select seal material to fit specific application requirements.

The series is available in both gauge- and differentialmeasurement versions with full-scale pressure ranges of 5,10 , and 15 psi . Repeatability and hysteresis accuracy translates to $\pm 0.75 \%$ FSO. Linearity is $0.5 \%$ full scale. The sensors operate from 10 V dc supplies and dissipate 20 mW . Their full-scale output, when operating from a 10 V dc supply, equals 250 mV . The price is $\$ 10.24$ (500).

## Handling harsh environments

Model 1224 pressure transducers from Foxboro/ICT boast a proprietary 316 stainless-steel/oil-media interface that isolates the solid-state piezoresistive sensing element from excessive shock and vibration. The units' 0 - to $100-\mathrm{mV}$ output has a $\pm 0.02 \%$ repeatability spec. Available in both gauge-pressure and absolute-pressure versions that cover full-scale pressure ranges from $15 \mathrm{psig} / \mathrm{psia}$ to 5000 psig , the sensors cost $\$ 135$. Standard devices have $\pm 0.5 \%$ FSO max static accuracy; $\pm 0.25 \%$ is available as an option. Maximum thermal error is $\pm 2 \%$ FSO over the -20 to $+180^{\circ} \mathrm{F}$ compensated temperature-operating range. The devices are designed specifically to accommmodate rugged operating conditions and come with a 3 -year warranty.

Revere's PT11 pressure sensor finds use as a primary control component in hydraulic systems, air-conditioning systems, engine monitoring, refrigeration units,

Designed for applications requiring surface-pressure measurements, Endevco's Model 8515B transducer features hybrid temperature-compensation circuitry for stable operation over -54 to $+121^{\circ} \mathrm{C}$.

compressors, and industrial-control systems. The seven models that make up the line feature full-scale measurement ranges spanning 100 to $10,000 \mathrm{psi}$. The average combined error for any unit equals $\pm 0.05 \%$.

The 100 -psi sensor is constructed of 2024-T351 aluminum; the $250-$, $500-$, $1000-$, $3000-$, $5000-$, and $10,000-\mathrm{psi}$ devices are constructed of \#17-4 PH stainless steel. The minimum burst-pressure limit for all models is greater than $500 \%$ of range, and the safe-overrange limit is $200 \%$ of full scale. The supply voltage can be 12 V ac or dc. The full-scale output is $2 \mathrm{mV} / \mathrm{V} \pm 20 \%$, and the operating range covers -40 to $+150^{\circ} \mathrm{F}$. Prices range from $\$ 40$ to $\$ 57$, depending on quantity.

The TJE high-precision, wet/wet differential-pressure transducer, available from Sensotec, is engineered for pressure ranges of 50 to 500 psi . The bidirectional device sells for $\$ 950$, specs $0.1 \%$ full-scale accuracy, and


Offering full control of variable-speed pump motors, the 150 B and 150C transducers from Robinson-Halpern are designed to solve measurement and control problems in sprinkler-system water tanks.

## Transducers with low power consumption and logic-level electrical outputs are ideal for remote or portable applications.

is designed to accommodate fluid or gas media in both ports. Features such as stainless-steel construction and built-in overload protection provide a high degree of durability in harsh industrial environments. The bond-ed-foil, strain-gauge-type Model TJE employs a 10 V dc supply to produce a standard $2-\mathrm{mV} / \mathrm{V}$ output; a 4 - to $20-\mathrm{mA}$ output is optional. The operating range extends from -65 to $+250^{\circ} \mathrm{F}$. Zero-offset and span temperature effects are both $0.25 \% /{ }^{\circ} \mathrm{F}$.

Viatran's 22 Series strain-gauge-type transducers are available with either a mV/V output (Model 122,
$\$ 479$ ) or 1 to 6 V dc output (Model 222, $\$ 525$ ). The use of military-grade components makes possible a -60 to $+250^{\circ} \mathrm{F}$ operating range. The units are available in versions with full-scale pressure-measurement capabilities of 50 to $10,000 \mathrm{psig}$ and psia . All feature an accuracy of better than $\pm 0.4 \%$ FSO.

The transducers are compensated for an operating range of 70 to $170^{\circ} \mathrm{F}$. Temperature effects on span and zero are $\pm 2 \%$ max. Model 122 operates from a 10 V ac/de supply voltage, whereas Model 222 operates from either 12 or 28 V dc. The full-scale output for the 122

## Manufacturers of pressure sensors and transducers

For more information on pressure sensors and tranducers such as those discussed in this article, contact the following manufacturers directly, circle the appropriate numbers on the Information Retrieval Service card, or use EDN's Express Request service.

varies from 1 to $2 \mathrm{mV} / \mathrm{V}$. Long-term stability equals $\pm 0.25 \%$ max for six months, and repeatability is better than $\pm 0.1 \%$ FSO. The units will mount in any position and have proof- and burst-pressure specs of 1.5 and 5 times rated pressure, respectively.

The SA devices from Data Instruments are designed to work in applications demanding a rugged, reliable transducer. The units' design isolates the pressure measurand from the electronic hybrid circuitry, allowing the transducers to accommodate media such as oil, gas, saline solutions, blood, hydraulic fluids, alcohol,
freon, ammonia, acids, and gasoline. Full-scale pressure ranges extend from 15 to 50 psia and 100 to 5000 psis (pounds per square in. sealed gauge). Each device is computer-trimmed for calibration and temperature compensation to ensure unit-to-unit interchangeability of $\pm 1 \%$. The transducers' 5 V output precludes the need for external amplification. Accuracy (including the effects of nonlinearity, hysteresis, and repeatability) is within $\pm 1 \%$ of span. Long-term stability measures $\pm 1 \%$ FSO over one year.

The SA transducers function from supply voltages in

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## Stainless-steel construction and built-in overload protection provide a high degree of durability in harsh industrial environments.

the 9 to 20 V dc range. Although the units have a 30 to $185^{\circ} \mathrm{F}$ compensated operating-temperature range, they will function from -67 to $+257^{\circ} \mathrm{F}$. Thermal effects on span and zero-offset voltages are $\pm 1 \%$. All units include reverse-polarity and RFI protection as standard features. They cost $\$ 209$.

## Foil-gauge units satisfy low-power needs

Boasting a current consumption of only 1 mA , Trans Metrics' P21L transducers are well-suited to applications where low power consumption is a primary design consideration. In addition, the units provide a 0 to 5 V output from 5 V de supplies. They employ foil strain gauges bonded to the nonwetted side of a machined stainless-steel diaphragm-type sensor. There is no contact between the measured fluids and the electrical circuits, so the transducers accommodate any corrosive fluid that does not attack stainless steel.
Relative-, absolute-, compound-, vacuum-, and differ-ential-pressure types are available in full-scale measurement ranges of 15 through $15,000 \mathrm{psi}$. All models cost $\$ 395$. The transducers will operate over the -30 to $+180^{\circ} \mathrm{F}$ range and are compensated for operation from 0 to $160^{\circ} \mathrm{F}$. Zero-balance and nonrepeatability-error specifications are $\pm 2 \%$ FSO and $0.2 \%$ FSO, respectively. The static error band (including nonlinearity, hysteresis, and nonrepeatability) measures $\pm 1 \%$ FSO for models below 150 psi and $\pm 0.5 \%$ FSO for all others.
The signal-conditioning circuitry consists of a low dropout regulator, a voltage doubler, a negative-voltage converter, and an amplifier. Because of the nega-


Developed for laboratory and biomedical use, PPS-1 sensors from Thermometrics monitor changes in both pressure and temperature.
tive-voltage converter, P21L transducers provide an output of 0 V at zero pressure, thereby eliminating the need to trim out any voltage offset. The signal-conditioning circuitry also contains field-adjustable zerobalance and span potentiometers. These adjustments are located behind weather-tight plugs.

The Model PA194SH from West Coast Research is a micropower transducer ideal for use in salt water, corrosive fluids, and hostile environments. Capable of working in a submerged condition, it operates from approximately 1 mW and provides a full-scale output of 1V or higher. High sensitivity is a key feature of the PA194SH. At 600 ft of water, for example, it is capable of resolving 1 part in $72,000(0.0014 \%)$. Output nonlinearity is less than $0.05 \%$. The unit is available with an output converter or transmitter to yield either BCD, 4to $20-\mathrm{mA}, 0$ to 5 V , or other control-signal outputs. The submersible version costs $\$ 1250$; a reduced-sensitivity version ( 1 part in 24,000 ) is available for $\$ 850$.

## Medical uses-no problem for foil units

The PPS- 1 from Thermometrics costs only $\$ 76$ and is designed for invasive use in measuring pressure at selected anatomical sites, and, more specifically, for use in thermal-dilution catheters and wedge-pressure catheters. It does accommodate noninvasive applications, too, however. The sensor features a distal tip mount, which attaches to a catheter tube. The silicone rubber diaphragm is at the front face of the sensor, and the


Offering wet/wet media compatibility, Micro Switch's 14PC Series is assembled using a TAB (tape automated bonding) process and elastomer O-ring seals.

## the beat goes on.... <br>  <br> . . . WITH SENSYM AT THE HEART.

## Pressure Sensors

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## Transformer-based transducers have long operational life and feature excellent resolution.

rear section serves as a shank that attaches to the catheter. Three wires and an atmospheric vent tube extend from the shank.
The sensor measures pressures over the range of -30 to +300 mm of mercury $(\mathrm{Hg})$, and it operates over the range of 15 to $45^{\circ} \mathrm{C}$. Overpressure capability extends from -100 to +1200 mm . The sensor operates from 5 to 10 V dc supply voltages. Sensitivity at $25^{\circ} \mathrm{C}$ equals 5 to $10 \mathrm{mV} / \mathrm{V} / \mathrm{mm} \mathrm{Hg}$. Zero balance is $\pm 75 \mathrm{~mm} \mathrm{Hg}$, and the combined linearity and hysteresis error equals $\pm 2 \mathrm{~mm}$ Hg . The zero-drift spec (over time and temperature) measures $0.3 \mathrm{~mm} \mathrm{Hg} /{ }^{\circ} \mathrm{C}$ max and $3 \mathrm{~mm} \mathrm{Hg} / 24$ hours, respectively.

The number and variety of available strain-gauge transducers attests to their popularity. Nonetheless, other technologies still command a piece of the market. LDVT transducers represent one such technology. Compared with strain-gauge types of sensors, LVDT transducers rely on a transformer coil for the sensing element. Such transducers are unaffected by mechanical strain, making them extremely reliable.

## LVDT transducers find an industrial home

Robinson-Halpern's Models 150B and 150C pressure transducers feature a wide zero- and span-adjustment capability that makes them well-suited to solving pres-sure-measurement and -control problems in auxiliary water tanks such as those used for sprinkler systems. The $\$ 325$ units can provide precise liquid-level measurement in elevated tanks or water towers even if the instrumentation is located at ground level. Using only a small portion of their pressure-range capability, they maintain full control of variable-speed pump motors. Users can compress the transducers' span from 100 to


Combining varied media compatibility with low cost, Model 151 and 154 sensors from IC Sensors measure full-scale pressures from 5 to 300 psi with $\pm 0.25 \%$ accuracy.
$15 \%$ of the rated range, and suppress the zero-offset voltage by as much as $85 \%$. Zero and span controls are externally accessible.
Models 150B and 150 C measure the pressure of liquid or gas media in gauge, differential, or absolute units. They are available in nine versions featuring full-scale capabilities of 1.5 to 300 psi ; accuracy is $\pm 0.5 \%$ of the user-calibrated range. The devices operate over the -40 to $+176^{\circ} \mathrm{F}$ range, and they are temperaturecompensated over the 25 to $150^{\circ} \mathrm{F}$ range. Their castaluminum housing has a NEMA-1 rating. An internal regulator, which features polarity-reversal protection, allows the devices to operate from unregulated power sources.

The P-3061, a member of the P-3000 Series from Schaevitz, has a low-pressure-measurement-range capability. It is available in gauge- and differentialpressure versions capable of measuring 0 to 2 in . of water to 100 psi . The transducer operates from an unregulated 10 to 32 V de supply, consumes only 1.2 mA , and provides a 0 to 5 V dc output. The P-3061's total static error is less than $\pm 0.5 \%$ of its full-range output (FRO); a $\pm 0.25 \%$ FRO spec is available as an option. The operating range extends from -40 to $+175^{\circ} \mathrm{F}$. The cost is $\$ 300$.

Kaman Instrumentation Corp also employs inductive technology in its pressure-sensor offerings. The KP1911 high-temperature pressure-measuring system is an integrated transducer system that operates at temperatures of $1100^{\circ} \mathrm{F}$. It includes a pressure sensor/ transducer (with 50,000-psi capability), cabling, and signal-conditioning circuitry. The cabling has two sections joined by a transition connector-a metalsheathed high-temperature section, and a flexible twisted-pair section that connects to the system's electronics.

The sensor can operate in media such as liquid sodium, steam, plastics, and petrochemicals. The system output is an analog voltage directly proportional to the applied static or dynamic pressure. The sensor uses internal impedance variations to measure static and dynamic absolute pressure. Gauge- and differential-pressure-measurement versions are also available. The gauge sensor has a vent that provides an atmospheric reference, and the differential sensor includes a tube that allows you to provide a dry-gas reference pressure.
The KP-1911's housing is alloy 718; the cable sheath is alloy 600 . The patented 2 -coil sensor design in essence cancels temperature effects and provides a compensated operating range of -100 to $+1000^{\circ} \mathrm{F}$. Each sys-

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# CAE software uses algorithms instead of schematics 

> Drawing logic schematics isn't the only way to create ASIC designs. You can also use algorithms to automatically generate the net list for your bardware design - and you can do so in about one-tenth of the time that it often takes to complete a schematic.

## Jay R Southard, Algorithmic Systems Corp

Using algorithms to design your hardware can simplify and clarify the design process as well as reduce by as much as $90 \%$ the amount of time and money you spend on it. Although engineers customarily associate schematic design with hardware and algorithmic design with software, this need not always be the case. Hardware and software are really just media for implementing your design. Schematics and algorithms, on the other hand, are design methods.

The Ascyn synthesizer transforms your algorithm directly into an ASIC foundry's net list. It also simulates designs, determines clock rates, and automatically generates test vectors. The package, written in LISP, is an artificial-intelligence logic synthesizer that can run on VAXs and personal computers.

Using this synthesizer, you can program your de-
sign's algorithm and then either compile it for a simulation run, or synthesize a net list for a specific ASIC foundry's process. After the software generates a net list, the ASIC foundry can tell you how much the circuit will cost to make.

The synthesizer is neither a schematic-capture package nor a silicon compiler. Silicon compilers work by converting high-level schematic specifications into ASIC layouts. The synthesizer, on the other hand, translates algorithms directly into net lists. How the manufacturer lays out the hardware devices is irrelevant to the algorithmic synthesizer's operation.

The algorithmic synthesizer isn't great for every type of design. Designs larger than 10 pages of Ascyn code probably won't fit in 10,000 -gate gate arrays, which are generally the largest available. Moreover, any synthesized ASIC will of course be faster than the corresponding algorithm running on a standard computer, but it's unlikely to be more than 50 times faster at best.

Synthesized designs can be larger and slower than carefully crafted, schematic designs. At the 1000 -gate level, the average synthesizer-generated ASIC is about $10 \%$ larger than the best schematic implementation of the same function. However, at about 2000 gates and above, the algorithmic synthesizer is better than schematic capture at minimizing chip size because a synthesizer looks over the entire IC and finds shareable resources.

You can't make a general, quantified comparison of
the speed of an algorithmically designed ASIC with the speed of a schematically designed ASIC. Rather you must evaluate them on a case-by-case basis.

## Three key traits characterize synthesizer

The synthesizer's ability to transform algorithms into net lists stems from three key characteristics of the package: its frequent use of master-slave registers, its method of assigning operators to operations, and the way it implements Boolean functions.

Although the master-slave register plays only a minor part in schematic designs, it's fundamental to algorithmic designs. If you write in any computer language:

$$
\mathrm{B}:=\mathrm{B}+\mathrm{A}
$$

what you really mean is:

- Sample the values of variables on the right-hand side of the assignment operator.
- Perform the indicated operation (in this case an addition).
- Hold the resulting value.
- Update the values of the variable on the left-hand side of the assignment operator with the held results.
This sequence defines exactly what a master-slave register does.

The synthesizer's master-slave registers have three inputs: data (word-wide), latch enable (a single-bit signal), and clock. If the latch-enable signal is 0 , then the register retains its stored value; if it's 1 , then the register gets updated in accordance with the data input, the clock, and the master-slave rules. In Figs 1, 2 , and 3 , the master-slave registers are boxes with parallel data inputs and a latch-enable single-bit signal. The clock line is implicit.

## Operators vs operations

The synthesizer's second key characteristic is its method of assigning operations to operators. For the purposes of this article, operations are the primitive elements of an algorithm. They are the instructions that apply to specific variables at specific times. In contrast, operators are the primitive elements of a schematic. They represent the objects that are eventually connected together to make an integrated circuit. Good logic design involves the art of packing the required operations into the least expensive, highestperformance set of operators.


Fig 1-The algorithmic logic synthesizer transforms a simple limit-checking code fragment into this circuit.


Fig 2-The synthesizer automatically condenses the redundant operations in (a) into a single operator in (b).

Using the synthesizer, you can try various design options. You can, for example, write a series of operations either sequentially or in parallel. Thus you can write two lines of operations sequentially:

$$
\mathrm{A}:=\mathrm{A}-1 \quad \mathrm{~B}:=\mathrm{B}+\mathrm{A}
$$

or in parallel:

$$
\begin{aligned}
&(\mathrm{par} \\
& \mathrm{A}:=\mathrm{A}-1 \\
& \mathrm{~B}:=\mathrm{B}+\mathrm{A})
\end{aligned}
$$

The parallel construction includes two operations in a single cycle. The synthesizer can implement the sequential operations with only a single physical operator that can alternately add or subtract (such as an ALU), but not do both at the same time. The parallel version, on the other hand, needs two physical operators to execute both operations simultaneously: It's faster but it requires more hardware.

In some cases, the synthesizer only has to assign a
separate operator to each operation. For instance, consider this small fragment of code in Fortran:

> .NOT. (LIMIT .EQ. (COUNT + 1))

The corresponding piece of Ascyn code is:
(not (= limit (inc1 count))

Working under the assumption that these operations must execute in one clock cycle, the synthesizer transforms this algorithm into the circuit in Fig 1 (the program can generate schematics of this sort at any point in the design cycle). The thick lines in Fig 1 highlight word-wide data paths; the thin line is an unregistered, single Boolean signal (a "wire" variable in the Ascyn language). For this simple algorithm, every operation has its own operator.

Now consider a slightly larger piece of code:

$$
\begin{aligned}
& \text { (if }(\text { not }(=\operatorname{limit}(\text { inc1 count))) }(\text { setn-word count } \\
& \text { (inc1 count))) }
\end{aligned}
$$

According to the rules of the Ascyn language, the synthesized hardware must execute these operations in one clock cycle. The hardware must therefore perform two "incl" operations in parallel. Fig 2 a meets this requirement by including two "incl" operators, but that solution is very inefficient.

The synthesizer solves the problem by recognizing that both "incl" operations increment the same variable ("count"); it then synthesizes by applying its fan-out rule and generates the circuit in Fig 2b.

To see how the synthesizer would handle a slightly more complicated design, consider the following set of operations that must be sequential (Program 1):
$($ setn-word A $(-(+$ A B) D)) (setn-bit BOOL
$(=\operatorname{LIMIT}(-\mathrm{C} D)))$

If you designed the circuit so that each step is performed separately, you would create the configuration in Fig 3a. The synthesizer, however, recognizes that because these operations are sequential and not parallel, they can share operators (Fig 3b). This streamlined circuit substitutes a multiplexer and a little more control wiring for a subtractor-a good tradeoff.

Note the control box in Fig 3a. Normally, a hardware designer designs everything in the data path except the code for the control section (the data path includes all
the registers and arithmetic operators through which data is processed). Programmers later specify the control code so that the data path executes the higher-level algorithm desired at the time. In contrast, the synthesizer automatically generates all code for the data path, including the control code.

When you want to estimate an ASIC's performance, algorithmic design facilitates identifying the critical path, which is difficult to do from schematics. In the Fig 3 designs, for example, the critical path depends on the control code. With one control code, the critical path might flow from A and B through the +, MUX, - , and $=$ operators all the way to BOOL. With another control code, the critical path might only flow through the +, MUX, and - operators (Program 1, in fact, defines this path). Thus you can easily determine the critical path if you look at the algorithm; but if you just look at the schematic, the critical path is obscure.

To further illustrate this point, consider that a second algorithm exists that generates the same data path as the previous algorithm but not the same control section:

```
(set n - word A (- C D))
(set n - bit BOOL (=(-(+ A B)D)LIMIT))
```

It's possible to determine the different critical paths by


Fig 3-The synthesizer maximizes the use of expensive operators, such as the subtractor in (a) by employing a multiplexer for sequential operations (b).

## A synthesized ASIC is not likely to be much more than 50 times faster than the same algorithm running on a standard computer.

analyzing the algorithms, but not if you look at the schematics-which are identical.

The third key element of the Ascyn synthsizer is its Boolean implementation, which differs from the more common Boolean reduction. Most Boolean reductions produce sets of sum-of-products expressions. By DeMorgan's law, these expressions can be realized as 2 -level NAND-NAND implementations. But NANDNAND implementations are not always practical. Consider the high fan-in NAND gate in Fig 4a. Such high fan-in gates pose two problems: They are slow, and they're not available in gate arrays. In fact, if you do find an 8 -input NAND-gate entry in your foundry's cell library, it's almost certain to be a macro that expands into the primitive cells in Fig 4b.

In many cases, the particular high fan-in gate that you need is neither a primitive nor a macro available in a given foundry's library. If you were using normal design methods, you would use the next highest size available and simply tie off some of the inputs. In comparison, if the Ascyn synthesizer can't find exactly the right size in the cell library, it generates from primitives the size needed. To do this, the synthesizer uses a series of rules that transforms and minimizes gates. For example, if an 8 -input NAND gate does not exist, the synthesizer generates the circuit in Fig 4b.

By analogy, you might expect that the synthesizer would transform a 5 -input NAND gate into the set of primitives in 5a. Instead it goes beyond simple Boolean reduction, and applies DeMorgan's law several more times to generate the circuit in Fig 5b. In this way, the synthesizer achieves optimal Boolean implementation. Not only does the design in Fig 5b have one less gate than the one in Fig 5a, but in some circumstances, it's also faster.

The step-by-step designing of a complex ASIC algorithmically provides a more practical example of how the synthesizer works. Consider a closest-threat


Fig 4-Simple Boolean reducers often produce results that require high fan-in NAND gates (a). In actuality, such gates in a foundry's cell library are macros that expand to the circuit in (b).


Fig 5-The synthesizer develops the exact-sized gate required. Although (a) is one possible expansion of a 5-input NAND gate, the synthesizer repeatedly applies DeMorgan's law to achieve an even simpler circuit (b).
alert subsystem. The system monitors a scanning device (such as a radar or video detector) and then determines and reports the object nearest to a selected set of X,Y coordinates. First, a master processor specifies a set of X,Y coordinates. The ASIC then checks these coordinates against the next frame of data from the scanning device and all subsequent frames until the master selects a new set of coordinates. When the ASIC detects a new closest threat, it alerts the master processor, which then can obtain the current closest object's X,Y location (the alert signal remains high until the next such interrogation) (Fig 6).

The closest-threat alert subsystem executes algorithms to perform five tasks:

- Input coordinates from the master processor.
- Output coordinates and alarm the master processor.
- Process the pixel data from the scanner.
- Compute the distance from detected objects to the coordinates supplied by the master processor.
- Update the closest threat and alert the master processor if necessary.
Before deciding that an ASIC is the best way to implement the closest-threat subsystem, you must first determine the throughput requirements of the subsystem. The distance calculation seems to be the most critical; the other algorithms seem much less computationally complex. The scanning device scans 256 lines of 256 points each. A frame consists of a framing bit followed by 65,536 data bits. A data bit of 1 indicates that an object exists at the point then being scanned in the frame.

If $\mathrm{X}_{\mathrm{L}}$ and $\mathrm{Y}_{\mathrm{L}}$ are the X and Y values of the coordinates selected by the master processor, then the distance from $\mathrm{X}_{\mathrm{L}}, \mathrm{Y}_{\mathrm{L}}$ to any $(\mathrm{X}, \mathrm{Y})$ coordinate is:

$$
\sqrt{\left.\left(\mathrm{X}-\mathrm{X}_{\mathrm{L}}\right)^{2}+\left(\mathrm{Y}-\mathrm{Y}_{\mathrm{L}}\right)^{2}\right)} .
$$

This simple formula is very computationally intensive however. The following algorithm approximates $\sqrt{\mathrm{a}^{2}+\mathrm{b}^{2}}$ without recourse to squaring or square roots:

$$
\begin{gathered}
\mathrm{g}=\text { MAXIMUM }\left\{\begin{array}{l}
|\mathrm{a}| \\
|\mathrm{b}|
\end{array}\right. \\
\mathrm{l}=\text { MINIMUM }\left\{\begin{array}{l}
|\mathrm{a}| \\
|\mathrm{b}|
\end{array}\right. \\
\mathrm{c}=\frac{7}{8} \mathrm{~g}+\frac{1}{2} \mathrm{l}
\end{gathered}
$$

APPROXIMATE SQUARE ROOT $=$ MAXIMUM $\left\{\begin{array}{l}\text { c } \\ g\end{array}\right.$
The subsystem must process 65,536 bits (or pixels) per frame. The number of frames per second depends on the scanning device; the device in this example displays 60 frames per second (a typical number). The subsystem must therefore have a throughput rate of about 4 M pixels/sec or better.

## Standard $\mu \mathrm{P}$ won't work

You can rule out using a standard $\mu \mathrm{P}$. The fast magnitude-approximation algorithm needs about 20 instructions. At a clock rate of 16 MHz , and assuming 4 cycles per instruction (both are optimistic estimates), an advanced standard $\mu \mathrm{P}$ can do no better than 0.2 M distance computations per second.

You could try other algorithms. For example, you could buffer a frame's worth of data and perform a spiral search. Or, rather than use an approximation, you could note that $(d+1)^{2}=d^{2}+2 d+1$, and compute successive values of distance recursively from the beginning of a scan. Using the synthesizer, you can evaluate both the advantages and disadvantages of these two algorithms.

Another hardware option to consider is a programmable logic device (PLD). Unfortunately, the arithmetic required for this application immediately disqualifies
standard PLDs because the carry action of multibit arithmetic implements poorly as a sum-of-products Boolean function.

Yet another possibility is to use a ROM. The distance calculation is completely combinational-it doesn't require saving any data from one input to the next. Because the X and Y inputs are 8 bits each, the resulting distance may be as large as 360 , thus requiring 9 bits. This magnitude implies a $64 \mathrm{k} \times 9$-bit ROM, which is hardly a standard size. Even if you did have such a part, you would still need some auxiliary logic to implement the other, system-oriented, parts of the problem (eg, comparison against current minimum, alerting the master processor).

## Code implements distance calculation

The code in Listing 1 implements only the distance calculations and does not include any I/O functions. The first line names the ASIC as "mag," and declares the default integer word size to be 9 bits wide. The next section, starting with the keyword "def," declares the types of the variables-in this case "input," "output," and "wire" ("wire" variables are unregistered, singlevalue Boolean variables).

The last section, starting with "always," defines the algorithm. The keyword "always" starts an endless loop that executes everything that follows continuously. The Ascyn language uses a prefix notation, so that ( $>=x$ xl ) denotes a conditional test that is true if x is greater than or equal to xl , and is false otherwise. Therefore, when the first three program lines following the "always" says:

$$
\begin{aligned}
& \text { (if }(>=\mathrm{xxl}) \text { (set-word dx }(-\mathrm{x} \mathrm{xl}) \text { ) } \\
& \text { (set-word dx }(-\mathrm{xl} \mathrm{x})))
\end{aligned}
$$



Fig 6-The closest-threat alert subsystem processes a high-speed bit stream from a scanning device and gives a master processor the coodinates of the closest object.

Three key characteristics of the syntbesizer give it the ability to transform algorithms into net lists.
this code fragment means that if $x$ is greater than or equal to $x l$, then $d x$ is assigned the value $x-x l$; otherwise, $d x$ is assigned the value $\mathrm{xl}-\mathrm{x}$. In other words, this code fragment guarantees that $d x$ is the positive difference between x and xl . (Comments start with a ";" and continue to the end of the line.)

The second program line treats dy similarly. The next program section finds the minimum and maximum of $d x$ and dy. In this case, if $d x$ is greater than or equal to dy, the code does two things in parallel: It assigns the value of $d x$ to $g$, and it assigns the value of $d y$ to $l$. The parallel action is exactly the opposite when dy is greater than dx .

The next section creates the approximation c according to the algorithm. The operator $\gg$ is a right-shift command and can take several options, including an option for the number of bits to shift. Note that $7 / 8=1-1 / 8$. Finally, the program selects the maximum of g or c to be the d output.

In order to submit this program to the synthesizer and discover how fast the algorithm will run, you obviously must choose a foundry's cell library. For the closest-threat subsystem, several are available. This
example uses the California Devices Inc (CDI) $2-\mu \mathrm{m}$ double-level metal-cell library.

To estimate the maximum frequency, the synthesizer analyzes the critical path, taking fan-out capacitive loading into account. This estimate stops short of actually converting the net list into a physical layout and calculating the speed based on the chip's geometry -that's the task of the foundry. The package's simulator estimates that the ASIC's maximum clock frequency will be 4.57 MHz .

Further, this algorithm consumes 742 physical cells of a CDI gate-array chip. The cell count indicates how expensive production is going to be. The average gatearray chip comprises 2,000 of these cells, and the foundry makes up to 10,000 -gate chips available (chips with more gates cost more of course).

These manufacturing specifications do not apply to other foundries. To see how large this design will be (and how much it will cost and how fast it will run) using other foundries' gate arrays or standard cells, you must resynthesize the algorithm using their libraries.

In this case the CDI chip is an acceptable size. If it weren't but you only needed a $10 \%$ improvement, you

```
LISTING 1-BASIC DISTANCE-CALCULATION ALGORITHM
(system mag 9
    (def
        (x input word *) (x| input word *) (dx wire word *)
        (y input word *) (yl input word *) (dy wire word *)
        (g wire word *) (I wire word *)
        (c wire word *) (d output word *)
        )
        (a)ways
; dx = absolute value of ( }x-x|
    (if ()=x x|) (set-word dx (-x x|)) (set-word dx (-x| x)))
; dy = absolute value of (y - yl)
    (if ()= y yl) (set-word dy (-y yl)) (set-word dy (-yl y)))
; g and l are max and min of absolute values
    (if (>= dx dy)
        (par (set-word g dx) (set-word I dy))
        (par (set-word g dy) (set-word l dx)))
; c=7/8g+1/2 1
        (set-word c (+ (>> 1). ; 1/2 1+
        (-g(>)gbit-0 3)))) ; 7/8 g
; approximate square root = max of g , c
    (if (>=cg) (set-word d c) (set-word d g))
        ))
```

might try altering the fan-out ratio. However a $10 \%$ change often isn't sufficient, and you must consider more radical measures.

## Pipelining shortens critical paths

One possible measure is pipelining. It gives you a larger design but a shorter critical path. Yet pipelining also changes the chip's operation because the outputs are delayed some number of cycles behind their associated inputs. The rest of the system must be able to accommodate this latency.

You have the option of pipelining the distance calculation. This sort of pipelining is very easy to effect. Just designate some of the variables to be "saved" variables rather than as unregistered ("wire") variables. Selecting variables $l$ and $g$ is a good choice because they are roughly in the middle of the algorithm (Listing 2). Thus pipelining here yields the greatest improvement in performance for the least cost. The first pipeline stage calculates l and g from dx and dy . The second stage uses $l$ and $g$ to calculate $c$ and d. Except for the declarations of the type of variables, the code is unchanged from Listing 1.

Selecting the same foundry's library as before, and
resynthesizing, yields a chip that runs at 7.72 MHz instead of 4.57 MHz . However, the chip now consumes 842 physical cells instead of 742 . So far, this result is exactly as you would expect: You've traded time for space. You could make more pipeline stages. For example, dx and dy could be pipeline registers-so could c. Each time you add a pipeline stage, you should improve the throughput at the expense of increased chip area.

One of the primary advantages of an ASIC over standard parts is that you can achieve exactly the right I/O configuration as well as exactly the right computation on a system-by-system basis. Depending on the processor bus to which you are connecting, the data bus could, for example, be bidirectional. If you are using one of the Intel processors, you might find that the address and data buses were multiplexed. A standard chip that does not meet the processor's interface requirements requires extra chips for the interface. On the other hand, you can change the I/O algorithm to accommodate such processor-bus variations.

The processor needs to be able to load the ASIC with new values of xl and yl (Listing 3). In specifying I/O, for example, you can assign some of the processor's

Text continued on pg 134

## LISTING 2-DISTANCE ALGORITHM WITH PIPELINING

```
(system mag 9
    (def
            (x input word *) (x| input word *) (dx wire word *)
            (y input word *) (yl input word *) (dy wire word *)
            (g saved word *) (l saved word *) ;pipeline registers
            (c wire word *) (d output word *)
            )
        (a)ways
; }dx=\mathrm{ absolute value of (x - xl
            (if (> = x x|) (set-word dx (-x x|)) (set-word dx (-x| x)))
; dy = absolute value of (y - y|)
    (if (>= y yl) (set-word dy (-y yl)) (set-word dy (-yl y)))
; g and l are max and min of absolute values
    (if (>= dx dy)
            (par (setn-word g dx) (setn-word l dy ))
            (par (setn-word g dy) (setn-word l dx)))
; c=7/8g}+1/2
            (set-word c (+ (>> 1) ; 1/2 1 +
            (-g(>>ggit-0 3)))) ; 7/8 g
; approximate square root = max of g , c
    (if (>= c g) (set-word d c) (set-word d g))
    ))
```


## LISTING 3-MEMORY-MAPPED I/O TO CUSTOMIZE ASIC FOR A $\mu$ P

;The processor expects to receive an interrupt when the closest threat changes.

```
(system com 9
    (def
        (ad-hi input word 8 0 *) ; The inputs are only
        (ad-10 input word 8 0 *) ; 8 bits wide,
        (do input word 80 *) ; When used in a 9-bit operations, msb will be 0
        (di tri-state word *)
    ; di will be hi-impedence unless explicitly set during a clock cycle
        (rd input bit *)
        (wr input bit *)
        (xc saved word *)
        (yc saved word *)
        (dc saved word *) ; saved value of current distance
        (intfl saved bit *) ; generate an interrupt if the newly calculated
        (int output bit *) ; distance chages from the old value of dc
        (x| saved word *)
        (yl saved word *)
        (y saved word*)
        (x saved word *)
        (fr input bit*)
        (pi input bit *)
        (restart saved bit *)) ; hold until next
    (always
        (set-bit int intfl) ; int is always the same as interrupt flag
        (if wr
            (if (reduce and ad-hi) ; ad-hi all ones
            (par (set-bit restart bit-1) ; resync calculation
                        (if (bit-lsb ad-lo)
                                (setn-word yl do) ; odd -> set new yl
                (setn-word xl do))))) ; even -> set new xl
    (if rd
            (if (reduce and ad-hi) ; ad-hi all ones
                (par (setn-bit intfl bit-0) ; clear pending interrupt
                        (if (bit-1sb ad-lo)
                                    (set-word di yc) ; odd -> send current yc
                                    (set-word di xc))))) ; even }->\mathrm{ send current xc
; now input frame sync and manage }x\mathrm{ and }y\mathrm{ counters,
    (if fr
            (par (setn-word x 0) (setn-word y 0)
            (if restart
                        (par (setn-bit restart bit-0) ; clear pending restart
                            (setn-word de 511)))) ; and set de to max.
            (if (bit-msb y) () ; too many rows - wait for frame sync
            (if (bit-msb x)
            (par (setn-word x 0) (setn-word>
```



Fig 7-Adding data and address buses to the ASIC customizes it for a particular $\mu P$.
memory-address locations to the ASIC. If the processor writes to an even address in the FFXX ${ }_{\text {HEX }}$ block, it loads a new xl value, and similarly loads a new yl value if the address is odd.
If the processor reads an even address in the FFXX $_{\text {HEX }}$ block, the chip will output the closest threat's $x$ coordinate for the duration of the read signal. If the address is odd, the chip outputs the yl coordinate, and the processor generates the "Read" and "Write" signals, which the chip then inputs.
The chip maintains the x and y values; these values are derived from the framing bit ("fr") that comes from the scanning device. Fig 7 reflects these modifications in the initial closest-threat-alert diagram. When synthesized, the completed chip uses 1342 gates and operates at 4 MHz .

EDN

## Author's biography

Jay $R$ Southard is the president of Al gorithmic Systems Corp (Braintree, MA). Previously he worked for MetaLogic, MIT"s Lincoln Lab, Draper Labs, and General Instruments. Jay received a BA from Grinnel College and an MSEE from Stanford. He belongs to the IEEE.


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## FEATURES

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$1.5 \mu$ s to $0.0025 \%$ for 10 V Step
Slew Rate: $75 \mathrm{~V} / \mu \mathrm{s}$
Total Harmonic Distortion (THD): 0.0003\%
$>1000$ pF Capacitive Load Drive Capability with $10 \mathrm{~V} / \mu \mathrm{s}$ Slew Rate

Input Offset Voltage: 0.25 mV max Input Offset Drift: $3 \mu V /{ }^{\circ} \mathrm{C}$ max
Open Loop Gain: $250 \mathrm{~V} / \mathrm{mV}$ min
Noise: $4 \mu V$ p-p max, 0.1 Hz to 10 Hz

## PRODUCT DESCRIPTION

The AD744 makes a breakthrough in the high speed BiFET market by offering guaranteed maximum settling to $0.01 \%$ in 750 ns . It also offers the excellent dc characteristics of the AD711 BiFET family with enhanced slew rate, bandwidth and load driving capability.
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The AD744 offers optional custom compensation for additional design flexibility. This external compensation allows the AD744 to drive capacitive loads up to 2000 pF and greater with full stability, making the AD744 outstanding for use as a coaxial cable driver. Alternatively, external decompensation may be used to increase the gain bandwidth of the AD744 to over 200 MHz at high gains. This makes the AD744 ideal for use as an ac preamp in digital signal processing (DSP) front ends.


AD744 Settling Characteristics 0 to +10 V Step
PRODUCT HIGHLIGHTS

1. The AD744 offers exceptional dynamic response. It settles to $0.01 \%$ in 500 ns and has a $100 \%$ tested minimum slew rate of $50 \mathrm{~V} / \mu \mathrm{s}$.
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3. The AD744 has a guaranteed and tested maximum voltage noise of $4 \mu \mathrm{~V}$ p-p, 0.1 to 10 Hz .
4. The AD744 is a high speed BiFET op amp that offers excellen performance at competitive prices.

## THIS PAGE SETTLES THE QUESTION OF WHO THI FISTESTSETHLIMCBH.10

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| TMM2464AF | $8 \mathrm{KX8}$ | NMOS |
| TMM24128AP | $16 \mathrm{KX8}$ | NMOS |


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| :---: | :---: | :---: |
| ONE TIME PROGRAMMABLE |  |  |
| TMM24128AF | $16 \mathrm{KX8}$ | NMOS |
| TMM24256BP | $32 \mathrm{KX8}$ | NMOS |
| TMM24256BF | $32 \mathrm{KX8}$ | NMOS |
| TC54256AP | 32KX8 | CMOS |
| TC54256AF | 32KX8 | CMOS |
| TMM24512AP | $64 \mathrm{KX8}$ | NMOS |
| TMM24512AF | 64KX8 | NMOS |
| TC541000P | $128 \mathrm{KX8}$ | CMOS |
| TC541001P | $128 \mathrm{KX8}$ | CMOS |
| ROM |  |  |
| TC53257P | 32KX8 | CMOS |
| TC53257F | 32KX8 | CMOS |
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## Single VLSI chip helps you implement EPA factory networks

The Enhanced Performance Architecture (EPA), a subset of the MAP specification, allows networks to provide fast, noise-immune communication over short distances. By applying a single VLSI chip that implements most of the EPA functions in silicon, you can speed your network's response time without a buge software-development effort.

## Rhonda Alexis Dirvin and Anne-Marie Larkin Motorola Inc

If you're building or extending a factory network that includes intelligence, programmable controllers, and systems that must respond quickly to their controlling computer, you'll find the Manufacturing Automation Protocol (MAP) too slow for your purposes. To let machines or sensors send status information and command acknowledgment to their controlling computer quickly, you can implement the Enhanced Performance Architecture (EPA), a streamlined version of the MAP specification that takes into account the networking requirements of the workstations, PCs, and intelligent machines in a factory environment-requirements that are vastly different from those of large computers.


Fig 1-A broadband network connects the major nodes of a MAP system, which uses all seven layers of the ISO/OSI model. Local carrierband networks connect the smaller computers to sensors and intelligent machines that require high performance and short response times.

The EPA, created by the SP-72 committee of the Instrument Society of America (ISA) and the MAP Enhanced Performance Architecture Task Group, accommodates a wide variety of machines by dividing the total network into a hierarchy of subnetworks, each of which could meet a different set of performance requirements (Fig 1). The Process Industries Special Interest Group also played an important part in pressing for and helping to define the EPA.

The initial MAP specification concentrated on creating a backbone network in the factory to provide

The Enbanced Performance Architecture provides a bigh-performance, low-cost link between process-control stations and a MAP network.
communication among the large computers at the administrative level. This type of communication did not need to be especially fast; several seconds might elapse between file transfers, for instance. On the lower level of the factory, however, machines such as numerical controllers and robots need to perform information transfers at much shorter intervals (tenths of milliseconds, in some cases). At present, it's not possible to obtain this high performance if all communications must pass through the full seven layers of the OSI protocol, which the MAP uses. Further, it takes significant memory and horsepower to run all seven layers. If your task is merely to provide a network for the low-cost machines at the factory's lower level, you probably won't be able to justify the expense of all that memory and horsepower.
The EPA solves that problem by specifying only layers 1 and 2 , an interface, and layer 7. The remaining layers ( 3 through 6 ) perform functions that are not needed in the low-level nodes served by the EPA. The interface between layer 2 and layer 7 puts the data into a format compatible with the Manufacturing Message Specification (MMS). Reducing the number of layers of ISO/OSI software not only yields an improvement in network performance (because EPA nodes execute fewer instructions to perform a given function), but also


Fig 2-Mini-MAP nodes serve small machines and programmable controllers. They communicate with backbone nodes via MAP/EPA nodes and a router.
reduces software-design costs, because the system designer has fewer layers to deal with.
For one thing, the specifying bodies chose broadband technology as the physical layer for the backbone of the MAP, because it can operate over long distances (miles) and because it has high noise immunity. EPA subnetworks, on the other hand, operate only over distances of 500 to 1000 ft ; they could therefore use either carrierband or baseband technology in the physical layer. Both of these technologies are far less expensive than broadband technology.

Today's broadband modems cost from $\$ 700$ to $\$ 1500$ and require a head-end remodulator (one per network), which sells for around $\$ 10,000$. Carrierband modems cost from $\$ 400$ to $\$ 600$ and require no head-end remodulator. This price comparison does not take into consideration the fact that carrierband technology will soon be available in silicon form-a fact that will drop the carrierband price to approximately $\$ 100$ per node. Carrierband technology also lets you save on space, because you'll probably need only one board instead of two to accommodate the hardware. The primary reason to choose carrierband over broadband, therefore, is economic.

When considering alternatives to broadband technology, the specifying bodies preferred carrierband over baseband for several reasons, including cost: Carrierband is somewhat less expensive than baseband. More important, however, is that carrierband has much better noise immunity, because it frequency-modulates a constant-amplitude carrier signal instead of ampli-tude-modulating a constant-frequency signal, as the baseband technique does. Its higher noise immunity makes carrierband more suitable for use in an electrically noisy factory environment.

## EPA subnetworks contain a variety of node types

The Enhanced Performance Architecture embraces several different types of stations that fulfill different purposes; the types include Mini-MAP, MAP/EPA, and router nodes (Fig 2). Mini-MAP nodes can communicate only with each other and with MAP/EPA nodes over the subnetwork, using the 3-layer EPA architecture. Direct communications of this type are easy and efficient. The MAP/EPA node is also a carrierband device, but it contains facilities for converting data from 3 -layer format to 7 -layer format and vice versa.

All the communications between a Mini-MAP node and a backbone node take place via a MAP/EPA node and a router. The MAP/EPA node performs all of the
conversions between the 3-layer EPA architecture and the 7 -layer MAP architecture. The router is a 3-layer device that can segment data frames to ensure that they conform to the maximum size limitations, both on the backbone side and on the EPA-segment side. It also contains both broadband hardware that connects to the backbone network, and carrierband hardware that connects to the EPA segment (subnetwork). Routers' frame-segmenting capability is the reason that the specifying bodies preferred routers over bridges or gateways as the links between the backbone and EPA segments. Bridges are too simple and can only filter addresses; gateways, which perform full protocol conversion, are more complex and more expensive than is necessary for the MAP/EPA link.

## Redundancy maintains network integrity

A subnetwork may contain multiple MAP/EPA nodes, both to provide redundancy and to improve performance. If one of the MAP/EPA nodes fails, another one can perform the same functions; it thus prevents the subnetwork from being isolated from the backbone. If you expect to have a substantial amount of network traffic between the subnetwork and the MAP back-
bone, you should consider providing more than one MAP/EPA node in order to maintain acceptable levels of performance. Indeed, the MAP specification requires that a subnetwork contain at least two MAP/EPA nodes before it can be classified as an EPA segment.

## EPA functions allow very simple nodes

One function that was not in the original MAP specification but was added to the EPA is the IEEE 802.2 Logical Link Control (LLC) type 3 Acknowledged Connectionless Service. This function allows EPA station A to ask station B a simple question, such as "What is your temperature?" While station A still holds the token, station B can respond immediately with a simple answer (for example, " $58^{\circ} \mathrm{C}^{\prime}$ "). This function can achieve response times of around $5 \mu \mathrm{sec}$, because station B doesn't have to obtain the token and then establish a new communication link.

By using the LLC type 3 feature, you can greatly reduce the cost of attaching simple devices to an EPA segment of the network. Temperature and pressure sensors, for example, do not need to participate in the token-passing ring; they only have to be able to respond properly when they're sent an LLC type 3 request

## ISO/OSI model specifies seven layers

The MAP specification is based on the International Standards Organization's Open Systems Interconnect (ISO/OSI) model primarily because the ISO/OSI model already has widespread international support. Now the US federal government, too, has chosen the ISO/OSI model for the new networking procurements described in the Government Open Systems Interconnection Profile (GOSIP) document.

The ISO/OSI model breaks the communications task into seven sections, or layers. Each layer has a duty to perform or a service that it must provide to the layer immediately above it. ISO specifies what services must be
provided, but it doesn't specify the manner of implementing them. Table A shows the name and function of each layer of the

ISO/OSI model. Immediately above layer 7 are any user application programs, which are not part of the ISO/OSI model.

TABLE A-THE SEVEN LAYERS OF THE ISO/OSI MODEL

| NAME | FUNCTION |
| :--- | :--- |
| LAYER 7-APPLICATION | PROVIDES ALL SERVICES THAT ARE DIRECTLY <br> COMPREHENSIBLE TO APPLICATION <br> PROGRAMS |
| LAYER 6-PRESENTATION | TRANSFORMS DATA TO AND FROM NEGOTIATED <br> STANDARDIZED FORMATS |
| LAYER 5-SESSION | SYNCHRONIZES AND MANAGES DIALOGUES |
| LAYER 4-TRANSPORT | PROVIDES TRANSPARENT RELIABLE DATA <br> TRANSFER FROM END NODE TO END NODE |
| LAYER 3-NETWORK | PERFORMS MESSAGE ROUTING FOR DATA <br> TRANSFER BETWEEN NODES |
| LAYER 2-DATA LINK | MANAGES ORDERED ACCESS TO THE MEDIUM <br> AND DETECTS ERRORS INVOLVED IN THIS <br> PROCESS |
| LAYER 1-PHYSICAL | ELECTRICALLY ENCODES AND PHYSICALLY |
|  | TRANSFERS DATA OVER THE MEDIUM |

EPA networks use carrierband technology as the medium because it's less expensive than broadband and bas better noise immunity than baseband.

## TABLE 1-LLC TYPE 3 PRIMITIVES PROVIDED BY THE MC68824

ACKNOWLEDGED CONNECTIONLESS DATA-UNIT TRANSMISSION SERVICE

| PRIMITIVE | DESCRIPTION |
| :--- | :--- |
| L_DATA ACK.request | USED TO SEND A DATA UNIT WITH |
|  | ACKNOWLEDGMENT TO ANOTHER NODE |
| L_DATA ACK.indication | USED TO INDICATE THE RECEPTION OF A |
|  | NON-NULL, NONDUPLICATE LSDU FROM A |
|  | REMOTE DATA-LINK NODE |
| L_DATA ACK_STATUS.indication | USED TO INDICATE WHETHER THE |
|  | PREVIOUS LSDU TRANSMISSION REQUEST |
|  | WAS SUCCESSFUL |

ACKNOWLEDGED CONNECTIONLESS DATA-UNIT EXCHANGE SERVICE

| PRIMITIVE | DESCRIPTION |
| :---: | :---: |
| L_REPLY.request | USED TO REQUEST AN ACKNOWLEDGED CONNECTIONLESS DATA-UNIT EXCHANGE |
| L REPLY.indication | USED TO INDICATE RECEPTION OF AN RWR FRAME |
| L_REPLY_STATUS.indication | USED TO CONFIRM RWR FRAMES |

REPLY DATA-UNIT PREPARATION

| PRIMITIVE | DESCRIPTION |
| :--- | :--- |
| L_REPLY_UPDATE.request |  <br> USED TO INDICATE THE NEED TO UPDATE A <br> RESPONSE |
| L_REPLY__UPDATE_STATUS.indication | USED TO CONFIRM THAT A RESPONSE WAS <br> UPDATED |
|  |  |

frame. Very simple hardware can provide them with this ability; the node can consist merely of an enhanced token-bus-controller chip, a carrierband chip, some memory to hold the response, and some mechanism for updating the memory as required.

## VLSI chip embodies the EPA functions

To meet the requirements for low cost and high performance that are driving the development of the EPA, you can apply a device such as the MC68824 token-bus controller (TBC), which implements the EPA functions in silicon. Motorola recently enhanced the chip so that it now implements both the Media Access Control (MAC) portion of the IEEE 802.4 standard and the Receiver portion of the IEEE 802.2 LLC type 3 Acknowledged Connectionless Service (Table 1 summarizes the latter).
These services provide a means, at the data-link level, of sending a message from one node to another and of receiving an acknowledgment message without establishing a separate data-link connection. This feature yields the very fast response time that the applications require. In addition, the MC68824 provides two
separate methods of bridging: hierarchical and IBMdefined source routing. The chip allows the user to choose between powerful address mechanisms that permit group-address recognition and multidrop operation.
Some of the MC68824's enhancements include a re-duced-data-structure mode for increased performance, a FIFO option for increased reliability, and a busanalyzer mode that lets you run the TBC as a powerful protocol analyzer. The chip can now perform serial data transfers over the network at speeds ranging from 10 k to 12.5 M bps, and it can accommodate system clock rates from 10 to 16.67 MHz .
The MC68824 functions as an intelligent peripheral device to a host microprocessor. The TBC is a full MC68000 bus master; it has on-chip DMA facilities for managing tables and frame buffers located in shared memory, and these DMA facilities transfer data frames to and from a shared memory with minimal intervention from the host $\mu \mathrm{P}$. A fully linked buf-fer-management scheme allows the queueing of frames during transmission and reception. You can configure the TBC bus interface to handle both 8 -bit and 16 -bit


Fig 3-An EPA node consists of an enhanced TBC, a host CPU, memory shared by the host and the TBC, and a modem. The node can be very simple, because it needs to implement only the lower three layers of the ISO/OSI model.
data transfers in either Motorola/DEC or Intel/IBM data formats. Fig 3 shows the MC68824 in a typical intelligent system environment.

## EPA is one of chip's three operational modes

The MC68824 has three main operational modes: the TBC mode, the EPA mode, and the bus-analyzer mode. In the TBC mode, which is the default mode, the MC68824 provides a full MAC implementation as well as support for LLC types 1 and 3 . Running the TBC in its default mode allows the host $\mu \mathrm{P}$ to perform the LLC functions (or equivalent functions) in software.

In the EPA mode, the MC68824 performs the receiver functions of the Acknowledged Connectionless Service of the LLC sublayer, as well as operating normally when it receives a nonrequest-with-response frame. The receiver portion of LLC type 3 is the most timecritical, and it performs much better when implemented on chip.

To use the EPA mode, the host must set a bit in the initialization table, and must also add an LSAP (Link Service Access Point) table to the end of the initialization table. This table has an entry for each of the 128 possible individual LSAPs and one for each of the 128 group LSAPs. Each LSAP entry contains user information as well as LSAP status information (such as whether or not the LSAP is active). For each active LSAP, the host prepares a response and puts the pointer to the response in the corresponding entry of the LSAP table. The host also updates the response when necessary.

When a receiving station running in EPA mode finds a Request with Response (RWR) encoded in the frame control of an incoming frame, the station looks at the corresponding LSAP entry in the LSAP table. If the

RWR frame requests data, the receiving station will return the data that the LSAP entry points to, provided that the LSAP is active. If the LSAP is not active, the station returns a NACK. If the RWR frame does not request data, the station merely returns an ACK (if it has received the frame correctly) or a NACK (if it has received the frame incorrectly).

In the bus-analyzer mode, the MC68824 is not part of the logical ring, but it receives all frames. This feature makes the TBC an ideal chip for use in a protocolanalyzer application. In this mode, the TBC timestamps all frames, so it can measure the elapsed time between network events.

## Host, TBC communicate through shared memory

The TBC also makes use of a simpler data structure. The MC68824 and the host communicate with each other through shared memory. This shared memory comprises two tables and some buffer structures that store received frames or frames awaiting transmission. The two tables are called the initialization table and the private area.

The initialization table contains the initial parameters that the host loads into the private area by giving the Initialize command at power-up or after a reset. The table also holds the command-parameter area, the interrupt status words, some statistics, and, if the TBC is running in the EPA mode, an LSAP table.

The command-parameter area has two main uses: The host employs it mainly to pass parameters to the TBC in conjunction with a command, and the TBC uses it to return parameters to the host in response to a command. The TBC continuously updates two Interrupt Status Words (ISWs) to indicate status changes. Flags in the ISW, filtered by an interrupt status mask,

You can now obtain a single VLSI chip that implements most of the EPA functions, as well as standard token-bus-controller functions, in silicon.
determine whether or not a particular event will generate an interrupt. ISW1 keeps track of events that relate to the local system. For example, one of the flags in ISW1 informs the host that the TBC has received a data frame. ISW2 keeps track of events that relate to the network. One of the flags in ISW2, for example, informs the host that the TBC has detected a duplicate MAC address on the network.

The statistics area of the initialization table holds network statistics, such as "number of who_follows events." The host may collect all the statistics or may disable the collection of two items (Tokens_passed and Number_of tokens_heard). Also, the TBC keeps track of various types of modem errors in the statistics area.

The private area contains mainly MAC operational parameters and statistics; therefore, the TBC uses it primarily as an extra storage area. Hi_Priority_ Token_Hold Timer, for example, is a MAC operational timer that you can specify in the private area if you use the priority option. IEEE 802.4 specifies this and many other timers. Among the MAC statistics kept in the private area is the Last Token_Rotation_Time, which is the observed time measured from token arrival to token arrival.

The private area also holds the pointers to the various buffer structures used for receiving and transmitting data frames. After power-up or reset, the host
initializes the private area by storing the appropriate values in the initialization table and then issuing the Initialize command. After initialization, the host can view or modify the private area by issuing the appropriate commands. However, in order to avoid conflicts, the host must not directly access the private area while the TBC is participating in a network.

## Linked buffers hold frame queues

The TBC uses linked buffer structures (Fig 4) to provide four transmit queues and four receive queues in accordance with the IEEE 802.4 message-priority scheme. The fully linked buffer structures consist of frame descriptors, buffer descriptors, and data buffers. Each frame descriptor contains control information pertaining to a frame sent or received, plus links to the next frame descriptor and to the buffer descriptor. The buffer descriptor contains the data buffer's attributes, such as the user-specified offset, which points directly to the data in the data buffer. This feature is useful for appending upper-layer protocol headers without moving the data. Each buffer descriptor also contains links to the next buffer descriptor (if there is one) and to the data buffer. Finally, the data buffers contain the actual data.

This linking scheme makes it easy for the host's software to build data frames: The scheme lets each

## MAP uses IEEE 802.4 token-bus protocol

At the physical and data-link layers, the MAP uses the tokenpassing bus protocol specified by IEEE standard 802.4. The protocol was chosen because it's deterministic; that is, the maximum allowable time for any station to access the medium is known. A token bus is also very flexible, because its topology allows you to add stations to or delete them from the network without disrupting network operations. The token-bus protocol also clearly defines error-related functions such as recovery from lost or multiple tokens, tokenpassing failures, inoperative
transmitters or receivers, and duplicate station addresses.

A token is a specific bit sequence in the frame-control portion of the frame. When a station holds the token, it has the right to transmit. When that station has finished transmitting all of its frames, or when the timer that tracks the station's maximum token-holding time (which is specified by the user) has timed out, the station passes the token to the station having the next lower address. This process continues until the station with the lowest address passes the token back to the sta-
tion with the highest address.
The station that holds the token periodically polls stations that are connected to the bus (but are not currently part of the logical ring) to see if any station wishes to join the logical ring. If a new station does wish to join the ring, the sending station patches the new station into the ring as its logical successor. For example, if the sending station, A, receives a valid reply from station E, station A patches station $E$ into the ring by making $E$ (instead of B) the logical successor to A.


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Providing a subnetwork with multiple MAP/EPA nodes not only yields higher performance, but prevents malfunctions from isolating the subnetwork.
software layer append the header information to the data by simply changing a pointer or offset. The MC68824 can also run in a reduced-data-structure mode that combines the frame descriptor, buffer descriptor, and data buffer as one structure.
The reduced-data-structure mode is well suited for receiving frames in systems that require very high throughput rates, because it requires fewer linking operations than the full data structure does, and, therefore, it's faster. In some applications, the higher performance that you get from the reduced data structure is more advantageous than the power and flexibility of the full data structure.

## The TBC has a flexible command set

The host processor commands the TBC to perform various functions by writing the appropriate command to the TBC command register. Initialization commands configure the TBC for operation after a hardware or software reset. These commands include Reset, Offline, Load Initialization Table Function Code, Initialize, and Idle. Idle causes the TBC to join the ring.
Mode commands set or clear various modes of TBC operation; the Set/Clear In_Ring_Desired command is typical of this category. Three commands pertain to the transmission of data frames: Start, Stop, and Restart. Parameter commands allow the host to modify or read TBC parameters such as function codes, the pad-timerpreset register, and some of the parameters that reside


Fig 4-The shared memory holds frame descriptors (FDs), buffer descriptors (BDs), and data buffers. The private area holds operational parameters and pointers to the first FD of each message to be queиed.
in the private area.
Test commands allow the host to run diagnostics while the TBC is in the off-line state; these commands test the host/TBC interface, the transmitter, the receiver, and the serial sections of the TBC, as well as the internal sections of the TBC.

Finally, two commands allow the TBC to provide management services to the physical layer of the node. The TBC's serial interface conforms to the IEEE 802.4 Standard Serial Interface. The Standard Serial Interface specifies a set of $10 \mathrm{I} / 0$ signals, which are used for data transfer when the TBC is in the MAC mode, and for passing station-management information when the TBC is in the station-management mode.

## EPA is gaining acceptance

The Enhanced Performance Architecture fulfills a requirement of the factory floor that the MAP did not previously address: the ability to provide sensors and intelligent machines with a low-cost, high-performance link to the MAP network. However, the fast response times that the EPA requires were initially very difficult to achieve. The EPA is only now starting to gain acceptance because of the emergence of devices such as the MC68824 Enhanced TBC.

EDN

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| Conversion Time ( $\mu \mathrm{Sec}$ ) Throughput Speed (kHz) | $\begin{aligned} & 16 \\ & 50 \end{aligned}$ | $\begin{aligned} & 14 \\ & 56 \end{aligned}$ | $\begin{aligned} & 7 \\ & 100 \end{aligned}$ | 1.3 | $\begin{aligned} & 1.25 \\ & 1000 \end{aligned}$ | 20 | $\begin{aligned} & 16 \\ & 50 \end{aligned}$ | $\begin{aligned} & 14 \\ & 56 \\ & \hline \end{aligned}$ | $\begin{aligned} & 7 \\ & 100 \\ & \hline \end{aligned}$ |
| Static Specifications: Linearity Error (\% FS, max) No Missing Codes (Bits) | $\begin{aligned} & +/-.0015 \\ & 16 \\ & \hline \end{aligned}$ | $\begin{aligned} & +1-.003 \\ & 14 \end{aligned}$ | $\begin{aligned} & +1-.012 \\ & 12 \end{aligned}$ | $8^{+1-.2}$ | ${ }_{12}^{+/-.01}$ | 16 | 16 | 14 | 12 |
| $\begin{aligned} & \text { Dynamic Specifications } \\ & \text { THD (\%) } \\ & \text { S/(N+D) (dB) } \\ & \hline \end{aligned}$ |  |  |  |  | $\begin{aligned} & .02 \\ & 70 \\ & \hline \end{aligned}$ | $\begin{aligned} & .007 \\ & 84 \\ & \hline \end{aligned}$ | $\begin{aligned} & .001 \\ & 92 \end{aligned}$ | $\begin{aligned} & .003 \\ & 83 \\ & \hline \end{aligned}$ | $\begin{aligned} & .008 \\ & 73 \\ & \hline \end{aligned}$ |
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# Static-system design exploits low-power CMOS 

> Static CMOS architectures let you customize a design for optimum performance and minimum power dissipation. Capitalizing on the characteristics of CMOS means addressing clock-frequency control before starting the initial design process.

## Walter J Niewierski, Harris Semiconductor

As every engineer knows, CMOS $\mu$ Ps combine low power with high performance: A CMOS $\mu$ P's power dissipation is directly related to its frequency of operation. Reducing the operating frequency from 16 MHz , say, to 1 MHz , allows the power-supply current to drop proportionately. An NMOS or bipolar $\mu$ P's supply current, on the other hand, remains the same regardless of how you manipulate a system's throughput-rate spec.

## An NMOS system can be bulky

The benefits of CMOS vs NMOS are readily apparent when you compare datasheet operating specifications. An NMOS $80286 \mu \mathrm{P}$, for example, has a typical powerdissipation spec of 3 W ( $\mathrm{I}_{\mathrm{IC}}=600 \mathrm{~mA}$ at 5.25 V ). In contrast, a CMOS $80 \mathrm{C} 286 \mu \mathrm{P}$, when running at a clock rate of 16 MHz , has a maximum operating current of 260 mA .
Operating power also has a ripple effect through a
system design. Usually you have to add NMOS memory and NMOS or bipolar peripherals to support an NMOS device. Depending on the size of your system, power requirements may be 25 to 30 W , and you may well have to specify a larger and heavier 50 W power supply to accommodate future expansion.

You also have to account for the fact that temperatures within an NMOS system can rise significantly. It's not unusual for the temperature of an NMOS or bipolar die to rise 40 to $60^{\circ} \mathrm{C}$ during operation, and fans and heat sinks are necessary to compensate for this increase. Oftentimes, the system also requires an air filter for the fan. Obviously, you have to design an enclosure large enough to hold all of this equipmentfans, bulky power supplies, and heat sinks.

Because CMOS devices dissipate less power, you can specify smaller power supplies. Lower power dissipation also means that the system will run at a lower ambient temperature. Heat sinks and fans are unnecessary, and vents can be shut to ensure a clean operating environment. Sealed enclosures can also be smaller and lighter.

For these reasons, many NMOS- and bipolar-device architectures are now available in CMOS. Furthermore, CMOS technology is amenable to static-circuit design, and NMOS is not. Static circuits allow you to customize your design to balance speed requirements and power considerations.

Static ICs run on any clock frequency from dc to the maximum circuit operating specification. In a dynamic circuit, if the clock frequency falls below a manufactur-er-specified minimum, the circuit can lose or alter data,

Static circuits run on any frequency from dc to the maximum circuit operating specification.
and thus the memory cells require refreshing at a certain rate to maintain valid data. You can put a static $\mu \mathrm{P}$ in a standby mode simply by stopping the clock signals, and even if you stop the external clock indefinitely, the device will retain data. When you reinitiate the CPU clock signal, the system starts up where it left off.

The ability to stop the clock also allows you to perform single-step operations during system development. This valuable method of debugging allows you to inspect the system bus and other functions for proper operation by manually stepping through the clock cycles. Single-stepping-and therefore the elimination of real-time data transfers-can simplify debugging significantly.

By taking advantage of opportunities to manipulate the system clock during your initial design, you can mix and match standby, low-frequency, and high-frequency
operations to achieve the most performance for the least power. You can design a $\mu \mathrm{P}$ system that allows the software executive routine to power down the entire system or just portions of the system not in use (certain I/O sections, for example, or file-maintenance areas).

## A clock controller eases design

To ease the implementation of static-operating modes in CMOS systems, clock controllers are available for various $\mu$ Ps. Fig 1 illustrates a CMOS system configuration using a Harris 82C85 static clock controller to interface directly to the maximum-mode status lines $\left(\overline{\mathrm{S}}_{2}, \overline{\mathrm{~S}}_{1}\right.$, and $\overline{\mathrm{S}}_{0}$ ) of either a Harris 80 C 86 or $80 \mathrm{C} 88 \mu \mathrm{P}$. The three control lines ( $\overline{\mathrm{S}_{2}} / \overline{\mathrm{Stop}}, \overline{\mathrm{S}}_{1}$, and $\overline{\mathrm{S}}_{0}$ ) on the 82 C 85 let you control the operation of the system clock with $80 \mathrm{C} 86 / 80 \mathrm{C} 88$ software. Essentially, the $\mu \mathrm{P}$ controls the clock by issuing a Halt instruction in software.

The status lines of the 80C86/80C88 identify the type


Fig 1-The 82C85 clock controller interfaces directly to an $80 C 86$ or $80 C 88$ CPU operating in the maximum mode. When the controller detects a Halt instruction, it stops the clock.


Fig 2-A CMOS peripheral such as the 82 C 55 can control a clock controller's stop, start, and slow or fast operations.
of bus cycle that the CPU is starting to execute, and are accessible only during maximum-mode operation. Typically, a bus controller uses these status lines to decode the current bus-cycle status of the CPU. Fig 1's truth table shows the status lines for different bus transactions. When operating in the maximum mode, the CPU places the $\overline{\mathrm{S}}_{2}, \overline{\mathrm{~S}}_{1}$, and $\overline{\mathrm{S}}_{0}$ lines in a 111 state followed by a 011 state to indicate a Halt instruction. The control lines on the 82 C 85 automatically decode this state sequence to halt the system clock.

You can use a peripheral such as a programmable interface (in this case, the 82 C 55 A ) to stop the clock by using the $\overline{\mathrm{S}}_{2} / \overline{\text { Stop }}$ line as a stand-alone Stop command (Fig 2). You must tie the $\overline{\mathrm{S}}_{0}$ and $\overline{\mathrm{S}}_{1}$ pins of the 82 C 85 to $\mathrm{V}_{\mathrm{CC}}$ to emulate a logic 111 state before applying a logic 0 to the $\overline{\mathrm{S}}_{2} / \overline{\text { Stop }}$ line to stop the clock. To place the 82C85 clock controller in a stop-oscillator mode, you tie the F/C line low. In this mode, the 82C85 stops the clock synchronously whenever it detects a Halt instruction. Regardless of whether the controller is operating in the fast or slow mode, it will decode a Halt instruction. Table 1 compares the four operating modes of the 82C85.

## Stop the clock!

The 82 C 85 also offers another way of stopping the system clock. The start-input line on the 82C85 provides independent clock control. A level-triggered ac-tive-high input overrides any stop condition. You can control the start-input line, for example, by using an 82C59A priority-interrupt controller (Fig 2). The INT

TABLE 1-COMPARISON OF 82C85's OPERATING MODES

| OPERATING <br> MODE | DESCRIPTION | POWER |
| :--- | :---: | :---: | :---: |
| LEVEL |  |  | PERFORMANCE

output pin of the 82 C 59 A connects directly to the start-input pin of the 82 C 85 and the INTR pin of the 80C86/80C88 $\mu$ P. An external event (such as a key depression) can produce an 82 C 59 A interrupt request, which causes the 82 C 85 to start the system clock.
If your system must operate continuously but power

During the initial design of a static CMOS system, the designer should anticipate circumstances in which the system can be stopped or run more slowly.
consumption is a concern, you can utilize the slow mode on the clock controller to divide the clock outputs by 256. The 82C85 has a SLO/FST pin that determines the speed of the system clock. When you apply a logic 1 to the $\overline{\mathrm{SLO}} / \mathrm{FST}$ pin, the frequency of the CLK and CLK50 outputs equals the crystal frequency divided by 3. A logic 0 on the $\overline{\text { SLO} / F S T ~ p i n ~ m a k e s ~ t h e ~ f r e q u e n c y ~ o f ~}$ the CLK and CLK50 outputs equal to the crystal frequency divided by 256.
If you are converting an NMOS or bipolar system to CMOS technology, each device you replace achieves a power savings. Table 2 shows typical power-supply currents for devices in an 80 C 88 -based system with an 82C85 static clock controller. When the CPU operates at 5 MHz (that is, with a $15-\mathrm{MHz}$ crystal frequency), the total system draws 72 mA (less than 400 mW ). The slow mode ( $20-\mathrm{kHz}$ CPU frequency) reduces the powersupply current by a factor of $3(24 \mathrm{~mA})$. This mode maintains continuous operation yet gives you the option to switch the system to full speed at any time.

The two stop modes offer the alternative of fast clock turn-on or minimum power dissipation. In the stopclock mode, the main system oscillator continues to run and permits full-speed operation to resume within three CPU clock cycles ( 600 nsec at 5 MHz ) after the fast mode is reinitiated. The main crystal oscillator


Fig 3-An on-chip bus-hold circuit maintains a valid CMOS logic level on its external pin when a driving source goes to a highimpedance state.
continues to run at 15 MHz (for a $5-\mathrm{MHz}$ system).
The stop-oscillator mode stops the system clock and the main crystal oscillator. The standby supply current for the system referred to in Table 2 is $752.7 \mu \mathrm{~A}$, which is less than 4 mW at 5 V . When you restart the crystal oscillator, the frequency and amplitude can vary significantly. After the amplitude reaches a specified amplitude, the 82 C 85 counts 8192 oscillator cycles to ensure crystal stabilization. The stabilization time guarantees that the clock outputs turn on within the specification limits and with the proper phase relationship.

The choice of operating modes depends upon system

TABLE 2-SYSTEM CURRENT DRAIN WITH AN 82C85 CLOCK CONTROLLER

| TYPICAL-SYSTEM DEVICES | MODES |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | STOP-OSCILLATOR | STOP-CLOCK | SLOW | FAST |
| 82C85 | $24.4 \mu \mathrm{~A}$ | 14.1 mA | 16.9 mA | 24.7 mA |
| 80 C 88 | $106.6 \mu \mathrm{~A}$ | $106.6 \mu \mathrm{~A}$ | $173.0 \mu \mathrm{~A}$ | 23.8 mA |
| 82C82 | $1.0 \mu \mathrm{~A}$ | $1.0 \mu \mathrm{~A}$ | $6.5 \mu \mathrm{~A}$ | 1.7 mA |
| 82C86 | $1.0 \mu \mathrm{~A}$ | $1.0 \mu \mathrm{~A}$ | $14.0 \mu \mathrm{~A}$ | 1.4 mA |
| 82C88 | $3.8 \mu \mathrm{~A}$ | $3.8 \mu \mathrm{~A}$ | $14.3 \mu \mathrm{~A}$ | 3.5 mA |
| 82C52 | $1.0 \mu \mathrm{~A}$ | $1.0 \mu \mathrm{~A}$ | $72.0 \mu \mathrm{~A}$ | $151.2 \mu \mathrm{~A}$ |
| 82C54 | $1.0 \mu \mathrm{~A}$ | $3.5 \mu \mathrm{~A}$ | $915.0 \mu \mathrm{~A}$ | $943.0 \mu \mathrm{~A}$ |
| 82C55A | $1.0 \mu \mathrm{~A}$ | $1.0 \mu \mathrm{~A}$ | $1.2 \mu \mathrm{~A}$ | $3.2 \mu \mathrm{~A}$ |
| 82C59A | $509.0 \mu \mathrm{~A}$ | $509.0 \mu \mathrm{~A}$ | $520.0 \mu \mathrm{~A}$ | $580.0 \mu \mathrm{~A}$ |
| HD-6406 | 4.97 mA | 4.97 mA | 5.09 mA | 5.12 mA |
| 74HCXX <br> PLUS OTHER ICs | $90.0 \mu \mathrm{~A}$ | $90.0 \mu \mathrm{~A}$ | $110.0 \mu \mathrm{~A}$ | 2.9 mA |
| HM-6516 | $1.9 \mu \mathrm{~A}$ | $1.9 \mu \mathrm{~A}$ | $132.0 \mu \mathrm{~A}$ | $820.0 \mu \mathrm{~A}$ |
| HM-6616 | $12.0 \mu \mathrm{~A}$ | $12.0 \mu \mathrm{~A}$ | $52.5 \mu \mathrm{~A}$ | 6.3 mA |
| TOTAL WITH 6406 | 5.72 mA | 19.8 mA | 24.0 mA | 71.9 mA |
| TOTAL <br> WITHOUT 6406 | $752.7{ }_{\mu} \mathrm{A}$ | - | - | - |
| CPU FREQUENCY | dc | dc | 20 kHz | 5 MHz |
| CRYSTAL FREQUENCY | dc | 15 MHz | 15 MHz | 15 MHz |

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## Clock control is the key to designing a min-

 imum-power system.

Fig 4-In CMOS devices with gated inputs, a low logic level on the $\overline{S T B}$ line turns the $Q_{1}$ and $Q_{2}$ FETs off so that a transition on the data-input line doesn't draw power-supply current.
priorities. The stop-clock mode delivers the fastest response and return from standby operation, but does so at the expense of higher power dissipation than the stop-oscillator mode. The stop-oscillator mode minimizes power dissipation, but has a longer response time.

Although clock control is key to low-power performance in a static CMOS system, there are other factors of which you should be aware. For example, when the input-voltage level on a CMOS device approaches the minimum $\mathrm{V}_{\text {IH }}$ (high-level voltage) or maximum $\mathrm{V}_{\mathrm{IL}}$ (low-level voltage), the p- and n-channel FETs can partially turn on. This condition produces a low-resistance path between $\mathrm{V}_{\mathrm{CC}}$ and GND and results in excessive power-supply current. Therefore, you must tie all CMOS inputs to $\mathrm{V}_{\mathrm{CC}}$ or GND in some manner when they are not in use.

Pull-up/down resistors are the most common method of ensuring CMOS input-voltage-level specifications. The resistors maintain valid CMOS voltage levels when a driving source is in a floating high-impedance state. This technique, however, suffers from some disadvantages. First, extra resistors are necessary, which increases both board real estate and cost. Second, the circuit's power dissipation increases because the driving source must supply extra current to the resistors. Needless to say, if you can avoid using pull-up/down resistors, you can improve overall system performance.

Fig 3 shows the bus-hold circuitry that the 80 C 86 , $80 \mathrm{C} 88,82 \mathrm{C} 55 \mathrm{~A}$, and 80 C 286 devices include to eliminate the need for pull-up/down resistors. When no driving source is present or when a driving source is in a floating high-impedance state, the on-chip circuitry maintains a valid CMOS logic level on specific inputs.

The circuit will maintain this valid CMOS level until an external source overdrives the affected input. The external device must source or sink approximately 400 $\mu \mathrm{A}$ to overdrive the input. Because the bus-hold circuits are active, their current requirements are similar to the leakage currents for an active device: The circuit typically draws less than $10 \mu \mathrm{~A}$ when it isn't switching. This current is in contrast to the current drawn through a $10 \mathrm{k}-\Omega$ pull-up resistor: $500 \mu \mathrm{~A}$ when the logic level is low.

Excessive current drain can also result when the input circuitry of a CMOS device responds to voltage transitions unrelated to its operation. A typical system has bus transceivers and latches attached to the system bus. The input circuitry of these devices undergoes many signal transitions unrelated to the functional operation of the devices themselves. These unrelated voltage transitions switch the input circuitry of a CMOS device, causing an increase in power dissipation. In addition, if the system bus experiences a high-impedance state (a floating condition), it could result in an indeterminate logic state internal to an attached CMOS device.

You can eliminate excessive current drain due to unrelated input transitions by using bus tranceivers and latches with gated inputs. The Harris Series 82 C 8 X bus transceivers and latches (such as the $82 \mathrm{C} 82 /$ 83/86/87) let you gate their data inputs when not in use (Fig 4). When the strobe line (STB) is in a active-low state, the upper p-channel FET ( $Q_{1}$ ) and the lower n-channel FET ( $Q_{2}$ ) turn off. This action removes $\mathrm{V}_{\mathrm{CC}}$ from the input circuitry so that no supply current flows to GND during unrelated input transitions and the circuit doesn't respond to floating inputs.

EDN

## Author's biography

Walter J Niewierski is a marketing manager at Harris's Semiconductor Product Div in Melbourne, FL. He previously worked for Ford Motor Co and Burroughs Corp before joining Harris eight years ago. Walt has a BSEE from the University of Michigan and enjoys coaching little-league baseball in his spare time.


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## Discover <br> Fuoronics Resources

# Fluorinert'Miquids-products that power Fluoronics Resources. 

## *Fluoronics Resources:

An exclusive 3M combination of innovative products backed by research and development, manufacturing expertise, technical data and service assistance built on more than 35
 years' experience of pioneering in fluorochemistry.

3 M has had a whole generation of experience in the development, manufacture and refinement of perfluorinated liquids. We first introduced these versatile liquids to electronics design, testing and production professionals in the fifties. Since then, Fluorinert Liquids have become the mainstays in electronic cooling, high reliability testing and vapor phase soldering.
Fluorinert Liquids, used as a direct contact heat transfer medium, offer a range of physical properties that make them particularly suitable for electronic uses. They are non-polar and exhibit no solvent action. They are colorless, low in toxicity, non-flammable and offer exceptionally high dielectric strength plus thermal and chemical stability. Most important, they have almost no chemical reactivity and they evaporate without leaving a residue on parts.

## Buy the numbers

Our FC ${ }^{\text {tw }}$ numbers - FC-40, FC-70, FC-77, etc. - are used to identify Fluorinert Liquids that offer certain physical characteristics to meet specific application needs. These FC numbers are solely 3 M designations for various fluorochemical products.

Fluorinert Liquids are being used cost-effectively in cooling, high reliability testing and vapor phase soldering operations. When you are interested in applying these versatile liquids in your own production, 3 M can provide an abundance of technical information and support.

## Technical assistance: the main benefit of Fluoronics Resources

3M offers prompt assistance to help you solve many production and testing problems. We provide comprehensive technical recommendations for specific fluids. We consult with you on the proper application equipment and help you evaluate production methods and results. Our service bulletins bring you up to date on the most recent advances in vapor phase soldering and high reliability testing. Ask us about 3M's audiovisual materials and on-site application training seminars.

## Discover Fluorinert ${ }^{\text {TM }}$ Liquids' heat transfer capability

What are your needs? A precise degree of temperature control? Fast, uniform heat transfer? High dielectric strength? Fluorinert Liquids offer the broad range of physical characteristics required in most applications.
Fluorinert Liquids are an effective direct contact heat transfer medium whether used in a liquid or vapor state. Their unique properties enable you to use them in contact with sensitive components and substrates.
Major differences between the various products in the Fluorinert Liquids family can be seen in their boiling points. These can range from $56^{\circ} \mathrm{C}$ to $253^{\circ} \mathrm{C}$. Should you need products with intermediate boiling temperatures, the 3M staff will work with you to fashion a product especially for your needs. It's an example of how 3M's Fluoronics Resources provide you with "customized" service to solve special problems.



## Fluorinert ${ }^{\text {TM }}$ Liquids achieve accurate high reliability testing

It's a small world you work in. Where time ticks in nanoseconds and dimension is measured in Angstrom units. And as circuitry becomes more complex, a greater demand is placed on testing capability - not only in speed, but in higher reliability and accuracy.

Fluorinert Liquids meet those requirements by providing a controlled temperature environment and a high degree of electrical protection. They offer maximum compatibility between

the heat transfer medium and the device under test. Fluorinert Liquids reduce testing costs by reducing testing time substantially. They do this by rapidly reaching test temperature and providing precise and uniform temperature control. You'll minimize the number of faulty units by detecting defects before they become rejects.

These liquids provide cost-effective tests such as gross leak, thermal shock, liquid burn-in, ceramic crack detection, electrical environmental, temperature calibration and failure analysis/short detection.
Fluorinert Liquids are specified in the MIL-STD's for thermal shock and gross leak testing.
THERMAL SHOCK TEST CONDITIONS

| Military Standard 883-1011 |  |  | Military Approved <br> Fluorinert Liquids |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Test <br> Condition | Hot Test <br> Step 1 | Cold Test <br> Step 2 | Hot Test <br> Step 1 | Cold Test <br> Step 2 |  |
| A | $100^{\circ} \mathrm{C}$ | $-0^{\circ} \mathrm{C}$ | Water, FC-40 | Water <br> FC-40, FC-77 |  |
| B | $125^{\circ} \mathrm{C}$ | $-55^{\circ} \mathrm{C}$ | FC-40, FC-70, <br> FC-5311 | FC-77 |  |
| C | $150^{\circ} \mathrm{C}$ | $-65^{\circ} \mathrm{C}$ | FC-40. FC-70, <br> FC-5311 | FC-77 |  |
| D | $200^{\circ} \mathrm{C}$ | $-65^{\circ} \mathrm{C}$ | FC-70, <br> FC-5311 | FC-77 |  |
| E | $150^{\circ} \mathrm{C}$ | $-195^{\circ} \mathrm{C}$ | FC-40, FC-70, <br> FC-5311 | Liq. N2 |  |
| F | $200^{\circ} \mathrm{C}$ | $-195^{\circ} \mathrm{C}$ | FC-70, <br> FC-5311 | Liq. N2 |  |

## GROSS LEAK TEST CONDITIONS

| Military Standards | Military Approved Fluorinert Liquids |  |  |
| :---: | :---: | :---: | :---: |
|  | Indicator Fluids | Detector Fluids | $\begin{aligned} & \text { Absorption } \\ & \text { Fluids } \end{aligned}$ |
| $\begin{aligned} & \overline{\text { MIL-STD }} \\ & 883-1014 \end{aligned}$ | FC-40, FC-43 | FC-72, FC-84 | Do not apply |
| $\begin{aligned} & \overline{\text { MIL-STD }} \\ & 750-1071 \end{aligned}$ | FC-40, FC-43 | FC-72, FC-84 | $\begin{gathered} \mathrm{FC}-43, \mathrm{FC}-75, \\ \mathrm{FC}-77 \end{gathered}$ |
| $\begin{aligned} & \text { MIL-STD } \\ & \text { 202-112 } \end{aligned}$ | FC-40, FC-43 | FC-72, FC-84 | Do not apply |



## Discover higher yields in vapor phase soldering

Fluorinert Liquids have been the industry's fluid of choice since the vapor phase reflow soldering (VPS) process was introduced in 1975. There are a number of good reasons for this universal acceptance. VPS with Fluorinert Liquids produces highly reliable solder joints. The system reduces reject rates, increases production, and lowers production costs. With Fluorinert Liquids, you can be assured that your products will never be exposed to a temperature higher than the selected liquid's boiling point. (See above)

You'll avoid those problems usually associated with other systems shadowing, uneven heating, and overheating. The liquids are non-flammable. Their low surface tension helps them evaporate quickly from the work pieces without leaving a residue.

VPS with Fluorinert Liquids is especially suited for boards with high mass or complex geometries. The liquid vapors completely surround the assembly and penetrate remote recesses to heat all surfaces evenly. The vapors are 15 to 20 times heavier than air so they can be contained easily within the work area. The system offers an oxy-gen-free, non-corrosive environment to minimize rejects from oxidation contamination.

Some typical applications using Fluorinert Liquids in VPS include surface mounted leaded or leadless components, through-hole leads and wire-wrap pins, lead frame attachment, reflow of electroplated solder or tin and miscellaneous metal joining.

VPS SELECTION GUIDE

| Fluorinert Liquid | Boiling Point | Typical Solders |
| :---: | :---: | :---: |
| FC- 43 | $174^{\circ} \mathrm{C} / 345^{\circ} \mathrm{F}$ | $70 \mathrm{Sn} / 18 \mathrm{~Pb} / 12 \mathrm{In}$ |
|  |  | 100 In |
|  |  | $58 \mathrm{Sn} / 42 \mathrm{In}$ |
|  |  | $58 \mathrm{Bi} / 42 \mathrm{Sn}$ |
| FC-70, FC-5311 | $215^{\circ} \mathrm{C} / 419^{\circ} \mathrm{F}$ | $63 \mathrm{Sn} / 37 \mathrm{~Pb}$ |
| FC-5312 |  | $60 \mathrm{Sn} / 40 \mathrm{~Pb}$ |
|  |  | $62 \mathrm{Sn} / 36 \mathrm{~Pb} / 2 \mathrm{Ag}$ |
| FC-71 | $253^{\circ} \mathrm{C} / 487^{\circ} \mathrm{F}$ | 100 Sn |
|  |  | $95 \mathrm{Sn} / 5 \mathrm{Ag}$ |
|  |  | $60 \mathrm{~Pb} / 40 \mathrm{Sn}$ |

## Discover the unique cooling benefits of Fluorinert ${ }^{\text {m" }}$ Liquids

As the package size decreases, your need for more efficient heat dissipation increases in proportion: 3M Fluorinert Liquids are very efficient as a direct contact heat transfer medium, with the added advantage of having the high dielectric characteristics needed to meet stringent demands of the diversified electronics industry. We offer 11 liquids with boiling points that range from $56^{\circ} \mathrm{C}$ to $253^{\circ} \mathrm{C}$.
These stable liquids allow you to maximize power density and miniaturize your package. Yet they reduce failure rates and increase reliability.
Fluorinert Liquids are used in such demanding applications as:

- Radar transmitters - Power supplies
- High voltage transformers - Lasers
- Radar klystrons • Computer modules
- Computer memories - Fuel cells

Typical properties of Fluorinert Liquids used in cooling are:

| Fluorinert Liquid FC-77 (English Units) | Liquid |  | Vapor |
| :---: | :---: | :---: | :---: |
|  | Room Temp. ( $77^{\circ} \mathrm{F}$ ) | Boiling Point (207 ${ }^{\circ} \mathrm{F}$ ) | Boiling Point $207^{\circ}$ @ $@ /$ ATM |
| Density <br> lb. $/ \mathrm{t}^{3}$ | 111 | 100 | 0.85 |
| Thermal Conductivity Btu(hr) ( $\mathrm{t}^{2}$ ) ${ }^{\circ} \mathrm{F}$ Fit) | 0.037 | 0.033 | 0.008 |
| Specific Heat Btw(lb.) ( $\left.{ }^{\circ} \mathrm{F}\right)$ | 0.25 | 0.28 | 0.23 |
| Viscosity c.p. | 1.42 | 0.46 | 0.02 |
| Coefficient of Thermal Expansion $\mathrm{tt}^{3} /\left(\mathrm{ft}^{3}\right)\left({ }^{\circ} \mathrm{F}\right)$ | 0.0008 | 0.0009 | 0.0015 |

## Discover heating/curing with Fluorinert ${ }^{\text {M }}$ Liquids

Because they maintain their vapor temperature with absolute precision, Fluorinert Liquids can be used in many heating and/or curing operations. They serve as heat transfer media in solder mask and polymer thick film applications and for polymer processing. The non-corrosive vapors will not support oxidation. Ideal where solvent flash-off is a problem.

## Discover Fluoronics Resources

3 M presents a unique short course in the use of Fluorinert ${ }^{\text {m }}$ Liquids for the electronics industry.
3 M is now offering a series of "Applied Fluoronics" tapes demonstrating how Fluorinert Liquids are used in a number of applications. See first hand how these remarkable products can improve overall electronic production.

Three cassettes are available:

1. "Applied Fluoronics: High Reliability Testing
2. "Applied Fluoronics: Vapor Phase Soldering"
3. "Applied Fluoronics: Direct Contact Cooling"
These informative VHS format tapes are available to qualified personnel in the electronics industry. Specify which cassette(s) you would like to view.

Write on your company letterhead, describing your general interest. Mail to: Fluoronics Resources, Industrial Chemical Products Division/3M, Building 223-6SE-04, 3M Center, St. Paul, MN 55144-1000.

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Number 13 in a series from Linear Technology Corporation

# Closed Loop Control with the LTC1090 Series of Data Acquisition Systems 

Guy Hoover<br>William Rempfer

## Introduction

The use of microprocessors in process control loops is quite common. A processor based control loop requires special design considerations as compared to traditional analog loops. Often a single centrally located processor will be used to control several remotely located processes. The outputs of the remote process sensors can be digitized at the sensor location and then be transmitted to the central processor. Unfortunately, transmitting digital signals typically requires one wire for each bit of resolution and requires expensive cabling. Alternatively, the sensor output can be transmitted as an analog signal to the central processor area for digitization. However, transmitting analog signals over distances can introduce errors because of noise and voltage drops in the wires.

The solution to these control loop problems can be found in the LTC1090 series of data acquisition systems. As can be seen in the schematic of Figure 2, ten bits of data can be digitized remotely and sent to the processor with only three wires plus ground. The single supply capability and the low DC current drain (1mA typ.) also simplify remote location. The LTC1090 series provides the user with blocks of $1,2,6$ or 8 10-bit channels which can be chosen according to how many sensors are located in each remote site.

The LTC1090 series is ideally suited for such process control loop applications as position control, temperature control, container filling and tension control.

## Circuit Description

The circuit of Figure 2 is a container filling control loop which has a resolution of .03 pounds with a 30 pound full scale. It
was designed to implement an automatic filling station for the model train shown in Figure 1. When S1 is closed the MC68HC05 processor reads the LTC1092. If the weight is below the preprogrammed limit in the processor then the motor drive line which controls the pump is turned on. The LTC1092 is continually read by the processor as the truck is filled, until the limit is reached. The motor drive line is then shut off. The limit may be derived in a number of ways. A fixed limit will result in filling to an absolute weight, while relative or tare weight filling can be implemented when the measured empty weight is used in the calculation of the limit. Code for this application is available upon request from Linear Technology Corporation.


Figure 1. A Typical Application. Automatic Filling at a Railroad Siding.


Figure 2. This Circuit Determines Small Weight Changes, Permitting Accurate Filling. Using the Appropriate Transducer, Containers may Range from Perfume Bottles to Railroad Cars.

The NCI 3220 strain gauge used in this circuit has a linearity specification of $.04 \%$ which makes it a good match for the .05\% linearity of the LTC1092. However, the offset and full scale of the strain gauge are only guaranteed to $10 \%$ so trims are required. The circuit is run ratiometrically so an absolute reference is not required. The strain gauge output is amplified by one-half of an LT1013 with the other half being used to buffer the resistor divider that is used for the LTC1092's V REF pin. Only one op amp is necessary to amplify the strain gauge output because of the differential inputs of the LTC1092. The 2.15M 2 resistor from pin 1 to 3 of the LT1013 is to balance the load on the strain gauge bridge. With the strain gauge zeroed, both inputs on the LTC1092 are at about 2.5V. As weight is added, the output of the LT1013 into the minus input of the LTC1092 swings toward ground. At the 30 pound full scale, the output of the LT1013 is about 100 mV above ground which results in a total swing of about 2.4V. The $2 \mu$ F mylar capacitor filters the LT1013 output eliminating the effects of vibration caused by filling the train car. (As the train car nears the full point, vibration induced noise can cause the processor to stop the filling too soon.) It is important that the processor monitors the filling process in a timely fashion to prevent overflow. The setup shown relied on a slow fill rate to solve the last problem but with the processor in the loop it is possible to give the fill algorithm some
intelligence so that it would run at a high speed to begin with and then run at a slower speed at some preset limit until the final limit is reached.

To calibrate the circuit, offset is first adjusted with no weight on the platform. Next, a known weight near full scale is used to adjust the gain. Once calibrated, variations in the supply voltage within the voltage limits of the LTC1092 should not cause additional errors.

## Summary

The LTC1090 series is well suited for use in closed loop control systems. Their low supply current and serial interface make them easy to locate remotely. With a total unadjusted error of $.05 \%$ over temperature the LTC1090 series is a good match to a wide variety of sensors. The differential inputs of the LTC1090 series can also simplify circuit design while a choice of $1,2,6$ or 8 inputs gives the user just the level of complexity that is needed.

For LTC1090 Series literature call 800.637.5545. For help with an application call (408) 432-1900, Ext. 361.

# Transistor powers low-dropout regulator 

James E Dekis<br>Maxim Integrated Products, Sunnyvale, CA and Terry Blake<br>Motorola, Schaumberg, IL

The monolithic regulator chip in Fig 1, combined with an external pnp transistor, forms a very-low-dropout regulator. The composite regulator can supply several hundred milliamps at 5 V from an input as low as 5.3 V . Such low-dropout performance suits battery-powered applications, because it extends the useful life of batteries having sloping discharge curves, such as sealed lead-acid and lithium batteries.

The monolithic regulator derives its supply current from the base circuit of the external pnp transistor. The feedback-resistor ratio sets the output voltage:

$$
\mathrm{V}_{\text {OUT }}=1.3 \mathrm{~V} \times\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right) / \mathrm{R}_{1} .
$$

If the output-voltage feedback to the chip's $\mathrm{V}_{\text {SET }}$ input is below the bandgap-reference voltage ( 1.3 V ), the supply current into $\mathrm{V}_{\text {IN }}$ (the pnp transistor's base current) increases. The transistor multiplies this base current by $\beta$ and delivers it to the load. The circuit's quiescent current is a function of the transistor's $\beta$ and load current.

When there's no load, the quiescent current is typically $10 \mu \mathrm{~A}$. For larger load currents, the quiescent current is simply the load current divided by the transistor's $\beta$. The regulator chip can sink 40 mA max. When you enable the chip's shut-down input, the circuit consumes $6 \mu \mathrm{~A}$ typ. $\mathrm{R}_{4}$ supplies current to the chip under no-load conditions.


Fig 1-A monolithic regulator chip driving a dummy load sets the base current of an external, series-pass pnp transistor; the result is a very-low-dropout regulator for batteries whose output voltage droops under load.
$R_{3}$ can limit the transistor's base current. The chip's $V_{\text {out }}$ pin will try to raise its voltage level to that of the $\mathrm{V}_{\mathrm{IN}}$ pin when the output voltage of the chip is low. Reducing $R_{3}$ has the effect of supplying larger base currents to the external transistor.

You can substitute a 2 N 2945 for the 2N2907 shown in Fig 1. With this substitution, the circuit will supply a $5 \mathrm{~V}, 100-\mathrm{mA}$ max output from a 5.1 V input.

EDN

To Vote For This Design, Circle No 750

# 8031 routines expand address space 

Robert J Ryan<br>Robert Merrill Inc, Tukwila, WA

This Design Idea corrects a problem with the Design Idea "Add two 16 -bit pointers to the $8031 \mu \mathrm{P}$ " (EDN, April 28, 1988, pg 238), by Noor Singh Khalsa. Khalsa uses the $\mu$ P's P2 port indirectly to form the upper byte, or page, of a 16 -bit address; he maintains the page
addresses in $\mu \mathrm{P}$ registers and writes the appropriate page address to P 2 just prior to moving data to or from the page.

Problems arise when you attempt to use this scheme in both your main routine and a subroutine that uses a different page. Obviously, the subroutine must save the page address being used by the main routine and restore that address to P 2 before returning control to

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2580.81 Functional Block Diagram

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## DESIGN IDEAS

the main routine. The problem is that the $\mu \mathrm{P}$ stores P2's output in a special-function register (SFR). Executing a simple read or stack push from P2 gives your subroutine the contents of P2 only when your program runs in an in-circuit emulator. In a target system, however, read and stack-push commands input the byte
on P2's pins and don't input the contents of its SFR. Only read-modify-write commands that have the SFR as a destination access P2's SFR directly when your program runs in your target system.

Figs 1a and 1b show two subroutine-entry stubs that use read-modify-write instructions, DJNZ and JBC, to


Fig 1-These three routines allow you to use an 8031's P2 special-function register to indirectly address pages of external memory in both your main routine and subroutines. The first routine (a) uses few instructions, but its execution time depends on P2's contents. The second routine (b) is faster but uses up more memory. The third routine (c) maintains a RAM image of P2's special-function register.


## DESIGN IDEAS

determine the contents of P2's SFR. Fig 1a's program uses few instructions, but the execution time depends on the SFR's contents; its loop can execute 255 times if P2's SFR holds FF $_{\text {HEx }}$. Fig 1b's program executes more quickly, but is not very memory efficient.

Fig 1c shows an alternative scheme that employs a RAM image of P2. Every time you write a page address to P2, you also update the byte P2RAM directly. Subroutines can then push the contents of P2RAM onto
the stack upon entry and restore the calling routine's page address to P2RAM and P2 upon exit. Note that you must always update P2RAM before updating P2 to ensure that the scheme in Fig 1c operates properly during interrupts.

EDN

To Vote For This Design, Circle No 746

# 555 timer turns beep into warble or chirp 

R E McCain<br>Avalon Products, Fremont, CA

The simple circuit in Fig 1 transforms the steady beep of an audible-signal device such as a Mallory Sonalert into a distinctive warble or chirp. The value of $\mathrm{C}_{2}$ determines just what tone color you'll get. With the $1-\mu \mathrm{F}$ value shown, the circuit produces a warble similar to the ring tone of an inexpensive, non-Bell phone. A $10-\mu \mathrm{F}$ value produces a chirp similar to a truck's back-up alarm. One elaboration of this circuit would be to use the second section of a 556 timer to drive a piezoelectric transducer instead of a Sonalert; that modification would vary the tone's pitch as well as the chirp rate.

EDN
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Fig 1-Adding a 555 timer to your audible-signal device's circuit will transform its plain-Jane beep into a distinctive warble or chirp.

## Single-chip $\mu \mathrm{P}$ controls resolver

Tadeusz Jarosinski<br>FMiK ERA, Warsaw, Poland

Without the aid of external circuitry, a single-chip $\mu \mathrm{P}$ can't handle all the overhead associated with controlling a resolver. The 8253 programmable-interval-timer
chips- $\mathrm{IC}_{1}$ and $\mathrm{IC}_{2}$ in Fig 1a-along with a handful of logic devices, provide sufficient speed and resolution for many applications.
The circuit in Fig 1 lowers the single-chip $\mu$ P's overhead because the $\mu \mathrm{P}$ only needs to respond to the $\overline{\text { INT signal from the motor controller, load the } 8253}$

## DESIGN IDEAS

programmable-interval timers with the appropriate divisors, and enable the timers via the START line. Normally, counter 0 of $\mathrm{IC}_{1}$ divides the system clock (CLK) by 8 . But if the single-chip $\mu \mathrm{P}$ reloads this counter for exactly one period with 16 or for two
periods with 4 , the drive circuit will subtract one clock pulse, or add one clock pulse, respectively (Fig 1b). (The single-chip $\mu \mathrm{P}$ must reload the counter with 8 immediately.) Adding or subtracting clock pulses has the effect of shifting the resolver's phase in $\pm 4-\mu \mathrm{sec}$


Fig 1-A single-chip $\boldsymbol{\mu} \boldsymbol{P}$ can control a resolver's phase in $4-\mu$ sec increments with the aid of external programmable counters and logic.

## DESIGN IDEAS

increments.
The feedback-control portion of this design begins with the feedback signal, or error, from the resolver's rotor winding. $\mathrm{IC}_{9}$ and $\mathrm{IC}_{10}$ filter this signal to a pure sinusoid and then turn it into a square wave. The phase
of this square wave, with respect to the stator-drive signal, corresponds to the rotor's position. Counter 2 in $\mathrm{IC}_{1}$ divides this square wave by 8 (mode 3 ). XOR gate $\mathrm{IC}_{4 \mathrm{~A}}$ compares the divided-feedback square wave with the rotor-drive signal from counters 0 and 1 of $\mathrm{IC}_{1}$


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## DESIGN IDEAS

(whose phase the single-chip $\mu \mathrm{P}$ can alter in $\pm 4-\mu$ sec increments). Therefore, when the BLOCK signal is active high, $\mathrm{IC}_{5 \mathrm{~A}}$ impresses the error-signal-modulated drive signal upon the motor-drive IC's Enable pin ( $\mathrm{IC}_{6}$ ). The width of this drive signal, in turn, determines the resolver's' current.
The D flip-flop, $\mathrm{IC}_{3 \mathrm{~A}}$, monitors the error signal's direction (positive or negative) and controls the drive IC's PHASE input. Note that counter 0 of $\mathrm{IC}_{2}$ can trigger the single-chip $\mu$ P's interrupt line. This counter triggers an interrupt if the error signal exceeds the allowable maximum by monitoring the output of $\mathrm{IC}_{4 \mathrm{~A}}$. A positive transition from $\mathrm{IC}_{4 \mathrm{~A}}$ starts the counter. Therefore, any error signal that lasts for less than half of the counter's period does not result in an interrupt.

The controller circuit's $2.5-\mathrm{kHz}$ stator-drive signals result from $\mathrm{IC}_{2}$ 's dividing the system clock by 200 (in mode 3). $\mathrm{IC}_{7}$ and $\mathrm{IC}_{8}$ develop the sine and cosine drive signals for the resolver's two stators.

The listing in Fig 2 is for a 12-MHz 8031 single-chip $\mu \mathrm{P}$. The major software problem to be solved is that of
synchronizing the reloading of the counter with 4 at the instant that $\mathrm{IC}_{1}$ 's counter 0 holds 7 or 6 (reloading with 16 is not as time critical as reloading with 4 is). The program achieves this synchronization by waiting exactly $6 \mu \mathrm{sec}$ (until counter 0 of $\mathrm{IC}_{1}$ contains 3 or 2) before beginning to reload the counter with 8 . The $2-\mathrm{MHz}$ resolver system clock is derived from the singlechip $\mu$ P's $12-\mathrm{MHz}$ clock; therefore, the loading of the counter and the resolver will remain synchronized (that is, the START signal ensures that the counters and flip-flops will start together), because this program ensures that two $2-\mu$ sec periods will elapse.

To synchronize the stator and rotor signals with an external event such as the closure of a limit switch, the $\mu \mathrm{P}$ must first set the BLOCK line, then incrementally shift the phase of $\mathrm{IC}_{1}$ 's counter 1 , and then read back the count of $\mathrm{IC}_{2}$ 's counter 0 to measure the phase difference between $\mathrm{IC}_{1}$ 's counters 1 and 2 .

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To Vote For This Design, Circle No 749


Fig 2-This program allows a 12-MHz 8031 single-chip $\mu P$ to control the circuit in Fig 1.

## DESIGN IDEAS

## Thermistor measures respiration rate

Ricardo Jimenez-G
San Diego State University, Calexico, CA
The circuit in Fig 1 uses a thermistor ( $\mathrm{R}_{\mathrm{TH}}$ ) that measures respiration rate when placed near a subject's
nostril. The thermistor responds to the relatively high temperature of exhaled air. The circuit measures respiration rates between 10 and 50 breaths/sec; above and below these limits, it sounds an alarm. To measure these low-frequency phenomena, the circuit uses the


Fig 1-This circuit uses a thermistor placed near a subject's nostril to measure respiration rate. The EPROMs translate the respiration period into rate.

## DESIGN IDEAS

respiration signal to gate a higher-frequency multivibrator on and off. A counter totals the multivibrator's output, and a combination of EPROMs and BCD- to 7 -segment decoders drives an LCD. Fig 2 shows the programs for the two EPROMs, $\mathrm{IC}_{7}$ and $\mathrm{IC}_{8}$.
$Q_{1}$ and its associated circuitry compose a constantcurrent generator that linearizes the response of thermistor $\mathrm{R}_{T H}$ (Fenwal UUT43J1). $\mathrm{R}_{4}$ is the equivalent resistance of the thermistor at $25^{\circ} \mathrm{C}$. The constant current is


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$$
\begin{gathered}
\mathrm{I}_{\mathrm{C}}=\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{b}}-0.7\right) / \mathrm{R}_{2} \\
\mathrm{I}_{\mathrm{C}}=0.62 \mathrm{~mA} .
\end{gathered}
$$

The resistance across the thermistor, therefore, equals this constant current multiplied by the parallel combination of $\mathrm{R}_{4}$ and the thermistor. $\mathrm{C}_{1}$ blocks dc levels from $\mathrm{IC}_{1}$, which amplifies voltage changes arising from changes in thermistor resistance. Potentiometer $\mathrm{R}_{7}$ adjusts $\mathrm{IC}_{2}$ 's gain.
$\mathrm{IC}_{2}$, a Schmitt trigger, triggers $\mathrm{IC}_{3}$, which is a one-shot having a period of approximately 0.2 sec ( $\mathrm{T}=\mathrm{R}_{8} \times \mathrm{C}_{2}$ ). The one-shot, in turn, gates on $\mathrm{IC}_{5}$, a
multivibrator operating at 10 Hz , via flip-flop $\mathrm{IC}_{4}$. The multivibrator drives $\mathrm{IC}_{6}$, a 12 -stage counter. $\mathrm{IC}_{7}$ through $\mathrm{IC}_{10}$ form a latched decoder for the counter's output. The EPROMs, $\mathrm{IC}_{7}$ and $\mathrm{IC}_{8}$, convert the respiration period measured by the counter into respiration rate or trigger the out-of-range alarm, whichever is appropriate. $\mathrm{IC}_{11}$ 's inverter gates reset the counter and latch the display drivers.


Fig 2-These EPROM programs for $I C_{7}(\boldsymbol{a})$ and $I C_{8}(\boldsymbol{b})$ effect the period-rate translations.

## DESIGN IDEAS

## Design Entry Blank

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[^11]
## NEW PRODUCTS

## INTEGRATED CIRCUITS



RAM CONTROLLERS

- Work with 40-nsec dynamic RAMs
- Dual-ported, 1M-bit capability

The 74 F 1764 and 74 F 1765 1M-bit dynamic RAM controllers offer synchronous single- and dual-port operation at 100 MHz . They provide arbitration, signal timing, and refresh address generation for $40-\mathrm{nsec}$ dynamic RAMs. The 74F1764 differs from the 74 F 1765 only in the addition of an on-chip input address latch-a useful feature for systems that employ unlatched or multiplexed address and data buses. Other features include the ability to address as much as 4 M -bits of memory via external address multiplexing and a separate refresh clock that supports adjustable refresh timing. 48-pin DIP, \$11; PLCC, \$11.75 (1000).

Signetics Corp, Box 3409, Sunnyvale, CA 94088. Phone (408) 9912000.

Circle No 351

## RESONANT CONTROL IC

- Uses frequency-modulation control scheme
- Operates to 3 MHz

Intended for resonant-mode powersupply control applications, the UC3860 features a frequency-modulated, fixed on-time control scheme. A precision 5 V reference and a highgain error amplifier controls a vari-able-frequency oscillator that operates to 3 MHz . The IC generates
temperature-stable pulses as short as 200 nsec , and contains two totempole outputs for driving the gates of power MOSFETs. Each output, which you can program to run alternately or in parallel, is capable of providing 2A current pulses. The chip also provides programmable control for soft start, undervoltage lockout, and fault management. MIL-STD-883 devices will be available in a 24 -pin DIP or a 28 -pin LCC in December 1988. A sample-pack containing a UC3860N, a multilevel pc board, a UC3611 Schottky-diode array, and descriptive material is available for $\$ 15$.

Unitrode Integrated Circuits, 7 Continental Blvd, Merrimack, NH 03054. Phone (603) 424-2410.

Circle No 352

## CUSTOM VLSI CHIP <br> Cusiom

- 80386 clock rates to 20 MHz
- 8088 clock rates to 10 MHz

The EL386-88 processor/converter chip is a high-speed CMOS IC that chip is a high-speed CMOS IC that
translates the control signals and 32-bit transfer sequences of an 32 -bit transfer sequences of an
$80386 \mu \mathrm{P}$ into the equivalent signals and 8 -bit transfer sequences of an and 8 -bit transfer sequences of an
$8088 \mu$. To the 80386 , the EL38688 appears as a 32 -bit memory or as a peripheral device operating at a $16-$ to $20-\mathrm{MHz} 80386$ clock rate. To the 8 -bit circuitry on the system's the 8-bit circuitry on the system's
mother board, the El386-88 appears as an 8-bit 8088 operating at its own

## MODEM IC

- Suitable for use in mobile radio systems
- Interfaces to $a \mu P$

The FX429 single-chip 1200-baud fast frequency-shift keying modem is targeted for use in trunked radio systems. It conforms to the UK's MPT1317/1327 Band-III trunked ra-dio-protocol specification, but you can use it in other radio- or cablebased data-link applications. The device operates in full-duplex mode at 1200 baud. While transmitting data, you can program the device to automatically generate and transmit preamble bit-reversals for link synchronization, and a 2 -byte checksum for the transmitted data. You can disable the checksum function to

clock rate. The two $\mu \mathrm{P}$ clocks can be completely asynchronous, allowing the 80386 to run at a full 20 MHz , yet address peripherals designed for $4.77-\mathrm{MHz} 8088$ IBM PC data rates. The EL386-88, which increases operating speed as much as 10 times that of the 8088 -based PC, automatically converts 32 -bit transfer requests into multiple 8 -bit transfers. The IC performs all the translation without any software changes. $\$ 50$ (1000).
Edsun Laboratories Inc, 9 Spring St, Waltham, MA 02154. Phone (617) 647-9300. TLX 853664.

Circle No 353

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CIRCLE NO 27

## DID YOU KNOW?

## EDN is distributed at every major

 electronics/computer show in the U.S., France, and Germany. EDN
## INTEGRATED CIRCUITS

allow continuous data transmission. In receive mode, the device detects the 16 -bit Sync (control channel frame synchronization) or Synt (traffic channel frame synchronization) word that's in front of the data packet. You can also program it to error-check the data, using the transmitted checksum. The FX429 has an on-chip clock oscillator that accepts a $4.032-\mathrm{MHz}$ crystal or external clock input and an 8 -bit $\mu \mathrm{P}$ compatible control/data interface. Its $1.008-\mathrm{MHz}$ output drives peripheral circuitry. Fabricated in CMOS, the device operates from a 5 V supply and draws a typical supply current of 5 mA when active. It is available in a 24 -pin DIP or a sur-face-mount package. $£ 6.78$ (1000).

Consumer Microcircuits Ltd, 1 Wheaton Rd, Witham, Essex CM8 3TD, UK. Phone (0376) 513833. TLX 99382.

Circle No 354
Mx-Com Inc, 4800 Bethania Station Rd, Winston-Salem, NC 27105. Phone (919) 744-5050.

Circle No 355


NMOS CONTROLLER

- 8-bit circuitry
- Includes on-chip ROM

Fabricated in NMOS, the SAB 80513 8-bit, single-chip microcontroller contains 16 k bytes of maskprogrammable ROM. The 80513 maintains all features of the 8051 family architecture. The controller is a low-cost upgrade for designers using the SAB 8052, particularly CMOS 16k-byte ROM users whose system doesn't need the low power

## INTEGRATED CIRCUITS

consumption of CMOS. In addition to the ROM, the SAB 80513 contains 256 bytes of RAM, four 8 -bit ports, an interrupt structure with six vectors and two programmable priority levels, a serial channel, and oscillator and clock circuitry. The SAB 80513 comes in a 40 -pin DIP and a 44-pin PLCC in both 12 - and $16-\mathrm{MHz}$ versions. $\$ 4.10$ to $\$ 4.55$ $(100,000)$.

Siemens Components, 186 Wood Ave S, Iselin, NJ 08830. Phone (201) 321-3400.

Circle No 356


## SYMMETRIC ROW DRIVER

- For AC-TFEL displays
- Data-transfer rate to 4 MHz

According to the manufacturer, the Si9560 is the first monolithic symmetric row driver to meet the requirements for ac thin-film electroluminescent (AC-TFEL) displays. Replacing two chips with one, the Si9560's symmetric drive scheme significantly reduces the differential aging effects that cause latent imaging in AC-TFEL displays, and nearly doubles the lifetime of EL displays. A front-end logic interface accepts 10.8 to 15 V serial inputs and transfers data at rates to 4 MHz . Thirty-four 230 V push-pull outputs on each driver provides a sink and source capability of greater than 100 mA . Although specifically designed for AC-TFEL displays, the high-
voltage, high-current outputs of the Si9560 can also drive large-area matrix vacuum-fluorescent and plasma displays. The IC is also useful for driving capacitive elements in inkjet printers. $\$ 25(100), \$ 5$ (OEM qty, available in September).

Siliconix Inc, 2201 Laurelwood Rd, Santa Clara, CA 95054. Phone (408) 988-8000.

Circle No 357


QUAD ANALOG SWITCH

- Contains four spdt switches
- Low On/Off leakage current

Designed for switching functions that require four spdt analog switches, the MAX333 operates from either single or dual supplies. Dual-supply operation ranges from $\pm 5$ to $\pm 18 \mathrm{~V}$. Where ground serves as the negative source, single-supply operation ranges from 10 to 30 V . The device does not require a separate 5 V logic supply and is TTL/CMOS compatible. The monolithic IC can switch signals to the supply rails. At a supply voltage of $\pm 15 \mathrm{~V}$, the MAX333 consumes 130 $\mu \mathrm{A}$ from the positive supply and 10 $\mu \mathrm{A}$ from the negative supply. The on-resistance is $140 \Omega$, and the onleakage is 0.2 nA . The off-leakage is 0.02 nA , and the turn-off time is 50 nsec. The device has break-beforemake switching action. 20-pin plastic DIP $\$ 2.95$; 20-pin military-grade ceramic DIP \$7.90, (100).

Maxim Integrated Products, 120 San Gabriel Dr, Sunnyvale, CA 94086. Phone (408) 737-7600.

Circle No 358

Because you're thinking fast... count on us for the speed you need.

## Now, 19ns settling op amps that survive saturations and shorts...

Comlinear's two new high-speed op amps bring you built-in protection against saturation. Plus simple shortcircuit protection. That means easy solutions for fast input and output amplifiers in systems where signal level or load can"t be controlled.

## use as little as $\mathbf{5 7 m W}$...

Our new 170 MHz CLC205 offers fast dynamic performance and power consumption down to 57 mW (with $\pm 5 \mathrm{~V}$ supplies). A settling time of 24 ns to $0.05 \%$ is complemented by the drive performance of a $\pm 12 \mathrm{~V}$ output swing and $\pm 50 \mathrm{~mA}$ output current.

## or drive up to $\pm 100 \mathrm{~mA}$.

For higher drive, call for our 180 MHz CLC206 which will drive up to $\pm 100 \mathrm{~mA}$ and settle in just 19 ns (to $0.1 \%$ ). It is coupled with a high slew rate of $3400 \mathrm{~V} / \mu \mathrm{s}$ and delivers a largesignal bandwidth of 70 MHz at $20 \mathrm{~V}_{\text {pp }}$.
Both of these new op amps give you saturation and short-circuit protection plus tested and guaranteed performance at half the price of other high-speed amps. Now you can be safe at high speed.

> vComlinear Corporation Solutions with speed 4800 Wheaton Drive Fort Collins, Colorado 80525 (303) 226-0500

## INTEGRATED CIRCUITS



GaAs MMIC

- 6- to 18-GHz bandwidth
- 5-dB small-signal gain

Designed for electronic-warfare applications, the HMM-11820 MMIC (monolithic microwave integrated circuit) has a small-signal gain of 5 dB over the bandwidth of 6 to 18 GHz . The chip consumes 35 mA of current and has an output power of 12 dBm at 1 dB of gain compression. The chip's VSWR is more than 2:1 at both the input and output, and the noise figure is 5.5 dB typ. The device includes de blocking at the RF input and output, and an on-chip source bias resistor. Using a distributed topology, the chip contains four $200-\mu \mathrm{m}$ MESFETs and all necessary capacitors, coils, and resistors. Standard wafer qualification exceeds MIL-STD-883C. $\$ 99$ (1000). Production quantities are scheduled for late 1988.

Harris Microwave Semiconductor, 1530 McCarthy Blvd, Milpitas, CA 95035. Phone (408) 433-2222.

Circle No 359

## QUAD COMPARATOR

## - Input impedance is $10^{12} \Omega$ <br> - Input bias current of 10 pA

An improved version of the industry standard LM339, the ALD-4302 CMOS quad comparator features an input impedance of $10^{12} \Omega$, an input bias current of 10 pA , and a TTLinput response time of 120 nsec . The comparator can sink an output current of 60 mA , yet requires only a $150-\mu \mathrm{A}$ supply current for each
comparator. The device works from a nominal 5 V supply, but will work with any single supply from 3 to 12 V , or from a dual supply of 1.5 to 6 V . Other features include an input voltage range that includes ground, and outputs that can sink or source current. You can use the outputs in a wired-OR connection without pullup resistors, or in a push-pull configuration with 2 mA of sourcing current. The ALD-4302 is also a standard cell in the company's linear ASIC program. The 14-pin device has the standard LM339 pinout. Plastic DIP, \$2.28; military-grade ceramic DIP, $\$ 4$; small-outline package, $\$ 2.80$; die form, $\$ 2.28$ (100).

Advanced Linear Devices, 1030 W Maude Ave, Sunnyvale, CA 94086. Phone (408) 720-8737. TWX 510-100-6588.

Circle No 360


## DISPLAY DRIVER

- 50 V output into 6 -pF load
- Operates at 175 MHz

Based on the company's QuickChip ASIC technology, the TKDD10P CRT (Z-axis) display driver can drive a $6-\mathrm{pF}$ load with 50 V p-p at 175 MHz . According to the company, it is the only monolithic IC capable of this performance. You can add an external impedance-matching network to extend the performance to 200 MHz . Two input connections allow either a single-ended or a differential input drive. The inputs accept RS-343 signals of $\pm 0.714 \mathrm{mV}$ p-p. By means of a $5-\mathrm{k} \Omega$ potentiometer, you can vary the gain linearly

Miniature basic switches provide versatility


The V Series of miniature, basic switches has a variety of low-cost solutions for your applications. The V7 product is a cost effective design that features printed wiring board terminals which reduce installation costs. The V10 product serves as an effective option for on-off/off-on applications and its non-snap design is interchangeable with other industry standard products.

Other features include special contacts for switching low energy circuits, and power handling capabilities of up to 25 amps . Some versions feature an operating force as low as 10 grams. The standard temperature tolerance range is $-20^{\circ}$ to $+180^{\circ} \mathrm{F}\left(-4^{\circ}\right.$ to $\left.+85^{\circ} \mathrm{C}\right)$.

For more information, contact MICRO SWITCH, The Sensor Consultants at 815-235-6600.

## Subminiature basic switches fit many different needs



The 9SM, a newer addition to the sub-miniature SM Series, was developed for applications requiring less precise switching characteristics. These miniature switches are costeffective, durable, and adaptable to a variety of different auxiliary or integral actuators.
The precision 11SM is a light weight switch and can be made with silver projection contacts for low energy circuit applications, providing a cost effective alternative to gold contacts.

The standard temperature range is $-65^{\circ}$ to $+180^{\circ} \mathrm{F}\left(-54^{\circ}\right.$ to $\left.+85^{\circ} \mathrm{C}\right)$. Certain versions are rated for logic level switching and handle power loads of up to $11 \mathrm{~A}-1 / 4$ HP 125 VAC.

For more information, contact MICRO SWITCH, The Sensor Consultants at 815-235-6600.

## C money saving ideas.



Take the long black strip of switches for example. It was the result of working together with a gaming machine manufacturer who needed to lower the cost of their product.

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customer's assembly costs in half.
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## INTROL

## CROSS DEVELOPMENT SYSTEMS

- INIROL-C Cross-Compilers
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All compiler systems include: Compiler • Cross-assembler • Support utilities • Runtime library, including multi-tasking executive $\cdot$ Linker • One year maintenance • User's manual, etc.
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## I N T R O L CORPORATION

## INTEGRATED CIRCUITS

from 0 to 80 . The TKDD10P operates from a $\pm 10 \mathrm{~V}$ power supply and comes in a 24 -pin power-tab package that features low thermal impedance. $\$ 35$ (1000).
Tektronix Inc, Integrated Circuits Div, Box 14928, Portland, OR 97214. Phone (800) 835-9433, ext 100.

Circle No 361


## PHASE SPLITTER

- Operates from 10 MHz to 3 GHz
- 2-dB typical insertion loss

Fabricated in GaAs, the APS30010 is an active $180^{\circ}$ phase splitter that operates to 3.5 GHz and features a single-port insertion loss of 2.0 dB typ. The chip has an amplitude balance of 0.3 dB , a phase balance of $1.0^{\circ}$, and a reverse isolation of 30 dB . Other features include an isolation of $20 \mathrm{~dB} \min$ between ports, a noise figure of 12 dB , a $3-\mathrm{dBm}$ compression level, and a third-order intercept point of 8 dBm . The input return loss is 20 dB typ, and the output return loss is 12 dB typ. Die form, $\$ 43.50$; 8-pin flatpack, $\$ 65$ (1000). Delivery, four to eight weeks ARO.

Anadigics Inc, 35 Technology Dr, Warren, NJ 07060. Phone (201) 6685000. TWX 510-600-5741.

Circle No 362


## 130 SERIES SWITCHING BIPOLAR TRANSISTORS YOUR ONLY ALTERNATIVE TO SWITCHMODE III*

## An alternative: Philips new switching bipolar

 transistors are the only alternative to Switchmode III by Motorola. Philips transistors are extremely rugged, have $1000 \mathrm{~V} / 40 \mathrm{~ns}$ capabilities, and meet all Switchmode III requirements. Three series are immediately available. BUT131 ( 5 amps ), BUW132 (8 amps) and BUW133 (15 amps).An improvement: Philips Statistical Process Control and advanced production techniques deliver a consistent, high-quality product, batch after batch. So you can produce a product that performs like the prototype.
Put us to the test. Send for free samples to replace the Motorola Switchmode III you're now using. We'll also include product specifications, a product cross-reference chart, and an application note.


Amperex Electronic Company, a Division of North American Philips Corporation George Washington Highway, Smithfield, RI 02917, (401) 232-0500 In Canada: Philips Electronics Ltd., 601 Milner Ave., Scarborough, Ontario M1B 1M8, (416) 292-5161
*Switchmode III is a registered trademark of Motorola, Inc.
CIRCLE NO 32

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## COMPONENTS \& POWER SUPPLIES

## DC/DC CONVERTERS

- Available in single- and multi-ple-output versions
- Powers as many as four disk drives

Operating from 48 V inputs, these de/dc converters can output as much as 350 W . Housed in packages measuring $9 \times 5 \times 2.5 \mathrm{in}$., they feature a high-efficiency flux-gate switching post regulator, monocoque construction to optimize heat transfer, a current-fed inverter topology, and surface-mount technology. The open-frame converters are available in either single- (DC) or multipleoutput (DCX) versions. The singleoutput models output from 5 to 48 V at currents in the 5.5 to 30 A range. The main output of some multi-output models can deliver as much as 50 A of tightly regulated ( $1 \%$ for line and load changes) 5 V power. In addition, multiple output versions fea-

ture two post-regulated mag-amp outputs and one low-power, 3 -terminal regulated output. The 12 V outputs have a peak current rating of 12 A , enabling them to power as many as four disk drives. You can expand units to include isolation diodes and modified current-sharing
for redundancy. Optional remote inhibit and dc power fail are available, as are cover/fan assemblies. $\$ 470$ (100).

Todd Products Corp, 50 Emjay Blvd, Brentwood, NY 11717. Phone (516) 231-3366. TWX 510-227-4905.

Circle No 365


CURRENT LOOP

- 3200-ft transmission distance capability
- Solves isolation problems

This 4 - to $20-\mathrm{mA}$ current-loop transmission system solves problems with erratic signals due to ground loops by using fiber-optic technology to isolate the loop and to ensure that it floats with respect to system ground. The loop has a $3200-\mathrm{ft}$ transmission distance capability. The Model 5911 link transmitter accepts the current input and converts it to light. The transmitter is cur-
rent-loop powered and requires no external power. The Model 5912 receiver accepts the optical signal and outputs a 4 - to $20-\mathrm{mA}$ output. The transmitter and receiver are housed in $2 \times 3$-in. aluminum blocks with mounting holes in the four corners. Terminal strips are provided for electrical (signal and power) inputs. Transmitter, $\$ 268$; receiver, $\$ 343$ (25).

Dymec Inc, 8 Lowell Ave, Winchester, MA 01890 . Phone (800) 2251511; in MA, (617) 729-7870.

Circle No 366

## REED RELAYS

- Design can withstand surfacemounting processes
- Offer a 10-VA max switching capability
Available with gull-wing or J-hook terminals, JS2 reed relays are designed specifically for surface-

mount applications. The units have a maximum switching capacity of 10 W or 10 VA . Maximum voltage and current-switching ratings for the spst NO reed switch equal 100 V and 0.5 A , respectively. The switch will carry 1A max. Standard contin-uous-duty coil voltages are $5,12,15$, and 24 V dc. Initial dielectric strength is 250 V dc min between open contacts and 500 V dc min between the relays' coil and contacts. $\$ 3.39$ (500). Delivery, stock to 10 weeks ARO.
Potter \& Brumfield Inc, 200 S Richland Creek Dr, Princeton, IN 47671. Phone (812) 386-2257.

Circle No 367


CIRCLE NO 33

## JRIE THIRD EDITION NFW Handbook For Data cquisition

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The New PCI Handbook, Burr-Brown Corp. Intelligent Instrumentation 1141 West Grant Rd., \#131, Tucson, Arizona 85705

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SWITCHING POWER SUPPLIES Engineered and Manufactured by Perfectionists!


CIRCLE NO 34


With a Pearson current monitor and an oscilloscope you can make precise amplitude and waveshape measurement of ac and pulse currents from milliamperes to kiloamperes. Currents can be measured in any conductor or beam of charged particles, including those at very high voltage levels.
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## TRANSCEIVER

- Provides an Ethernet connection to a coaxial cable
- Conforms to Ethernet IEEE802.3 specifications

The ANC-15 thin-net transceiver provides an Ethernet connection to a coaxial cable. It conforms to Ethernet/IEEE-802.3 transceiver specifications and $10 \mathrm{M}-\mathrm{bps}$ Ethernet CSMA/CD operational requirements. The unit has an integrated BNC-type T-connector that facilitates the network connection. Two integral LEDs serve as SQE
(heartbeat) and power indicators. The ANC-15 is housed in a compact aluminum die-cast package. $\$ 240$.

American Network Connections Inc, 462 Oakmead Pkwy, Sunnyvale, CA 94086. Phone (408) 7371511.

Circle No 368

## LINE CHOKES

- Feature UL recognition
- Employ high-saturation core material

The RL1283 and RL1284 Series compact power-line chokes are ULrecognized devices. The axial-lead units employ high-saturation core material, which allows them to accommodate high current levels. The chokes are available with inductance values ranging from 3.9 to $100,000 \mu \mathrm{H}$. Saturation-current values range from 81 mA to 15.5 A . The series includes almost 100 standard

values, and nonstandard values are available. The chokes are available on tape and reel to accommodate auto-insertion equipment, or with their leads preformed to user specifications. The operating range spans -55 to $+125^{\circ} \mathrm{C}$. All units are supplied with insulated sleeves. MIL-STD-27D devices are also available. $\$ 0.50(10,000)$. Delivery, stock to eight weeks ARO.
Renco Electronics Inc, 60 Jefryn Blvd E, Deer Park, NY 11729. Phone (516) 586-5566.

Circle No 369

## INDUCTOR

- Accommodates reflow soldering processes
- Available in bulk or tape-andreel packaging
The CM 1812 chip inductor has a ferrite core encased in a molded body made of heat-resistant resin. The core is mounted to solder-
coated metal terminals, which attach to pe boards. The unit can accommodate reflow soldering processes and requires no special handling. It is available in values ranging from 0.18 to $1000 \mu \mathrm{H}$, and minimum Q specs at 50 . From 0.18 to $8.2 \mu \mathrm{H}$, the standard tolerance equals $\pm 20 \%$; above $8.2 \mu \mathrm{H}$, the standard tolerance equals $\pm 10 \%$.

The unit is available in bulk or tape-and-reel type packaging . From $\$ 0.25$.

ICS Manufacturing Inc, 11661 Martens River Circle, Fountain Valley, CA 92708. Phone (800) 642-2645; in CA, (800) 247-7864.

Circle No 370


## REED RELAYS

- Feature a 1-billion cycle operating life
- Available in SIP and DIP housings
These miniature mercury-wetted reed relays can switch 50 W for a minimum of 1 billion cycles. Operating from a 5 V supply, they consume only 125 mW . Offered in either SIPs or DIPs, the spst NO relays spec a maximum switching capability of 500 V dc and 1 A . Release time, over a 0 - to $250-\mathrm{kHz}$ frequency range, specs at 1.5 msec . The relays are available with 5,12 , or 24 V dc coilvoltage ratings. Contact resistance specs at $50 \mathrm{~m} \Omega$. Clamp diodes are available as an option. $\$ 4.65$ (500). Delivery, stock to six weeks ARO.

Gordos Corp, 1000 N Second St, Rogers, AR 72756. Phone (800) 6433500 ; in AR, (501) 526-4415.

Circle No 371

## TRIMPOTS

- Have 15-turn adjustment screws
- Surface-mount devices

TS6 and TS63 Series $0.25-\mathrm{in}$. square surface-mount multiturn cermet trimmers with nominal values $\geq 100 \Omega$ feature a typical temperature coefficient of $\pm 50 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$
( $\pm 100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ maximum) over their operating temperature range of -55 to $+155^{\circ} \mathrm{C}$. Their typical contact resistance variation is $2 \%$ of their nominal value or $2 \Omega$, and their end resistance is typically $1 \Omega$. The TS6 version meets military specifications. The trimmers have a 15 -turn mechanical rotation, and a clutch mechanism that prevents damage if you overrotate them. Four variants, with different adjustment screw and contact positions, are available. The trimmers are available in E3 Series values between $10 \Omega$ and $2.2 \mathrm{M} \Omega$. The standard resistance value tolerance is $\pm 10 \%$, and $\pm 5 \%$ tolerance

trimmers are available on request. Their power rating is 0.25 W at $85^{\circ}$ C. TS6 Series, $\$ 0.95$; TS63 Series, $\$ 0.73$ (1000).

Sfernice, 199 blvd de la Madeleine, 06021 Nice Cedex, France. Phone 93446262. TLX 470261.

Circle No 372
Ohmtek, 2160 Liberty Dr, Niagara Falls, NY 14304. Phone (716) 283-4025. TWX 710-524-1653.

Circle No 373

## POWER SUPPLIES

- Provide as many as four outputs
- UL, CSA, and VDE approved

Low-profile (under 1 in.) ESPO22A 25 W switching power supplies offer as many as four outputs that are 5 and 12 V dc, and $\pm 5$ and $\pm 12 \mathrm{~V}$ dc. Available models offer dual $90-130$ / $180-260 \mathrm{~V}$ ac jumper-selectable in-

puts; a universal 90 to 260 V ac input is available as an option. All models feature $0.1 \%$ line regulation. Load regulation specs at $0.5 \%$ on the main 5 V output and $0.2 \%$ on all auxiliary outputs. Ripple and noise equals $1 \%$ typ. The operating range spans 0 to $50^{\circ} \mathrm{C}$, and the MTBF specs at 60,000 hours. All models are UL, CSA, and VDE approved. From $\$ 52$ (100).

Total Power International Inc, 418 Bridge St, Lowell, MA 01850. Phone (617) 453-7272. TLX 948617.

Circle No 374

# HICH DENSI-D 

15 through 104 poles Sub-D Connectors


76\% CONTACT DENSITY INCREASE: Qualified to MIL-C24308 and U.L. recognized. CONTACTS: Removable, machined copper alloy, gold over nickel, 5 amp. "closed entry" female, MIL-C-39029. TERMINATIONS: Crimp, solder, printed board, straight or $90^{\circ}$. INSULATOR: Polyester glass, U.L. 94 V,$~ 6$ variants of $15,26,44,62,78$ and 104 poles. SHELLS: Steel or brass with tin plate (dimpled); zinc or cadmium plate with dichromate seal, stainless steel, passivated. MOUNTING: Panel and printed board. COUPLING: Jackscrews and slide lock system. HOODS: Metal, plastic. EMI/RFI.


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## MOSFETs

- Feature a fully isolated package
- Mount directly to heat sinks

CPX200 Series H-Bridge HEXFET modules are designed to replace four TO-220-type power MOSFET devices in power supplies, motordrive applications, and servo amplifiers. On-resistances spec at $0.28 \Omega$ for the $10 \mathrm{~A}, 250 \mathrm{~V}$ CPX234A module and $0.85 \Omega$ for the $5.8 \mathrm{~A}, 500 \mathrm{~V}$ CPX254A device. Both are housed in a 9 -pin SIP and feature an electrically isolated base that allows you to mount the units directly to heat sinks or a chassis with only two screws or a body clip. Available options include zener-protected gates to eliminate overvoltage gate stressing, a Kelvin-connected source, and HEXSense lower switches for onboard current sensing. CPX234A, $\$ 10.82$; CPX254A, $\$ 10.99$ (1000). Delivery, four to six weeks.

International Rectifier, 233 Kansas St, El Segundo, CA 90245. Phone (213) 607-8939.

Circle No 375


SUPPRESSORS

- Pass transient voltages to ground within 5 nsec
- Available in 9-, 15-, 25-, and 37position versions
Data Line Protectors provide compact, reliable defense against dataline transient voltages that can damage computers and peripherals.

These subminiature D-type protectors are a connectorized package of silicon avalanche diodes and pass any transient voltage exceeding 25 V to ground in less than 5 nsec . The maximum surge power equals 2000 W , and the maximum peak current measures 75A. The protectors simply plug in between the port needing protection and the appropriate cable assembly. Standard units are available in $9-, 15-$, $25-$-, and 37 -position versions. You can specify the units to protect either all or selectively loaded pins. $\$ 30$ (1000). Delivery, 10 weeks ARO.

AMP Inc, Box 3608, Harrisburg, PA 17105. Phone (717) 564-0100.

Circle No 376


## CARD CONNECTORS

- Withstand $200^{\circ} \mathrm{C}$ temperature
- Available in a variety of termination styles
EB7D and EB8 dual-row edgeboard connectors are designed for hightemperature applications, including component burn-in ovens. A special high-temperature thermoset molding compound for the body and be-ryllium-copper material for the contacts helps them withstand $200^{\circ} \mathrm{C}$ operating temperatures. The connectors are available in a variety of terminations, including dip-solder, solder-eyelet, and card-extender styles. EB7D units are available with $6,10,12,15,18,22,36$, and 43 contacts per row; EB8 connectors offer a choice of $6,10,12,15,18,22$, 24 , or 25 contacts per side. EB8 connector with dual rows of $22 / 24$ contact positions and card-extender
terminations, $\$ 4.80$ (1000). Delivery, four to six weeks ARO.
Dale Electronics Inc, Dept 860, Box 609, Columbus NE 68601. Phone (605) 665-9301.

Circle No 377


## RELAYS

- Can switch 40A loads
- Spec a 100,000 cycle lifetime

VF4 Series electromechanical relays provide a 40A switching capability in a package measuring approximately $1 \mathrm{in} .^{3}$ The units are available in pc-board or socketmount versions. The relays feature silver contacts and are available in spst NO and spdt contact configurations. Voltage-dependent maximum switching power capability ranges from 50 to 500 W for de loads and 900 VA for ac loads. The expected life, at 40 A resistive, specs at 100,000 operations. Continuous-duty coils are available for 6,12 , and 24 V dc operation. Nominal coil-power dissipation equals 1.6 W , and initial coil-tocontact breakdown voltage equals 500 V dc. Standard VF4 relays are equipped with a black plastic dust cover. Weatherproof enclosures and cases, with integral mounting brackets, are available as options. $\$ 1.33(10,000)$ for an spst version. Delivery, stock to 10 weeks ARO.
Potter \& Brumfield Inc, 200 S Richland Creek Dr, Princeton, IN 47671. Phone (812) 386-2272.

Circle No 378

## NEW PRODUCTS

## COMPUTERS \& PERIPHERALS



## RAM DISKS

- Come in full- and half-height 5114-in. form factors
- SCSI-compatible units have access time of $1 / 2 \mathrm{msec}$
The RAMstor family of RAM disks provides full- and half-height $51 / 4-\mathrm{in}$. form factors. Their SCSI ports can sustain a 5M-byte/sec transfer rate. The disks feature an average access time of less than $1 / 2 \mathrm{msec}$. This figure includes overhead for processing the SCSI commands. The fullheight Model 5150 comes with 8 Mto 80 M -byte storage. The half-
height Model 4150 comes with 8 M to 32 M -byte capacities. You can upgrade each model in 8 M -byte increments. A processor performs disk emulation, diagnostics, and error correction. It can do single-bit and multiple-bit error detection. An optional internal battery backs up 32M bytes of memory for as long as two hours. The unrecoverable data-error rate equals 1 in $10^{5}$ power-on hours. The 8 M -byte model, $\$ 6895$.

Western Automation, 1700 N 55th St, Boulder, CO 80301. Phone (303) 449-6400. TWX 710-111-1401.

Circle No 380

## MEMORY BOARD

- Provides 512k to 16M bytes of RAM
- Operates with version 4.0 of LIM/EMS specification

The JustRAM/AT16 memory expansion board for the IBM PC/AT and compatibles contains 512 k to 16 M bytes of RAM, depending on the configuration. The board operates with version 4.0 of the Lotus/Intel/ MicroSoft Expanded Memory Specification (LIM/EMS). With three boards you can get a 32 M -byte expanded memory and 16M-byte extended memory. The board oper-

ates at 10 MHz with zero wait state and at 12 MHz with one wait state. In addition, it features switch-selectable starting addresses on 128 k byte boundaries from 0 to 16M bytes. An EMS 4.0 driver and diagnostics are standard. When operat-
ing at the 16M-byte capacity, the board requires 15 W . The board comes with a 5 -year warranty. The 2 M -byte version, $\$ 1888$; the 16 M byte version, $\$ 11,097$.

Monolithic Systems Corp, 84 Inverness Circle E, Englewood, CO 80112. Phone (303) 790-7400. TLX 221126.

Circle No 381


## IMAGE SCANNER

- Has 100- to 200-dot/in. resolution in ambient light
- Offers optional optical-charac-ter-recognition software

The Deskscan 2000 image scanner performs overhead scanning in ambient light conditions. The user-selectable resolution ranges from 100 to 200 dots/in. The unit weighs less than 6 lbs and consumes 15 W . It provides an RS-232C with 19,200baud communications and a bidirectional parallel interface. A configuration that includes cabling and a utility disk costs $\$ 659$. For a turnkey publishing system, with PC Paintbrush+ software for an IBM PC, PC/XT, or PC/AT with at least 384 k bytes of memory, the price is $\$ 695$. A package with FrontPage Personal Publisher, PC Paintbrush +, TopDOS Version 3.0, and Haba

Finder software costs $\$ 795$. Optional optical-character-recognition software is available for an additional $\$ 895$.

Chinon America Inc, 6374 Arizona Circle, Los Angeles, CA 90045. Phone (213) 216-7611. FAX 213-2167646.

Circle No 382

## CPU BOARD

- Features a $68030 \mu P$ that runs up to 30 MHz
- Provides a $4 M$-byte DRAM and a 16k-byte cache memory

The MZ7130 CPU board for the VME Bus features a $68030 \mu \mathrm{P}$ that runs to 30 MHz . It provides a 4 M byte, dual-ported dynamic RAM with up to 2 M bytes of static RAM and up to 1 M byte of EPROM. In addition, the board has a 16k-byte cache memory and a memory expansion interface. At 25 MHz it oper-

ates with zero wait states. The board's I/O lines include two RS232 Cs , a SCSI port, and a parallel I/O expansion port. The board comes with a real-time clock with battery backup, an interrupt handler, an interrupt generator, and a mailbox. The board has a 32 -bit master or slave interface to the bus and a system controller with four levels of arbitration. The board provides sockets for an optional 68881 or 68882 floating-point coprocessor. The software development packages
include Ready System's VRTX32, Wind River System's VxWorks, and Microware's System OS-9. \$5995.

Mizar Inc, 1419 Dunn Dr, Carrollton, TX 75006. Phone (214) 446-2664. TWX 510-600-4272.

Circle No 383

## A/D BOARD

- Offers four channels for the IBM PC, PC/XT, and PC/AT
- Features a 12-bit A/D converter sampling at 1 MHz

The DAS-50 A/D board provides four channels for the IBM PC, $\mathrm{PC} / \mathrm{XT}, \mathrm{PC} / \mathrm{AT}$ and compatibles. It features a 12 -bit A/D converter with a rate of 1 M samples/sec max. You can trigger a series of samples with software commands, an external trigger, or a voltage level input. The board comes with a 1M-word onboard buffer memory. It employs 16 consecutive locations in the I/O


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address space that you can start at any base address. It has five soft-ware-programmable, unipolar or bipolar input ranges: 0 to $5 \mathrm{~V}, 0$ to $10 \mathrm{~V}, \pm 2.5 \mathrm{~V}, \pm 5 \mathrm{~V}$, or $\pm 10 \mathrm{~V}$. The linear-error figure specs at 1 LSB max. There are no missing codes over the full input range. A pop-up display controls the board with such features as diagnostics, memory checking, and on-line help screens. The 256 k -word version, $\$ 1999$; the 1M-word version, $\$ 2449$.

MetraByte Corp, 440 Myles Standish Blvd, Taunton, MA 02780. Phone (617) 880-3000. TLX 503989.

Circle No 384

## DSP WORKSTATION

- Has a 20-MHz $80386 \mu P$ and an 80387 coprocessor
- Provides a 42M-byte hard disk

The DSP Workstation computing system employs a $20-\mathrm{MHz} 80386 \mu \mathrm{P}$ and a $20-\mathrm{MHz} 80387$ coprocessor. The system uses a DSP and an acquisition board based on AT\&T's WE DSP32 chip and an $8-\mathrm{kHz}$ codec. The system contains a 42 M byte hard-disk drive with an average access time of 28 msec ; a 44 M byte tape backup; a 1.2M-byte 514-in. floppy-disk drive; a choice of an additional 360 k - or a 720 k -byte $5^{1 / 4}$-in. floppy-disk drive; a Keytronics 101 keyboard; an NEC GB-1 EGA card; an NEC Multisync II high-resolution color monitor; a 2Mbyte static RAM; and HypersignalWorkstation DSP software. This software lets you acquire real-time data at 150 kHz to a RAM disk and at 70 kHz continuously using an
optional Data Translation 2821 or 2823 acquisition board. $\$ 9995$.

Hyperception Inc, 9550 Skillman St, LB 125, Dallas, TX 75243. Phone (214) 343-8525.

Circle No 385


## V. 32 MODEM

- Provides full-duplex operation at 9600 bps
- Has a port adapter for synchronous communications

The Sync-Up V. 32 IBM PC modem board conforms to the CCITT V. 32 standard. It permits full-duplex operation at 9600 bps through the use !of an echo-cancellation technique


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## COMPUTERS \& PERIPHERALS

and a trellis-code error-correction scheme. The board features a synchronous port adpater that permits synchronous communications without a separate adapter. It runs the AT command set in asynchronous mode. This lets you use communications software such as Crosstalk, Smartcom, and MirrorII. In the synchronous mode, a Sync-Up synchronous autodialing language supports both Bisync and SDLC protocols. The board is compatible with third-party IBM 3270 and 3770 terminal emulation packages. Other features include call-progress monitoring, answer and line disconnect, an onboard speaker with volume control, and built-in diagnostics. From $\$ 1295$.

Universal Data Systems, 5000 Bradford Dr, Huntsville, AL 35805. Phone (205) 721-8000. TLX 752602.

Circle No 386


## CPU CARD

- Employs a $68020 \mu$ P and 68881 math coprocessor
- Provides FIFO registers for message passing

The vendor of the PG2050 32-bit computer board offers either a stand-alone, single-board computer or a CPU card for VME Bus systems. The board runs a $68020 \mu \mathrm{P}$ and 68881 math coprocessor at 16 MHz . The sockets feature a 1 M -byte max, 150 -nsec, 16-bit, dual-ported static RAM. The 32 -bit local memory provides a 4M-byte EPROM or a 2M-byte, zero-wait-state, static RAM and a 2M-byte EPROM. The board employs a 16 -byte FIFO reg-

## COMPUTERS \& PERIPHERALS

ister for message passing. The board offers three serial I/O ports: two are front-panel asynchronous RS-232Cs and the third provides either a synchronous/asynchronous RS-232C or an RS-485 interface to the board's two P2 connectors. In addition, the board features two 8 -bit parallel I/Os, six programmable timers, and a real-time clock calendar. Gld 6000.

Philips, Industrial \& Electroacoustic Systems Div, 5600 MD Eindhoven, The Netherlands. Phone (040) 788620. TLX 35000.

Circle No 387
Philips Electronic Instruments Inc, 85 McKee Dr, Mahwah, NJ 07430. Phone (201) 529-3800.

Circle No 388

## VGA BOARD

- Provides a $800 \times 600$-pixel mode on IBM PCs
- Displays 132 text columns with a 256k-byte memory

The AST-VGA Plus VGA-adapter board for the IBM PC, PC/XT, PC/AT, and compatibles offers several graphics resolution modes. One mode provides $640 \times 480$-pixel resolution and can display 256 colors simultaneously. You can use the $800 \times 600$-pixel resolution mode with 16 simultaneously displayed colors for CAD/CAM and desktop publishing applications. In monochrome text mode, the board provides $720 \times 400$-pixel resolution with a $9 \times 16$ dot-matrix box. It displays 132 columns of text and features a 256k-byte graphics memory, but you can get up to 512 k bytes. The $3 / 4$-size board is compatible with EGA, CGA, MDA, and HGC (Hercules Graphic Card) graphic standards. The board runs with the IBM 8503-, 8512-, 8513-, and 8514-compatible displays and with NEC Multisync monitors. \$599.

AST Reasearch Inc, 2121 Alton Ave, Irvine, CA 92714. Phone (714) 863-9991. FAX 714-863-9478.

Circle No 389

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## NEW PRODUCTS

## TEST \& MEASUREMENT INSTRUMENTS



## CALCULATOR

- Solves equations repeatedly without re-entry
- Uses loops, tests, and flags to make logical decisions

When you need to solve an equation for more than one variable, the HP-32S, RPN (reverse-Polish-notation) scientific calculator for engineering professionals saves you from having to enter an equation repeatedly. Instead of rewriting the equation to isolate the dependent variable, you simply set it equal to zero and supply values for all of the
independent variables. The calculator performs numeric integration and manipulates complex numbers; it allows keystroke programming of equations for which you need repetitive solutions. The unit's 4-level stack and 27 storage registers, as well as its ability to execute program loops, to perform tests, and to use flags allow you to use it for solving problems that require logical decisions. $\$ 69.95$

Hewlett-Packard Co, 19310 Pruneridge Ave, Cupertino, CA 95014. Phone local office.

Circle No 395

## LOGIC ANALYZER

- Provides 32 25-MHz clock-rate channels
- Offers disassembler support for $13 \mu$ Ps

The K25 logic analyzer provides 32 channels with a 1 k -word-deep capture memory for each channel. It operates with internal or external clocks at rates to 25 MHz and provides four levels of triggering with restart capability as well as event and delay counting. Disassembler

packages support the following 8and 16 -bit $\mu$ Ps: Z80, 6800, 8085, 8031, 8051, 8086, 8088, 68000,

64180, 8048, and 8049. In timing mode, you can display any 17 channels; you can expand the display as much as 16 times; and you can search for any word. In list mode, you can group 32 channels and assign them a single label. You can choose among five formats for the list-mode display. The unit includes IEEE-488, RS-232C, and Centronics interfaces; it stores data and setup information in battery-backed RAM. \$2995. Delivery, 60 days ARO.

Gould Inc, 3631 Perkins Ave, Cleveland, OH 44114. Phone (800) 538-9320. TLX 196113.

Circle No 396


## MEASURING UNIT

- Adds parameter measurement to ASIC-verification tester
- Verifies continuity and measures impedance

The Parametric Measuring Unit (PMU) provides an option to the vendor's STM 5200 ASIC-verification test system. Basically, the PMU is a bipolar de source-andmeasurement subsystem that can force voltage onto one or more device pins and measure the current that flows through the pin(s); or force current through a pin and measure the voltage that appears at the pin. To prevent damage to the device under test, you can separately program the forcing quantity and
the maximum permissible values of the measured quantity. If application of the programmed forcing voltage or current causes the measured quantity to exceed its programmed limits, the unit will not allow application of the full requested forcing quantity. The voltage range is -3 to 9.995 V with a $4.88-\mathrm{mV}$ resolution. The unit can force and measure cur-
rents from 150 pA to 100 mA in six ranges. Among other things, these capabilities permit you to use the PMU to verify continuity and measure impedance. You can also use the PMU to load a device's output and measure its maximum currentdrive capability. $\$ 12,500$. STM 5200 , including 80386 -based workstation, from $\$ 39,500$.

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Circle No 397


## PC-BASED SCOPE

- Can acquire 40 M samples/sec
- Stores 256k samples in onboard RAM

The Compuscope 220 is a single-slot board for the IBM PC bus. It allows the computer to function as a 20 MHz -bandwidth oscilloscope that simultaneously samples two analog waveforms at any of seven full-scale sensitivities from 0.1 to 10 V . The board allows you to independently select the sweep rate for each chan-nel-you can set the rate from 10 nsec to $10 \mathrm{sec} / \mathrm{div}$. The maximum acquisition speed and acquisitionmemory depth on channels A and B, respectively, are 40 M and 20 M samples/sec and 256 k and 128 k bytes. The unit stores acquired data on board-not in the computer's memory. It allows mid-, pre-, and posttriggering initiated by either channel, an external source, or a keyboard command. The software supplied performs addition, subtraction, multiplication, division, integration, and differentiation of waveforms. It also performs FFTs and curve fitting, and allows you to use the board as a digital voltmeter or a frequency counter. $\$ 1995$.

Gage Applied Sciences Inc, 5905 St-François Rd, Montreal, Quebec, Canada H4S 1B6. Phone (514) 3376893. FAX (514) 337-8411.

Circle No 398

## PLD PROGRAMMER

- Connects to PC via parallel printer port
- Incorporates 23 universal pin drivers
The PLD-1100 programmer connects to an IBM PC or compatible computer through a parallel printer port. The unit, which utilizes 23 identical universal pin drivers capable of producing 0 to 25 V in $100-\mathrm{mV}$ steps and delivering peak currents as large as 5 A , can handle nearly all programmable logic devices in 20 and 24-pin DIPs without hardware modifications or the use of personality modules. The vendor distributes device-library updates on MS-DOSformat disks. An internal $\mu \mathrm{P}$ automatically calibrates the unit, performs a number of self tests, and executes diagnostic routines. A fullscreen editor allows you to edit device fuse maps, and the vector sets the programmer uses for device testing. $\$ 798$.

BP Microsystems, 10681 Haddington, Suite 190, Houston, TX 77043. Phone (800) 225-2102; in TX (713) 461-9430. TLX 1561477.

Circle No 399


## 612-DIGIT DMM CARD

- Makes 250 measurements/sec
- Measures 5 quantities

The DMM20 6½-digit multimeter on a card plugs into the vendor's PCbased modular automatic-test systems. Depending on the resolution you program, the meter will make from 0.25 to 250 measurements/sec. Quantities measured include dc and
ac (true rms) voltage and current, resistance ( 2 - or 4 -wire), and the rms value of a voltage or current consisting of de plus superimposed ac. On its most sensitive ranges, the unit can resolve 100 nV and $100 \mu \Omega$; on its least sensitive ranges, it can measure $300 \mathrm{~V} \mathrm{rms}, 500 \mathrm{~V} \mathrm{dc}$, and 20 $\mathrm{m} \Omega$. For dc measurements on the 2 V range, the error is less than
$0.001 \%$ of reading plus 20 counts for 24 hours after calibration at $\pm 1^{\circ} \mathrm{C}$ of calibration temperature. The unit specs ac-measurement accuracy from 20 Hz to $100 \mathrm{kHz}: \$ 2500$. Delivery, 90 days ARO.
Summation Inc, 11335 NE 122nd Way, Kirkland, WA 98034. Phone (206) 823-8688. TLX 152219.

Circle No 400
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## CAE \& SOFTWARE DEVELOPMENT TOOLS

## DESIGN VERIFIER

- Simulator lets you perform realtime "what if?" analysis
- Accepts input in VHDL format and shows results of changes
The VantageSpreadsheet lets you use VHDL (VLSI Hardware Description Language) statements to create a schematic, simulate the design's operation, and display the results in real time. You can select multiple windows for displaying your definitions. You can also display all or part of the resulting schematic and see waveforms at critical nodes. The package includes a complete implementation of the VHDL standard defined in the IEEE specification 1076. The package comes with VHDL-compatible library models. You can import schematics created by other CAE tools that use existing standards such as Unix, VHDL, and EDIF (Electronic Design Interchange Format). However, because the simulator uses incremental compiling

techniques, you can make changes on-the-fly and see the results immediately without going through timeconsuming, start-up procedures such as net-list extractions. When you're satisfied with the performance of the design, you can reexport the modified schematic to your standard CAE tools for full
processing. The program runs on Apollo Domain workstations and costs $\$ 30,000$ to $\$ 60,000$, depending on the host's configuration.

Vantage Analysis Systems Inc, 428-40 Christy St, Suite 201, Fremont, CA 94538. Phone (415) 6590901. FAX (415) 659-0129.

Circle No 405

## IEEE-488 SOFTWARE

- Uses command set of the HP-41 scientific calculator
- Has 15-digit precision

The Eli-488 program runs on the IBM PC, PC/XT, PC/AT and compatibles. It provides engineers and scientists with calculation and control facilities that are completely compatible with the HP-41 scientific calculator. The program can use an 8087/80287 math coprocessor, if you have one in your system, and provides 15 -digit precision for all calculations. If you have a National Instruments GPIB-PC interface board, you can control as many as 30 IEEE-488-based instruments. The program also includes additional functions to control I/O ports such as digital I/O boards, relay boards,
and ADCs and DACs. All HP-41 user-solution libraries are available on disk; they give you access to a range of fully debugged programs that were originally written for the HP-41. \$295.

Eclipse Logic Inc, Box 2003, Huntington Park, CA 90255. Phone (213) 569-6020.

Circle No 406

## DSP DESIGN TOOL

- Lets you generate, analyze, and test a DSP design
- Provides an interface to boardlevel DSP products

The Monarch menu-driven, DSP design tool provides a user interface with pull-down menus and onscreen help. The Design module lets you design and analyze FIR, IIR,
and Pole-zero filters with user-specified architecture and time- or frequency-domain properties. The View module lets you perform an FFT on specified files and display the result as 2 - or 3-dimensional plots. The Siglab module lets you generate and analyze a wide variety of test signals and systems without leaving the main program. An applications window provides access to applications programs such as the TMS320 package. Finally, the OS Shell module lets you perform standard PC-DOS file-management operations without leaving the program. Basic Monarch package, $\$ 399$.

The Athena Group Inc, 3424 NW 31st St, Gainesville, FL 32605. Phone (904) 371-2567.

Circle No 407

## REAL-TIME KERNEL

- Functions as an OS for an embedded 8051 controller system
- Performs task-timing and synchronization functions

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[^12]
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| TLX-1181-G3B | $640 \times 400$ | $0.35 \times 0.35$ | $276 \times 168 \times 12$ | T7779 | $\star$ |
| TLX-561 | $640 \times 200$ | $0.35 \times 0.49$ | $275 \times 126 \times 14$ | T7779 | TLX-562-EO |
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## Circle No 415

## Folder features timers and relays

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The vendor's $152-\mathrm{pg}$ catalog calls attention to its enlarged customerservice sections located at three separate distribution centers in Pittsburgh, PA; Rochester, MN; and Los Angeles, CA. The publication is divided into nine parts covering coaxial and multiconductor cables, cabling systems, patch panels, cabinets, connectors, cables, converters, modems, multiplexers, power equipment, switches, and other related products for installing, expanding, or modifying computer communications systems. Specific sections on local-area networking offer coax, twinax, Ethernet, cabling system, twistedpair, and multimode fiber-optic systems.

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Circle No 419

# Milt Scovill's long and winding road 

Deborah Asbrand, Associate Editor

Although Milt Scovill left Pacific Northwest Bell 13 years ago, the old Bell system's presence is never far away. Vestiges of what was once the world's largest company are scattered through the office that Scovill built adjacent to his Seattle, WA, home. Along a window ledge sits a row of glass insulators. Collectors items now, they
once perched atop the telephone poles that remain the most visible signs of the vast Bell network. A bookshelf holds early versions of the cordless telephone and the first personal paging system, or "beeper." And mounted on the wall by the door is an old bronze plaque, a genuine Bell artifact that bears the phone company's motto: "No job is
so important, no service so urgent, that we can't take time to perform our jobs safely."

In these days of fast-track careers and the pursuit of engineering as a stepping stone to marketing or management, the slow and careful immersion into engineering that the Bell system offered Scovill is long gone. So is the system itself, which


Milt Scovill parlayed an interest in radio communications and a job as a messenger boy into an engineering career. For the next 33 years, he doggedly remained an engineer first, an employee second. Now a successful consultant at age 66 , he devotes all his working hours to the engineering he loves.
was his nurturer, his teacher, and, eventually, his adversary. Throughout his 33-year career at Pacific Northwest Bell, Scovill doggedly refused to acquire the notorious "Bell-shaped head" or to acquiesce to policies that he believed compromised his engineering designs. Retired from the company since 1975, he attributes his ability to command \$140/hour consulting fees to the richness of his engineering experience during his long and stormy Bell career.
Ever since Pacific Telephone and Telegraph opened its Northwest operations in the 1890s, Seattle has been Bell country. Scovill's family lived across the street from Pacific Telephone's north garage, where the company's installers would go each day to sign out a truck and begin their rounds. His first association with the phone company came when he was a boy. Telephone in-
stallers paid him 25 cents to crawl beneath porches and thread the thick phone wires through holes drilled in the wooden slats.

Radio fascinated Scovill when he was a teenager, and his acquisition of amateur and commercial radio operator's licenses led to future associations with Bell that were more interesting and profitable. When he graduated from high school in 1939, Scovill, like most Seattlites, filled out a job application for the phone company. Within a short time, Pacific Telephone hired him as a messenger. For a while, he took engineering classes at the University of Washington by day and worked for the phone company at night.

After a hitch in the Navy during World War II, Scovill returned to Seattle and to his job at the phone company. When Pacific Telephone found out about his radio interest and his radio operator's licenses, it

${ }^{\text {cc }}$ Opportunities kept popping up for me one after the other. The company put me on a rotational training program. It was an excellent program, the best learning vehicle you could find."
relieved Scovill of his message-shuttling duties and transferred him to a job maintaining mobile-telephone equipment. Shortly after that, Scovill moved to the main training center, where he conducted FCClicense training classes.

## The start of a career

Scovill and Pacific Telephone quickly developed a mutual liking. The company's efforts to devise new means of communication fired his imagination, and the young man's interest and enthusiasm made him an ideal training candidate. "Opportunities kept popping up for me one after the other," he remembers. "The company put me on a rotational training program. It was an excellent program, the best learning vehicle you could find." College learning paled by comparison, he decided, and he never returned to campus.

In 1950, opportunity knocked again. "I got a call from the engineering department to work on the layout of the first microwave system in this part of the country," Scovill says. The engineers he'd been assigned to work with were "graduate engineers," he recalls, adding "I was more of a maverick than they were." First impressions aside, Scovill and the others quickly jelled into a team. His transfer to engineering, originally scheduled for just six weeks, became permanent.

Throughout the 1950s he continued to work for Pacific Telephone and remained involved in its microwave and mobile communications operations. Now married and supporting his wife and two daughters, Scovill began to think more about his financial security. He began buying into the Bell systems' stock options plan, which allowed employees to purchase shares for 85 cents on the dollar and without the assistance of a broker. "It was generous, and it was an excellent deal," he says. "Dividends were solid as a rock. You couldn't lose."

Over time, though, Scovill began
to realize that the stock options actually carried a much higher price. It was common knowledge that Western Electric-AT\&T's manufacturing arm-supplied all the parts that Pacific Telephone and the other Bell operating companies used. Stamped onto each Bell product were the words "Made by Western Electric." Scovill found some Western Electric products to be of the highest quality and the best purchase for his engineering designs. Western Electric's cable, for example, was among the best. "They could spin cable like you wouldn't believe, and build good, rugged black telephones," Scovill says. Other products, however, were less reliable, and for these items, he turned to other vendors.

## The road not taken

Scovill wasn't naive. He knew that the Bell system's profits didn't come from the local-service fees that the company charged consumers. The real moneymakers were longdistance service and Western Electric. What he didn't know was how far Bell employees would go to defend the policy of purchasing only Western Electric equipment once they qualified for the company's stock plan, which gave them a financial stake in the company's prosperity. He also underestimated the amount of antipathy that his actions would arouse.
Word spread about Scovill's purchases, and at times life became difficult for him. A Pacific Telephone manager once buttonholed Scovill at a party and chastised him for appearing not to understand "how this game is played." Then, after several years of satisfaction with Scovill's performance in the mic owave department, Pacific Telephone transferred him to another department. A dyed-in-the-wool company man became Scovill's new supervisor. The company's message was unmistakable: His new manager would show him how to be a true company employee.

Although he was transferred back to the microwave department two years later, Scovill's days at Bell remained "turbulent," he says. He continued to protest the mandatory Western Electric purchases. "I was under a tremendous amount of pressure." He wasn't a lone voice, however: He always had the support of his staff and his supervisor. When he's asked why he remained with the company, answers elude Scovill at first, but then he sums up his decision to stay in one word: "Stubbornness."

> After hearing descriptions of the 1950 's-era, 24-channel vacuum-tube radios used by Washington's state-police patrol, Scovill accepted the patrol's invitation to redesign the system and postponed his other plans. "It made all the old adrenaline run again."

Indeed, Scovill turned down opportunities that others might have seized. The most tempting one came in 1958 from Bill Farinon, an engineering colleague who worked for a microwave component company. Farinon planned to start his own microwave company in the Silicon Valley area, and he was recruiting engineers to join him. He put in a call to Scovill, who turned him down. "I had just dug the hole for the foundation to this house, and Bill said 'Well, fill it in and get down here,'" Scovill recounts with a smile. His Seattle roots, however, run deep. He's a third-generation Seattlite; his grandfather was born there in 1861. Leaving, he decided, was out of the question. Undoubtedly , it was a similar resoluteness that kept him at Bell.

But in 1975, when Scovill was manager of the microwave department at Pacific Northwest Bell (the name that Pacific Telephone took on in 1965), he finally tired of the conflict. For the first time in 33 years, he laid plans that didn't include Pa cific Northwest Bell. At 53, he planned to retire and finance a venture of his own, using his healthy pension to tide him over. "I had every intention of never having anything to do with the telephone business again," he says adamantly. His great love is flying, and the object of that passion for the past 25 years has been the C-2 Colonial Skimmer amphibian aircraft that he houses at Paine Field. As his first post-retirement project, he intended to manufacture and market an electronic aircraft checklist that he'd devised over the years.

## A new career

A true engineer, Scovill can't resist the lure of a problem that needs solving. So when Washington's state-police patrol approached him shortly after his retirement about consulting on a microwave project, he politely explained that he was going to market his avionics device -but he kept one ear cocked. The state patrol planned to abandon its antiquated radio system in favor of a microwave system that would more efficiently link its dispatchers with its fleet of patrol cars. The patrol invited Scovill to redesign the system on a consulting basis. After hearing descriptions of the 1950sera, 24 -channel vacuum-tube radios the patrol used, Scovill caved in. "When I saw what they wanted me to do, it made all the old adrenaline run again."

Shortly after Scovill took on the state-patrol project, Oregon Public Broadcasting called to seek his services for a microwave project it had tried and failed twice to implement. Scovill paused. He hadn't planned on making consulting a second career, but he relished working on his own terms. Even more satisfying

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was the fact that he could make a lot of money at it. "I started getting serious about it when my daily income from consulting was rapidly approaching my monthly pension check. I began to think maybe this wasn't a bad business to be in."
The final chapter in the relationship between Scovill and the telephone company came in 1980 , when happenstance brought Scovill to the attention of two lawyers working on the Justice department's antitrust case against AT\&T. The department's lawyers needed assistance in wading through the piles of technical documents associated with the case. They hired Scovill as an expert witness to decipher some of the technical material as well as AT\&T's explanation of it. As he reviewed the documents, Scovill became irate. "I began to get madder and madder at the chicanery that was being played by AT\&T," he says. In the spring of 1981, Scovill flew to Washington DC and recounted for the trial record the ramifications of the Bell system's parts procurement from Western Electric. His former supervisor at Pacific Northwest Bell also received a subpoena. In 1982, Judge Harold Greene ended the Bell era by ruling that the monopoly be dismembered.
Scovill's tumultuous days are behind him, and so is the corporate scene. His office now is a secondstory aerie that overlooks Seattle's Lake Washington. From the room's picture window, he points out the big-leaf maples, cedars, and hawthorns. Atop his L-shaped desk are a pair of binoculars for bird watching, and a ham radio. There's also a 1917 edition of Audel's Easy Lessons in Wireless Telegraphy in mint condition.
Scovill, now 66, continues to handle about four consulting projects at a time-few enough to allow him to play hooky occasionally and take his aircraft for a spin. Retirement, though, is not something he considers. "I can see now that even though I'd like to do more flying, I'll never

The final chapter in Scovill's relationship with the telephone company came in 1980, when Justice department lawyers hired him as an expert witness to decipher some of the technical material involved in the department's antitrust case against AT\&T.
totally retire. If I didn't enjoy what I was doing and feel that I was of value to my clients, I wouldn't do it." Not one to spend time looking back, he's unsentimental about the passing of the Bell era and the breakup of AT\&T.
The $5 \times 5$-ft topographical map mounted on one of Scovill's office walls bears mute testimony to his past and present relationship with the now-defunct Bell system. Scovill had bought the map from Pacific Northwest Bell when it sold its older office furniture and other equipment after relocating to new headquarters. The map once hung in the engineering department where Scovill worked, and neat black lines on the map's surface trace the labyrinthine pattern of the company's microwave facilities throughout the northwest.

Undaunted by the magnitude of his former employer's presence, Scovill has added to the map since he acquired it. In addition to the original lines, he's plotted a second set of lines that represent his own, post-Bell microwave projects. The added lines are fewer in number than the original, but they're marked just as boldly. EDN

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## LOOKING AHEAD

EDITED BY CYNTHIA B RETTIG

## Surveys quiz CASE users and vendors

What do CASE-tool users want? And what do vendors think they ought to want? Strategic Focus, a consulting firm based in Milpitas, CA, interviewed 400 CASE-tool users and some of the sellers of those tools. It found that users think the tools often lack some very highly desirable features. They said that the majority of individual packages doesn't address the entire range of requirements for developing software from beginning to end. Users also wanted to see more interfaces between tools. And they wanted better integration of various tools needed throughout the development cycle, including a uniform interface to database management systems.

Learning how to use CASE tools takes too long, some users said. Others pointed out that the initial costs are high enough to require some kind of formal justification to superiors prior to purchase and a great deal of pressure to produce afterward. Users would also like better means for measuring productivity when using CASE packages. Because most of the CASE-tool companies are relatively small, support for customers at remote sites is generally very weak.

On the positive side, users identified the five top reasons why they turn to CASE tools in the first place. Of those polled, $37 \%$ said they use these tools to track the users' requirements throughout the entire development period. And $35 \%$ found that when using CASE tools, they can communicate better with end users. Among the other benefits derived from using CASE tools were improved quality of the final products, reusable software modules, and reduction in the time required

## HOW USERS RATE CASE TOOLS

| LIFE CYCLE PHASE | SATISFIED (\%) | DISSATISFIED (\%) |
| :--- | :---: | :---: |
| STRATEGIC PLANNING | 30 | 70 |
| ANALYSIS | 65 | 35 |
| DESIGN | 70 | 30 |
| PROTOTYPING | 66 | 34 |
| CODE GENERATION | 50 | 50 |
| DEBUGGING | 54 | 46 |
| PERFORMANCE TUNING | 37 | 63 |
| SYSTEM BUILDING | 50 | 50 |
| OPERATION/MAINTENANCE | 50 | 50 |
| DOCUMENTATION | 61 | 39 |
| REVERSE ENGINEERING | 0 | 100 |
| PROJECT MANAGEMENT | 50 | 50 |
| CONFIGURATION MANAGEMENT | 46 | 54 |
| SOURCE CODE CONTROL | 50 | 50 |
|  |  |  |

to complete a given project.
When Strategic Focus surveyed the vendors themselves, it found them fairly aware of the relative importance of various features and services that will be keys to their success in the next few years. They said they considered customer education and support very significant. They also believe that it's good to include in the packages ways to measure improved productivity. Building and maintaining a strong
direct sales force and establishing a strong company image are also critical. Strategic Focus sees a major shakedown coming in the CASE-tool marketplace, and it predicts that, of the more than 55 CASE vendors currently vying for market share in the US, it's very possible that only a third of them will still be around in 1991.

## PC-board ATE market to reach $\$ 1.1 \mathrm{~B}$ by ' 93

Market Intelligence Research Company (Mountain View, CA) estimates that the world market for automatic test equipment (ATE) available on printed-circuit boards and subassemblies will reach $\$ 1.1$ billion by 1993. MIRC expects that the fastest growing area of the market over the next few years will be the segment that makes and sells defect analyzers for manufactured goods. This segment should grow at a compound annual growth rate of $24 \%$. Although functional-testing
products has recently lost market share, the market researchers predict that there will be renewed interest in those functional-testing products in the near future.

MIRC characterizes the pc-board ATE market as highly competitive. It discerns a trend toward combining and generally upgrading the types of boards currently available. These types include bareboard, incircuit, and performance boards, as well as functional devices and defect analyzers for manufacturers' goods.

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